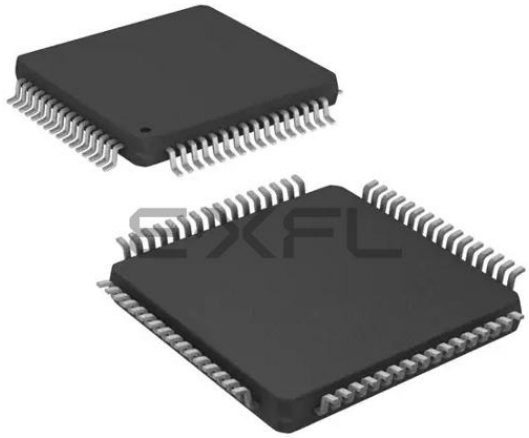

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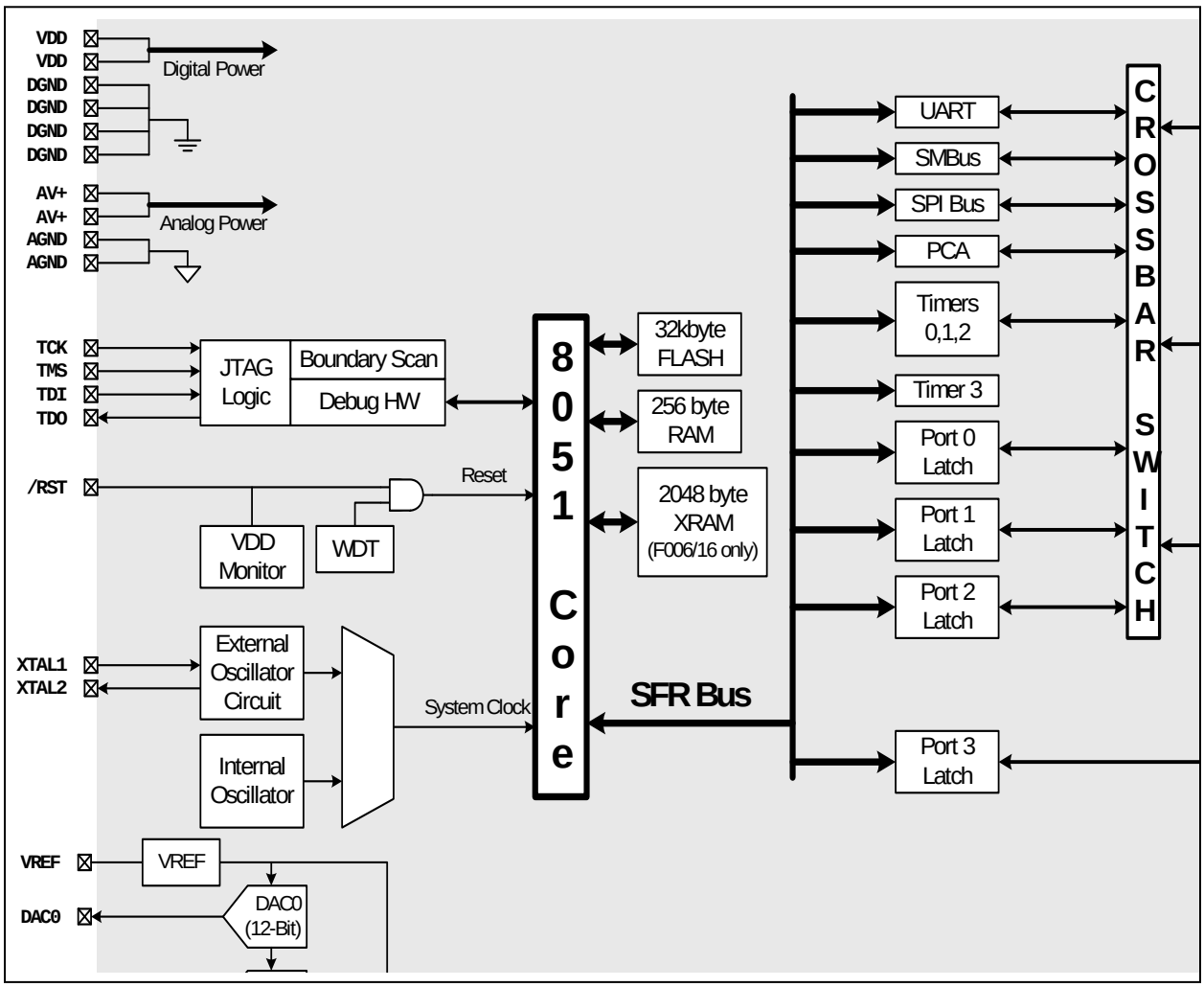
What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	8051
Core Size	8-Bit
Speed	25MHz
Connectivity	SMBus (2-Wire/I²C), SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, Temp Sensor, WDT
Number of I/O	32
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	2.25K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 3.6V
Data Converters	A/D 8x12b; D/A 2x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	64-TQFP
Supplier Device Package	64-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/c8051f005

Figure 1.2. C8051F001/06/11/16 Block Diagram



1.1.3. Additional Features

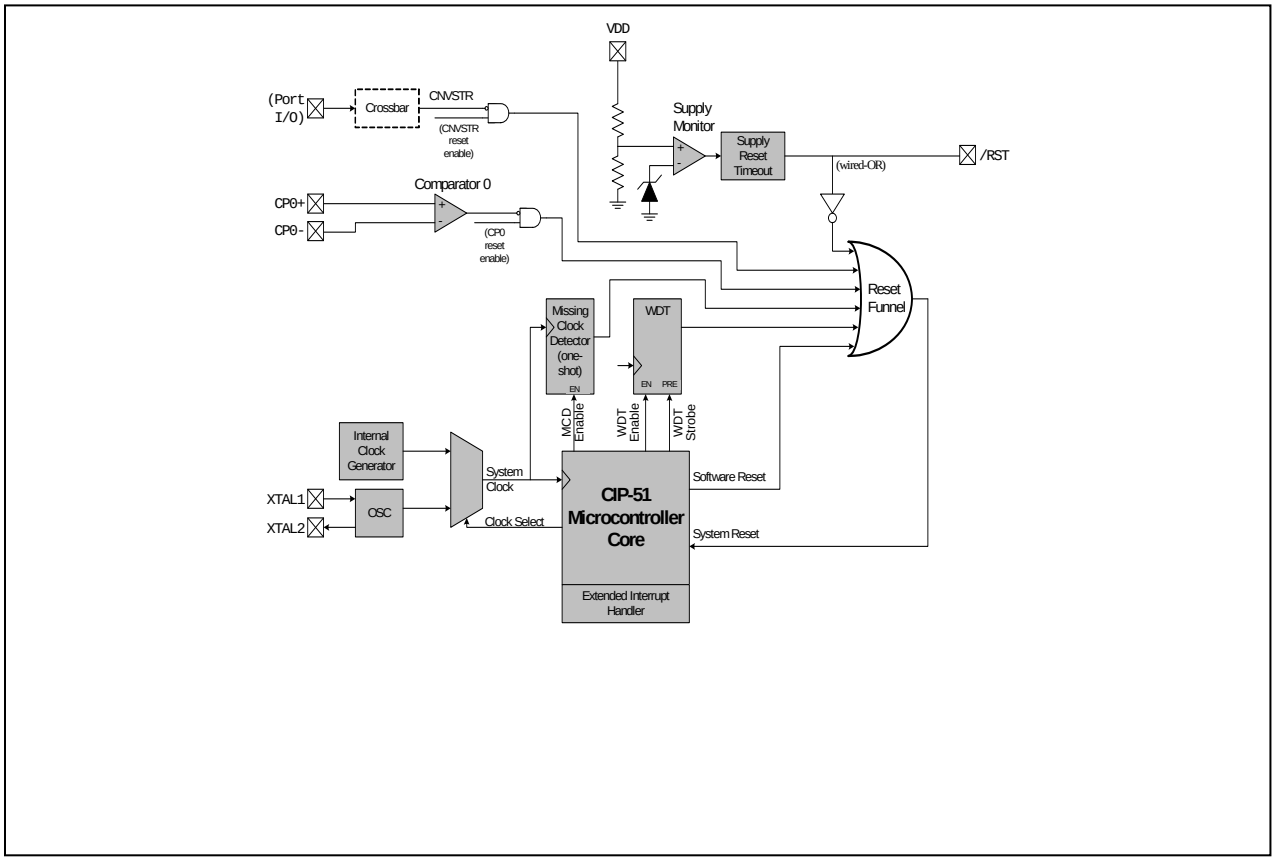
The C8051F000 MCU family has several key enhancements both inside and outside the CIP-51 core to improve its overall performance and ease of use in the end applications.

