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### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Obsolete
Core Processor	Coldfire V2
Core Size	32-Bit Single-Core
Speed	100MHz
Connectivity	CANbus, EBI/EMI, I <sup>2</sup> C, SPI, UART/USART
Peripherals	DMA, WDT
Number of I/O	61
Program Memory Size	-
Program Memory Type	ROMless
EEPROM Size	-
RAM Size	64K x 8
Voltage - Supply (Vcc/Vdd)	1.4V ~ 1.6V
Data Converters	-
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	256-LBGA
Supplier Device Package	256-MAPBGA (17x17)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/nxp-semiconductors/mcf5233cvm100j">https://www.e-xfl.com/product-detail/nxp-semiconductors/mcf5233cvm100j</a>

Table 2. MCF523x Signal Information and Muxing (continued)

Signal Name	GPIO	Alternate 1	Alternate 2	Dir. <sup>1</sup>	MCF5232 160 QFP	MCF5232 196 MAPBGA	MCF5233 256 MAPBGA	MCF5234 256 MAPBGA	MCF5235 256 MAPBGA
SD_WE	PSDRAM5	—	—	O	93	K13	L13	L13	L13
SD_SCAS	PSDRAM4	—	—	O	92	K12	M15	M15	M15
SD_SRAS	PSDRAM3	—	—	O	91	K11	M14	M14	M14
SD_CKE	PSDRAM2	—	—	O	—	E8	C10	C10	C10
SD_CS[1:0]	PSDRAM[1:0]	—	—	O	—	L12, L13	N15, M13	N15, M13	N15, M13
External Interrupts Port									
IRQ[7:3]	PIRQ[7:3]	—	—	I	IRQ7=64 IRQ4=65	N7, M7, L7, P8, N8	R8, T8, N9, P9, R9	R8, T8, N9, P9, R9	R8, T8, N9, P9, R9
IRQ2	PIRQ2	DREQ2	—	I	—	M8	T9	T9	T9
IRQ1	PIRQ1	—	—	I	66	L8	N10	N10	N10
eTPU									
TPUCH31	—	ECOL	—		—	—	F3	—	F3
TPUCH30	—	ECRS	—		—	—	F4	—	F4
TPUCH29	—	ERXCLK	—		—	—	E3	—	E3
TPUCH28	—	ERXDV	—		—	—	E4	—	E4
TPUCH[27:24]	—	ERXD[3:0]	—		—	—	D3, D4, C3, C4	—	D3, D4, C3, C4
TPUCH23	—	ERXER	—		—	—	D5	—	D5
TPUCH22	—	ETXCLK	—		—	—	C5	—	C5
TPUCH21	—	ETXEN	—		—	—	D6	—	D6
TPUCH20	—	ETXER	—		—	—	C6	—	C6
TPUCH[19:16]	—	ETXD[3:0]	—		—	—	B6, B5, A5, B7	—	B6, B5, A5, B7
TPUCH[15:0]	—	—	—		11, 10, 7:2, 159:154, 152, 151	E2, E1, D1 D2, D3, C1, C2, B1, B2, A2, C3, B3, A3, A4, C4, BR	F2, E1, E2, D1, D2, C1, C2, B1, B2, A2, B3, A3, B4, A4, A6, A7	F2, E1, E2, D1, D2, C1, C2, B1, B2, A2, B3, A3, B4, A4, A6, A7	F2, E1, E2, D1, D2, C1, C2, B1, B2, A2, B3, A3, B4, A4, A6, A7
TCRCLK	PETPU2	—	—		12	E3	F1	F1	F1
UTPUODIS	PETPU1	—	—		—	H10	J13	J13	J13
LTPUODIS	PETPU0	—	—		—	G10	J14	J14	J14
FEC									
EMDIO	PFECI2C2	I2C_SDA	U2RXD	I/O	—	—	—	C7	C7
EMDC	PFECI2C3	I2C_SCL	U2TXD	O	—	—	—	D7	D7
ECOL	—	—	—	I	—	—	—	F3	F3

Table 2. MCF523x Signal Information and Muxing (continued)

Signal Name	GPIO	Alternate 1	Alternate 2	Dir. <sup>1</sup>	MCF5232 160 QFP	MCF5232 196 MAPBGA	MCF5233 256 MAPBGA	MCF5234 256 MAPBGA	MCF5235 256 MAPBGA
ECRS	—	—	—	I	—	—	—	F4	F4
ERXCLK	—	—	—	I	—	—	—	E3	E3
ERXDV	—	—	—	I	—	—	—	E4	E4
ERXD[3:0]	—	—	—	I	—	—	—	D3, D4, C3, C4	D3, D4, C3, C4
ERXER	—	—	—	I	—	—	—	D5	D5
ETXCLK	—	—	—	I	—	—	—	C5	C5
ETXEN	—	—	—	O	—	—	—	D6	D6
ETXER	—	—	—	O	—	—	—	C6	C6
ETXD[3:0]	—	—	—	O	—	—	—	B6, B5, A5, B7	B6, B5, A5, B7
<b>Feature Control</b>									
eTPU/ $\overline{\text{EthENB}}$	—	—	—	I	—	—	—	—	M4
<b>I<sup>2</sup>C</b>									
I2C_SDA	PFECI2C1	CAN0RX	—	I/O	—	J12	L15	L15	L15
I2C_SCL	PFECI2C0	CAN0TX	—	I/O	—	J11	L14	L14	L14
<b>DMA</b>									
DACK[2:0] and $\overline{\text{DREQ}}$ [2:0] do not have a dedicated bond pads. Please refer to the following pins for muxing: $\overline{\text{TS}}$ and DT2OUT for DACK2, TSIZ1 and DT1OUT for DACK1, TSIZ0 and DT0OUT for DACK0, $\overline{\text{IRQ2}}$ and DT2IN for $\overline{\text{DREQ2}}$ , $\overline{\text{TEA}}$ and DT1IN for $\overline{\text{DREQ1}}$ , and $\overline{\text{TIP}}$ and DT0IN for $\overline{\text{DREQ0}}$ .					—	—	—	—	—
<b>QSPI</b>									
QSPI_CS1	PQSPI4	SD_CKE	—	O	139	B7	B10	B10	B10
QSPI_CS0	PQSPI3	—	—	O	147	A6	D9	D9	D9
QSPI_CLK	PQSPI2	I2C_SCL	—	O	148	C5	B8	B8	B8
QSPI_DIN	PQSPI1	I2C_SDA	—	I	149	B5	C8	C8	C8
QSPI_DOUT	PQSPI0	—	—	O	150	A5	D8	D8	D8
<b>UARTs</b>									
U2TXD	PUARTH1	CAN1TX	—	O	—	A8	D11	D11	D11
U2RXD	PUARTH0	CAN1RX	—	I	—	A7	D10	D10	D10
$\overline{\text{U1CTS}}$	PUARTL7	$\overline{\text{U2CTS}}$	—	I	—	B8	C11	C11	C11
$\overline{\text{U1RTS}}$	PUARTL6	$\overline{\text{U2RTS}}$	—	O	—	C8	B11	B11	B11
U1TXD	PUARTL5	CAN0TX	—	O	135	D9	A12	A12	A12

