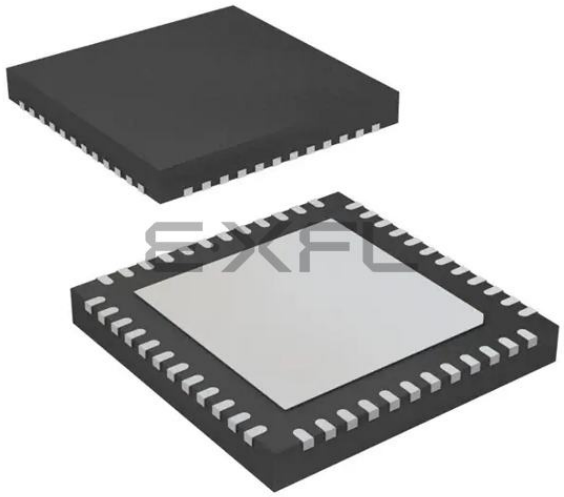

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I²C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	44
Program Memory Size	28KB (16K x 14)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	2.3V ~ 5.5V
Data Converters	A/D 43x10b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-UFQFN Exposed Pad
Supplier Device Package	48-UQFN (6x6)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16f15386t-i-mv

TABLE 5: 48-PIN ALLOCATION TABLE (PIC16(L)F15385, PIC16(L)F15386) (CONTINUED)

I/O[2]	48-Pin UQFN/TQFP	ADC	Reference	Comparator	NCO	DAC	Timers	CCP	PWM	CWG	MSSP	ZCD	EUSART	CLC	CLKR	Interrupt	Pull-up	Basic
RF5	13	ANF5	—	—	—	—	—	—	—	—	—	—	—	—	—	—	Y	—
RF6	14	ANF6	—	—	—	—	—	—	—	—	—	—	—	—	—	—	Y	—
RF7	15	ANF7	—	—	—	—	—	—	—	—	—	—	—	—	—	—	Y	—
VDD	30	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	Y	VDD
VDD	7	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	VDD
VSS	6	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	VSS
VSS	31	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	VSS
OUT[2]	—	—	—	C1OUT	NCO1OUT	—	TMR0	CCP1	PWM3OUT	CWG1A CWG2A	SDO1 SDO2	—	DT[3]	CLC1OUT	CLKR	—	—	—
	—	—	—	C2OUT	—	—	—	CCP2	PWM4OUT	CWG1B CWG2B	SCK1 SCK2	—	CK1 CK2	CLC2OUT	—	—	—	—
	—	—	—	—	—	—	—	—	PWM5OUT	CWG1C CWG2C	SCK1[3,4] SCL2[3,4]	—	TX1 TX2	CLC3OUT	—	—	—	—
	—	—	—	—	—	—	—	—	PWM6OUT	CWG1D CWG2D	SDA1[3,4] SDA2[3,4]	—	—	CLC4OUT	—	—	—	—

Note 1: This is a PPS re-mappable input signal. The input function may be moved from the default location shown to one of several other PORTx pins.

2: All digital output signals shown in this row are PPS re-mappable. These signals may be mapped to output onto one of several PORTx pin options.

3: This is a bidirectional signal. For normal module operation, the firmware should map this signal to the same pin in both the PPS input and PPS output registers.

4: These pins are configured for I^2C logic levels. PPS assignments to the other pins will operate, but input logic levels will be standard TTL/ST as selected by the INLVL register, instead of the I^2C specific or SMBUS input buffer thresholds.

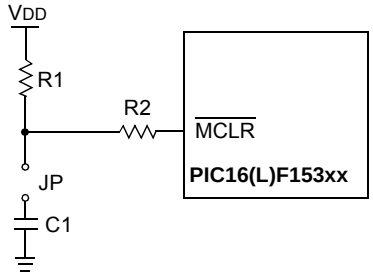
2.3 Master Clear ($\overline{MCLR}$) Pin

The $\overline{MCLR}$ pin provides two specific device functions: Device Reset, and Device Programming and Debugging. If programming and debugging are not required in the end application, a direct connection to V_{DD} may be all that is required. The addition of other components, to help increase the application's resistance to spurious Resets from voltage sags, may be beneficial. A typical configuration is shown in Figure 2-1. Other circuit designs may be implemented, depending on the application's requirements.

During programming and debugging, the resistance and capacitance that can be added to the pin must be considered. Device programmers and debuggers drive the $\overline{MCLR}$ pin. Consequently, specific voltage levels (V_{IH} and V_{IL}) and fast signal transitions must not be adversely affected. Therefore, specific values of R1 and C1 will need to be adjusted based on the application and PCB requirements. For example, it is recommended that the capacitor, C1, be isolated from the $\overline{MCLR}$ pin during programming and debugging operations by using a jumper (Figure 2-2). The jumper is replaced for normal run-time operations.

Any components associated with the $\overline{MCLR}$ pin should be placed within 0.25 inch (6 mm) of the pin.

FIGURE 2-2: EXAMPLE OF $\overline{MCLR}$ PIN CONNECTIONS



Note 1: R1 ≤ 10 kΩ is recommended. A suggested starting value is 10 kΩ. Ensure that the $\overline{MCLR}$ pin V_{IH} and V_{IL} specifications are met.

2: R2 ≤ 470Ω will limit any current flowing into $\overline{MCLR}$ from the external capacitor, C1, in the event of $\overline{MCLR}$ pin breakdown, due to Electrostatic Discharge (ESD) or Electrical Overstress (EOS). Ensure that the $\overline{MCLR}$ pin V_{IH} and V_{IL} specifications are met.

2.4 ICSP™ Pins

The PGC and PGD pins are used for In-Circuit Serial Programming™ (ICSP™) and debugging purposes. It is recommended to keep the trace length between the ICSP connector and the ICSP pins on the device as short as possible. If the ICSP connector is expected to experience an ESD event, a series resistor is recommended, with the value in the range of a few tens of ohms, not to exceed 100Ω.

Pull-up resistors, series diodes and capacitors on the PGC and PGD pins are not recommended as they will interfere with the programmer/debugger communications to the device. If such discrete components are an application requirement, they should be removed from the circuit during programming and debugging. Alternatively, refer to the AC/DC characteristics and timing requirements information in the respective device Flash programming specification for information on capacitive loading limits, and pin input voltage high (V_{IH}) and input low (V_{IL}) requirements.

For device emulation, ensure that the "Communication Channel Select" (i.e., PGCx/PGDx pins), programmed into the device, matches the physical connections for the ICSP to the Microchip debugger/emulator tool.

For more information on available Microchip development tools connection requirements, refer to **Section 39.0 "Development Support"**.