The extended interrupt handler provides 21 interrupt sources into the CIP-51 (as opposed to 7 for the standard 8051), allowing the numerous analog and digital peripherals to interrupt the controller. An interrupt driven system requires less intervention by the MCU, giving it more effective throughput. The extra interrupt sources are very useful when building multi-tasking, real-time systems.

There are up to seven reset sources for the MCU: an on-board VDD monitor, a Watchdog Timer, a missing clock detector, a voltage level detection from Comparator 0, a forced software reset, the CNVSTR pin, and the /RST pin. The /RST pin is bi-directional, accommodating an external reset, or allowing the internally generated POR to be output on the /RST pin. Each reset source except for the VDD monitor and Reset Input Pin may be disabled by the user in software. The WDT may be permanently enabled in software after a power-on reset during MCU initialization.

The MCU has an internal, stand alone clock generator which is used by default as the system clock after any reset. If desired, the clock source may be switched on the fly to the external oscillator, which can use a crystal, ceramic resonator, capacitor, RC, or external clock source to generate the system clock. This can be extremely useful in low power applications, allowing the MCU to run from a slow (power saving) external crystal source, while periodically switching to the fast (up to 16MHz) internal oscillator as needed.

Figure 1.5. On-Board Clock and Reset



1.8. Comparators and DACs

The C8051F000 MCU Family has two 12-bit DACs and two comparators on chip (the second comparator, CP1, is not bonded out on the F002, F007, F012, and F017). The MCU data and control interface to each comparator and DAC is via the Special Function Registers. The MCU can place any DAC or comparator in low power shutdown mode.

The comparators have software programmable hysteresis. Each comparator can generate an interrupt on its rising edge, falling edge, or both. The comparators' output state can also be polled in software. These interrupts are capable of waking up the MCU from idle mode. The comparator outputs can be programmed to appear on the Port I/O pins via the Crossbar.

The DACs are voltage output mode and use the same voltage reference as the ADC. They are especially useful as references for the comparators or offsets for the differential inputs of the ADC.

Figure 1.11. Comparator and DAC Diagram

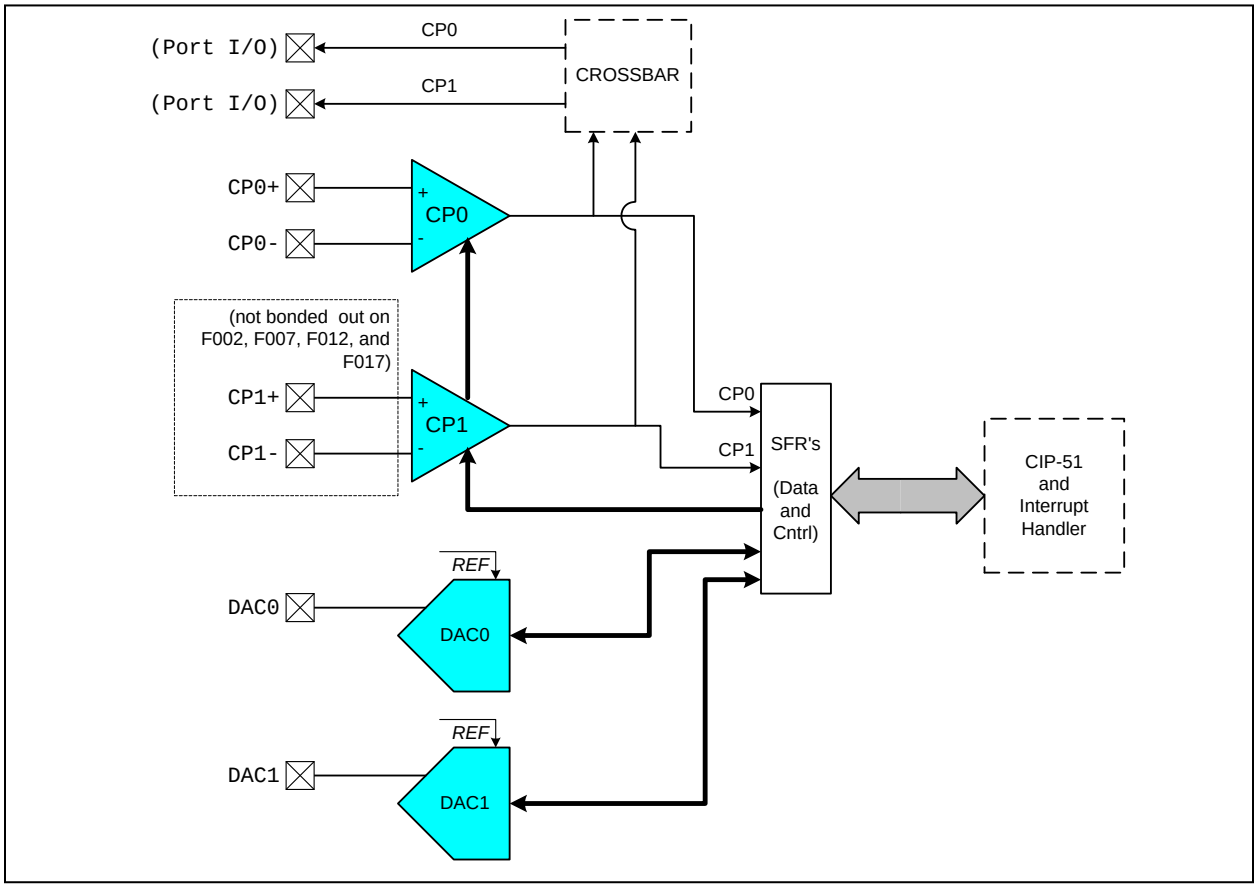
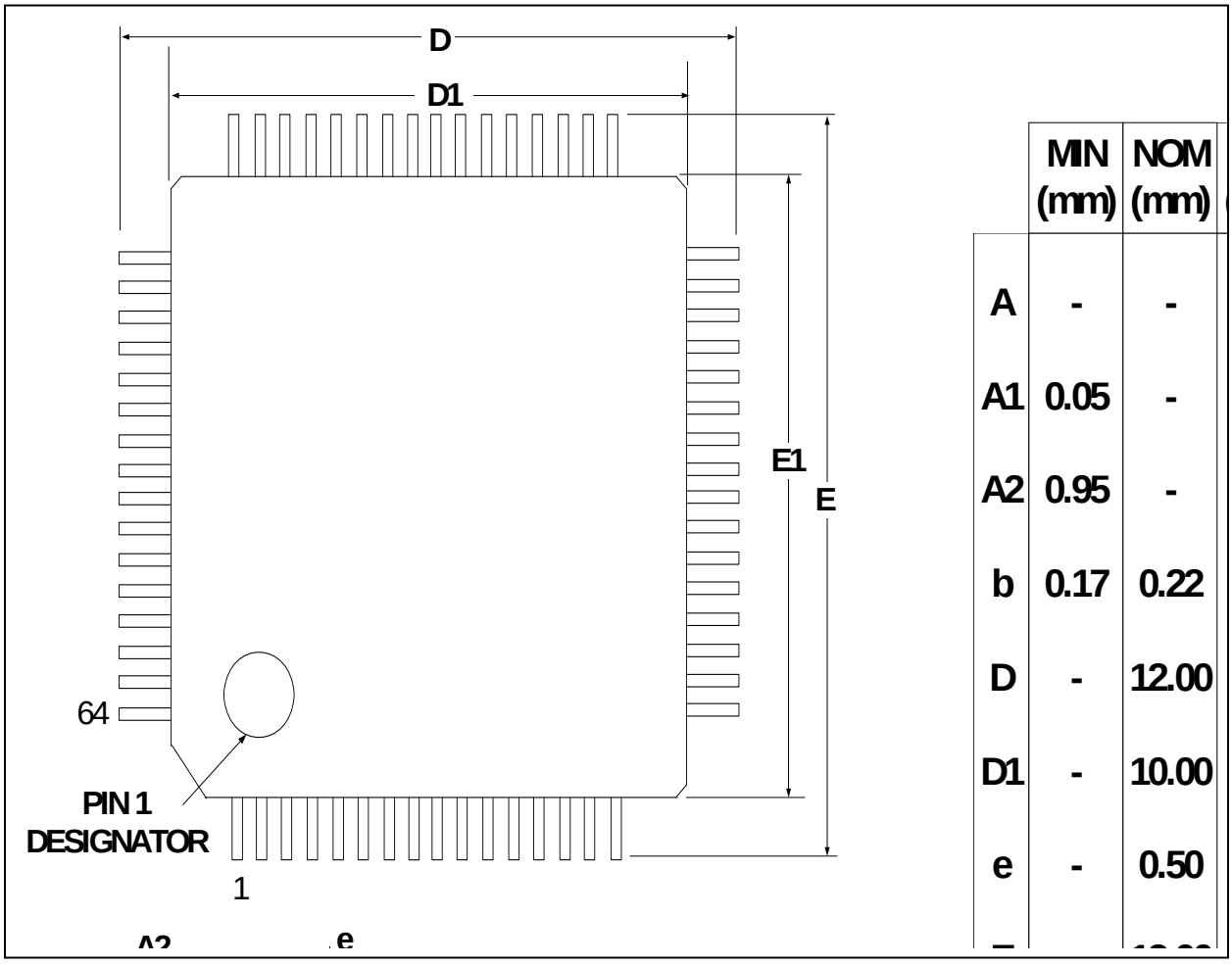


