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### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Not For New Designs
Core Processor	8051
Core Size	8-Bit
Speed	25MHz
Connectivity	SMBus (2-Wire/I <sup>2</sup> C), SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, Temp Sensor, WDT
Number of I/O	16
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	2.25K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 3.6V
Data Converters	A/D 8x12b; D/A 2x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-TQFP
Supplier Device Package	48-TQFP (7x7)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/silicon-labs/c8051f006-gq">https://www.e-xfl.com/product-detail/silicon-labs/c8051f006-gq</a>

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## 1.1. CIP-51™ CPU

### 1.1.1. Fully 8051 Compatible

The C8051F000 family utilizes Silicon Laboratories' proprietary CIP-51 microcontroller core. The CIP-51 is fully compatible with the MCS-51™ instruction set. Standard 803x/805x assemblers and compilers can be used to develop software. The core has all the peripherals included with a standard 8052, including four 16-bit counter/timers, a full-duplex UART, 256 bytes of internal RAM space, 128 byte Special Function Register (SFR) address space, and four byte-wide I/O Ports.

### 1.1.2. Improved Throughput

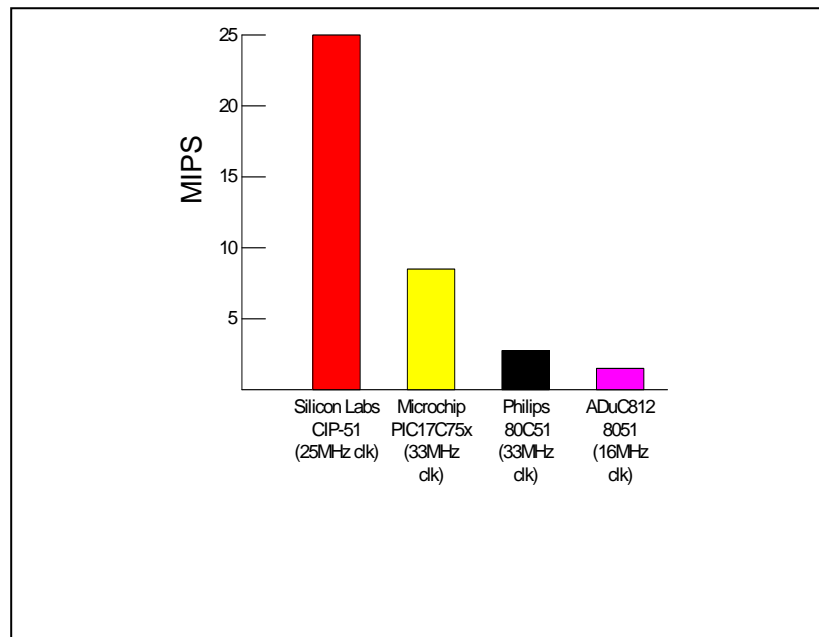
The CIP-51 employs a pipelined architecture that greatly increases its instruction throughput over the standard 8051 architecture. In a standard 8051, all instructions except for MUL and DIV take 12 or 24 system clock cycles to execute with a maximum system clock of 12-to-24MHz. By contrast, the CIP-51 core executes 70% of its instructions in one or two system clock cycles, with only four instructions taking more than four system clock cycles.

The CIP-51 has a total of 109 instructions. The number of instructions versus the system clock cycles to execute them is as follows:

<b>Instructions</b>	26	50	5	14	7	3	1	2	1
<b>Clocks to Execute</b>	1	2	2/3	3	3/4	4	4/5	5	8

With the CIP-51's maximum system clock at 25MHz, it has a peak throughput of 25MIPS. Figure 1.4 shows a comparison of peak throughputs of various 8-bit microcontroller cores with their maximum system clocks.

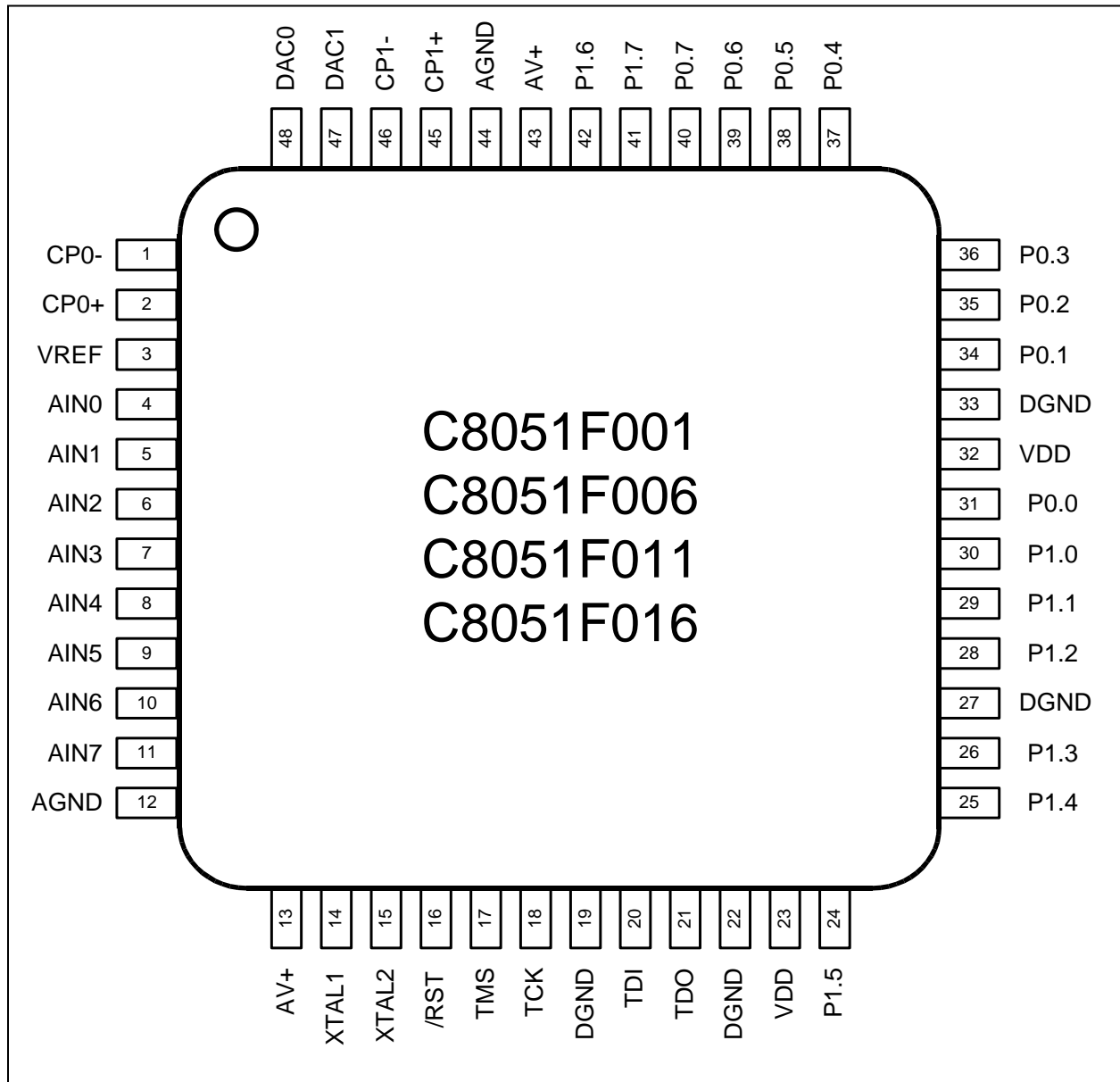
**Figure 1.4. Comparison of Peak MCU Execution Speeds**