TABLE 4-9: PIC16(L)F15356/75/76/85/86 MEMORY MAP, BANKS 60, 61, 62, AND 63

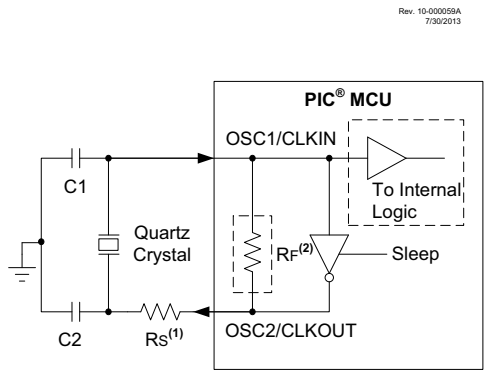
	Bank 60		Bank 61		Bank 62		Bank 63
1E0Ch	—	1E8Ch	RF7PPS(2)	1F0Ch	—	1F8Ch	—
1E0Dh	—	1E8Dh	—	1F0Dh	—	1F8Dh	—
1E0Eh	—	1E8Eh	—	1F0Eh	—	1F8Eh	—
1E0Fh	CLCDATA	1E8Fh	PPSLOCK	1F0Fh	—	1F8Fh	—
1E10h	CLC1CON	1E90h	INTPPS	1F10h	RA0PPS	1F90h	—
1E11h	CLC1POL	1E91h	T0CKIPPS	1F11h	RA1PPS	1F91h	—
1E12h	CLC1SEL0	1E92h	T1CKIPPS	1F12h	RA2PPS	1F92h	—
1E13h	CLC1SEL1	1E93h	T1GPPS	1F13h	RA3PPS	1F93h	—
1E14h	CLC1SEL2	1E94h	—	1F14h	RA4PPS	1F94h	—
1E15h	CLC1SEL3	1E95h	—	1F15h	RA5PPS	1F95h	—
1E16h	CLC1GLS0	1E96h	—	1F16h	RA6PPS	1F96h	—
1E17h	CLC1GLS1	1E97h	—	1F17h	RA7PPS	1F97h	—
1E18h	CLC1GLS2	1E98h	—	1F18h	RB0PPS	1F98h	—
1E19h	CLC1GLS3	1E99h	—	1F19h	RB1PPS	1F99h	—
1E1Ah	CLC2CON	1E9Ah	—	1F1Ah	RB2PPS	1F9Ah	—
1E1Bh	CLC2POL	1E9Bh	—	1F1Bh	RB3PPS	1F9Bh	—
1E1Ch	CLC2SEL0	1E9Ch	T2INPPS	1F1Ch	RB4PPS	1F9Ch	—
1E1Dh	CLC2SEL1	1E9Dh	—	1F1Dh	RB5PPS	1F9Dh	—
1E1Eh	CLC2SEL2	1E9Eh	—	1F1Eh	RB6PPS	1F9Eh	—
1E1Fh	CLC2SEL3	1E9Fh	—	1F1Fh	RB7PPS	1F9Fh	—
1E20h	CLC2GLS0	1EA0h	—	1F20h	RC0PPS	1FA0h	—
1E21h	CLC2GLS1	1EA1h	CCP1PPS	1F21h	RC1PPS	1FA1h	—
1E22h	CLC2GLS2	1EA2h	CCP2PPS	1F22h	RC2PPS	1FA2h	—
1E23h	CLC2GLS3	1EA3h	—	1F23h	RC3PPS	1FA3h	—
1E24h	CLC3CON	1EA4h	—	1F24h	RC4PPS	1FA4h	—
1E25h	CLC3POL	1EA5h	—	1F25h	RC5PPS	1FA5h	—
1E26h	CLC3SEL0	1EA6h	—	1F26h	RC6PPS	1FA6h	—
1E27h	CLC3SEL1	1EA7h	—	1F27h	RC7PPS	1FA7h	—
1E28h	CLC3SEL2	1EA8h	—	1F28h	RD0PPS(1)	1FA8h	—
1E29h	CLC3SEL3	1EA9h	—	1F29h	RD1PPS(1)	1FA9h	—
1E2Ah	CLC3GLS0	1EAAh	—	1F2Ah	RD2PPS(1)	1FAAh	—
1E2Bh	CLC3GLS1	1EABh	—	1F2Bh	RD3PPS(1)	1FABh	—
1E2Ch	CLC3GLS2	1EACh	—	1F2Ch	RD4PPS(1)	1FACh	—
1E2Dh	CLC3GLS3	1EADh	—	1F2Dh	RD5PPS(1)	1FADh	—
1E2Eh	CLC4CON	1EAEh	—	1F2Eh	RD6PPS(1)	1FAEh	—
1E2Fh	CLC4POL	1EAFh	—	1F2Fh	RD7PPS(1)	1FAFh	—
1E30h	CLC4SEL0	1EB0h	—	1F30h	RE0PPS(1)	1FB0h	—
1E31h	CLC4SEL1	1EB1h	CWG1PPS	1F31h	RE1PPS(1)	1FB1h	—
1E32h	CLC4SEL2	1EB2h	—	1F32h	RE2PPS(1)	1FB2h	—
1E33h	CLC4SEL3	1EB3h	—	1F33h	—	1FB3h	—
1E34h	CLC4GLS0	1EB4h	—	1F34h	—	1FB4h	—
1E35h	CLC4GLS1	1EB5h	—	1F35h	—	1FB5h	—
1E36h	CLC4GLS2	1EB6h	—	1F36h	—	1FB6h	—
1E37h	CLC4GLS3	1EB7h	—	1F37h	—	1FB7h	—
1E38h	RF0PPS(2)	1EB8h	—	1F38h	ANSELA	1FB8h	—
1E39h	RF1PPS(2)	1EB9h	—	1F39h	WPUA	1FB9h	—
1E3Ah	RF2PPS(2)	1EBAh	—	1F3Ah	ODCONA	1FBAh	—
1E3Bh	RF3PPS(2)	1EBBh	CLCIN0PPS	1F3Bh	SLRCONA	1FBBh	—
1E3Ch	RF4PPS(2)	1EBCh	CLCIN1PPS	1F3Ch	INLVLA	1FBCh	—
1E3Dh	RF5PPS(2)	1EBDh	CLCIN2PPS	1F3Dh	IOCAP	1FBDh	—
1E3Eh	RF6PPS(2)	1EBEh	CLCIN3PPS	1F3Eh	IOCAN	1FBEh	—

Legend: = Unimplemented data memory locations, read as '0'

Note 1: Present only on PIC16(L)F15375/76/85/86.

2: Present only on PIC16(L)F15385/86

FIGURE 9-3: QUARTZ CRYSTAL OPERATION (LP, XT OR HS MODE)



Note 1: A series resistor (Rs) may be required for quartz crystals with low drive level.

2: The value of RF varies with the Oscillator mode selected (typically between 2 MΩ and 10 MΩ).

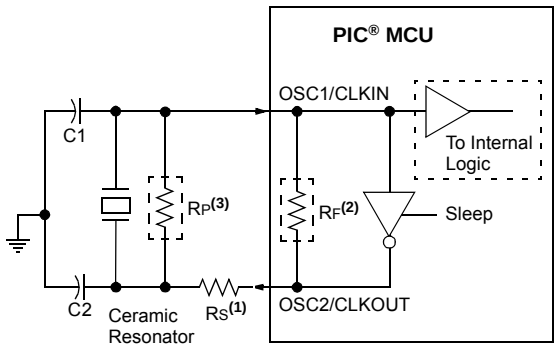
Note 1: Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.

2: Always verify oscillator performance over the VDD and temperature range that is expected for the application.

3: For oscillator design assistance, reference the following Microchip Application Notes:

- AN826, "*Crystal Oscillator Basics and Crystal Selection for rfPIC® and PIC® Devices*" (DS00826)
- AN849, "*Basic PIC® Oscillator Design*" (DS00849)
- AN943, "*Practical PIC® Oscillator Analysis and Design*" (DS00943)
- AN949, "*Making Your Oscillator Work*" (DS00949)

FIGURE 9-4: CERAMIC RESONATOR OPERATION (XT OR HS MODE)



Note 1: A series resistor (RS) may be required for ceramic resonators with low drive level.

2: The value of RF varies with the Oscillator mode selected (typically between 2 MΩ to 10 MΩ).

3: An additional parallel feedback resistor (RP) may be required for proper ceramic resonator operation.

9.2.1.3 Oscillator Start-up Timer (OST)

If the oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) counts 1024 oscillations from OSC1. This occurs following a Power-on Reset (POR), Brown-out Reset (BOR) or a wake-up from Sleep. The OST ensures that the oscillator circuit, using a quartz crystal resonator or ceramic resonator, has started and is providing a stable system clock to the oscillator module.