Figure 4.2. TQFP-64 Package Drawing



	MIN (mm)	NOM (mm)
A	-	-
A1	0.05	-
A2	0.95	-
b	0.17	0.22
D	-	12.00
D1	-	10.00
e	-	0.50

Figure 5.8. ADC0H: ADC Data Word MSB Register (C8051F00x)

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xBF

Bits7-0: ADC Data Word Bits
For ADLJST = 1: Upper 8-bits of the 12-bit ADC Data Word.
For ADLJST = 0: Bits7-4 are the sign extension of Bit3. Bits 3-0 are the upper 4-bits of the 12-bit ADC Data Word.

Figure 5.9. ADC0L: ADC Data Word LSB Register (C8051F00x)

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xBE

Bits7-0: ADC Data Word Bits
For ADLJST = 1: Bits7-4 are the lower 4-bits of the 12-bit ADC Data Word. Bits3-0 will always read 0.
For ADLJST = 0: Bits7-0 are the lower 8-bits of the 12-bit ADC Data Word.

NOTE: Resulting 12-bit ADC Data Word appears in the ADC Data Word Registers as follows:
ADC0H[3:0]:ADC0L[7:0], if ADLJST = 0
(ADC0H[7:4] will be sign extension of ADC0H.3 if a differential reading, otherwise = 0000b)

ADC0H[7:0]:ADC0L[7:4], if ADLJST = 1
(ADC0L[3:0] = 0000b)

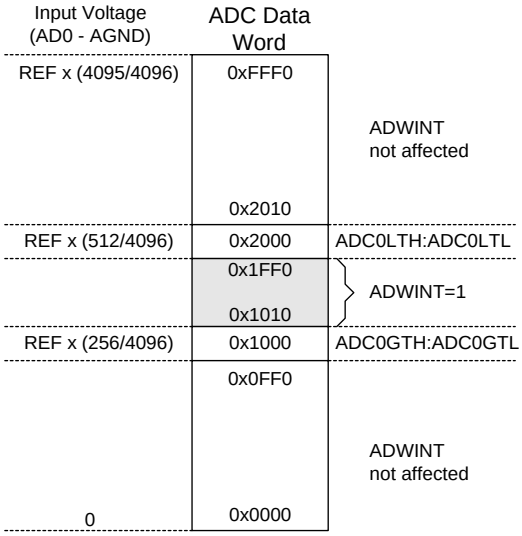
EXAMPLE: ADC Data Word Conversion Map, AIN0 Input in Single-Ended Mode
(AMX0CF=0x00, AMX0SL=0x00)

AIN0 – AGND (Volts)	**ADC0H:ADC0L** (ADLJST = 0)	**ADC0H:ADC0L** (ADLJST = 1)
REF x (4095/4096)	0x0FFF	0xFFF0
REF x ½	0x0800	0x8000
REF x (2047/4096)	0x07FF	0x7FF0
0	0x0000	0x0000

EXAMPLE: ADC Data Word Conversion Map, AIN0-AIN1 Differential Input Pair
(AMX0CF=0x01, AMX0SL=0x00)

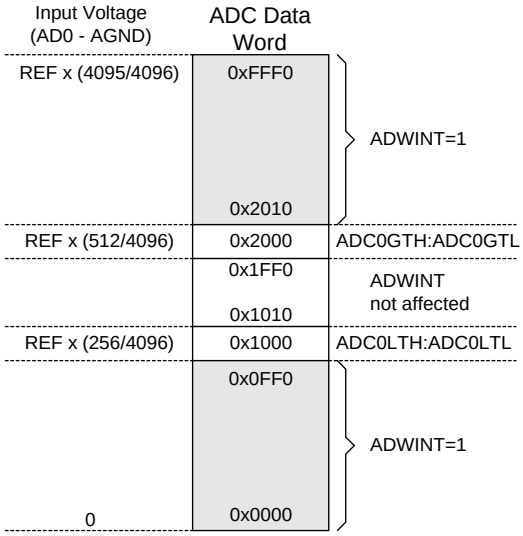
AIN0 – AIN1 (Volts)	**ADC0H:ADC0L** (ADLJST = 0)	**ADC0H:ADC0L** (ADLJST = 1)
REF x (2047/2048)	0x07FF	0x7FF0
0	0x0000	0x0000
-REF x (1/2048)	0xFFFF	0xFFF0
-REF	0xF800	0x8000

Figure 5.15. 12-Bit ADC Window Interrupt Examples, Left Justified Data



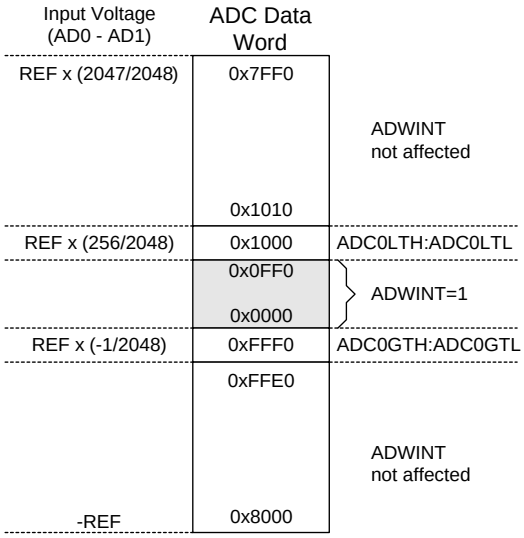
Given:
AMX0SL = 0x00, AMX0CF = 0x00, ADLJST = 1,
ADC0LTH:ADC0LTL = 0x2000,
ADC0GTH:ADC0GTL = 0x1000.

An ADC End of Conversion will cause an ADC Window Compare Interrupt (ADWINT=1) if the resulting ADC Data Word is < 0x2000 and > 0x1000.



Given:
AMX0SL = 0x00, AMX0CF = 0x00, ADLJST = 1,
ADC0LTH:ADC0LTL = 0x1000,
ADC0GTH:ADC0GTL = 0x2000.