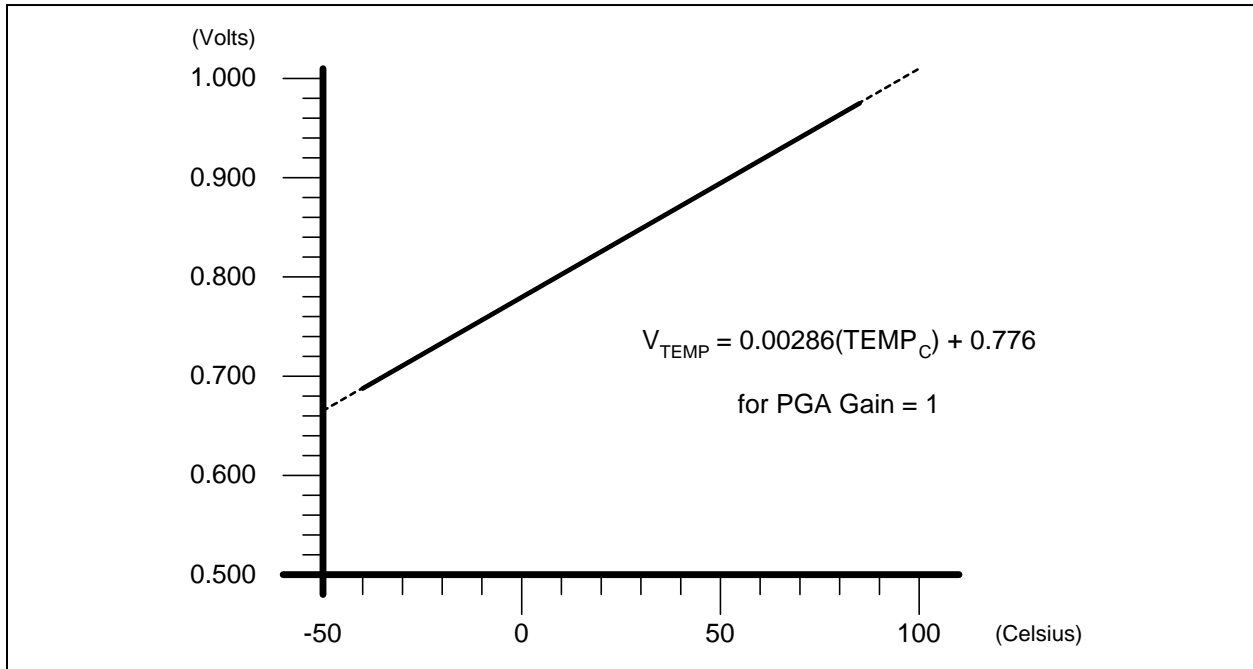
# C8051F000/1/2/5/6/7

# C8051F010/1/2/5/6/7

Figure 4.3. TQFP-48 Pinout Diagram



**Figure 5.3. Temperature Sensor Transfer Function**



**Figure 5.4. AMX0CF: AMUX Configuration Register (C8051F00x)**

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
-	-	-	-	AIN67IC	AIN45IC	AIN23IC	AIN01IC	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xBA

Bits7-4: UNUSED. Read = 0000b; Write = don't care

Bit3: AIN67IC: AIN6, AIN7 Input Pair Configuration Bit  
 0: AIN6 and AIN7 are independent singled-ended inputs  
 1: AIN6, AIN7 are (respectively) +, - differential input pair

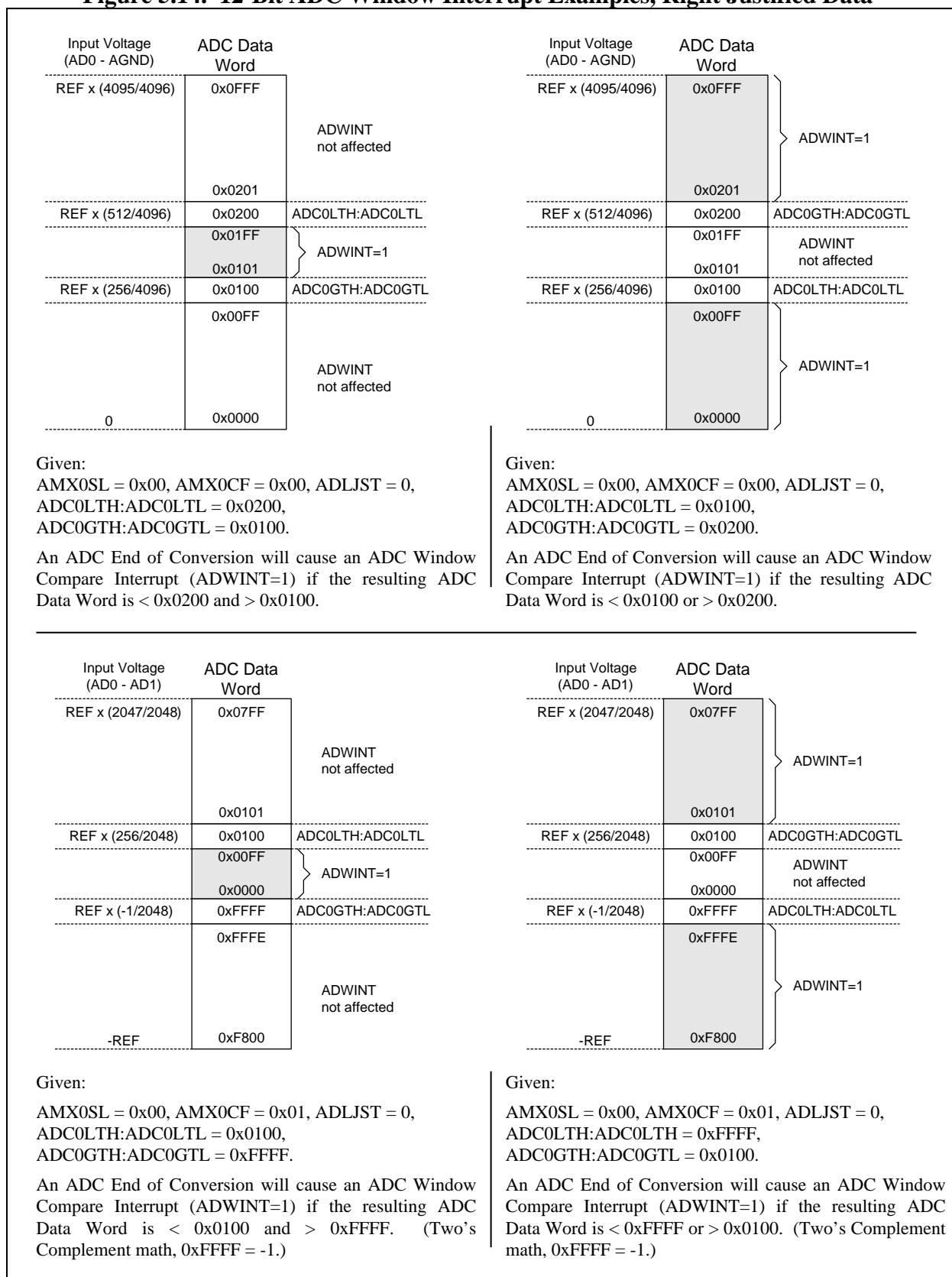
Bit2: AIN45IC: AIN4, AIN5 Input Pair Configuration Bit  
 0: AIN4 and AIN5 are independent singled-ended inputs  
 1: AIN4, AIN5 are (respectively) +, - differential input pair

Bit1: AIN23IC: AIN2, AIN3 Input Pair Configuration Bit  
 0: AIN2 and AIN3 are independent singled-ended inputs  
 1: AIN2, AIN3 are (respectively) +, - differential input pair

Bit0: AIN01IC: AIN0, AIN1 Input Pair Configuration Bit  
 0: AIN0 and AIN1 are independent singled-ended inputs  
 1: AIN0, AIN1 are (respectively) +, - differential input pair

NOTE: The ADC Data Word is in 2's complement format for channels configured as differential.