Changing the clock post-divider without changing the clock source (e.g., changing FOSC from 1 MHz to 2 MHz) is handled in the same manner as a clock source change, as described previously. The clock source will already be active, so the switch is relatively quick. CSWHOLD must be clear (CSWHOLD = 0) for the switch to complete.

The current COSC and CDIV are indicated in the OSCCON2 register up to the moment when the switch actually occurs, at which time OSCCON2 is updated and ORDY is set. NOSCR is cleared by hardware to indicate that the switch is complete.

9.3.2 PLL INPUT SWITCH

Switching between the PLL and any non-PLL source is managed as described above. The input to the PLL is established when NOSC selects the PLL, and maintained by the COSC setting.

When NOSC and COSC select the PLL with different input sources, the system continues to run using the COSC setting, and the new source is enabled per NOSC. When the new oscillator is ready (and CSWHOLD = 0), system operation is suspended while the PLL input is switched and the PLL acquires lock.

Note:	If the PLL fails to lock, the FSCM will trigger.

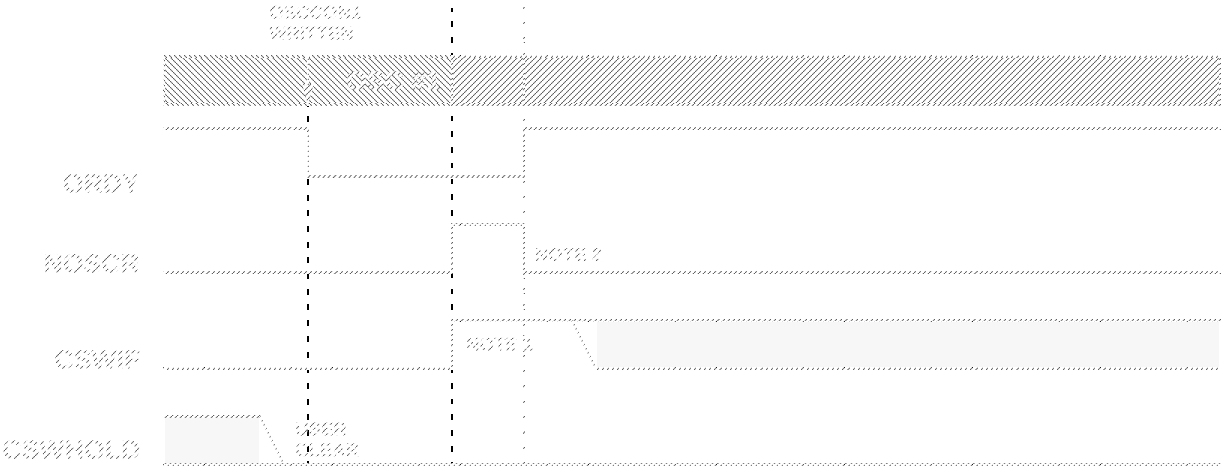
9.3.3 CLOCK SWITCH AND SLEEP

If OSCCON1 is written with a new value and the device is put to Sleep before the switch completes, the switch will not take place and the device will enter Sleep mode.

When the device wakes from Sleep and the CSWHOLD bit is clear, the device will wake with the 'new' clock active, and the clock switch interrupt flag bit (CSWIF) will be set.

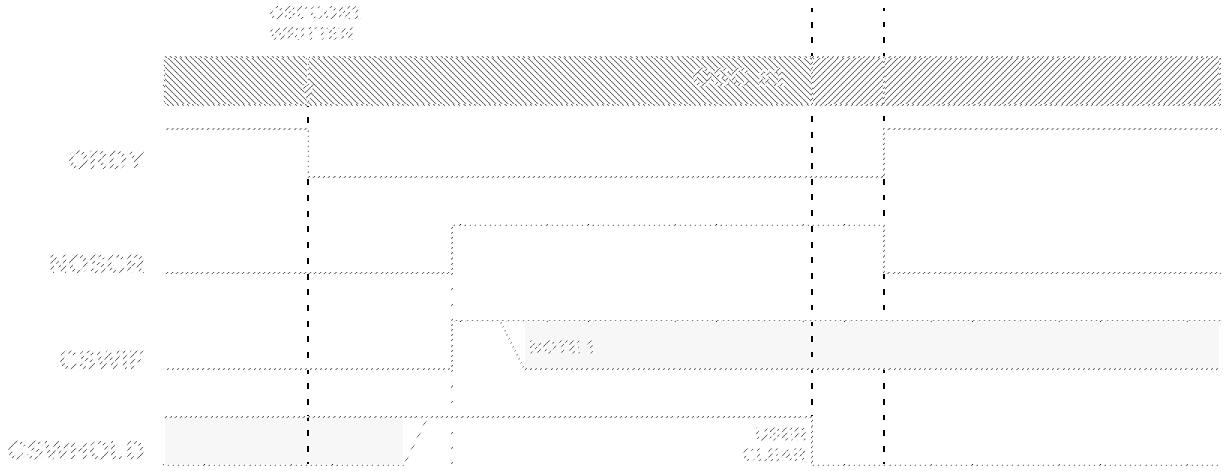
When the device wakes from Sleep and the CSWHOLD bit is set, the device will wake with the 'old' clock active and the new clock will be requested again.

FIGURE 9-6: CLOCK SWITCH (CSWHOLD = 0)



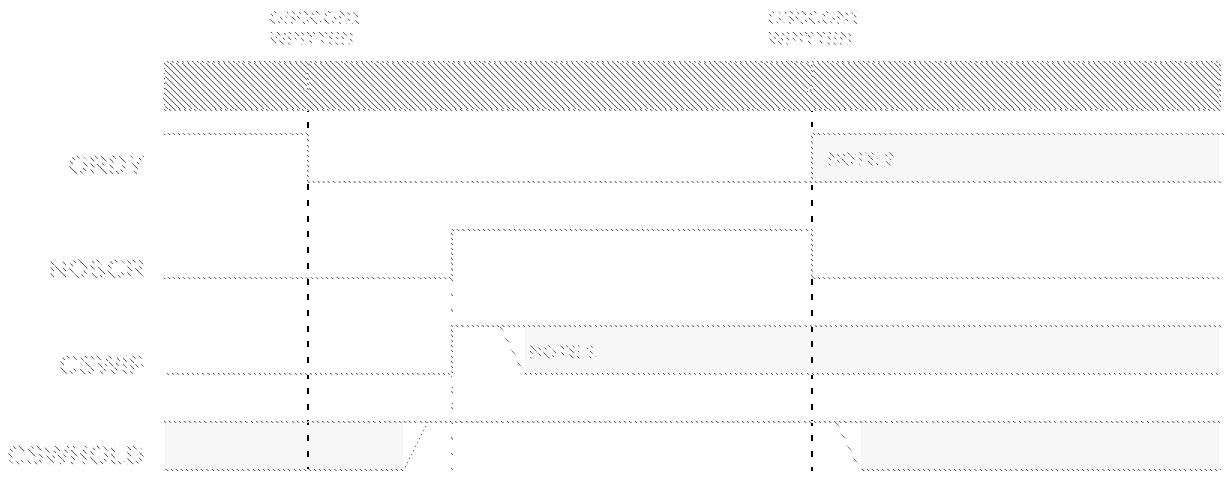
Note 1: CSWIF is asserted coincident with NOSCR; interrupt is serviced at OSC#2 speed.
2: The assertion of NOSCR is hidden from the user because it appears only for the duration of the switch.

FIGURE 9-7: CLOCK SWITCH (CSWHOLD = 1)



Note 1: CSWIF is asserted coincident with NOSCR, and may be cleared before or after clearing CSWHOLD = 0.