Table 2. MCF523x Signal Information and Muxing (continued)

Signal Name	GPIO	Alternate 1	Alternate 2	Dir. <sup>1</sup>	MCF5232 160 QFP	MCF5232 196 MAPBGA	MCF5233 256 MAPBGA	MCF5234 256 MAPBGA	MCF5235 256 MAPBGA
U1RXD	PUARTL4	CAN0RX	—	I	136	D8	A11	A11	A11
$\overline{U0CTS}$	PUARTL3	—	—	I	—	F3	G1	G1	G1
$\overline{U0RTS}$	PUARTL2	—	—	O	—	G3	H3	H3	H3
U0TXD	PUARTL1	—	—	O	14	F1	H2	H2	H2
U0RXD	PUARTL0	—	—	I	13	F2	G2	G2	G2
<b>DMA Timers</b>									
DT3IN	PTIMER7	$\overline{U2CTS}$	QSPI_CS2	I	—	H14	J15	J15	J15
DT3OUT	PTIMER6	$\overline{U2RTS}$	QSPI_CS3	O	—	G14	J16	J16	J16
DT2IN	PTIMER5	$\overline{DREQ2}$	DT2OUT	I	—	M9	P10	P10	P10
DT2OUT	PTIMER4	DACK2	—	O	—	L9	R10	R10	R10
DT1IN	PTIMER3	$\overline{DREQ1}$	DT1OUT	I	—	L6	P7	P7	P7
DT1OUT	PTIMER2	DACK1	—	O	—	M6	R7	R7	R7
DT0IN	PTIMER1	$\overline{DREQ0}$	—	I	—	E4	G4	G4	G4
DT0OUT	PTIMER0	DACK0	—	O	—	F4	G3	G3	G3
<b>BDM/JTAG<sup>2</sup></b>									
DSCLK	—	TRST	—	I	70	N9	N11	N11	N11
PSTCLK	—	TCLK	—	O	68	P9	T10	T10	T10
$\overline{BKPT}$	—	TMS	—	I	71	P10	P11	P11	P11
DSI	—	TDI	—	I	73	M10	T11	T11	T11
DSO	—	TDO	—	O	72	N10	R11	R11	R11
JTAG_EN	—	—	—	I	78	K9	N13	N13	N13
DDATA[3:0]	—	—	—	O	—	M12, N12, P12, L11	N14, P14, T13, R13	N14, P14, T13, R13	N14, P14, T13, R13
PST[3:0]	—	—	—	O	77:74	M11, N11, P11, L10	T12, R12, P12, N12	T12, R12, P12, N12	T12, R12, P12, N12

Table 2. MCF523x Signal Information and Muxing (continued)

Signal Name	GPIO	Alternate 1	Alternate 2	Dir. <sup>1</sup>	MCF5232 160 QFP	MCF5232 196 MAPBGA	MCF5233 256 MAPBGA	MCF5234 256 MAPBGA	MCF5235 256 MAPBGA
Test									
TEST	—	—	—	I	18	F5	J4	J4	J4
PLL_TEST	—	—	—	I	—		R14	R14	R14
Power Supplies									
VDDPLL	—	—	—	I	87	M13	P15		
VSSPLL	—	—	—	I	84	L14	R15		
OVDD	—	—	—	I	1, 9, 17, 32, 41, 55, 62, 69, 81, 90, 95, 105, 114, 128, 132, 138, 146	E5, E7, E10, F7, F9, G6, G8, H7, H8, H9, J6, J8, J10, K5, K6, K8	E6:11, F5, F7:10, F12, G5, G6, G11, G12, H5, H6, H11, H12, J5, J6, J11, J12, K5, K6, K11, K12, L5, L7:10, L12, M6:M11		
VSS	—	—	—	I	8, 16, 25, 31, 40, 54, 61, 67, 80, 88, 94, 104, 113, 127, 131, 137, 145, 153, 160	A1, A14, E6, E9, F6, F8, F10, G7, G9, H6, J5, J7, J9, K7, P1, P14	A1, A16, E5, E12, F6, F11, F16, G7:10, H7: 10, J1, J7:10, K7:10, L6, L11, M5, M12, N16, T1, T6, T16		
VDD	—	—	—	I	15, 53, 103, 144	D6, F11, G4, L4	A8, G16, H1, T5		

<sup>1</sup> Refers to pin's primary function. All pins which are configurable for GPIO have a pullup enabled in GPIO mode with the exception of PBUSCTL[7], PBUSCTL[4:0], PADDR, PBS, PSDRAM.

<sup>2</sup> If JTAG\_EN is asserted, these pins default to Alternate 1 (JTAG) functionality. The GPIO module is not responsible for assigning these pins.

## 5 Design Recommendations

### 5.1 Layout

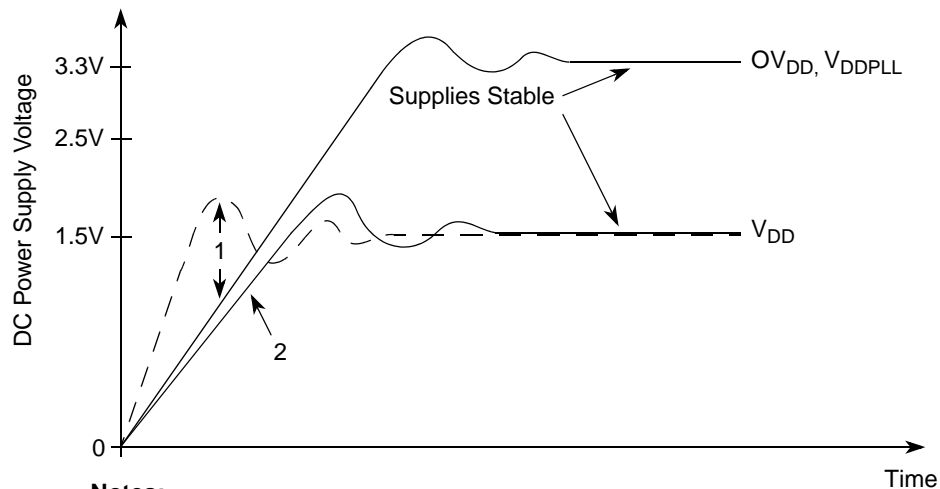
- Use a 4-layer printed circuit board with the VDD and GND pins connected directly to the power and ground planes for the MCF523x.
- See application note AN1259, *System Design and Layout Techniques for Noise Reduction in Processor-Based Systems*.
- Match the PC layout trace width and routing to match trace length to operating frequency and board impedance. Add termination (series or therein) to the traces to dampen reflections. Increase the PCB impedance (if possible) keeping the trace lengths balanced and short. Then do cross-talk analysis to separate traces with significant parallelism or are otherwise "noisy". Use 6 mils trace and separation. Clocks get extra separation and more precise balancing.