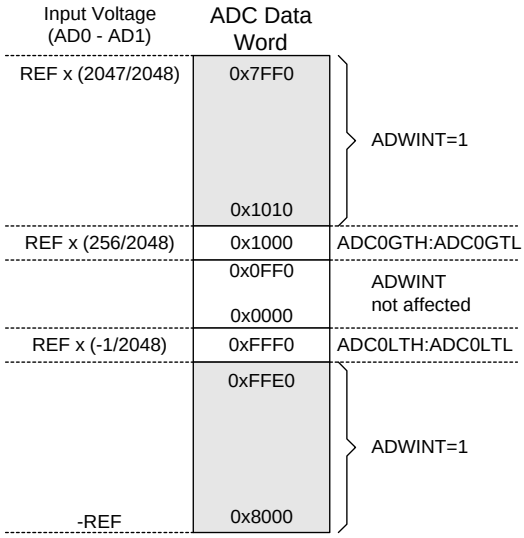
An ADC End of Conversion will cause an ADC Window Compare Interrupt (ADWINT=1) if the resulting ADC Data Word is < 0x1000 or > 0x2000.



Given:

AMX0SL = 0x00, AMX0CF = 0x01, ADLJST = 1,
ADC0LTH:ADC0LTL = 0x1000,
ADC0GTH:ADC0GTL = 0xFFF0.

An ADC End of Conversion will cause an ADC Window Compare Interrupt (ADWINT=1) if the resulting ADC Data Word is < 0x1000 and > 0xFFF0. (Two's Complement math.)



Given:

AMX0SL = 0x00, AMX0CF = 0x01, ADLJST = 1,
ADC0LTH:ADC0LTH = 0xFFF0,
ADC0GTH:ADC0GTL = 0x1000.

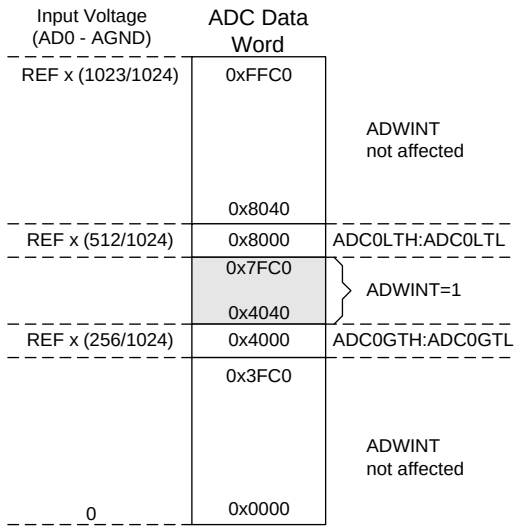
An ADC End of Conversion will cause an ADC Window Compare Interrupt (ADWINT=1) if the resulting ADC Data Word is < 0xFFF0 or > 0x1000. (Two's Complement math.)

Table 5.1. 12-Bit ADC Electrical Characteristics

VDD = 3.0V, AV+ = 3.0V, VREF = 2.40V (REFBE=0), PGA Gain = 1, -40°C to +85°C unless otherwise specified.

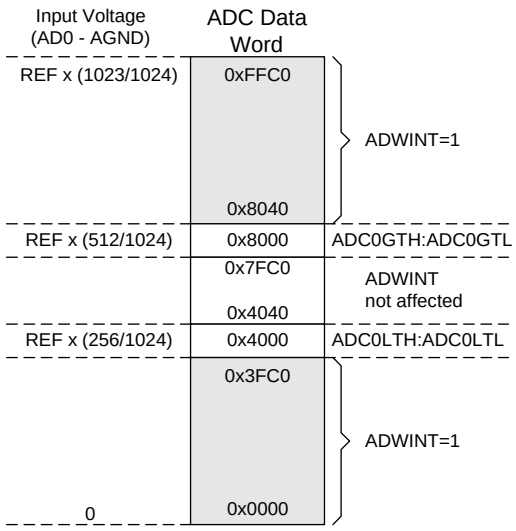
PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
DC ACCURACY					
Resolution			12		bits
Integral Nonlinearity				± 1	LSB
Differential Nonlinearity	Guaranteed Monotonic			± 1	LSB
Offset Error			-3 ± 1		LSB
Full Scale Error	Differential mode		-7 ± 3		LSB
Offset Temperature Coefficient			± 0.25		ppm/°C
DYNAMIC PERFORMANCE (10kHz sine-wave input, 0 to –1dB of full scale, 100ksps)					
Signal-to-Noise Plus Distortion		66	69		dB
Total Harmonic Distortion	Up to the 5th harmonic		-75		dB
Spurious-Free Dynamic Range			80		dB
CONVERSION RATE					
Conversion Time in SAR Clocks		16			clocks
SAR Clock Frequency	C8051F000, ‘F001, ‘F002 C8051F005, ‘F006, ‘F007			2.0 2.5	MHz MHz
Track/Hold Acquisition Time		1.5			µs
Throughput Rate				100	ksps
ANALOG INPUTS					
Voltage Conversion Range	Single-ended Mode (AINn – AGND) Differential Mode |(AINn+) – (AINm-)|	0		VREF - 1LSB	V
Input Voltage	Any AINn pin	AGND		AV+	V
Input Capacitance			10		pF
TEMPERATURE SENSOR					
Linearity			± 0.20		°C
Absolute Accuracy			± 3		°C
Gain	PGA Gain = 1		2.86		mV/°C
Gain Error (±1σ)	PGA Gain = 1		± 33.5		µV/°C
Offset	PGA Gain = 1, Temp = 0°C		776		mV
Offset Error (±1σ)	PGA Gain = 1, Temp = 0°C		± 8.51		mV
POWER SPECIFICATIONS					
Power Supply Current (AV+ supplied to ADC)	Operating Mode, 100ksps		450	900	µA
Power Supply Rejection			± 0.3		mV/V

Figure 6.15. 10-Bit ADC Window Interrupt Examples, Left Justified Data



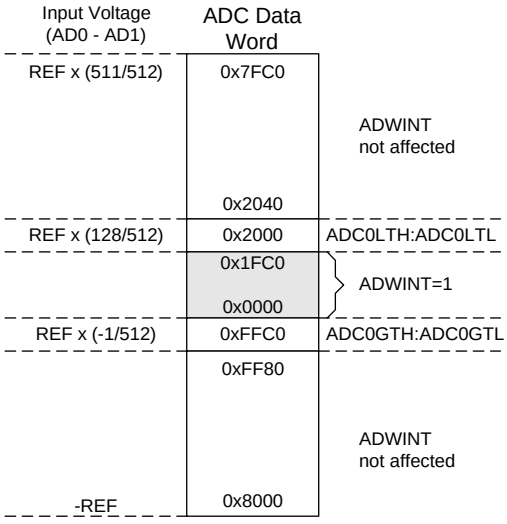
Given:
AMX0SL = 0x00, AMX0CF = 0x00, ADLJST = 1,
ADC0LTH:ADC0LTL = 0x8000,
ADC0GTH:ADC0GTL = 0x4000.

An ADC End of Conversion will cause an ADC Window Compare Interrupt (ADWINT=1) if the resulting ADC Data Word is < 0x8000 and > 0x4000.



Given:
AMX0SL = 0x00, AMX0CF = 0x00, ADLJST = 1,
ADC0LTH:ADC0LTL = 0x4000,
ADC0GTH:ADC0GTL = 0x8000.

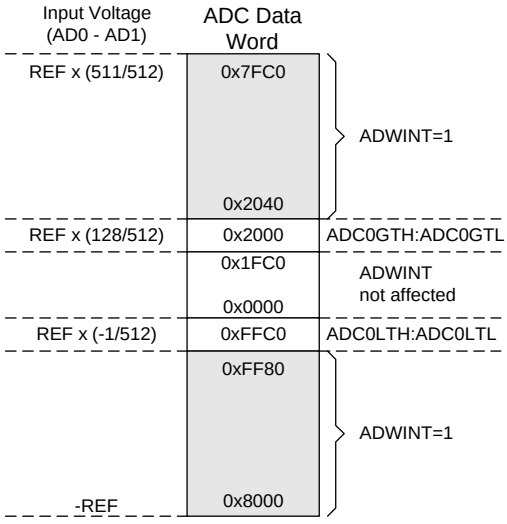
An ADC End of Conversion will cause an ADC Window Compare Interrupt (ADWINT=1) if the resulting ADC Data Word is < 0x4000 or > 0x8000.



Given:

AMX0SL = 0x00, AMX0CF = 0x01, ADLJST = 1,
ADC0LTH:ADC0LTL = 0x2000,
ADC0GTH:ADC0GTL = 0xFFC0.