FIGURE 9-8: CLOCK SWITCH ABANDONED



Note 1: CSWIF may be cleared before or after rewriting OSCCON1; CSWIF is not automatically cleared.
2: ORDY = 0 if OSCCON1 does not match OSCCON2; a new switch will begin.

10.0 INTERRUPTS

The interrupt feature allows certain events to preempt normal program flow. Firmware is used to determine the source of the interrupt and act accordingly. Some interrupts can be configured to wake the MCU from Sleep mode.

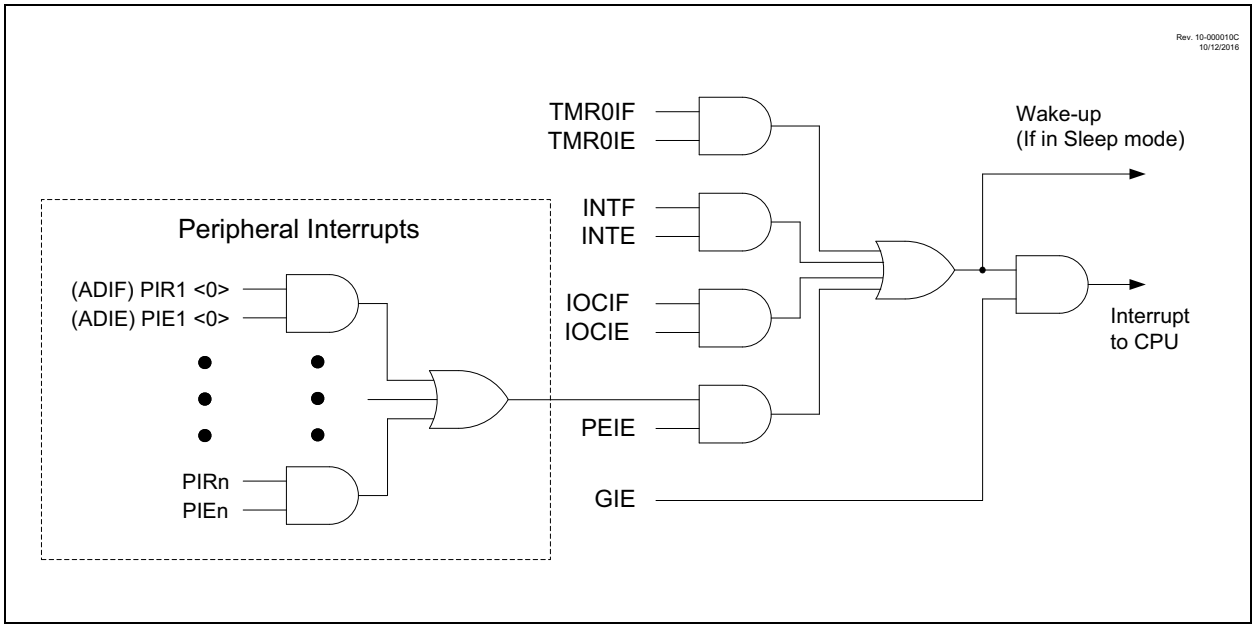
This chapter contains the following information for Interrupts:

- Operation
- Interrupt Latency
- Interrupts During Sleep
- INT Pin
- Automatic Context Saving

Many peripherals produce interrupts. Refer to the corresponding chapters for details.

A block diagram of the interrupt logic is shown in Figure 10-1.

FIGURE 10-1: INTERRUPT LOGIC



12.1 Independent Clock Source

The WDT can derive its time base from either the 31 kHz LFINTOSC or 31.25 kHz MFINTOSC internal oscillators, depending on the value of either the WDTCCS<2:0> Configuration bits or the WDTCS<2:0> bits of WDTCON1. Time intervals in this chapter are based on a minimum nominal interval of 1 ms. See **Section 37.0 "Electrical Specifications"** for LFINTOSC and MFINTOSC tolerances.

12.2 WDT Operating Modes

The Watchdog Timer module has four operating modes controlled by the WDTE<1:0> bits in Configuration Words. See Table 12-1.

12.2.1 WDT IS ALWAYS ON

When the WDTE bits of Configuration Words are set to '11', the WDT is always on.

WDT protection is active during Sleep.

12.2.2 WDT IS OFF IN SLEEP

When the WDTE bits of Configuration Words are set to '10', the WDT is on, except in Sleep.

WDT protection is not active during Sleep.

12.2.3 WDT CONTROLLED BY SOFTWARE

When the WDTE bits of Configuration Words are set to '01', the WDT is controlled by the SWDTEN bit of the WDTCON0 register.

12.2.4 WDT IS OFF

When the WDTE bits of the Configuration Word are set to '00', the WDT is always OFF.

WDT protection is unchanged by Sleep. See Table 12-1 for more details.

TABLE 12-1: WDT OPERATING MODES

WDTE<1:0>	SWDTEN	Device Mode	WDT Mode
11	X	X	Active
10	X	Awake	Active
		Sleep	Disabled
01	1	X	Active
	0	X	Disabled
00	X	X	Disabled

12.3 Time-Out Period

The WDTPS bits of the WDTCON0 register set the time-out period from 1 ms to 256 seconds (nominal). After a Reset, the default time-out period is two seconds.

12.4 Watchdog Window

The Watchdog Timer has an optional Windowed mode that is controlled by the WDTCWS<2:0> Configuration bits and WINDOW<2:0> bits of the WDTCON1 register. In the Windowed mode, the CLRWDT instruction must occur within the allowed window of the WDT period. Any CLRWDT instruction that occurs outside of this window will trigger a window violation and will cause a WDT Reset, similar to a WDT time out. See Figure 12-2 for an example.

The window size is controlled by the WDTCWS<2:0> Configuration bits, or the WINDOW<2:0> bits of WDTCON1, if WDTCWS<2:0> = 111.

In the event of a window violation, a Reset will be generated and the WDTWV bit of the PCON register will be cleared. This bit is set by a POR or can be set in firmware.

12.5 Clearing the WDT

The WDT is cleared when any of the following conditions occur:

- Any Reset
- Valid CLRWDT instruction is executed
- Device enters Sleep
- Device wakes up from Sleep
- WDT is disabled
- Oscillator Start-up Timer (OST) is running
- Any write to the WDTCON0 or WDTCON1 registers

12.5.1 CLRWDT CONSIDERATIONS (WINDOWED MODE)

When in Windowed mode, the WDT must be armed before a CLRWDT instruction will clear the timer. This is performed by reading the WDTCON0 register. Executing a CLRWDT instruction without performing such an arming action will trigger a window violation.

See Table 12-2 for more information.

12.6 Operation During Sleep

When the device enters Sleep, the WDT is cleared. If the WDT is enabled during Sleep, the WDT resumes counting. When the device exits Sleep, the WDT is cleared again.

The WDT remains clear until the OST, if enabled, completes. See **Section 9.0 "Oscillator Module (with Fail-Safe Clock Monitor)"** for more information on the OST.

When a WDT time-out occurs while the device is in Sleep, no Reset is generated. Instead, the device wakes up and resumes operation. The TO and PD bits in the STATUS register are changed to indicate the event. The RWDT bit in the PCON register can also be used. See **Section 4.3.2.1 "STATUS Register"** for more information.

TABLE 14-4: SUMMARY OF REGISTERS ASSOCIATED WITH PORTC

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	211
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	211
LATC	LATC7	LATC6	LATC5	LATC4	LATC3	LATC2	LATC1	LATC0	211
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	ANSC1	ANSC0	212
WPUC	WPUC7	WPUC6	WPUC5	WPUC4	WPUC3	WPUC2	WPUC1	WPUC0	212
ODCONC	ODCC7	ODCC6	ODCC5	ODCC4	ODCC3	ODCC2	ODCC1	ODCC0	213
SLRCONC	SLRC7	SLRC6	SLRC5	SLRC4	SLRC3	SLRC2	SLRC1	SLRC0	213
INLVLC	INLVLC7	INLVLC6	INLVLC5	INLVLC4	INLVLC3	INLVLC2	INLVLC1	INLVLC0	213

Legend: – = unimplemented locations read as '0'. Shaded cells are not used by PORTC.