## 5.2 Power Supply

- 33  $\mu$ F, 0.1  $\mu$ F, and 0.01  $\mu$ F across each power supply

### 5.2.1 Supply Voltage Sequencing and Separation Cautions

Figure 1 shows situations in sequencing the I/O  $V_{DD}$  ( $OV_{DD}$ ), PLL  $V_{DD}$  ( $V_{DDPLL}$ ), and Core  $V_{DD}$  ( $V_{DD}$ ).  $OV_{DD}$  is specified relative to  $V_{DD}$ .



**Notes:**

1.  $V_{DD}$  should not exceed  $OV_{DD}$  or  $V_{DDPLL}$  by more than 0.4 V at any time, including power-up.
2. Recommended that  $V_{DD}$  should track  $OV_{DD}/V_{DDPLL}$  up to 0.9 V, then separate for completion of ramps.
3. Input voltage must not be greater than the supply voltage ( $OV_{DD}$ ,  $V_{DD}$ , or  $V_{DDPLL}$ ) by more than 0.5 V at any time, including during power-up.
4. Use 1 ms or slower rise time for all supplies.

**Figure 1. Supply Voltage Sequencing and Separation Cautions**

#### 5.2.1.1 Power Up Sequence

If  $OV_{DD}$  is powered up with  $V_{DD}$  at 0 V, then the sense circuits in the I/O pads cause all pad output drivers connected to the  $OV_{DD}$  to be in a high impedance state. There is no limit on how long after  $OV_{DD}$  powers up before  $V_{DD}$  must power up.  $V_{DD}$  should not lead the  $OV_{DD}$  or  $V_{DDPLL}$  by more than 0.4 V during power ramp-up, or there will be high current in the internal ESD protection diodes. The rise times on the power supplies should be slower than 1  $\mu$ s to avoid turning on the internal ESD protection clamp diodes.

The recommended power up sequence is as follows:

1. Use 1 ms or slower rise time for all supplies.
2.  $V_{DD}$  and  $OV_{DD}/V_{DDPLL}$  should track up to 0.9 V, then separate for the completion of ramps with  $OV_{DD}$  going to the higher external voltages. One way to accomplish this is to use a low drop-out voltage regulator.

## 5.7 Interface Recommendations

### 5.7.1 SDRAM Controller

#### 5.7.1.1 SDRAM Controller Signals in Synchronous Mode

Table 3 shows the behavior of SDRAM signals in synchronous mode.

**Table 3. Synchronous DRAM Signal Connections**

Signal	Description
$\overline{\text{SD\_SRAS}}$	Synchronous row address strobe. Indicates a valid SDRAM row address is present and can be latched by the SDRAM. $\overline{\text{SD\_SRAS}}$ should be connected to the corresponding SDRAM $\overline{\text{SD\_SRAS}}$ . Do not confuse $\overline{\text{SD\_SRAS}}$ with the DRAM controller's $\overline{\text{SD\_CS}}[1:0]$ , which should not be interfaced to the SDRAM $\overline{\text{SD\_SRAS}}$ signals.
$\overline{\text{SD\_SCAS}}$	Synchronous column address strobe. Indicates a valid column address is present and can be latched by the SDRAM. $\overline{\text{SD\_SCAS}}$ should be connected to the corresponding signal labeled $\overline{\text{SD\_SCAS}}$ on the SDRAM.
$\overline{\text{DRAMW}}$	DRAM read/write. Asserted for write operations and negated for read operations.
$\overline{\text{SD\_CS}}[1:0]$	Row address strobe. Select each memory block of SDRAMs connected to the MCF523x. One $\overline{\text{SD\_CS}}$ signal selects one SDRAM block and connects to the corresponding $\overline{\text{CS}}$ signals.
$\overline{\text{SD\_CKE}}$	Synchronous DRAM clock enable. Connected directly to the CKE (clock enable) signal of SDRAMs. Enables and disables the clock internal to SDRAM. When CKE is low, memory can enter a power-down mode where operations are suspended or they can enter self-refresh mode. $\overline{\text{SD\_CKE}}$ functionality is controlled by DCR[COC]. For designs using external multiplexing, setting COC allows $\overline{\text{SD\_CKE}}$ to provide command-bit functionality.
$\overline{\text{BS}}[3:0]$	Column address strobe. For synchronous operation, $\overline{\text{BS}}[3:0]$ function as byte enables to the SDRAMs. They connect to the DQM signals (or mask qualifiers) of the SDRAMs.
CLKOUT	Bus clock output. Connects to the CLK input of SDRAMs.

#### 5.7.1.2 Address Multiplexing

See the SDRAM controller module chapter in the *MCF5235 Reference Manual* for details on address multiplexing.

### 5.7.2 Ethernet PHY Transceiver Connection

The FEC supports both an MII interface for 10/100 Mbps Ethernet and a seven-wire serial interface for 10 Mbps Ethernet. The interface mode is selected by R\_CNTRL[MII\_MODE]. In MII mode, the 802.3 standard defines and the FEC module supports 18 signals. These are shown in Table 4.

**Table 4. MII Mode**

Signal Description	MCF523x Pin
Transmit clock	ETXCLK
Transmit enable	ETXEN
Transmit data	ETXD[3:0]

# 6.2.2 Package Dimensions—256 MAPBGA

Figure 7 shows MCF5235CVMxxx, MCF5234CVMxxx, and MCF5233CVMxx package dimensions.