**Figure 5.14. 12-Bit ADC Window Interrupt Examples, Right Justified Data**



**Figure 7.5. DAC1H: DAC1 High Byte Register**

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	00000000
								SFR Address: 0xD6

Bits7-0: DAC1 Data Word Most Significant Byte.

**Figure 7.6. DAC1L: DAC1 Low Byte Register**

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	00000000
								SFR Address: 0xD5

Bits7-0: DAC1 Data Word Least Significant Byte.

**Figure 7.7. DAC1CN: DAC1 Control Register**

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
DAC1EN	-	-	-	-	DAC1DF2	DAC1DF1	DAC1DF0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xD7

Bit7: DAC1EN: DAC1 Enable Bit  
 0: DAC1 Disabled. DAC1 Output pin is disabled; DAC1 is in low power shutdown mode.  
 1: DAC1 Enabled. DAC1 Output is pin active; DAC1 is operational.

Bits6-3: UNUSED. Read = 0000b; Write = don't care

Bits2-0: DAC1DF2-0: DAC1 Data Format Bits  
 000: The most significant nybble of the DAC1 Data Word is in DAC1H[3:0], while the least significant byte is in DAC1L.

DAC1H				DAC1L			
			MSB				LSB

001: The most significant 5-bits of the DAC1 Data Word is in DAC1H[4:0], while the least significant 7-bits is in DAC1L[7:1].

DAC1H				DAC1L			
			MSB				LSB

010: The most significant 6-bits of the DAC1 Data Word is in DAC1H[5:0], while the least significant 6-bits is in DAC1L[7:2].

DAC1H				DAC1L			
			MSB				LSB

011: The most significant 7-bits of the DAC1 Data Word is in DAC1H[6:0], while the least significant 5-bits is in DAC1L[7:3].

DAC1H				DAC1L			
			MSB				LSB

1xx: The most significant byte of the DAC1 Data Word is in DAC1H, while the least significant nybble is in DAC1L[7:4].

DAC1H				DAC1L			
			MSB				LSB



**Figure 8.3. CPT0CN: Comparator 0 Control Register**

R/W	R	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
CPOEN	CPOOUT	CPORIF	CP0FIF	CP0HYP1	CP0HYP0	CP0HYN1	CP0HYN0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0x9E
<p>Bit7: CPOEN: Comparator 0 Enable Bit 0: Comparator 0 Disabled. 1: Comparator 0 Enabled.</p> <p>Bit6: CPOOUT: Comparator 0 Output State Flag 0: Voltage on CP0+ &lt; CP0- 1: Voltage on CP0+ &gt; CP0-</p> <p>Bit5: CPORIF: Comparator 0 Rising-Edge Interrupt Flag 0: No Comparator 0 Rising-Edge Interrupt has occurred since this flag was cleared 1: Comparator 0 Rising-Edge Interrupt has occurred since this flag was cleared</p> <p>Bit4: CP0FIF: Comparator 0 Falling-Edge Interrupt Flag 0: No Comparator 0 Falling-Edge Interrupt has occurred since this flag was cleared 1: Comparator 0 Falling-Edge Interrupt has occurred since this flag was cleared</p> <p>Bit3-2: CP0HYP1-0: Comparator 0 Positive Hysteresis Control Bits 00: Positive Hysteresis Disabled 01: Positive Hysteresis = 2mV 10: Positive Hysteresis = 4mV 11: Positive Hysteresis = 10mV</p> <p>Bit1-0: CP0HYN1-0: Comparator 0 Negative Hysteresis Control Bits 00: Negative Hysteresis Disabled 01: Negative Hysteresis = 2mV 10: Negative Hysteresis = 4mV 11: Negative Hysteresis = 10mV</p>								

# C8051F000/1/2/5/6/7

# C8051F010/1/2/5/6/7

**Figure 9.2. REF0CN: Reference Control Register**

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
-	-	-	-	-	TEMPE	BIASE	REFBE	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xD1

Bits7-3: UNUSED. Read = 00000b; Write = don't care

Bit2: TEMPE: Temperature Sensor Enable Bit  
0: Internal Temperature Sensor Off.  
1: Internal Temperature Sensor On.

Bit1: BIASE: Bias Enable Bit for ADC and DAC's  
0: Internal Bias Off.  
1: Internal Bias On (required for use of ADC or DAC's).

Bit0: REFBE: Internal Voltage Reference Buffer Enable Bit  
0: Internal Reference Buffer Off. System reference can be driven from external source on VREF pin.  
1: Internal Reference Buffer On. System reference provided by internal voltage reference.

**Table 9.1. Reference Electrical Characteristics**

VDD = 3.0V, AV+ = 3.0V, -40°C to +85°C unless otherwise specified.

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
<b>INTERNAL REFERENCE (REFBE = 1)</b>					
Output Voltage	25°C ambient	2.34	2.43	2.50	V
VREF Short Circuit Current				30	mA
VREF Power Supply Current (supplied by AV+)			50		μA
VREF Temperature Coefficient			15		ppm/°C
Load Regulation	Load = (0-to-200μA) to AGND (Note 1)		0.5		ppm/μA
VREF Turn-on Time1	4.7μF tantalum, 0.1μF ceramic bypass		2		ms
VREF Turn-on Time2	0.1μF ceramic bypass		20		μs
VREF Turn-on Time3	no bypass cap		10		μs
<b>EXTERNAL REFERENCE (REFBE = 0)</b>					
Input Voltage Range		1.00		(AV+) - 0.3V	V
Input Current			0	1	μA

Note 1: The reference can only source current. When driving an external load, it is recommended to add a load resistor to AGND.

Mnemonic	Description	Bytes	Clock Cycles
<b>PROGRAM BRANCHING</b>			
ACALL addr11	Absolute subroutine call	2	3
LCALL addr16	Long subroutine call	3	4
RET	Return from subroutine	1	5
RETI	Return from interrupt	1	5
AJMP addr11	Absolute jump	2	3
LJMP addr16	Long jump	3	4
SJMP rel	Short jump (relative address)	2	3
JMP @A+DPTR	Jump indirect relative to DPTR	1	3
JZ rel	Jump if A equals zero	2	2/3
JNZ rel	Jump if A does not equal zero	2	2/3
CJNE A,direct,rel	Compare direct byte to A and jump if not equal	3	3/4
CJNE A,#data,rel	Compare immediate to A and jump if not equal	3	3/4
CJNE Rn,#data,rel	Compare immediate to register and jump if not equal	3	3/4
CJNE @Ri,#data,rel	Compare immediate to indirect and jump if not equal	3	4/5
DJNZ Rn,rel	Decrement register and jump if not zero	2	2/3
DJNZ direct,rel	Decrement direct byte and jump if not zero	3	3/4
NOP	No operation	1	1

#### Notes on Registers, Operands and Addressing Modes:

**Rn** - Register R0-R7 of the currently selected register bank.

**@Ri** - Data RAM location addressed indirectly through register R0-R1

**rel** - 8-bit, signed (two's complement) offset relative to the first byte of the following instruction. Used by SJMP and all conditional jumps.