An ADC End of Conversion will cause an ADC Window Compare Interrupt (ADWINT=1) if the resulting ADC Data Word is < 0x2000 and > 0xFFC0. (Two's Complement math.)



Given:

AMX0SL = 0x00, AMX0CF = 0x01, ADLJST = 1,
ADC0LTH:ADC0LTH = 0xFFC0,
ADC0GTH:ADC0GTL = 0x2000.

An ADC End of Conversion will cause an ADC Window Compare Interrupt (ADWINT=1) if the resulting ADC Data Word is < 0xFFC0 or > 0x2000. (Two's Complement math.)

10. CIP-51 CPU

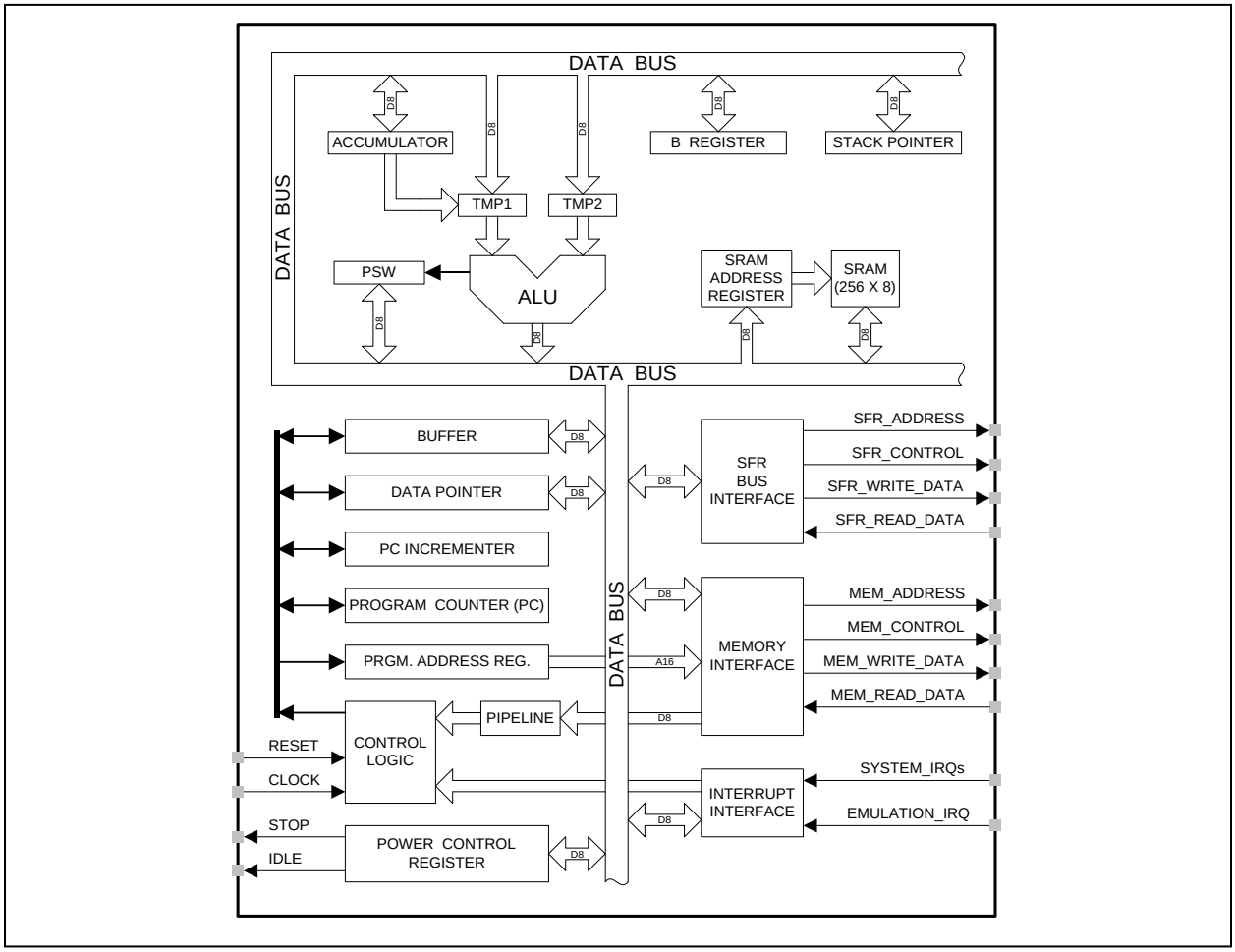
The MCUs' system CPU is the CIP-51. The CIP-51 is fully compatible with the MCS-51™ instruction set. Standard 803x/805x assemblers and compilers can be used to develop software. The MCU family has a superset of all the peripherals included with a standard 8051. Included are four 16-bit counter/timers (see description in Section 19), a full-duplex UART (see description in Section 18), 256 bytes of internal RAM, 128 byte Special Function Register (SFR) address space (see Section 10.3), and four byte-wide I/O Ports (see description in Section 14). The CIP-51 also includes on-chip debug hardware (see description in Section 21), and interfaces directly with the MCUs' analog and digital subsystems providing a complete data acquisition or control-system solution in a single integrated circuit.

Features

The CIP-51 Microcontroller core implements the standard 8051 organization and peripherals as well as additional custom peripherals and functions to extend its capability (see Figure 10.1 for a block diagram). The CIP-51 includes the following features:

- Fully Compatible with MCS-51 Instruction Set
- 25 MIPS Peak Throughput with 25MHz Clock
- 0 to 25MHz Clock Frequency (on 'F0x5/6/7)
- Four Byte-Wide I/O Ports
- Extended Interrupt Handler
- Reset Input
- Power Management Modes
- On-chip Debug Circuitry
- Program and Data Memory Security

Figure 10.1. CIP-51 Block Diagram



10.3. SPECIAL FUNCTION REGISTERS

The direct-access data memory locations from 0x80 to 0xFF constitute the special function registers (SFRs). The SFRs provide control and data exchange with the CIP-51's resources and peripherals. The CIP-51 duplicates the SFRs found in a typical 8051 implementation as well as implementing additional SFRs used to configure and access the sub-systems unique to the MCU. This allows the addition of new functionality while retaining compatibility with the MCS-51™ instruction set. Table 10.3 lists the SFRs implemented in the CIP-51 System Controller.

The SFR registers are accessed any time the direct addressing mode is used to access memory locations from 0x80 to 0xFF. SFRs with addresses ending in 0x0 or 0x8 (e.g. P0, TCON, P1, SCON, IE, etc.) are bit-addressable as well as byte-addressable. All other SFRs are byte-addressable only. Unoccupied addresses in the SFR space are reserved for future use. Accessing these areas will have an indeterminate effect and should be avoided. Refer to the corresponding pages of the datasheet, as indicated in Table 10.3, for a detailed description of each register.

Table 10.2. Special Function Register Memory Map

F8	SPI0CN	PCA0H	PCA0CPH0	PCA0CPH1	PCA0CPH2	PCA0CPH3	PCA0CPH4	WDTCN
F0	B						EIP1	EIP2
E8	ADC0CN	PCA0L	PCA0CPL0	PCA0CPL1	PCA0CPL2	PCA0CPL3	PCA0CPL4	RSTSRC
E0	ACC	XBR0	XBR1	XBR2			EIE1	EIE2
D8	PCA0CN	PCA0MD	PCA0CPM0	PCA0CPM1	PCA0CPM2	PCA0CPM3	PCA0CPM4	
D0	PSW	REF0CN	DAC0L	DAC0H	DAC0CN	DAC1L	DAC1H	DAC1CN
C8	T2CON		RCAP2L	RCAP2H	TL2	TH2		SMB0CR
C0	SMB0CN	SMB0STA	SMB0DAT	SMB0ADR	ADC0GTL	ADC0GTH	ADC0LTL	ADC0LTH
B8	IP		AMX0CF	AMX0SL	ADC0CF		ADC0L	ADC0H
B0	P3	OSCXCN	OSCICN				FLSCL	FLACL***
A8	IE					PRT1IF		EMI0CN***
A0	P2				PRT0CF	PRT1CF	PRT2CF	PRT3CF
98	SCON	SBUF	SPI0CFG	SPI0DAT		SPI0CKR	CPT0CN	CPT1CN
90	P1	TMR3CN	TMR3RLL	TMR3RLH	TMR3L	TMR3H		
88	TCON	TMOD	TL0	TL1	TH0	TH1	CKCON	PSCTL
80	P0	SP	DPL	DPH				PCON
	0(8)	1(9)	2(A)	3(B)	4(C)	5(D)	6(E)	7(F)

Bit Addressable

Table 10.3. Special Function Registers

SFRs are listed in alphabetical order. All undefined SFR locations are reserved.
* Refers to a register in the C8051F000/1/2/5/6/7 only.
** Refers to a register in the C8051F010/1/2/5/6/7 only.
*** Refers to a register in the C8051F005/06/07/15/16/17 only.