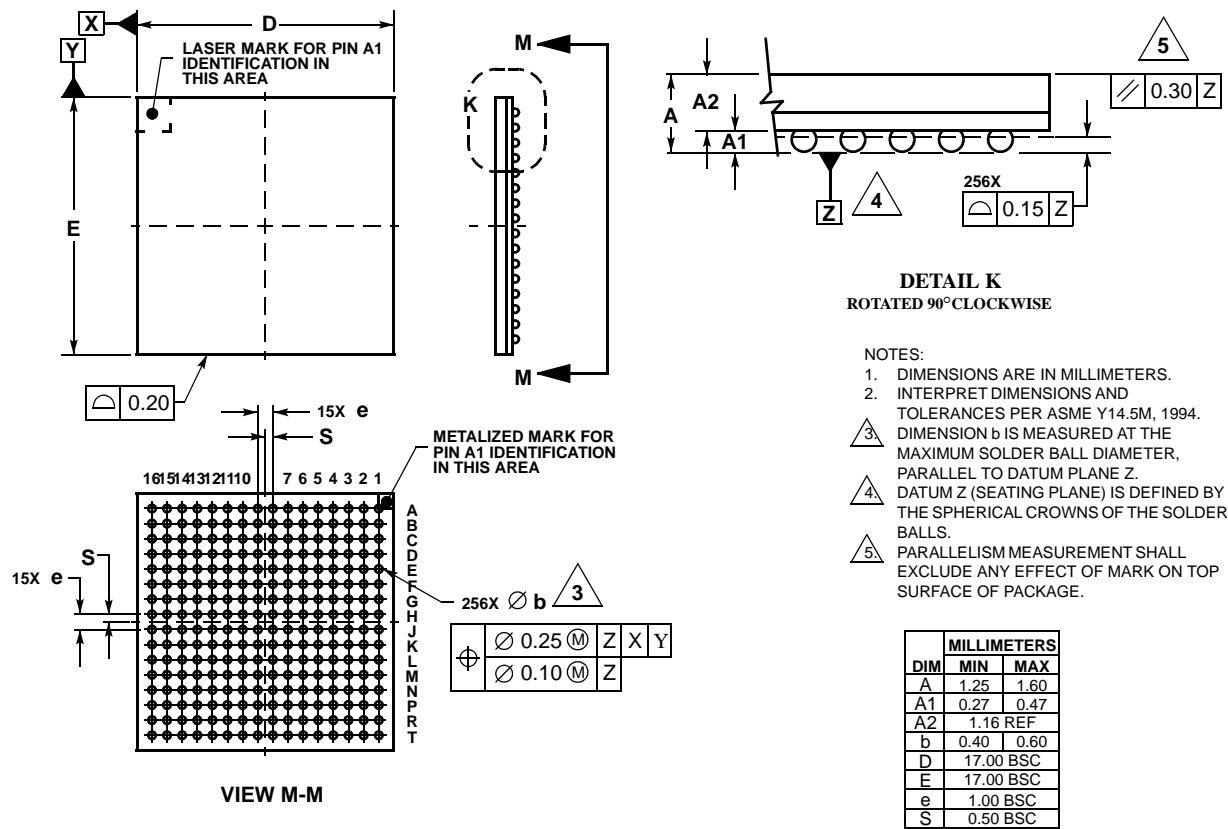


Figure 7. 256 MAPBGA Package Outline

## 6.3 Pinout—160 QFP

Figure 8 shows a pinout of the MCF5232CABxxx package.

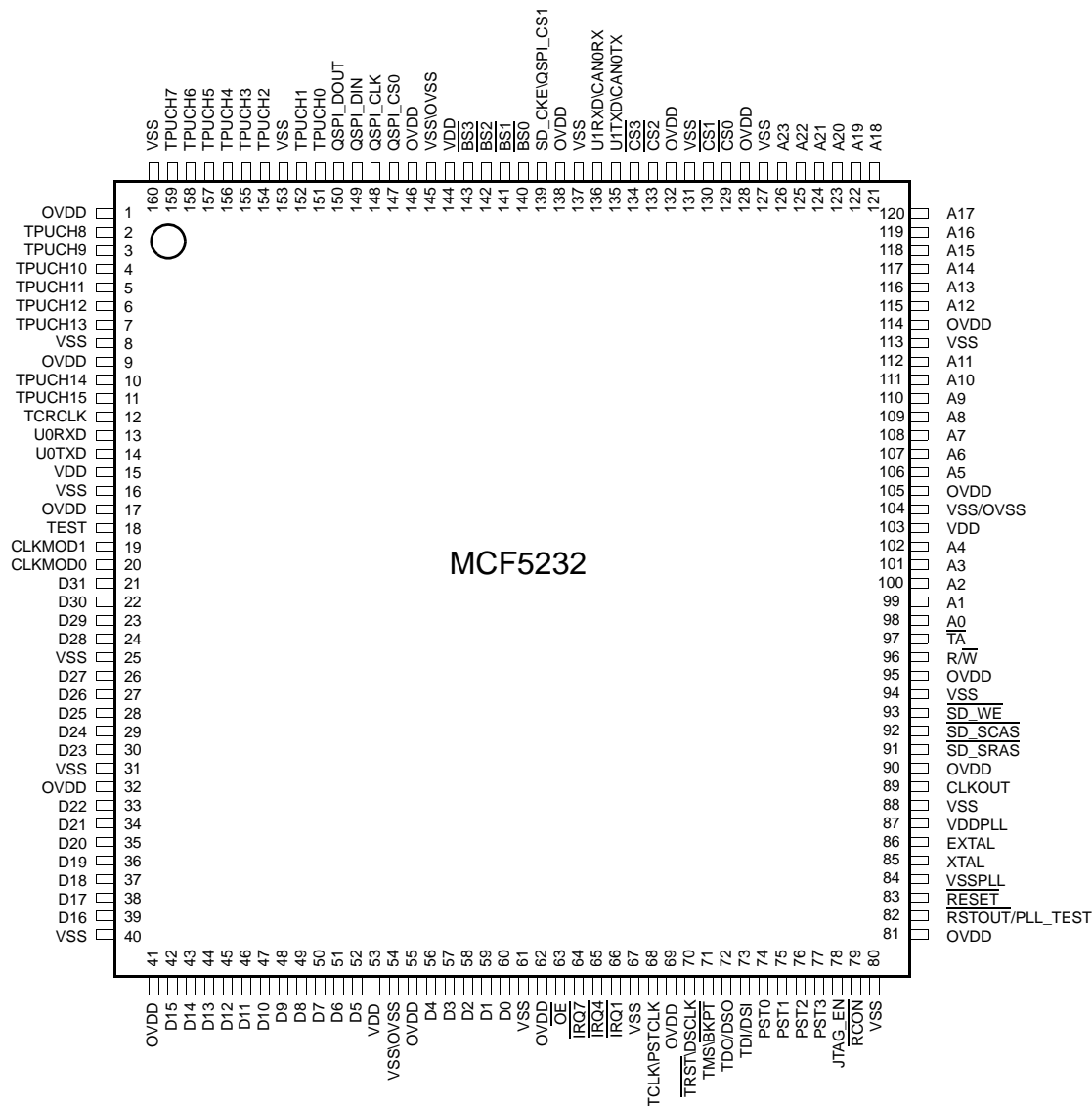
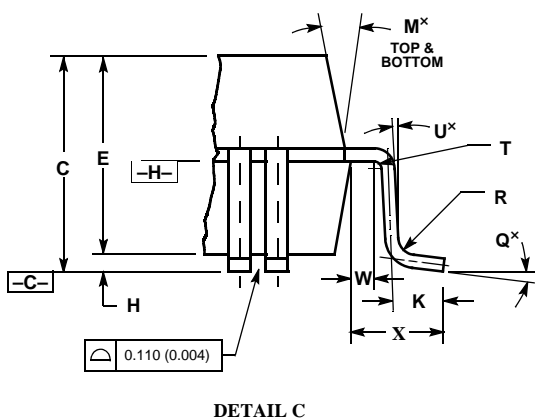


Figure 8. MCF5232CABxxx Pinout (160 QFP)

[illegible]

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.