REGISTER 16-2: PMD1: PMD CONTROL REGISTER 1

R/W-0/0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0
NCO1MD	—	—	—	—	TMR2MD	TMR1MD	TMR0MD
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	q = Value depends on condition

bit 7 **NCO1MD:** Disable Numerically Control Oscillator bit
1 = NCO1 module disabled
0 = NCO1 module enabled

bit 6-3 **Unimplemented:** Read as ‘0’

bit 2 **TMR2MD:** Disable Timer TMR2 bit
1 = Timer2 module disabled
0 = Timer2 module enabled

bit 1 **TMR1MD:** Disable Timer TMR1 bit
1 = Timer1 module disabled
0 = Timer1 module enabled

bit 0 **TMR0MD:** Disable Timer TMR0 bit
1 = Timer0 module disabled
0 = Timer0 module enabled

18.0 FIXED VOLTAGE REFERENCE (FVR)

The Fixed Voltage Reference, or FVR, is a stable voltage reference, independent of VDD, with 1.024V, 2.048V or 4.096V selectable output levels. The output of the FVR can be configured to supply a reference voltage to the following:

- ADC input channel
- ADC positive reference
- Comparator positive and negative input
- Digital-to-Analog Converter (DAC)

The FVR can be enabled by setting the FVREN bit of the FVRCON register.

Note:	Fixed Voltage Reference output cannot exceed VDD.

18.1 Independent Gain Amplifiers

The output of the FVR, which is connected to the ADC, comparators, and DAC, is routed through two independent programmable gain amplifiers. Each amplifier can be programmed for a gain of 1x, 2x or 4x, to produce the three possible voltage levels.

The ADFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the ADC module. Reference **Section 20.0 "Analog-to-Digital Converter (ADC) Module"** for additional information.

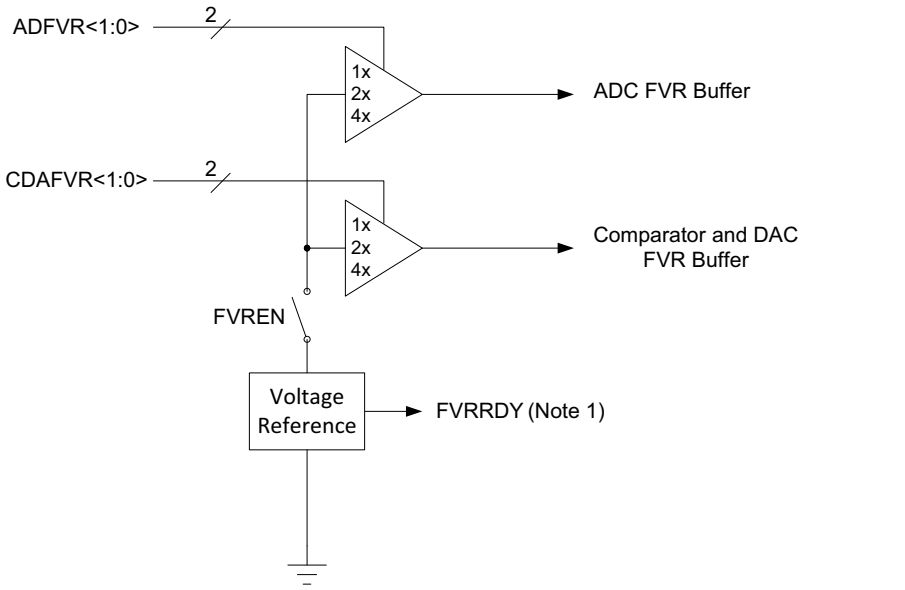
The CDAFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the DAC and comparator module. Reference **Section 21.0 "5-Bit Digital-to-Analog Converter (DAC1) Module"** and **Section 23.0 "Comparator Module"** for additional information.

18.2 FVR Stabilization Period

When the Fixed Voltage Reference module is enabled, it requires time for the reference and amplifier circuits to stabilize.

FVRRDY is an indicator of the reference being ready. In the case of an LF device, or a device on which the BOR is enabled in the Configuration Word settings, then the FVRRDY bit will be high prior to setting FVREN as those module require the reference voltage.

FIGURE 18-1: VOLTAGE REFERENCE BLOCK DIAGRAM



Note 1: FVRRDY is always '1'.
2: Any peripheral requiring the Fixed Reference (See Table 18-1).