**direct** - 8-bit internal data location's address. This could be a direct-access Data RAM location (0x00-0x7F) or an SFR (0x80-0xFF).

**#data** - 8-bit constant

**#data 16** - 16-bit constant

**bit** - Direct-addressed bit in Data RAM or SFR.

**addr 11** - 11-bit destination address used by ACALL and AJMP. The destination must be within the same 2K-byte page of program memory as the first byte of the following instruction.

**addr 16** - 16-bit destination address used by LCALL and LJMP. The destination may be anywhere within the 64K-byte program memory space.

There is one unused opcode (0xA5) that performs the same function as NOP.  
All mnemonics copyrighted © Intel Corporation 1980.

# C8051F000/1/2/5/6/7

# C8051F010/1/2/5/6/7

Address	Register	Description	Page No.
0xC7	ADC0LTH	ADC Less-Than Data Word (High Byte)	36*, 47**
0xC6	ADC0LTL	ADC Less-Than Data Word (Low Byte)	36*, 47**
0xBA	AMX0CF	ADC MUX Configuration	31*, 42**
0xBB	AMX0SL	ADC MUX Channel Selection	32*, 43**
0xF0	B	B Register	76
0x8E	CKCON	Clock Control	144
0x9E	CPT0CN	Comparator 0 Control	56
0x9F	CPT1CN	Comparator 1 Control	58
0xD4	DAC0CN	DAC 0 Control	52
0xD3	DAC0H	DAC 0 Data Word (High Byte)	52
0xD2	DAC0L	DAC 0 Data Word (Low Byte)	52
0xD7	DAC1CN	DAC 1 Control	53
0xD6	DAC1H	DAC 1 Data Word (High Byte)	53
0xD5	DAC1L	DAC 1 Data Word (Low Byte)	53
0x83	DPH	Data Pointer (High Byte)	74
0x82	DPL	Data Pointer (Low Byte)	74
0xE6	EIE1	Extended Interrupt Enable 1	81
0xE7	EIE2	Extended Interrupt Enable 2	82
0xF6	EIP1	External Interrupt Priority 1	83
0xF7	EIP2	External Interrupt Priority 2	84
0xAF	EMI0CN	External Memory Interface Control	92***
0xB7	FLACL	Flash Access Limit	90***
0xB6	FLSCL	Flash Memory Timing Prescaler	91
0xA8	IE	Interrupt Enable	79
0xB8	IP	Interrupt Priority Control	80
0xB2	OSCICN	Internal Oscillator Control	100
0xB1	OSCXCN	External Oscillator Control	101
0x80	P0	Port 0 Latch	109
0x90	P1	Port 1 Latch	110
0xA0	P2	Port 2 Latch	111
0xB0	P3	Port 3 Latch	112
0xD8	PCA0CN	Programmable Counter Array 0 Control	160
0xFA	PCA0CPH0	PCA Capture Module 0 Data Word (High Byte)	163
0xFB	PCA0CPH1	PCA Capture Module 1 Data Word (High Byte)	163
0xFC	PCA0CPH2	PCA Capture Module 2 Data Word (High Byte)	163
0xFD	PCA0CPH3	PCA Capture Module 3 Data Word (High Byte)	163
0xFE	PCA0CPH4	PCA Capture Module 4 Data Word (High Byte)	163
0xEA	PCA0CPL0	PCA Capture Module 0 Data Word (Low Byte)	163
0xEB	PCA0CPL1	PCA Capture Module 1 Data Word (Low Byte)	163
0xEC	PCA0CPL2	PCA Capture Module 2 Data Word (Low Byte)	163
0xED	PCA0CPL3	PCA Capture Module 3 Data Word (Low Byte)	163

## **10.4. INTERRUPT HANDLER**

The CIP-51 includes an extended interrupt system supporting a total of 22 interrupt sources with two priority levels. The allocation of interrupt sources between on-chip peripherals and external inputs pins varies according to the specific version of the device. Each interrupt source has one or more associated interrupt-pending flag(s) located in an SFR. When a peripheral or external source meets a valid interrupt condition, the associated interrupt-pending flag is set to logic 1.

If interrupts are enabled for the source, an interrupt request is generated when the interrupt-pending flag is set. As soon as execution of the current instruction is complete, the CPU generates an LCALL to a predetermined address to begin execution of an interrupt service routine (ISR). Each ISR must end with an RETI instruction, which returns program execution to the next instruction that would have been executed if the interrupt request had not occurred. If interrupts are not enabled, the interrupt-pending flag is ignored by the hardware and program execution continues as normal. (The interrupt-pending flag is set to logic 1 regardless of the interrupt's enable/disable state.)

Each interrupt source can be individually enabled or disabled through the use of an associated interrupt enable bit in an SFR (IE-EIE2). However, interrupts must first be globally enabled by setting the EA bit (IE.7) to logic 1 before the individual interrupt enables are recognized. Setting the EA bit to logic 0 disables all interrupt sources regardless of the individual interrupt-enable settings.

Some interrupt-pending flags are automatically cleared by the hardware when the CPU vectors to the ISR. However, most are not cleared by the hardware and must be cleared by software before returning from the ISR. If an interrupt-pending flag remains set after the CPU completes the return-from-interrupt (RETI) instruction, a new interrupt request will be generated immediately and the CPU will re-enter the ISR after the completion of the next instruction.

### **10.4.1. MCU Interrupt Sources and Vectors**

The MCUs allocate 12 interrupt sources to on-chip peripherals. Up to 10 additional external interrupt sources are available depending on the I/O pin configuration of the device. Software can simulate an interrupt by setting any interrupt-pending flag to logic 1. If interrupts are enabled for the flag, an interrupt request will be generated and the CPU will vector to the ISR address associated with the interrupt-pending flag. MCU interrupt sources, associated vector addresses, priority order and control bits are summarized in Table 10.4. Refer to the datasheet section associated with a particular on-chip peripheral for information regarding valid interrupt conditions for the peripheral and the behavior of its interrupt-pending flag(s).

### **10.4.2. External Interrupts**

Two of the external interrupt sources (/INT0 and /INT1) are configurable as active-low level-sensitive or active-low edge-sensitive inputs depending on the setting of IT0 (TCON.0) and IT1 (TCON.2). IE0 (TCON.1) and IE1 (TCON.3) serve as the interrupt-pending flag for the /INT0 and /INT1 external interrupts, respectively. If an /INT0 or /INT1 external interrupt is configured as edge-sensitive, the corresponding interrupt-pending flag is automatically cleared by the hardware when the CPU vectors to the ISR. When configured as level sensitive, the interrupt-pending flag follows the state of the external interrupt's input pin. The external interrupt source must hold the input active until the interrupt request is recognized. It must then deactivate the interrupt request before execution of the ISR completes or another interrupt request will be generated.