2. CONTROLLING DIMENSION: MILLIMETER
3. DATUM PLAN -H- IS LOCATED AT BOTTOM OF LEAD AND IS COINCIDENT WITH THE LEAD WHERE THE LEAD EXITS THE PLASTIC BODY AT THE BOTTOM OF THE PARTING LINE.
4. DATUMS -A-, -B-, AND -D- TO BE DETERMINED AT DATUM PLANE -H-.
5. DIMENSIONS S AND V TO BE DETERMINED AT SEATING PLANE -C-.
6. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS 0.25 (0.010) PER SIDE. DIMENSIONS A AND B DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT DATUM PLANE -H-.
7. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08 (0.003) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION. DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OR THE FOOT.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	27.90	28.10	1.098	1.106
B	27.90	28.10	1.098	1.106
C	3.35	3.85	0.132	0.166
D	0.22	0.38	0.009	0.015
E	3.20	3.50	0.126	0.138
F	0.22	0.33	0.009	0.013
G	0.65 BSC		0.026 REF	
H	0.25	0.35	0.010	0.014
J	0.11	0.23	0.004	0.009
K	0.70	0.90	0.028	0.035
L	25.35 BSC		0.996 REF	
M	5	16	5	16
N	0.11	0.19	0.004	0.007
P	0.325 BSC		0.013 REF	
Q	0	7	0	7
R	0.13	0.30	0.005	0.012
S	31.00	31.40	1.220	1.236
T	0.13	—	0.005	—
U	0	—	0	—
V	31.00	31.40	1.220	1.236
W	0.4	—	0.016	—
X	1.60 REF		0.063 REF	
Y	1.33 REF		0.052 REF	
Z	1.33 REF		0.052 REF	

**Case 864A-03**

**Figure 9. 160 QFP Package Dimensions**

Table 9. DC Electrical Specifications<sup>1</sup> (continued)

Characteristic	Symbol	Min	Typical	Max	Unit
Input Capacitance <sup>3</sup> All input-only pins All input/output (three-state) pins	$C_{in}$	— —	—	7 7	pF
Load Capacitance <sup>4</sup> Low drive strength High drive strength	$C_L$		— —	25 50	pF pF
Core Operating Supply Current <sup>5</sup> Master Mode	$I_{DD}$	—	135	150	mA
Pad Operating Supply Current Master Mode Low Power Modes	$O_{I_{DD}}$	— —	100 TBD	— —	mA $\mu$ A
DC Injection Current <sup>3, 6, 7, 8</sup> $V_{NEGCLAMP} = V_{SS} - 0.3$ V, $V_{POSCLAMP} = V_{DD} + 0.3$ Single Pin Limit Total processor Limit, Includes sum of all stressed pins	$I_{IC}$	—1.0 —10		1.0 10	mA mA

<sup>1</sup> Refer to Table 10 for additional PLL specifications.

<sup>2</sup> Refer to the MCF5235 signals section for pins having weak internal pull-up devices.

<sup>3</sup> This parameter is characterized before qualification rather than 100% tested.

<sup>4</sup> pF load ratings are based on DC loading and are provided as an indication of driver strength. High speed interfaces require transmission line analysis to determine proper drive strength and termination. See [High Speed Signal Propagation: Advanced Black Magic](#) by Howard W. Johnson for design guidelines.

<sup>5</sup> Current measured at maximum system clock frequency, all modules active, and default drive strength with matching load.

<sup>6</sup> All functional non-supply pins are internally clamped to  $V_{SS}$  and their respective  $V_{DD}$ .

<sup>7</sup> Input must be current limited to the value specified. To determine the value of the required current-limiting resistor, calculate resistance values for positive and negative clamp voltages, then use the larger of the two values.

<sup>8</sup> Power supply must maintain regulation within operating  $V_{DD}$  range during instantaneous and operating maximum current conditions. If positive injection current ( $V_{in} > V_{DD}$ ) is greater than  $I_{DD}$ , the injection current may flow out of  $V_{DD}$  and could result in external power supply going out of regulation. Insure external  $V_{DD}$  load will shunt current greater than maximum injection current. This will be the greatest risk when the processor is not consuming power. Examples are: if no system clock is present, or if clock rate is very low which would reduce overall power consumption. Also, at power-up, system clock is not present during the power-up sequence until the PLL has attained lock.

## Electrical Characteristics

- <sup>7</sup> This specification applies to the period required for the PLL to relock after changing the MFD frequency control bits in the synthesizer control register (SYNCR).
- <sup>8</sup> Assuming a reference is available at power up, lock time is measured from the time  $V_{DD}$  and  $V_{DDSYN}$  are valid to  $\overline{RSTOUT}$  negating. If the crystal oscillator is being used as the reference for the PLL, then the crystal start up time must be added to the PLL lock time to determine the total start-up time.
- <sup>9</sup>  $t_{PLL} = (64 \times 4 \times 5 + 5 \times \tau) \times T_{ref}$ , where  $T_{ref} = 1/F_{ref\_crystal} = 1/F_{ref\_ext} = 1/F_{ref\_1:1}$ , and  $\tau = 1.57 \times 10^{-6} \times 2(MFD + 2)$ .
- <sup>10</sup> PLL is operating in 1:1 PLL mode.
- <sup>11</sup> Jitter is the average deviation from the programmed frequency measured over the specified interval at maximum  $f_{sys/2}$ . Measurements are made with the device powered by filtered supplies and clocked by a stable external clock signal. Noise injected into the PLL circuitry via  $V_{DDSYN}$  and  $V_{SSSYN}$  and variation in crystal oscillator frequency increase the Cjitter percentage for a given interval.
- <sup>12</sup> Values are with frequency modulation disabled. If frequency modulation is enabled, jitter is the sum of Cjitter+Cmod.
- <sup>13</sup> Modulation percentage applies over an interval of 10 $\mu$ s, or equivalently the modulation rate is 100KHz.
- <sup>14</sup> Modulation rate selected must not result in  $f_{sys/2}$  value greater than the  $f_{sys/2}$  maximum specified value. Modulation range determined by hardware design.
- <sup>15</sup>  $f_{sys/2} = f_{ico} / (2 \times 2^{RFD})$

## 7.5 External Interface Timing Characteristics

Table 11 lists processor bus input timings.

### NOTE

All processor bus timings are synchronous; that is, input setup/hold and output delay with respect to the rising edge of a reference clock. The reference clock is the CLKOUT output.

All other timing relationships can be derived from these values.

**Table 11. Processor Bus Input Timing Specifications**

Name	Characteristic <sup>1</sup>	Symbol	Min	Max	Unit
freq	System bus frequency	$f_{sys/2}$	50	75	MHz
B0	CLKOUT period	$t_{cyc}$	—	1/75	ns
<b>Control Inputs</b>					
B1a	Control input valid to CLKOUT high <sup>2</sup>	$t_{CVCH}$	9	—	ns
B1b	$\overline{BKPT}$ valid to CLKOUT high <sup>3</sup>	$t_{BKVCH}$	9	—	ns
B2a	CLKOUT high to control inputs invalid <sup>2</sup>	$t_{CHCII}$	0	—	ns
B2b	CLKOUT high to asynchronous control input $\overline{BKPT}$ invalid <sup>3</sup>	$t_{BKNCH}$	0	—	ns
<b>Data Inputs</b>					
B4	Data input (D[31:0]) valid to CLKOUT high	$t_{DIVCH}$	4	—	ns
B5	CLKOUT high to data input (D[31:0]) invalid	$t_{CHDII}$	0	—	ns

<sup>1</sup> Timing specifications are tested using full drive strength pad configurations in a 50ohm transmission line environment..

<sup>2</sup> TEA and TA pins are being referred to as control inputs.

<sup>3</sup> Refer to figure A-19.