FIGURE 24-2: SIMPLIFIED ZCD BLOCK DIAGRAM

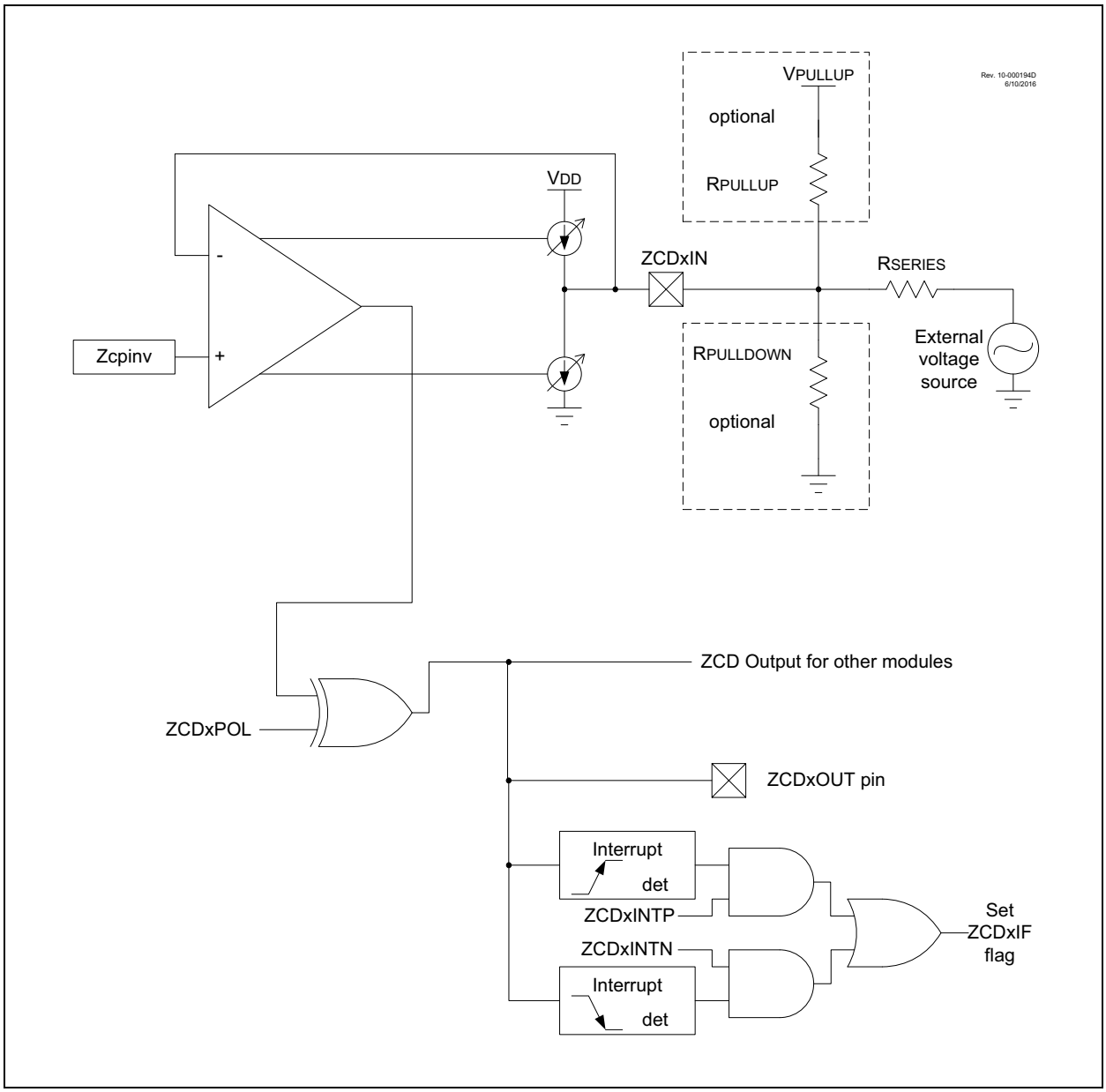
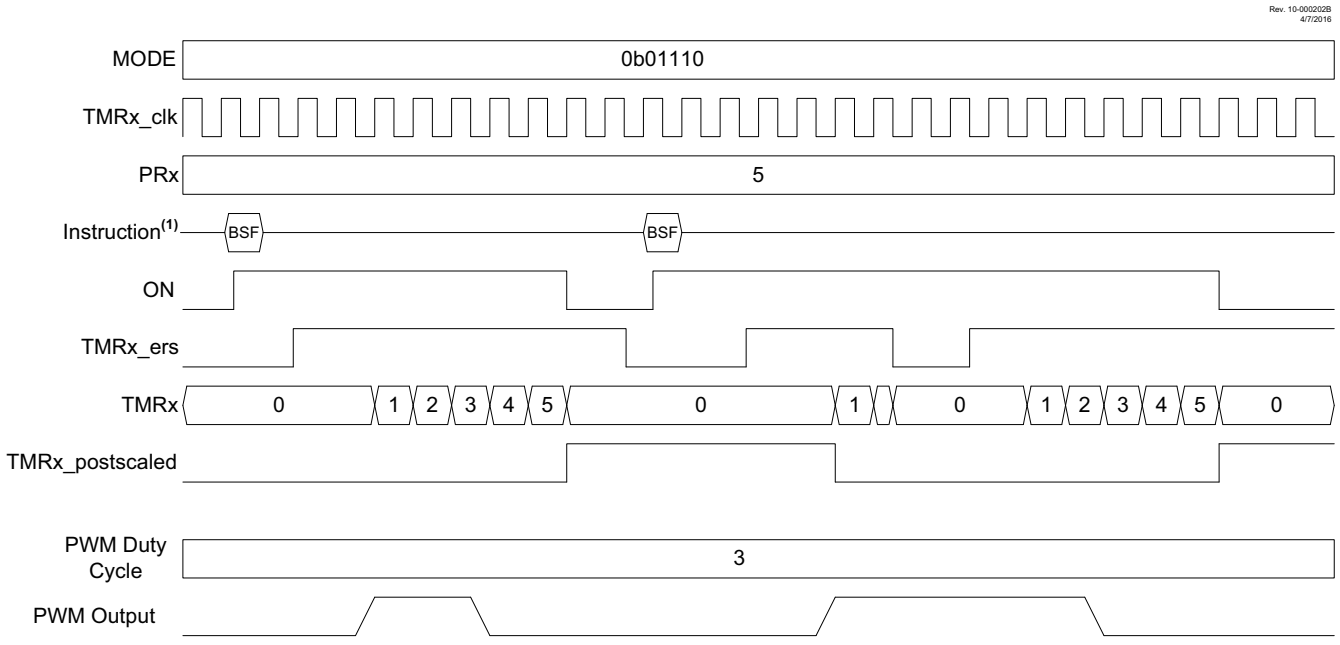
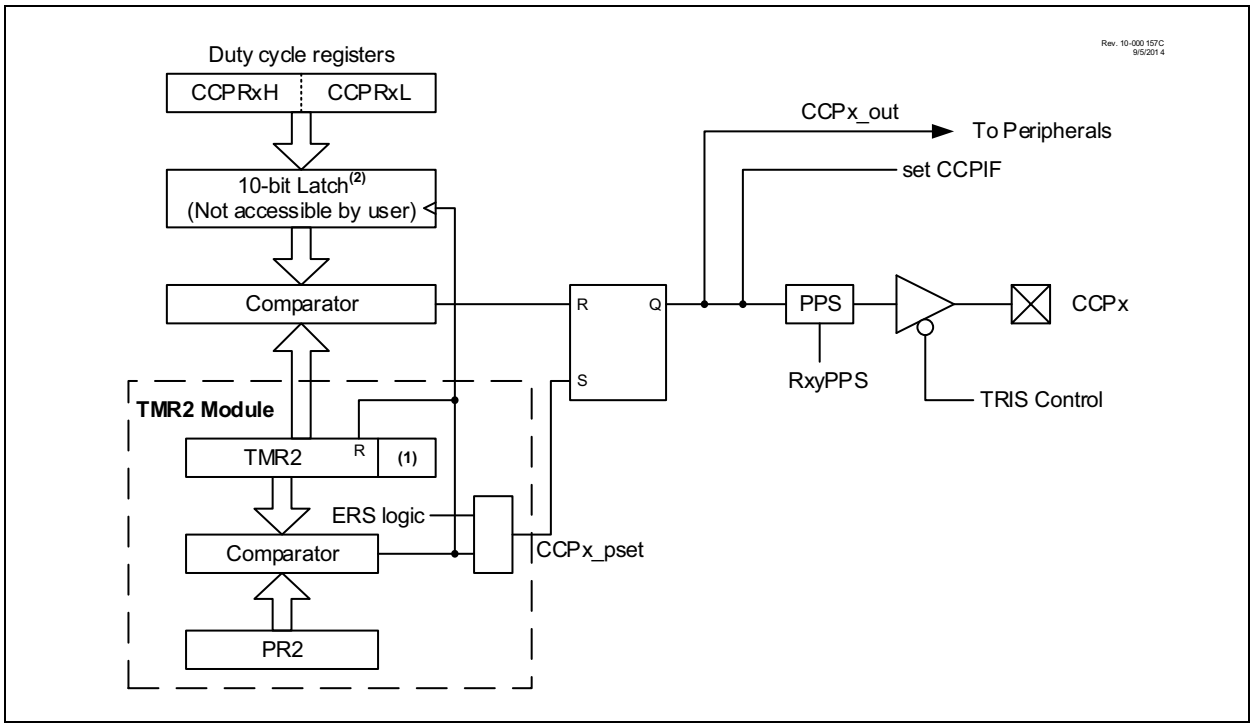


FIGURE 27-11: LOW LEVEL RESET, EDGE-TRIGGERED HARDWARE LIMIT ONE-SHOT MODE TIMING DIAGRAM (MODE = 01110)



Note 1: BSF and BCF represent Bit-Set File and Bit-Clear File instructions executed by the CPU to set or clear the ON bit of TxCON. CPU execution is asynchronous to the timer clock input.