Address	Register	Description	Page No.
0xE0	ACC	Accumulator	76
0xBC	ADC0CF	ADC Configuration	33*, 42**
0xE8	ADC0CN	ADC Control	34*, 45**
0xC5	ADC0GTH	ADC Greater-Than Data Word (High Byte)	36*, 47**
0xC4	ADC0GTL	ADC Greater-Than Data Word (Low Byte)	36*, 47**
0xBF	ADC0H	ADC Data Word (High Byte)	35*, 46**
0xBE	ADC0L	ADC Data Word (Low Byte)	35*, 46**

Figure 11.4. FLSCL: Flash Memory Timing Prescaler

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
FOSE	FRAE	-	-	FLASCL				10001111
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xB6

Bit7: FOSE: Flash One-Shot Timer Enable
0: Flash One-shot timer disabled.
1: Flash One-shot timer enabled
Bit6: FRAE: Flash Read Always Enable
0: Flash reads per one-shot timer
1: Flash always in read mode
Bits5-4: UNUSED. Read = 00b, Write = don't care.
Bits3-0: FLASCL: Flash Memory Timing Prescaler.
This register specifies the prescaler value for a given system clock required to generate the correct timing for Flash write/erase operations. If the prescaler is set to 1111b, Flash write/erase operations are disabled.
0000: System Clock < 50kHz
0001: 50kHz ≤ System Clock < 100kHz
0010: 100kHz ≤ System Clock < 200kHz
0011: 200kHz ≤ System Clock < 400kHz
0100: 400kHz ≤ System Clock < 800kHz
0101: 800kHz ≤ System Clock < 1.6MHz
0110: 1.6MHz ≤ System Clock < 3.2MHz
0111: 3.2MHz ≤ System Clock < 6.4MHz
1000: 6.4MHz ≤ System Clock < 12.8MHz
1001: 12.8MHz ≤ System Clock < 25.6MHz
1010: 25.6MHz ≤ System Clock < 51.2MHz *
1011, 1100, 1101, 1110: Reserved Values
1111: Flash Memory Write/Erase Disabled

The prescaler value is the smallest value satisfying the following equation:
FLASCL > $\log_2$(System Clock / 50kHz)

* For test purposes. The C8051F000 family is not guaranteed for operation over 25MHz.

Table 15.1. Crossbar Priority Decode

	P0								P1								P2							
PIN I/O	0	1	2	3	4	5	6	7	0	1	2	3	4	5	6	7	0	1	2	3	4	5	6	7
SDA	●																							
SCL		●																						
SCK	●		●																					
MISO		●		●																				
MOSI			●		●																			
NSS				●		●																		
TX	●		●		●		●																	
RX		●		●		●		●																
CEX0	●		●		●		●		●															
CEX1		●		●		●		●		●														
CEX2			●		●		●		●		●													
CEX3				●		●		●		●		●												
CEX4					●		●		●		●		●											
ECI	●	●	●	●	●	●	●	●	●	●	●	●	●	●										
CP0	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●									
CP1	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●								
T0	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●							
/INT0	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●						
T1	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●					
/INT1	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●				
T2	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●			
T2EX	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●		
/SYSCLK	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	
CNVSTR	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●	●

In the Priority Decode Table, a dot (•) is used to show the external Port I/O pin (column) to which each signal (row) can be assigned by the user application code via programming registers XBR2, XBR1, and XBR0.

Figure 15.4. XBR1: Port I/O CrossBar Register 1

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
SYSCKE	T2EXE	T2E	INT1E	T1E	INT0E	T0E	CP1OEN	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xE2

Bit7: SYSCKE: SYSCLK Output Enable Bit
0: SYSCLK unavailable at Port pin.
1: SYSCLK output routed to Port Pin.

Bit6: T2EXE: T2EX Enable Bit
0: T2EX unavailable at Port pin.
1: T2EX routed to Port Pin.

Bit5: T2E: T2 Enable Bit
0: T2 unavailable at Port pin.
1: T2 routed to Port Pin.

Bit4: INT1E: /INT1 Enable Bit
0: /INT1 unavailable at Port pin.
1: /INT1 routed to Port Pin.

Bit3: T1E: T1 Enable Bit
0: T1 unavailable at Port pin.
1: T1 routed to Port Pin.

Bit2: INT0E: /INT0 Enable Bit
0: /INT0 unavailable at Port pin.
1: /INT0 routed to Port Pin.

Bit1: T0E: T0 Enable Bit
0: T0 unavailable at Port pin.
1: T0 routed to Port Pin.

Bit0: CP1OEN: Comparator 1 Output Enable Bit
0: CP1 unavailable at Port pin.
1: CP1 routed to Port Pin.

15.3. General Purpose Port I/O

Each MCU has four byte-wide, bi-directional parallel ports that can be used general purpose I/O. Each port is accessed through a corresponding special function register (SFR) that is both byte addressable and bit addressable. When writing to a port, the value written to the SFR is latched to maintain the output data value at each pin. When reading, the logic levels of the port's input pins are returned regardless of the XBRn settings (i.e. even when the pin is assigned to another signal by the Crossbar, the Port Register can always still read its corresponding Port I/O pin). The exception to this is the execution of the *read-modify-write* instructions. The *read-modify-write* instructions when operating on a port SFR are the following: ANL, ORL, XRL, JBC, CPL, INC, DEC, DJNZ and MOV, CLR or SET, when the destination is an individual bit in a port SFR. For these instructions, the value of the port register (not the pin) is read, modified, and written back to the SFR.

15.4. Configuring Ports Which are not Pinned Out

P2 and P3 are not pinned out on the F001/06/11/16. P1, P2, and P3 are not pinned out on the F002/07/12/17. These port registers (and corresponding interrupts, where applicable) are still available for software use in these reduced pin count MCUs. Whether used or not in software, it is recommended not to let these port drivers go to high impedance state. This is prevented after reset by having the weak pull-ups enabled as described in the XBR2 register. It is recommended that each output driver for ports not pinned out should be configured as push-pull using the corresponding PRTnCF register. This will inhibit a high impedance state even if the weak pull-up is disabled.

Figure 15.6. P0: Port0 Register

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
P0.7	P0.6	P0.5	P0.4	P0.3	P0.2	P0.1	P0.0	11111111
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0 (bit addressable)	SFR Address: 0x80

Bits7-0: P0.[7:0]
(Write – Output appears on I/O pins per XBR0, XBR1, and XBR2 Registers)
0: Logic Low Output.
1: Logic High Output (high-impedance if corresponding PRT0CF.n bit = 0)
(Read – Regardless of XBR0, XBR1, and XBR2 Register settings).
0: P0.n pin is logic low.
1: P0.n pin is logic high.