The remaining four external interrupts (External Interrupts 4-7) are active-low, edge-sensitive inputs. The interrupt-pending flags for these interrupts are in the Port 1 Interrupt Flag Register shown in Figure 15.10.

**Figure 10.14. EIP2: Extended Interrupt Priority 2**

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
PXVLD	-	PX7	PX6	PX5	PX4	PADC0	PT3	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xF7

Bit7: PXVLD: External Clock Source Valid (XTLVLD) Interrupt Priority Control.  
This bit sets the priority of the XTLVLD interrupt.  
0: XTLVLD interrupt set to low priority level.  
1: XTLVLD interrupt set to high priority level.

Bit6: Reserved: Must write 0. Reads 0.

Bit5: PX7: External Interrupt 7 Priority Control.  
This bit sets the priority of the External Interrupt 7.  
0: External Interrupt 7 set to low priority level.  
1: External Interrupt 7 set to high priority level.

Bit4: PX6: External Interrupt 6 Priority Control.  
This bit sets the priority of the External Interrupt 6.  
0: External Interrupt 6 set to low priority level.  
1: External Interrupt 6 set to high priority level.

Bit3: PX5: External Interrupt 5 Priority Control.  
This bit sets the priority of the External Interrupt 5.  
0: External Interrupt 5 set to low priority level.  
1: External Interrupt 5 set to high priority level.

Bit2: PX4: External Interrupt 4 Priority Control.  
This bit sets the priority of the External Interrupt 4.  
0: External Interrupt 4 set to low priority level.  
1: External Interrupt 4 set to high priority level.

Bit1: PADC0: ADC End of Conversion Interrupt Priority Control.  
This bit sets the priority of the ADC0 End of Conversion Interrupt.  
0: ADC0 End of Conversion interrupt set to low priority level.  
1: ADC0 End of Conversion interrupt set to high priority level.

Bit0: PT3: Timer 3 Interrupt Priority Control.  
This bit sets the priority of the Timer 3 interrupts.  
0: Timer 3 interrupt set to low priority level.  
1: Timer 3 interrupt set to high priority level.

### 13.1. Power-on Reset

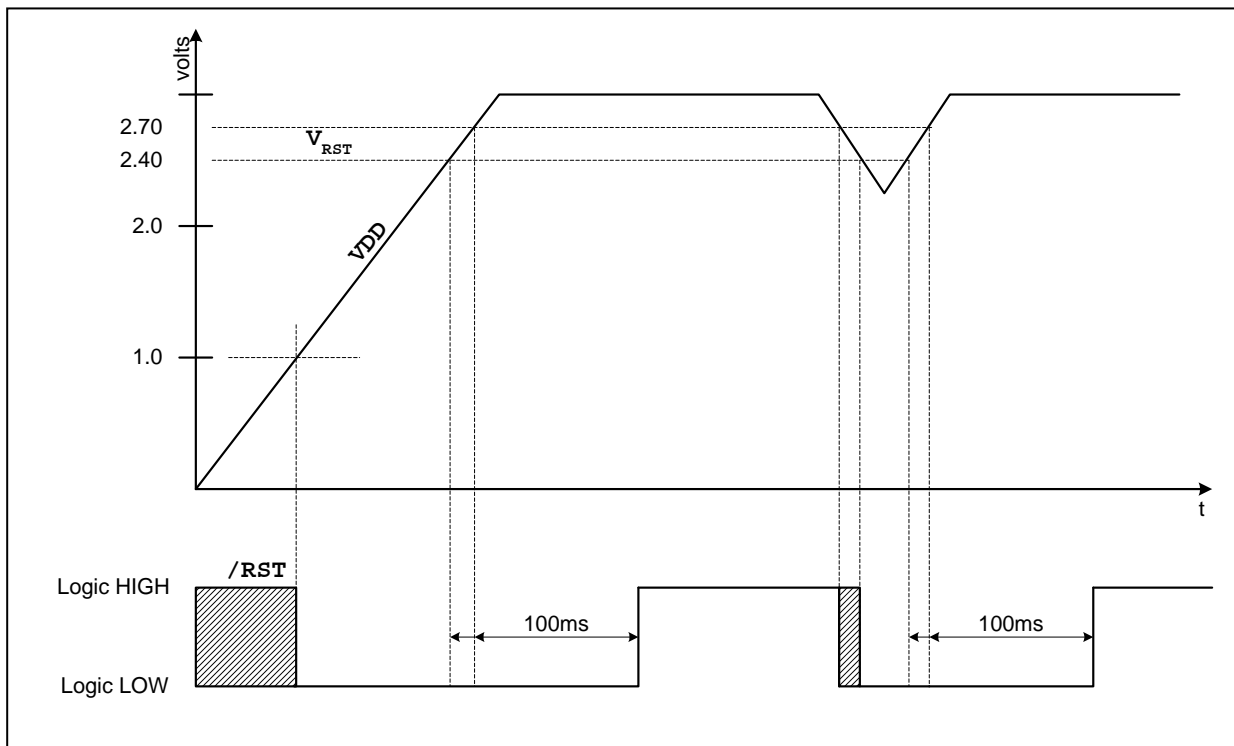
The C8051F000 family incorporates a power supply monitor that holds the MCU in the reset state until VDD rises above the  $V_{RST}$  level during power-up. (See Figure 13.2 for timing diagram, and refer to Table 13.1 for the Electrical Characteristics of the power supply monitor circuit.) The  $\overline{RST}$  pin is asserted (low) until the end of the 100ms VDD Monitor timeout in order to allow the VDD supply to become stable.

On exit from a power-on reset, the PORSF flag (RSTSRC.1) is set by hardware to logic 1. All of the other reset flags in the RSTSRC Register are indeterminate. PORSF is cleared by a reset from any other source. Since all resets cause program execution to begin at the same location (0x0000), software can read the PORSF flag to determine if a power-up was the cause of reset. The content of internal data memory should be assumed to be undefined after a power-on reset.

### 13.2. Software Forced Reset

Writing a 1 to the PORSF bit forces a Power-On Reset as described in Section 13.1.

Figure 13.2. VDD Monitor Timing Diagram



### 13.3. Power-fail Reset

When a power-down transition or power irregularity causes VDD to drop below  $V_{RST}$ , the power supply monitor will drive the  $\overline{RST}$  pin low and return the CIP-51 to the reset state (see Figure 13.2). When VDD returns to a level above  $V_{RST}$ , the CIP-51 will leave the reset state in the same manner as that for the power-on reset. Note that even though internal data memory contents are not altered by the power-fail reset, it is impossible to determine if VDD dropped below the level required for data retention. If the PORSF flag is set, the data may no longer be valid.

**Figure 14.3. OSCXCN: External Oscillator Control Register**

R	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
XTLVLD	XOSCMD2	XOSCMD1	XOSCMD0	-	XFCN2	XFCN1	XFCN0	00110000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xB1

Bit7: XTLVLD: Crystal Oscillator Valid Flag  
**(Valid only when XOSCMD = 1xx.)**  
0: Crystal Oscillator is unused or not yet stable  
1: Crystal Oscillator is running and stable (should read 1ms after Crystal Oscillator is enabled to avoid transient condition).