FIGURE 28-4: SIMPLIFIED PWM BLOCK DIAGRAM



28.3.2 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for standard PWM operation:

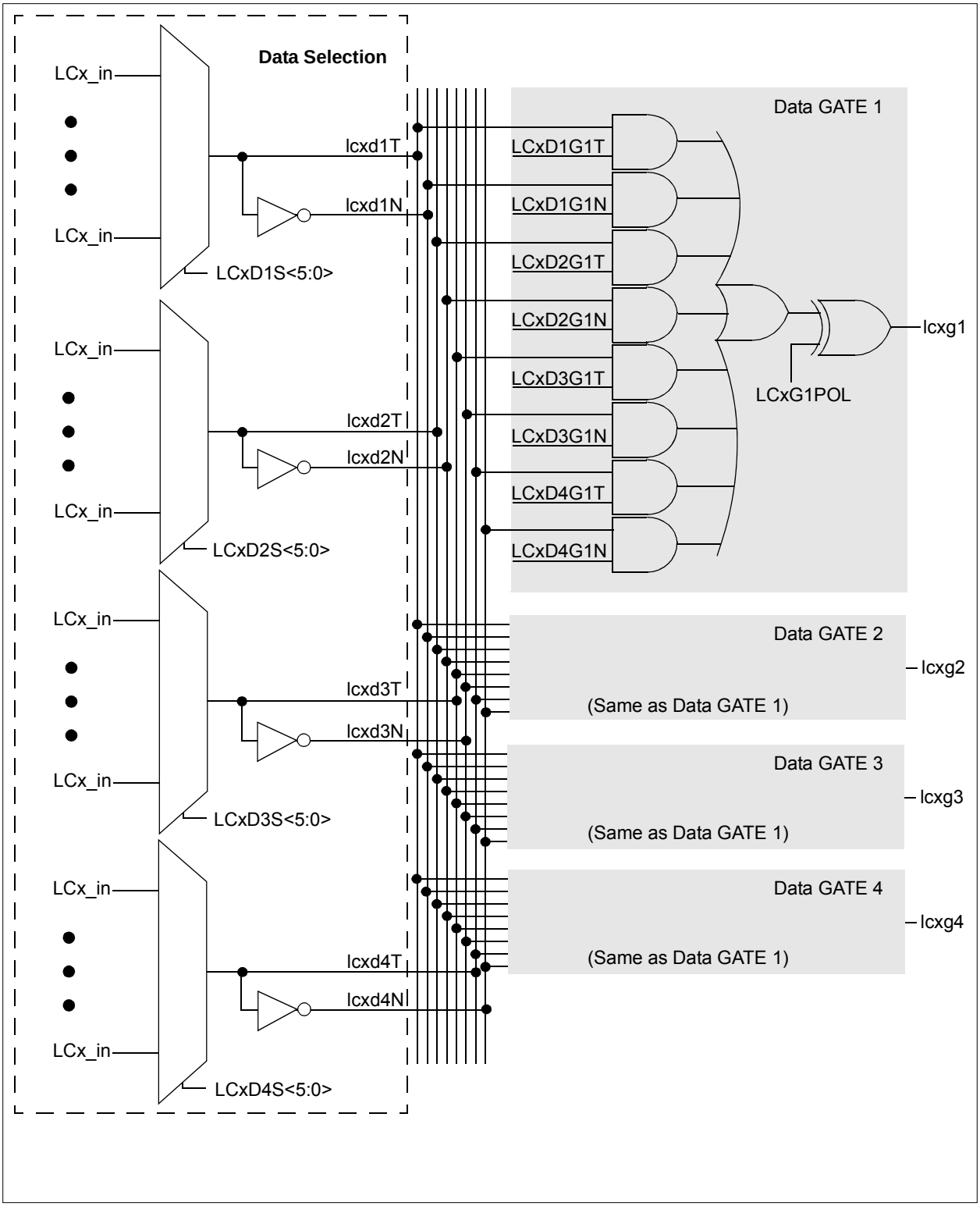
1. Use the desired output pin RxyPPS control to select CCPx as the source and disable the CCPx pin output driver by setting the associated TRIS bit.
2. Load the PR2 register with the PWM period value.
3. Configure the CCP module for the PWM mode by loading the CCPxCON register with the appropriate values.
4. Load the CCPRxL register, and the CCPRxH register with the PWM duty cycle value and configure the CCPxFMT bit of the CCPxCON register to set the proper register alignment.
5. Configure and start Timer2:
   - Clear the TMR2IF interrupt flag bit of the PIR4 register. See Note below.
   - Configure the CKPS bits of the T2CON register with the Timer prescale value.
   - Enable the Timer by setting the Timer2 ON bit of the T2CON register.
6. Enable PWM output pin:
   - Wait until the Timer overflows and the TMR2IF bit of the PIR4 register is set. See Note below.
   - Enable the CCPx pin output driver by clearing the associated TRIS bit.

Note:	In order to send a complete duty cycle and period on the first PWM output, the above steps must be included in the setup sequence. If it is not critical to start with a complete PWM signal on the first output, then step 6 may be ignored.

28.3.3 CCP/PWM CLOCK SELECTION

The PIC16(L)F15356/75/76/85/86 allows each individual CCP and PWM module to select the timer source that controls the module. Each module has an independent selection.

FIGURE 31-2: INPUT DATA SELECTION AND GATING



REGISTER 31-10: CLCxGLS3: GATE 3 LOGIC SELECT REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
LCxG4D4T	LCxG4D4N	LCxG4D3T	LCxG4D3N	LCxG4D2T	LCxG4D2N	LCxG4D1T	LCxG4D1N
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7 **LCxG4D4T:** Gate 3 Data 4 True (non-inverted) bit
1 = CLCIN3 (true) is gated into CLCx Gate 3
0 = CLCIN3 (true) is not gated into CLCx Gate 3

bit 6 **LCxG4D4N:** Gate 3 Data 4 Negated (inverted) bit
1 = CLCIN3 (inverted) is gated into CLCx Gate 3
0 = CLCIN3 (inverted) is not gated into CLCx Gate 3

bit 5 **LCxG4D3T:** Gate 3 Data 3 True (non-inverted) bit
1 = CLCIN2 (true) is gated into CLCx Gate 3
0 = CLCIN2 (true) is not gated into CLCx Gate 3

bit 4 **LCxG4D3N:** Gate 3 Data 3 Negated (inverted) bit
1 = CLCIN2 (inverted) is gated into CLCx Gate 3
0 = CLCIN2 (inverted) is not gated into CLCx Gate 3

bit 3 **LCxG4D2T:** Gate 3 Data 2 True (non-inverted) bit
1 = CLCIN1 (true) is gated into CLCx Gate 3
0 = CLCIN1 (true) is not gated into CLCx Gate 3

bit 2 **LCxG4D2N:** Gate 3 Data 2 Negated (inverted) bit
1 = CLCIN1 (inverted) is gated into CLCx Gate 3
0 = CLCIN1 (inverted) is not gated into CLCx Gate 3

bit 1 **LCxG4D1T:** Gate 4 Data 1 True (non-inverted) bit
1 = CLCIN0 (true) is gated into CLCx Gate 3
0 = CLCIN0 (true) is not gated into CLCx Gate 3

bit 0 **LCxG4D1N:** Gate 3 Data 1 Negated (inverted) bit
1 = CLCIN0 (inverted) is gated into CLCx Gate 3
0 = CLCIN0 (inverted) is not gated into CLCx Gate 3

33.1 EUSART Asynchronous Mode

The EUSART transmits and receives data using the standard non-return-to-zero (NRZ) format. NRZ is implemented with two levels: a VOH Mark state which represents a '1' data bit, and a VOL Space state which represents a '0' data bit. NRZ refers to the fact that consecutively transmitted data bits of the same value stay at the output level of that bit without returning to a neutral level between each bit transmission. An NRZ transmission port idles in the Mark state. Each character transmission consists of one Start bit followed by eight or nine data bits and is always terminated by one or more Stop bits. The Start bit is always a space and the Stop bits are always marks. The most common data format is eight bits. Each transmitted bit persists for a period of 1/(Baud Rate). An on-chip dedicated 8-bit/16-bit Baud Rate Generator is used to derive standard baud rate frequencies from the system oscillator. See Table 33-3 for examples of baud rate configurations.

The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent, but share the same data format and baud rate. Parity is not supported by the hardware, but can be implemented in software and stored as the ninth data bit.