Figure 15.7. PRT0CF: Port0 Configuration Register

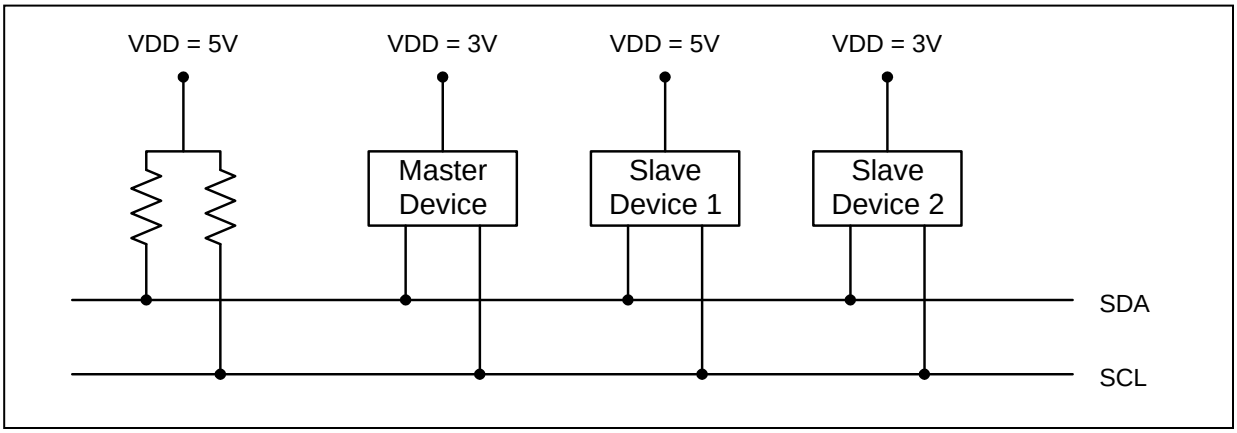
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xA4

Bits7-0: PRT0CF.[7:0]: Output Configuration Bits for P0.7-P0.0 (respectively)
0: Corresponding P0.n Output mode is Open-Drain.
1: Corresponding P0.n Output mode is Push-Pull.

(Note: When SDA, SCL, and RX appear on any of the P0 I/O, each are open-drain regardless of the value of PRT0CF).

Figure 16.2 shows a typical SMBus configuration. The SMBus interface will work at any voltage between 3.0V and 5.0V and different devices on the bus may operate at different voltage levels. The SCL (serial clock) and SDA (serial data) lines are bi-directional. They must be connected to a positive power supply voltage through a pull-up resistor or similar circuit. When the bus is free, both lines are pulled high. Every device connected to the bus must have an open-drain or open-collector output for both the SCL and SDA lines. The maximum number of devices on the bus is limited only by the requirement that the rise and fall times on the bus will not exceed 300ns and 1000ns, respectively.

Figure 16.2. Typical SMBus Configuration

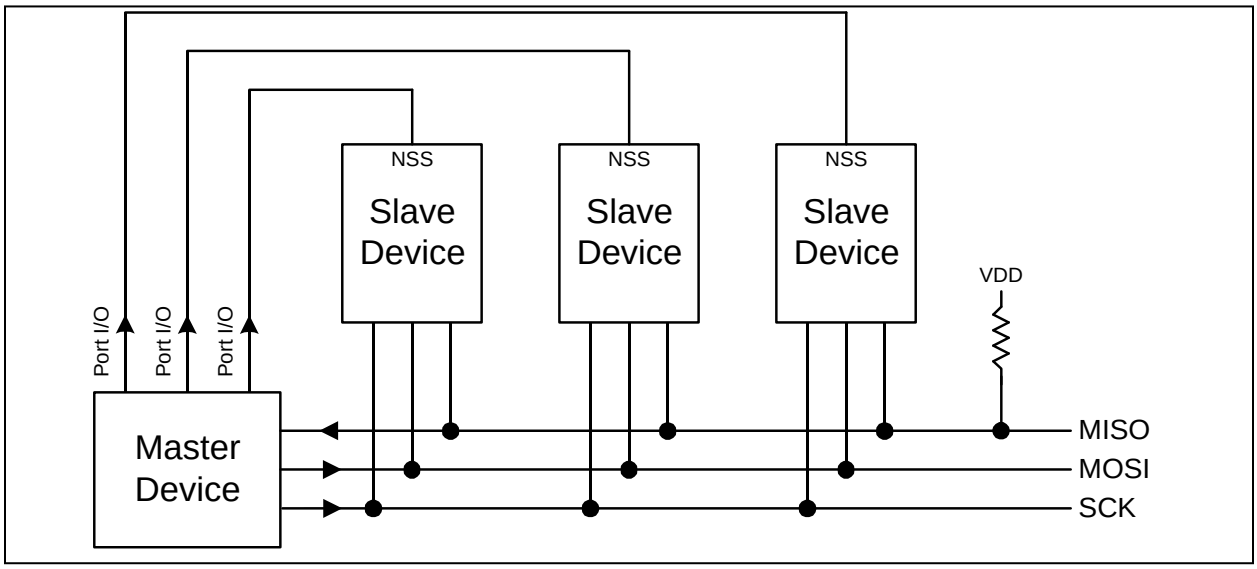


16.1. Supporting Documents

It is assumed the reader is familiar with or has access to the following supporting documents:

1. *The I²C-bus and how to use it (including specifications)*, Philips Semiconductor.
2. *The I²C-Bus Specification -- Version 2.0*, Philips Semiconductor.
3. *System Management Bus Specification -- Version 1.1*, SBS Implementers Forum.

Figure 17.2. Typical SPI Interconnection



17.1. Signal Descriptions

The four signals used by the SPI (MOSI, MISO, SCK, NSS) are described below.

17.1.1. Master Out, Slave In

The master-out, slave-in (MOSI) signal is an output from a master device and an input to slave devices. It is used to serially transfer data from the master to the slave. Data is transferred most-significant bit first.

17.1.2. Master In, Slave Out

The master-in, slave-out (MISO) signal is an output from a slave device and an input to the master device. It is used to serially transfer data from the slave to the master. Data is transferred most-significant bit first. A SPI slave places the MISO pin in a high-impedance state when the slave is not selected.

17.1.3. Serial Clock

The serial clock (SCK) signal is an output from the master device and an input to slave devices. It is used to synchronize the transfer of data between the master and slave on the MOSI and MISO lines.

17.1.4. Slave Select

The slave select (NSS) signal is an input used to select the SPI module when in slave mode by a master, or to disable the SPI module when in master mode. When in slave mode, it is pulled low to initiate a data transfer and remains low for the duration of the transfer.

19.2. Timer 2

Timer 2 is a 16-bit counter/timer formed by the two 8-bit SFRs: TL2 (low byte) and TH2 (high byte). As with Timers 0 and 1, Timer 2 can use either the system clock or transitions on an external input pin as its clock source. The Counter/Timer Select bit C/T2 bit (T2CON.1) selects the clock source for Timer 2. Clearing C/T2 selects the system clock as the input for the timer (divided by either one or twelve as specified by the Timer Clock Select bit T2M in CKCON). When C/T2 is set to 1, high-to-low transitions at the T2 input pin increment the counter/timer register. (Refer to Section 14 for information on selecting and configuring external I/O pins.) Timer 2 can also be used to start an ADC Data Conversion.

Timer 2 offers capabilities not found in Timer 0 and Timer 1. It operates in one of three modes: 16-bit Counter/Timer with Capture, 16-bit Counter/Timer with Auto-Reload or Baud Rate Generator Mode. Timer 2's operating mode is selected by setting configuration bits in the Timer 2 Control (T2CON) register. Below is a summary of the Timer 2 operating modes and the T2CON bits used to configure the counter/timer. Detailed descriptions of each mode follow.

RCLK	TCLK	CP/RL2	TR2	Mode
0	0	1	1	16-bit Counter/Timer with Capture
0	0	0	1	16-bit Counter/Timer with Auto-Reload
0	1	X	1	Baud Rate Generator for TX
1	0	X	1	Baud Rate Generator for RX
1	1	X	1	Baud Rate Generator for TX and RX
X	X	X	0	Off