Table 12. External Bus Output Timing Specifications (continued)

Name	Characteristic	Symbol	Min	Max	Unit
<b>Address and Attribute Outputs</b>					
B8	CLKOUT high to address (A[23:0]) and control ( $\overline{TS}$ , $\overline{TSIZ}[1:0]$ , $\overline{TIP}$ , $\overline{R/W}$ ) valid	$t_{CHAV}$	—	9	ns
B9	CLKOUT high to address (A[23:0]) and control ( $\overline{TS}$ , $\overline{TSIZ}[1:0]$ , $\overline{TIP}$ , $\overline{R/W}$ ) invalid	$t_{CHAI}$	1.5	—	ns
<b>Data Outputs</b>					
B11	CLKOUT high to data output (D[31:0]) valid	$t_{CHDOV}$	—	9	ns
B12	CLKOUT high to data output (D[31:0]) invalid	$t_{CHDOI}$	1.5	—	ns
B13	CLKOUT high to data output (D[31:0]) high impedance	$t_{CHDOZ}$	—	9	ns

<sup>1</sup>  $\overline{CS}$  transitions after the falling edge of CLKOUT.<sup>2</sup>  $\overline{BS}$  transitions after the falling edge of CLKOUT.<sup>3</sup>  $\overline{OE}$  transitions after the falling edge of CLKOUT.

Figure 13 shows an SRAM bus cycle terminated by  $\overline{\text{TEA}}$  showing timings listed in Table 12.

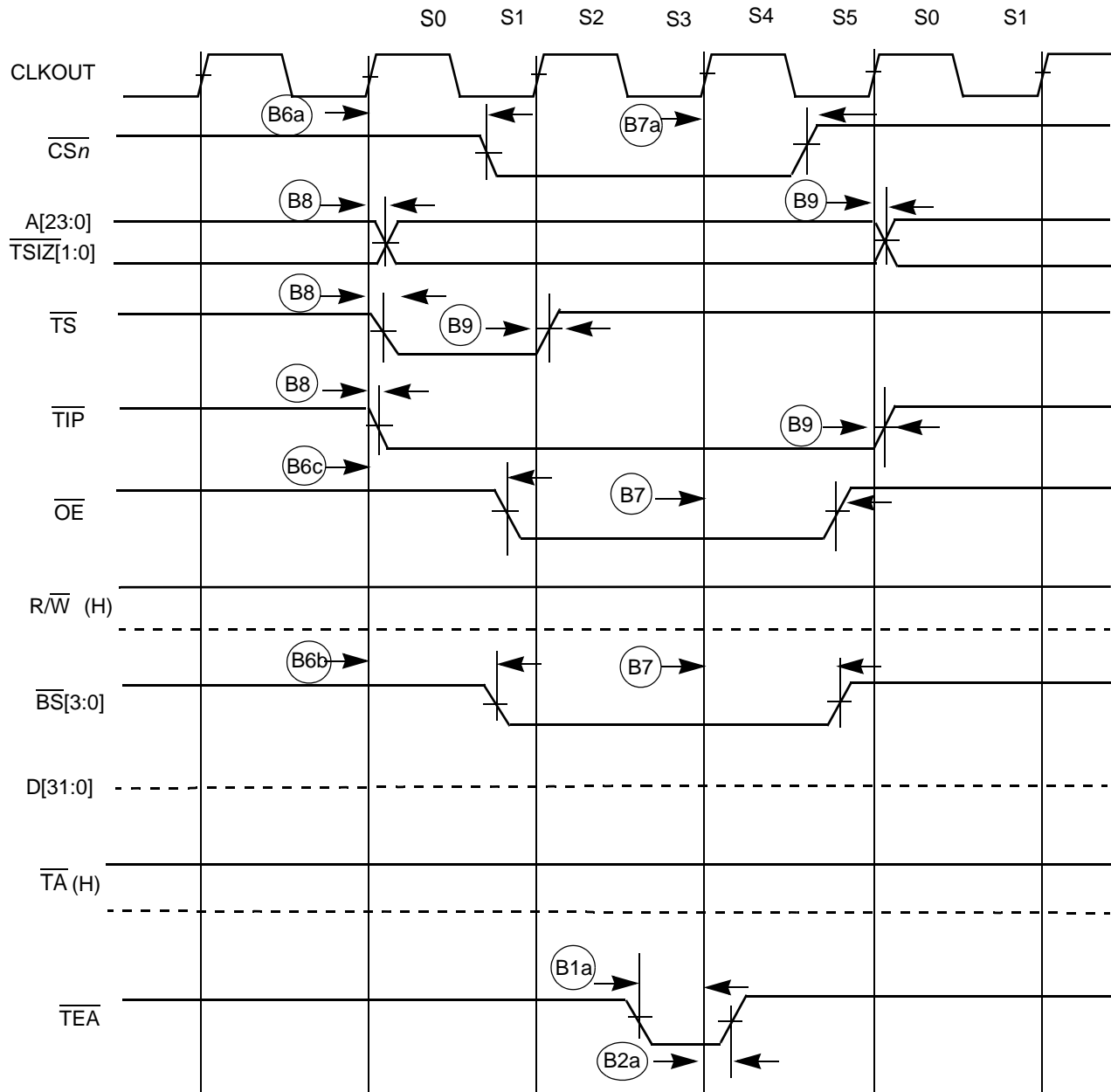


Figure 13. SRAM Read Bus Cycle Terminated by  $\overline{\text{TEA}}$

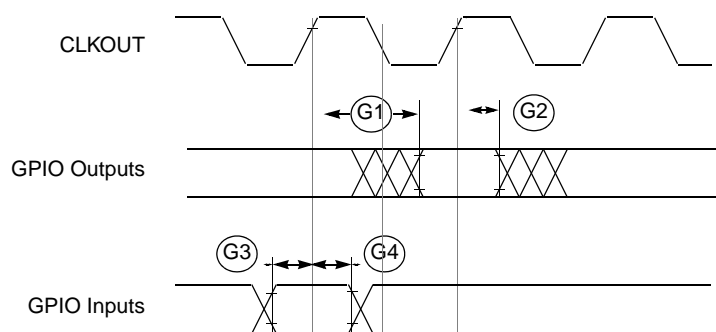


Figure 16. GPIO Timing

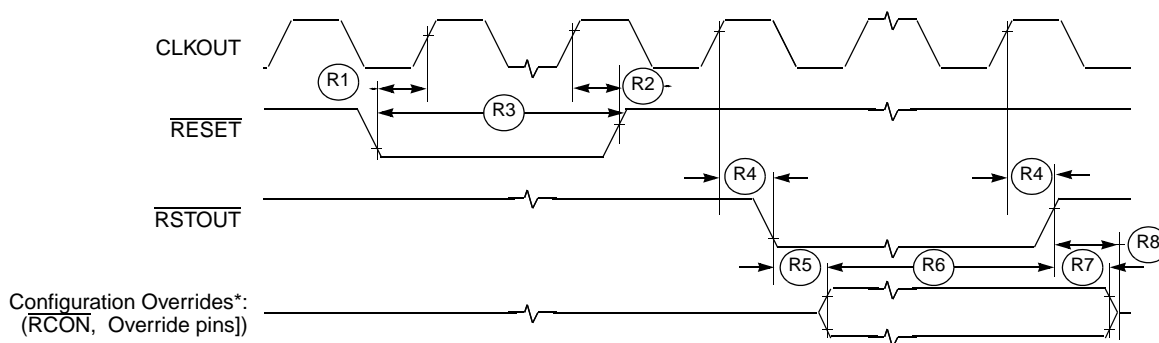
## 7.8 Reset and Configuration Override Timing