Bits6-4: XOSCMD2-0: External Oscillator Mode Bits  
00x: Off. XTAL1 pin is grounded internally.  
010: System Clock from External CMOS Clock on XTAL1 pin.  
011: System Clock from External CMOS Clock on XTAL1 pin divided by 2.  
10x: RC/C Oscillator Mode with divide by 2 stage.  
110: Crystal Oscillator Mode  
111: Crystal Oscillator Mode with divide by 2 stage.

Bit3: RESERVED. Read = undefined, Write = don't care

Bits2-0: XFCN2-0: External Oscillator Frequency Control Bits  
000-111: see table below

XFCN	Crystal (XOSCMD = 11x)	RC (XOSCMD = 10x)	C (XOSCMD = 10x)
000	$f \leq 12.5\text{kHz}$	$f \leq 25\text{kHz}$	K Factor = 0.44
001	$12.5\text{kHz} < f \leq 30.3\text{kHz}$	$25\text{kHz} < f \leq 50\text{kHz}$	K Factor = 1.4
010	$30.35\text{kHz} < f \leq 93.8\text{kHz}$	$50\text{kHz} < f \leq 100\text{kHz}$	K Factor = 4.4
011	$93.8\text{kHz} < f \leq 267\text{kHz}$	$100\text{kHz} < f \leq 200\text{kHz}$	K Factor = 13
100	$267\text{kHz} < f \leq 722\text{kHz}$	$200\text{kHz} < f \leq 400\text{kHz}$	K Factor = 38
101	$722\text{kHz} < f \leq 2.23\text{MHz}$	$400\text{kHz} < f \leq 800\text{kHz}$	K Factor = 100
110	$2.23\text{MHz} < f \leq 6.74\text{MHz}$	$800\text{kHz} < f \leq 1.6\text{MHz}$	K Factor = 420
111	$f > 6.74\text{MHz}$	$1.6\text{MHz} < f \leq 3.2\text{MHz}$	K Factor = 1400

**CRYSTAL MODE** (Circuit from Figure 14.1, Option 1; XOSCMD = 11x)  
Choose XFCN value to match the crystal or ceramic resonator frequency.

**RC MODE** (Circuit from Figure 14.1, Option 2; XOSCMD = 10x)  
Choose oscillation frequency range where:  
 $f = 1.23(10^3) / (R * C)$ , where  
f = frequency of oscillation in MHz  
C = capacitor value in pF  
R = Pull-up resistor value in kΩ

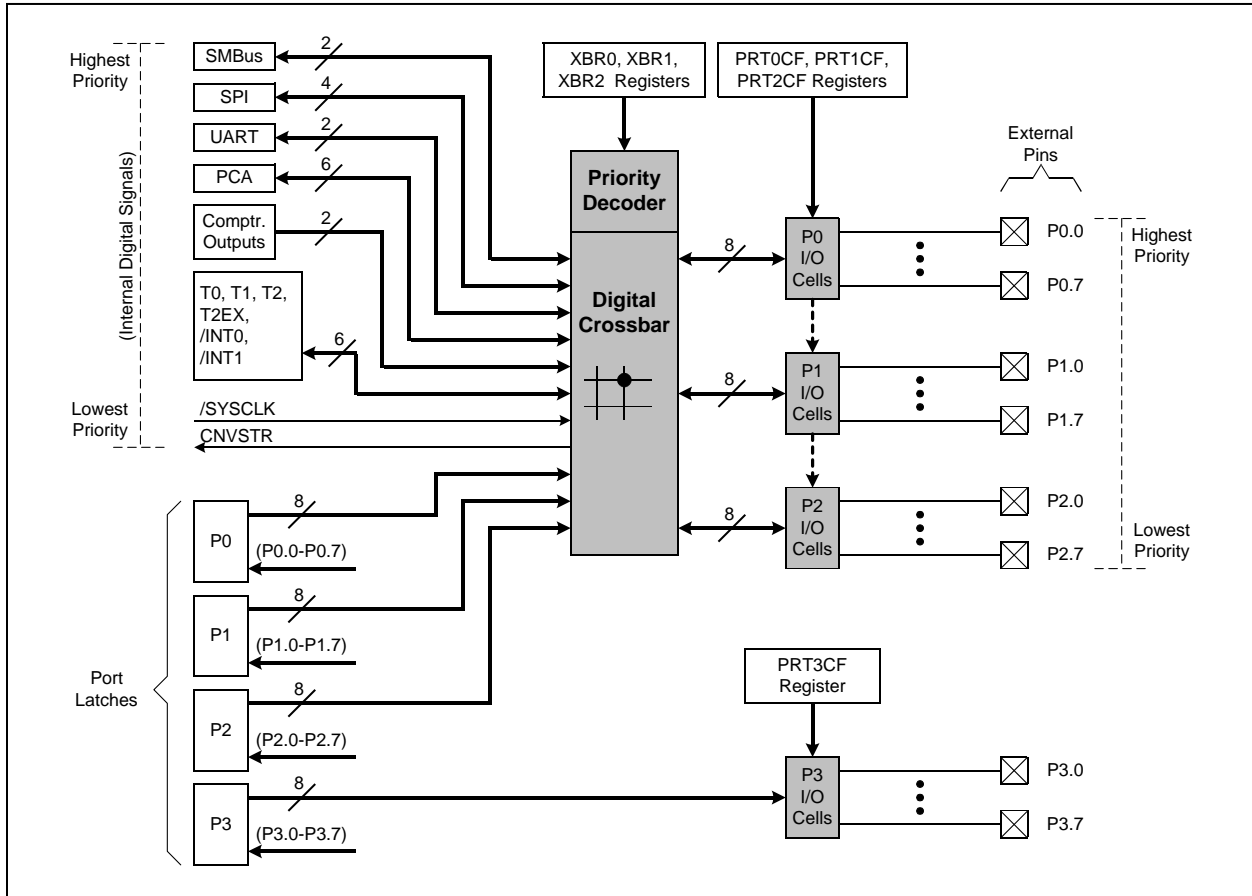
**C MODE** (Circuit from Figure 14.1, Option 3; XOSCMD = 10x)  
Choose K Factor (KF) for the oscillation frequency desired:  
 $f = KF / (C * AV+)$ , where  
f = frequency of oscillation in MHz  
C = capacitor value on XTAL1, XTAL2 pins in pF  
AV+ = Analog Power Supply on MCU in volts



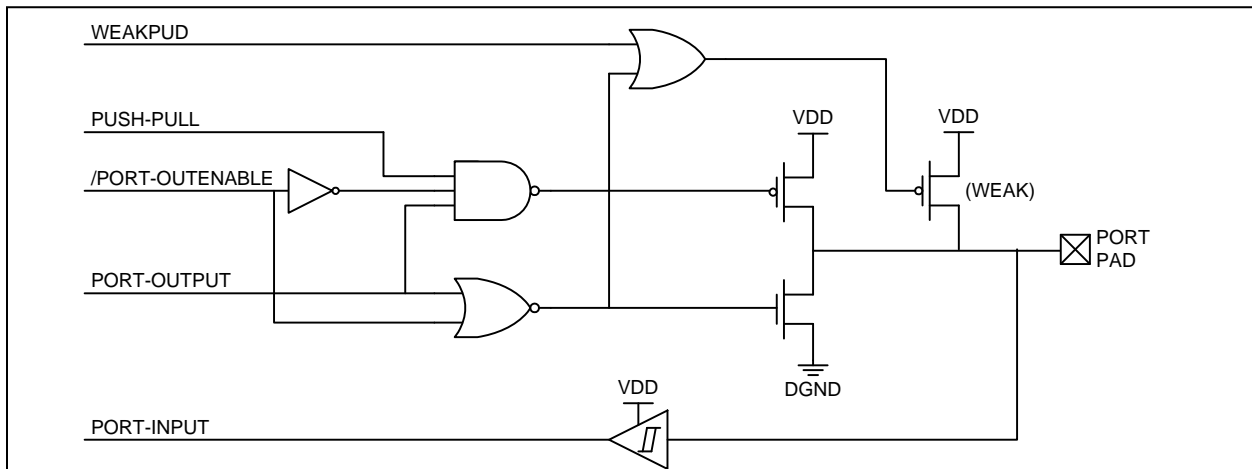
not affect the push-pull Port I/O. Furthermore, the weak pullup is turned off on an open-drain output that is driving a 0 to avoid unnecessary power dissipation.