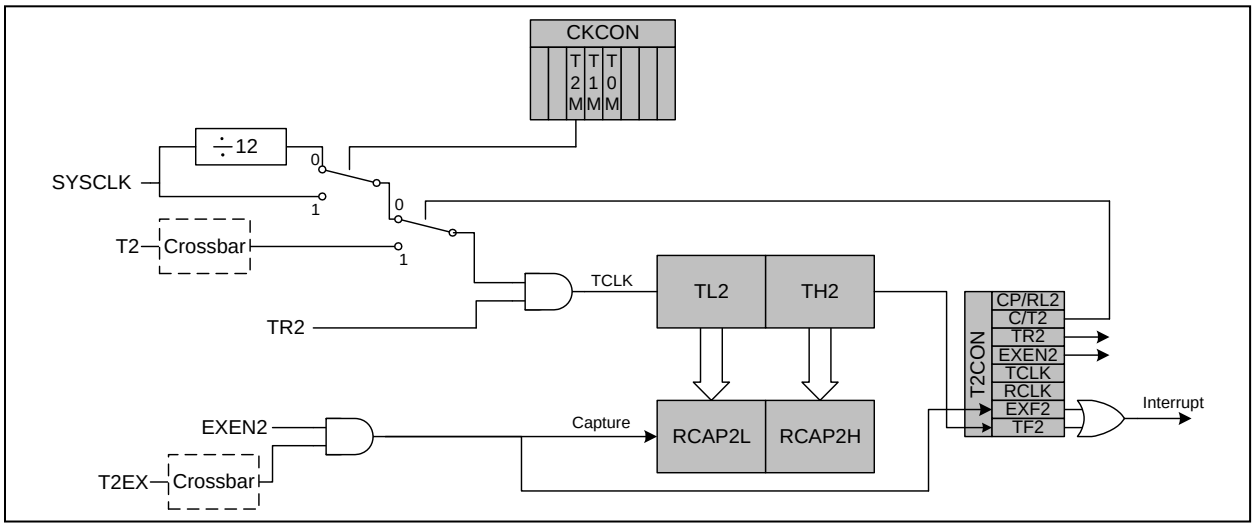
19.2.1. Mode 0: 16-bit Counter/Timer with Capture

In this mode, Timer 2 operates as a 16-bit counter/timer with capture facility. A high-to-low transition on the T2EX input pin causes the 16-bit value in Timer 2 (TH2, TL2) to be loaded into the capture registers (RCAP2H, RCAP2L).

Timer 2 can use either SYSCLK, SYSCLK divided by 12, or high-to-low transitions on the external T2 pin as its clock source when operating in Counter/Timer with Capture mode. Clearing the C/T2 bit (T2CON.1) selects the system clock as the input for the timer (divided by one or twelve as specified by the Timer Clock Select bit T2M in CKCON). When C/T2 is set to logic 1, a high-to-low transition at the T2 input pin increments the counter/timer register. As the 16-bit counter/timer register increments and overflows from 0xFFFF to 0x0000, the TF2 timer overflow flag (T2CON.7) is set and an interrupt will occur if the interrupt is enabled.

Counter/Timer with Capture mode is selected by setting the Capture/Reload Select bit CP/RL2 (T2CON.0) and the Timer 2 Run Control bit TR2 (T2CON.2) to logic 1. The Timer 2 External Enable EXEN2 (T2CON.3) must also be set to logic 1 to enable a capture. If EXEN2 is cleared, transitions on T2EX will be ignored.

Figure 19.11. T2 Mode 0 Block Diagram

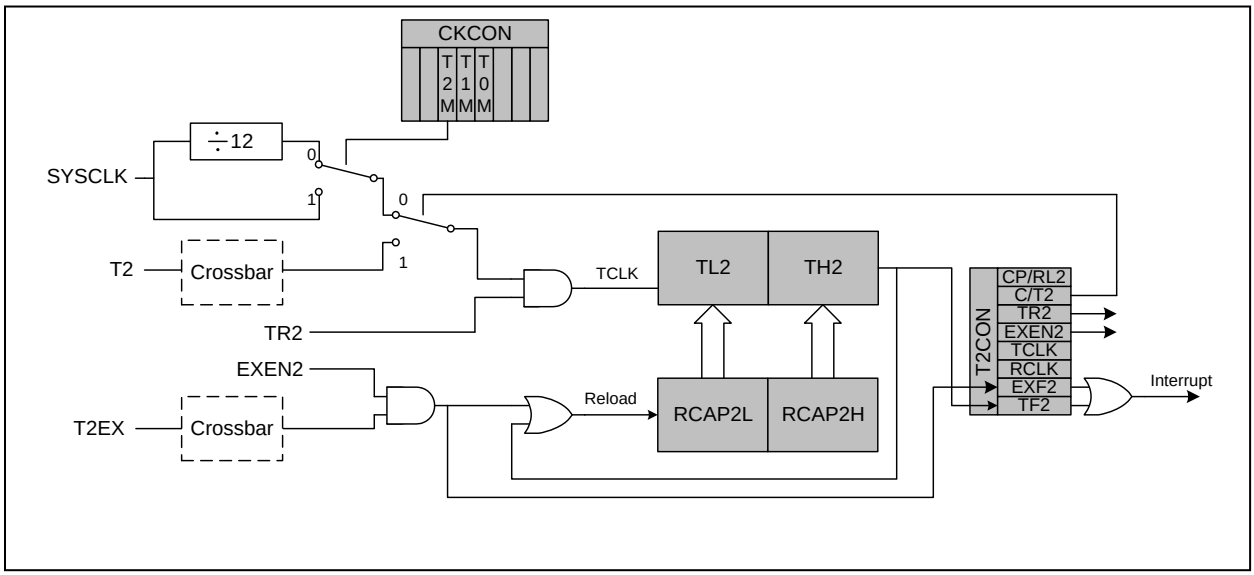


19.2.2. Mode 1: 16-bit Counter/Timer with Auto-Reload

The Counter/Timer with Auto-Reload mode sets the TF2 timer overflow flag when the counter/timer register overflows from 0xFFFF to 0x0000. An interrupt is generated if enabled. On overflow, the 16-bit value held in the two capture registers (RCAP2H, RCAP2L) is automatically loaded into the counter/timer register and the timer is restarted.

Counter/Timer with Auto-Reload mode is selected by clearing the CP/RL2 bit. Setting TR2 to logic 1 enables and starts the timer. Timer 2 can use either the system clock or transitions on an external input pin as its clock source, as specified by the C/T2 bit. If EXEN2 is set to logic 1, a high-to-low transition on T2EX will also cause Timer 2 to be reloaded. If EXEN2 is cleared, transitions on T2EX will be ignored.

Figure 19.12. T2 Mode 1 Block Diagram



21.2. Flash Programming Commands

The Flash memory can be programmed directly over the JTAG interface using the Flash Control, Flash Data, Flash Address, and Flash Scale registers. These Indirect Data Registers are accessed via the JTAG Instruction Register. Read and write operations on indirect data registers are performed by first setting the appropriate DR address in the IR register. Each read or write is then initiated by writing the appropriate Indirect Operation Code (IndOpCode) to the selected data register. Incoming commands to this register have the following format:

19:18	17:0
IndOpCode	WriteData

IndOpCode: These bit set the operation to perform according to the following table:

IndOpCode	Operation
0x	Poll
10	Read
11	Write

The Poll operation is used to check the Busy bit as described below. Although a Capture-DR is performed, no Update-DR is allowed for the Poll operation. Since updates are disabled, polling can be accomplished by shifting in/out a single bit.

The Read operation initiates a read from the register addressed by the IR. Reads can be initiated by shifting only 2 bits into the indirect register. After the read operation is initiated, polling of the Busy bit must be performed to determine when the operation is complete.

The write operation initiates a write of WriteData to the register addressed by the IR. Registers of any width up to 18 bits can be written. If the register to be written contains fewer than 18 bits, the data in WriteData should be left-justified, i.e. its MSB should occupy bit 17 above. This allows shorter registers to be written in fewer JTAG clock cycles. For example, an 8-bit register could be written by shifting only 10 bits. After a Write is initiated, the Busy bit should be polled to determine when the next operation can be initiated. The contents of the Instruction Register should not be altered while either a read or write operation is in progress.

Outgoing data from the indirect Data Register has the following format:

19	18:1	0
0	ReadData	Busy

The Busy bit indicates that the current operation is not complete. It goes high when an operation is initiated and returns low when complete. Read and Write commands are ignored while Busy is high. In fact, if polling for Busy to be low will be followed by another read or write operation, JTAG writes of the next operation can be made while checking for Busy to be low. They will be ignored until Busy is read low, at which time the new operation will initiate. This bit is placed at bit 0 to allow polling by single-bit shifts. When waiting for a Read to complete and Busy is 0, the following 18 bits can be shifted out to obtain the resulting data. ReadData is always right-justified. This allows registers shorter than 18 bits to be read using a reduced number of shifts. For example, the result from a byte-read requires 9 bit shifts (Busy + 8 bits).