Table 15. Reset and Configuration Override Timing  
( $V_{DD} = 2.7$  to  $3.6$  V,  $V_{SS} = 0$  V,  $T_A = T_L$  to  $T_H$ )<sup>1</sup>

NUM	Characteristic	Symbol	Min	Max	Unit
R1	$\overline{\text{RESET}}$ Input valid to CLKOUT High	$t_{RVCH}$	9	—	ns
R2	CLKOUT High to $\overline{\text{RESET}}$ Input invalid	$t_{CHRI}$	1.5	—	ns
R3	$\overline{\text{RESET}}$ Input valid Time <sup>2</sup>	$t_{RIVT}$	5	—	$t_{CYC}$
R4	CLKOUT High to $\overline{\text{RSTOUT}}$ Valid	$t_{CHROV}$	—	10	ns
R5	$\overline{\text{RSTOUT}}$ valid to Config. Overrides valid	$t_{ROVCV}$	0	—	ns
R6	Configuration Override Setup Time to $\overline{\text{RSTOUT}}$ invalid	$t_{COS}$	20	—	$t_{CYC}$
R7	Configuration Override Hold Time after $\overline{\text{RSTOUT}}$ invalid	$t_{COH}$	0	—	ns
R8	$\overline{\text{RSTOUT}}$ invalid to Configuration Override High Impedance	$t_{ROICZ}$	—	1	$t_{CYC}$

<sup>1</sup> All AC timing is shown with respect to 50%  $V_{DD}$  levels unless otherwise noted.

<sup>2</sup> During low power STOP, the synchronizers for the  $\overline{\text{RESET}}$  input are bypassed and  $\overline{\text{RESET}}$  is asserted asynchronously to the system. Thus,  $\overline{\text{RESET}}$  must be held a minimum of 100 ns.

Figure 17.  $\overline{\text{RESET}}$  and Configuration Override Timing

Refer to the chip configuration module (CCM) chapter in the device's reference manual for more information.

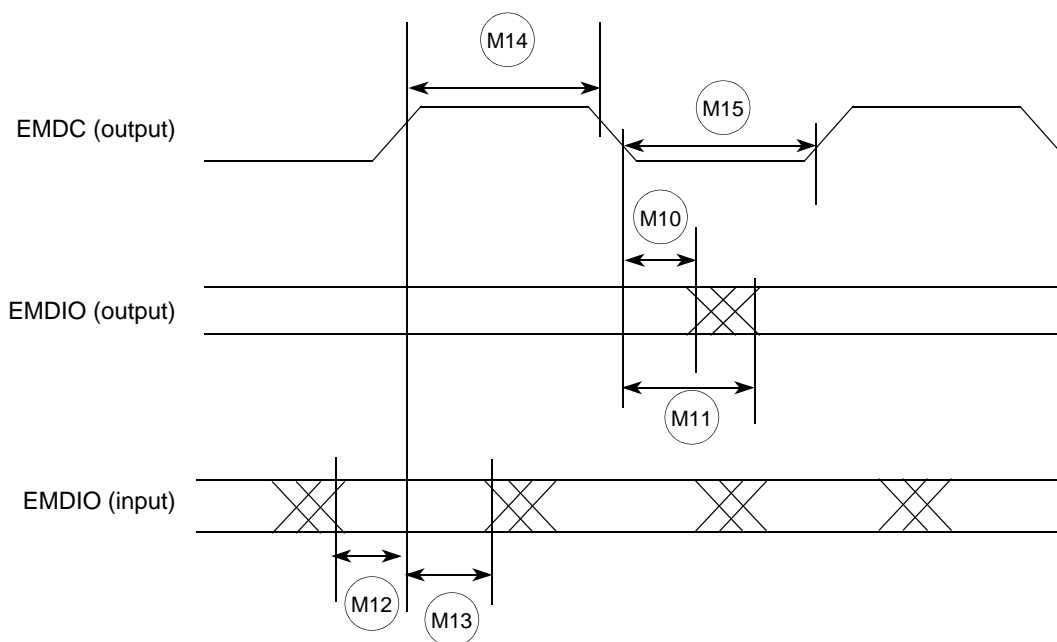
### 7.10.4 MII Serial Management Channel Timing (EMDIO and EMDC)

Table 21 lists MII serial management channel timings. The FEC functions correctly with a maximum MDC frequency of 2.5 MHz.

**Table 21. MII Serial Management Channel Timing**

Num	Characteristic	Min	Max	Unit
M10	EMDC falling edge to EMDIO output invalid (minimum propagation delay)	0	—	ns
M11	EMDC falling edge to EMDIO output valid (max prop delay)	—	25	ns
M12	EMDIO (input) to EMDC rising edge setup	10	—	ns
M13	EMDIO (input) to EMDC rising edge hold	0	—	ns
M14	EMDC pulse width high	40%	60%	MDC period
M15	EMDC pulse width low	40%	60%	MDC period

Figure 22 shows MII serial management channel timings listed in Table 21.



**Figure 22. MII Serial Management Channel Timing Diagram**

## 7.13 JTAG and Boundary Scan Timing

Table 24. JTAG and Boundary Scan Timing

Num	Characteristics <sup>1</sup>	Symbol	Min	Max	Unit
J1	TCLK Frequency of Operation	$f_{JCYC}$	DC	1/4	$f_{sys/2}$
J2	TCLK Cycle Period	$t_{JCYC}$	4	—	$t_{CYC}$
J3	TCLK Clock Pulse Width	$t_{JCW}$	26	—	ns
J4	TCLK Rise and Fall Times	$t_{JCRF}$	0	3	ns
J5	Boundary Scan Input Data Setup Time to TCLK Rise	$t_{BSDST}$	4	—	ns
J6	Boundary Scan Input Data Hold Time after TCLK Rise	$t_{BSDHT}$	26	—	ns
J7	TCLK Low to Boundary Scan Output Data Valid	$t_{BSDV}$	0	33	ns
J8	TCLK Low to Boundary Scan Output High Z	$t_{BSDZ}$	0	33	ns
J9	TMS, TDI Input Data Setup Time to TCLK Rise	$t_{TAPBST}$	4	—	ns
J10	TMS, TDI Input Data Hold Time after TCLK Rise	$t_{TAPBHT}$	10	—	ns
J11	TCLK Low to TDO Data Valid	$t_{TDODV}$	0	26	ns
J12	TCLK Low to TDO High Z	$t_{TDODZ}$	0	8	ns
J13	$\overline{TRST}$ Assert Time	$t_{TRSTAT}$	100	—	ns
J14	$\overline{TRST}$ Setup Time (Negation) to TCLK High	$t_{TRSTST}$	10	—	ns

<sup>1</sup> JTAG\_EN is expected to be a static signal. Hence, specific timing is not associated with it.