The third and final step is to initialize the individual resources selected using the appropriate setup registers. Initialization procedures for the various digital resources may be found in the detailed explanation of each available function. The reset state of each register is shown in the figures that describe each individual register.

**Figure 15.1. Port I/O Functional Block Diagram**



**Figure 15.2. Port I/O Cell Block Diagram**



**Figure 15.11. P2: Port2 Register**

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
P2.7	P2.6	P2.5	P2.4	P2.3	P2.2	P2.1	P2.0	11111111
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0 (bit addressable)	SFR Address: 0xA0

Bits7-0: P2.[7:0]  
 (Write – Output appears on I/O pins per XBR0, XBR1, and XBR2 registers)  
 0: Logic Low Output.  
 1: Logic High Output (high-impedance if corresponding PRT2CF.n bit = 0)  
 (Read – Regardless of XBR0, XBR1, and XBR2 Register settings).  
 0: P2.n is logic low.  
 1: P2.n is logic high.

**Figure 15.12. PRT2CF: Port2 Configuration Register**

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xA6

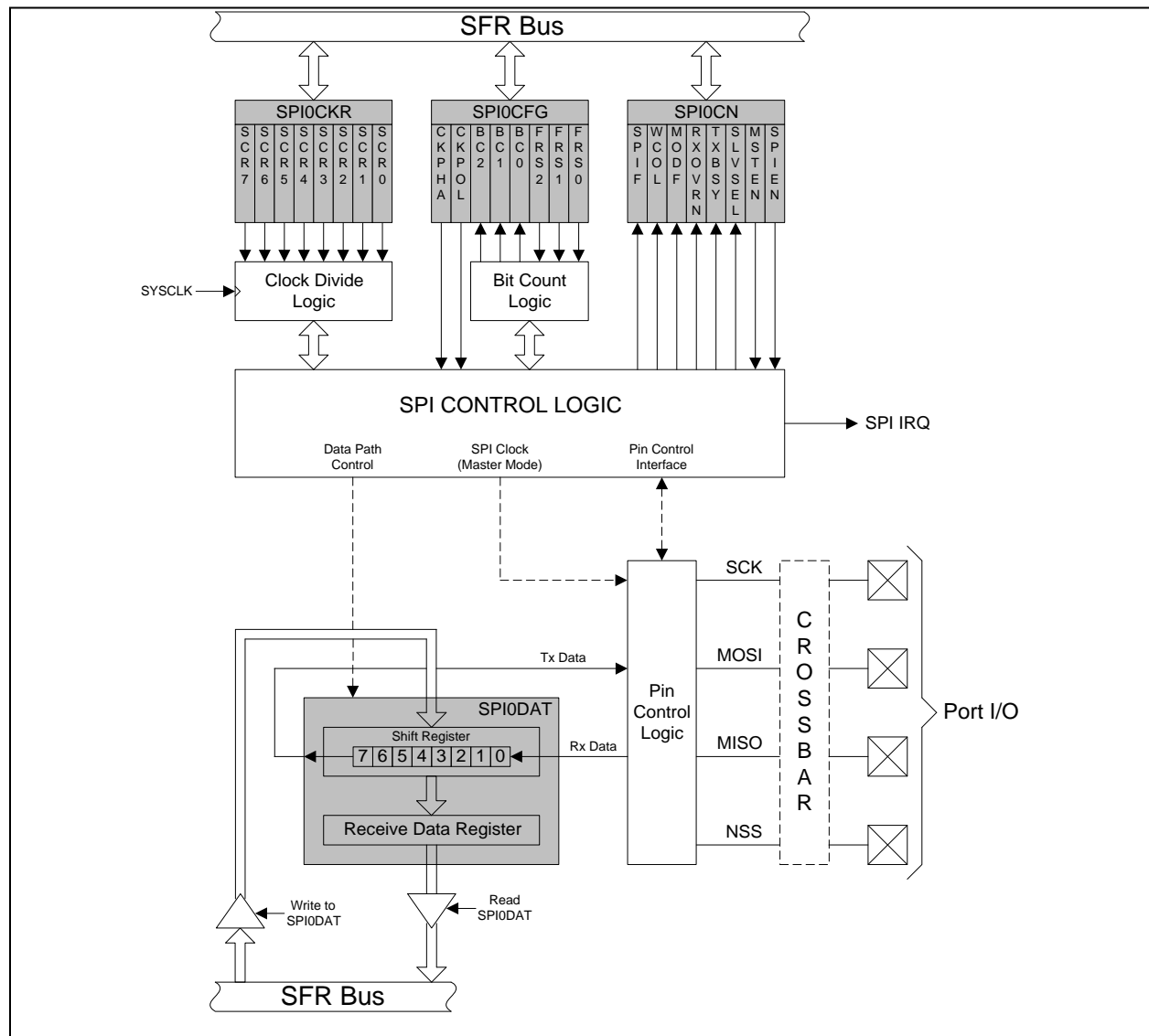
Bits7-0: PRT2CF.[7:0]: Output Configuration Bits for P2.7-P2.0 (respectively)  
 0: Corresponding P2.n Output mode is Open-Drain.  
 1: Corresponding P2.n Output mode is Push-Pull.

## 17. SERIAL PERIPHERAL INTERFACE BUS

The Serial Peripheral Interface (SPI) provides access to a four-wire, full-duplex, serial bus. SPI supports the connection of multiple slave devices to a master device on the same bus. A separate slave-select signal (NSS) is used to select a slave device and enable a data transfer between the master and the selected slave. Multiple masters on the same bus are also supported. Collision detection is provided when two or more masters attempt a data transfer at the same time. The SPI can operate as either a master or a slave. When the SPI is configured as a master, the maximum data transfer rate (bits/sec) is one-half the system clock frequency.

When the SPI is configured as a slave, the maximum data transfer rate (bits/sec) for full-duplex operation is 1/10 the system clock frequency, provided that the master issues SCK, NSS, and the serial input data synchronously with the system clock. If the master issues SCK, NSS, and the serial input data asynchronously, the maximum data transfer rate (bits/sec) must be less than 1/10 the system clock frequency. In the special case where the master only wants to transmit data to the slave and does not need to receive data from the slave (i.e. half-duplex operation), the SPI slave can receive data at a maximum data transfer rate (bits/sec) of 1/4 the system clock frequency. This is provided that the master issues SCK, NSS, and the serial input data synchronously with the system clock.