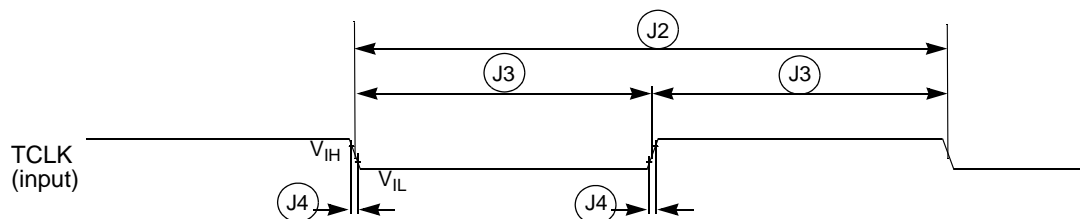


Figure 24. Test Clock Input Timing

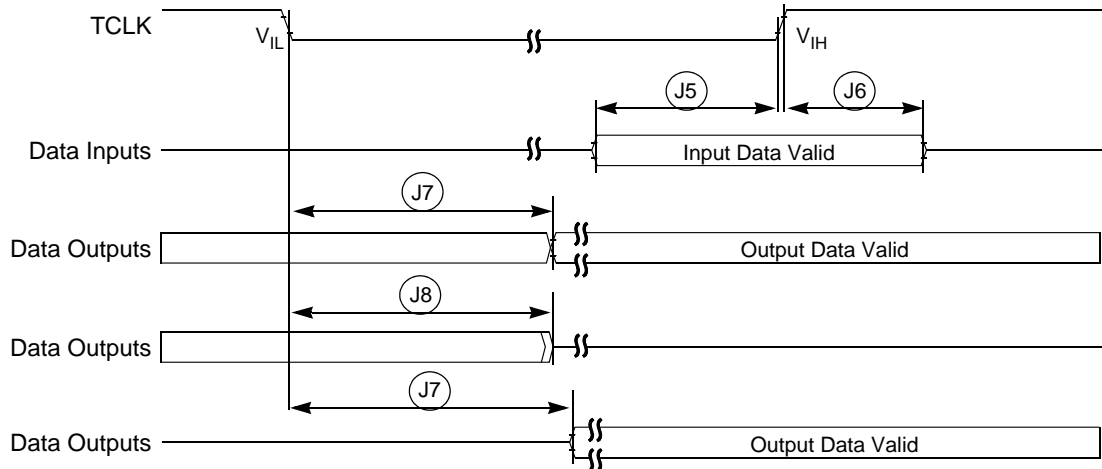


Figure 25. Boundary Scan (JTAG) Timing

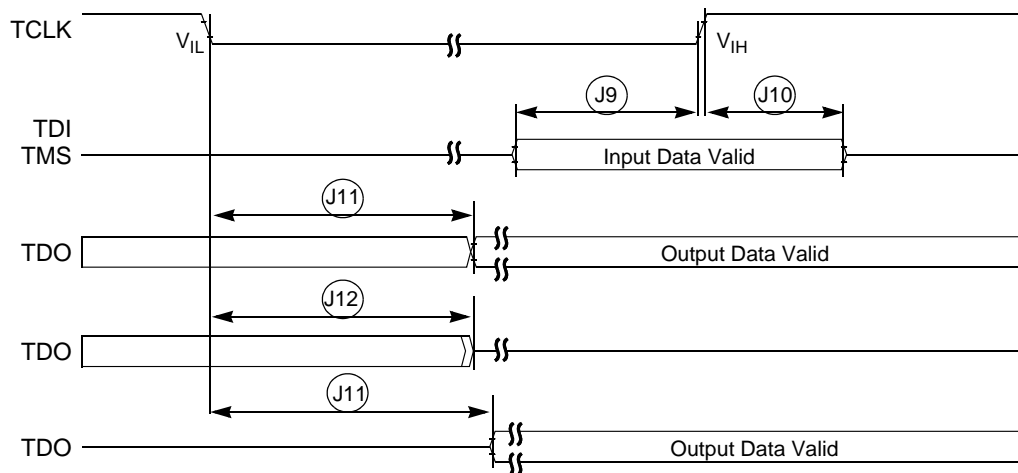


Figure 26. Test Access Port Timing

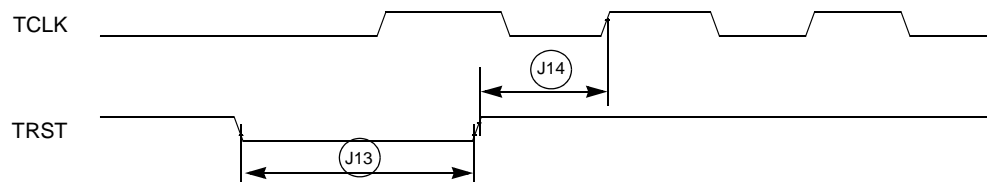


Figure 27. TRST Timing

## 7.14 Debug AC Timing Specifications

Table 25 lists specifications for the debug AC timing parameters shown in Figure 29.

Table 25. Debug AC Timing Specification

Num	Characteristic	150 MHz		Units
		Min	Max	
DE0	PSTCLK cycle time	—	0.5	$t_{cyc}$
DE1	PST valid to PSTCLK high	4	—	ns
DE2	PSTCLK high to PST invalid	1.5	—	ns
DE3	DSCLK cycle time	5	—	$t_{cyc}$
DE4	DSI valid to DSCLK high	1	—	$t_{cyc}$
DE5 <sup>1</sup>	DSCLK high to DSO invalid	4	—	$t_{cyc}$
DE6	$\overline{BKPT}$ input data setup time to CLKOUT rise	4	—	ns
DE7	CLKOUT high to $\overline{BKPT}$ high Z	0	10	ns

<sup>1</sup> DSCLK and DSI are synchronized internally. D4 is measured from the synchronized DSCLK input relative to the rising edge of CLKOUT.

Figure 28 shows real-time trace timing for the values in Table 25.

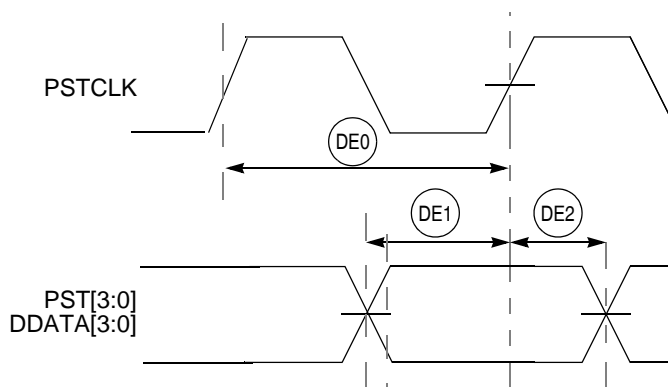


Figure 28. Real-Time Trace AC Timing

Figure 29 shows BDM serial port AC timing for the values in Table 25.



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