Figure 17.1. SPI Block Diagram



## 20.1. Capture/Compare Modules

Each module can be configured to operate independently in one of four operation modes: Edge-triggered Capture, Software Timer, High Speed Output, or Pulse Width Modulator. Each module has Special Function Registers (SFRs) associated with it in the CIP-51 system controller. These registers are used to exchange data with a module and configure the module's mode of operation.

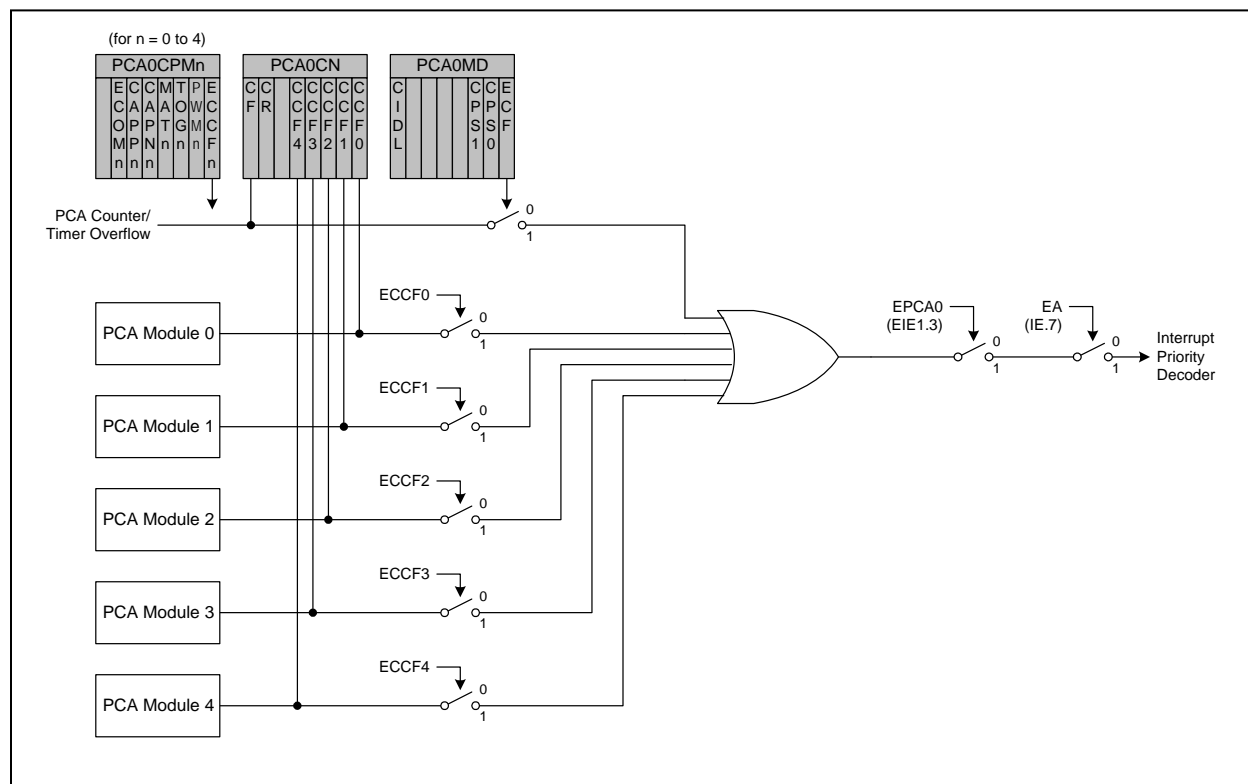
Table 20.1 summarizes the bit settings in the PCA0CPMn registers used to place the PCA capture/compare modules into different operating modes. Setting the ECCFn bit in a PCA0CPMn register enables the module's CCFn interrupt. Note: PCA0 interrupts must be globally enabled before individual CCFn interrupts are recognized. PCA0 interrupts are globally enabled by setting the EA bit (IE.7) and the EPCA0 bit (EIE1.3) to logic 1. See Figure 20.2 for details on the PCA interrupt configuration.

**Table 20.1. PCA0CPM Register Settings for PCA Capture/Compare Modules**

ECOM	CAPP	CAPN	MAT	TOG	PWM	ECCF	Operation Mode
X	1	0	0	0	0	X	Capture triggered by positive edge on CEXn
X	0	1	0	0	0	X	Capture triggered by negative edge on CEXn
X	1	1	0	0	0	X	Capture triggered by transition on CEXn
1	0	0	1	0	0	X	Software Timer
1	0	0	1	1	0	X	High Speed Output
1	0	0	X	0	1	X	Pulse Width Modulator

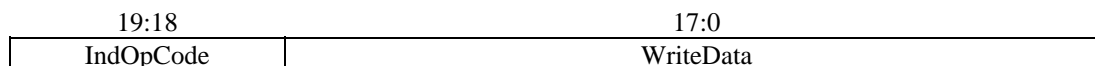
X = Don't Care

**Figure 20.2. PCA Interrupt Block Diagram**



## 21.2. Flash Programming Commands

The Flash memory can be programmed directly over the JTAG interface using the Flash Control, Flash Data, Flash Address, and Flash Scale registers. These Indirect Data Registers are accessed via the JTAG Instruction Register. Read and write operations on indirect data registers are performed by first setting the appropriate DR address in the IR register. Each read or write is then initiated by writing the appropriate Indirect Operation Code (IndOpCode) to the selected data register. Incoming commands to this register have the following format:



IndOpCode: These bit set the operation to perform according to the following table:

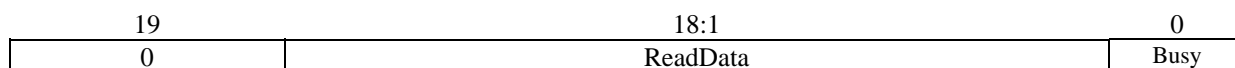
IndOpCode	Operation
0x	Poll
10	Read
11	Write

The Poll operation is used to check the Busy bit as described below. Although a Capture-DR is performed, no Update-DR is allowed for the Poll operation. Since updates are disabled, polling can be accomplished by shifting in/out a single bit.

The Read operation initiates a read from the register addressed by the IR. Reads can be initiated by shifting only 2 bits into the indirect register. After the read operation is initiated, polling of the Busy bit must be performed to determine when the operation is complete.

The write operation initiates a write of WriteData to the register addressed by the IR. Registers of any width up to 18 bits can be written. If the register to be written contains fewer than 18 bits, the data in WriteData should be left-justified, i.e. its MSB should occupy bit 17 above. This allows shorter registers to be written in fewer JTAG clock cycles. For example, an 8-bit register could be written by shifting only 10 bits. After a Write is initiated, the Busy bit should be polled to determine when the next operation can be initiated. The contents of the Instruction Register should not be altered while either a read or write operation is in progress.

Outgoing data from the indirect Data Register has the following format:



The Busy bit indicates that the current operation is not complete. It goes high when an operation is initiated and returns low when complete. Read and Write commands are ignored while Busy is high. In fact, if polling for Busy to be low will be followed by another read or write operation, JTAG writes of the next operation can be made while checking for Busy to be low. They will be ignored until Busy is read low, at which time the new operation will initiate. This bit is placed at bit 0 to allow polling by single-bit shifts. When waiting for a Read to complete and Busy is 0, the following 18 bits can be shifted out to obtain the resulting data. ReadData is always right-justified. This allows registers shorter than 18 bits to be read using a reduced number of shifts. For example, the result from a byte-read requires 9 bit shifts (Busy + 8 bits).