33.1.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 33-1. The heart of the transmitter is the serial Transmit Shift Register (TSR), which is not directly accessible by software. The TSR obtains its data from the transmit buffer, which is the TXxREG register.

33.1.1.1 Enabling the Transmitter

The EUSART transmitter is enabled for asynchronous operations by configuring the following three control bits:

- TXEN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the TXEN bit of the TXxSTA register enables the transmitter circuitry of the EUSART. Clearing the SYNC bit of the TXxSTA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RCxSTA register enables the EUSART and automatically configures the TX/CK I/O pin as an output. If the TX/CK pin is shared with an analog peripheral, the analog I/O function must be disabled by clearing the corresponding ANSEL bit.

Note:	The TXxIF Transmitter Interrupt flag is set when the TXEN enable bit is set.

33.1.1.2 Transmitting Data

A transmission is initiated by writing a character to the TXxREG register. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TXxREG is immediately transferred to the TSR register. If the TSR still contains all or part of a previous character, the new character data is held in the TXxREG until the Stop bit of the previous character has been transmitted. The pending character in the TXxREG is then transferred to the TSR in one TCY immediately following the Stop bit transmission. The transmission of the Start bit, data bits and Stop bit sequence commences immediately following the transfer of the data to the TSR from the TXxREG.

33.1.1.3 Transmit Data Polarity

The polarity of the transmit data can be controlled with the SCKP bit of the BAUDxCON register. The default state of this bit is '0' which selects high true transmit idle and data bits. Setting the SCKP bit to '1' will invert the transmit data resulting in low true idle and data bits. The SCKP bit controls transmit data polarity in Asynchronous mode only. In Synchronous mode, the SCKP bit has a different function. See **Section 33.4.1.2 "Clock Polarity"**.

33.1.1.4 Transmit Interrupt Flag

The TXxIF interrupt flag bit of the PIR3 register is set whenever the EUSART transmitter is enabled and no character is being held for transmission in the TXxREG. In other words, the TXxIF bit is only clear when the TSR is busy with a character and a new character has been queued for transmission in the TXxREG. The TXxIF flag bit is not cleared immediately upon writing TXxREG. TXxIF becomes valid in the second instruction cycle following the write execution. Polling TXxIF immediately following the TXxREG write will return invalid results. The TXxIF bit is read-only, it cannot be set or cleared by software.

The TXxIF interrupt can be enabled by setting the TXxIE interrupt enable bit of the PIE3 register. However, the TXxIF flag bit will be set whenever the TXxREG is empty, regardless of the state of TXxIE enable bit.

To use interrupts when transmitting data, set the TXxIE bit only when there is more data to send. Clear the TXxIE interrupt enable bit upon writing the last character of the transmission to the TXxREG.

33.4.2 SYNCHRONOUS SLAVE MODE

The following bits are used to configure the EUSART for synchronous slave operation:

- SYNC = 1
- CSRC = 0
- SREN = 0 (for transmit); SREN = 1 (for receive)
- CREN = 0 (for transmit); CREN = 1 (for receive)
- SPEN = 1

Setting the SYNC bit of the TXxSTA register configures the device for synchronous operation. Clearing the CSRC bit of the TXxSTA register configures the device as a slave. Clearing the SREN and CREN bits of the RCxSTA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RCxSTA register enables the EUSART.

33.4.2.1 EUSART Synchronous Slave Transmit

The operation of the Synchronous Master and Slave modes are identical (see **Section 33.4.1.3 "Synchronous Master Transmission")**, except in the case of the Sleep mode.

If two words are written to the TXxREG and then the SLEEP instruction is executed, the following will occur:

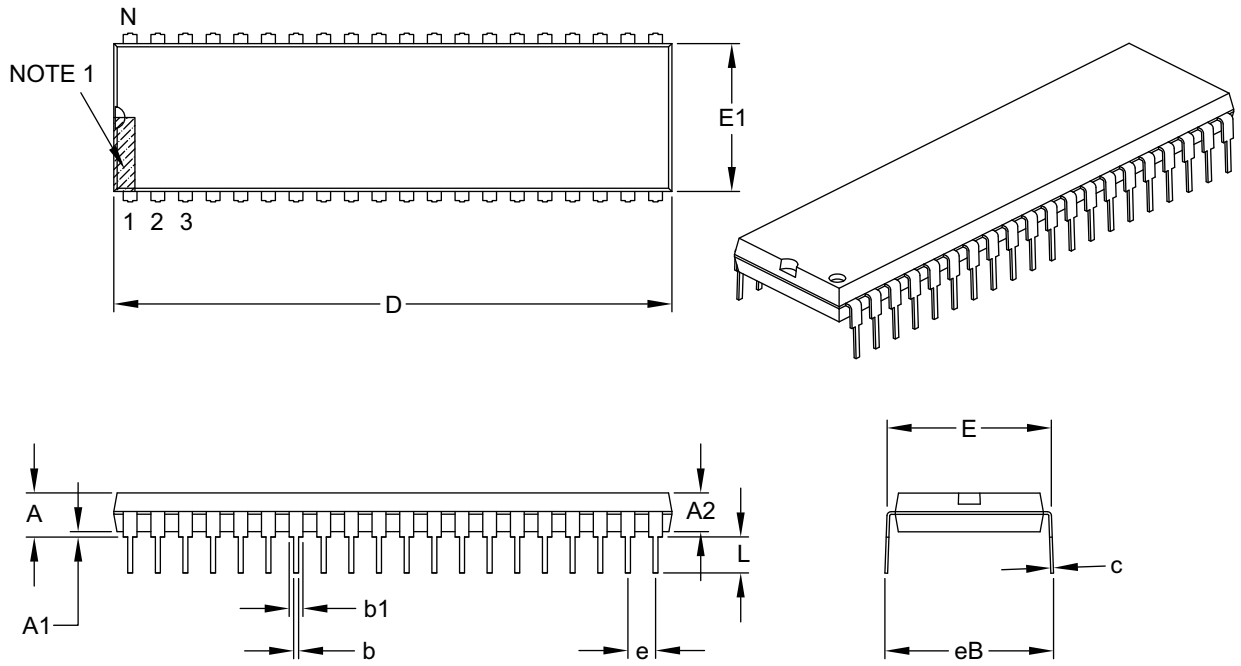
1. The first character will immediately transfer to the TSR register and transmit.
2. The second word will remain in the TXxREG register.
3. The TXxIF bit will not be set.
4. After the first character has been shifted out of TSR, the TXxREG register will transfer the second character to the TSR and the TXxIF bit will now be set.
5. If the PEIE and TXxIE bits are set, the interrupt will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will call the Interrupt Service Routine.

33.4.2.2 Synchronous Slave Transmission Set-up:

1. Set the SYNC and SPEN bits and clear the CSRC bit.
2. Clear the ANSEL bit for the CK pin (if applicable).
3. Clear the CREN and SREN bits.
4. If interrupts are desired, set the TXxIE bit of the PIE3 register and the GIE and PEIE bits of the INTCON register.
5. If 9-bit transmission is desired, set the TX9 bit.
6. Enable transmission by setting the TXEN bit.
7. If 9-bit transmission is selected, insert the Most Significant bit into the TX9D bit.
8. Start transmission by writing the Least Significant eight bits to the TXxREG register.

40-Lead Plastic Dual In-Line (P) – 600 mil Body [PDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		INCHES		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		40	
Pitch	e		.100 BSC	
Top to Seating Plane	A	–	–	.250
Molded Package Thickness	A2	.125	–	.195
Base to Seating Plane	A1	.015	–	–
Shoulder to Shoulder Width	E	.590	–	.625
Molded Package Width	E1	.485	–	.580
Overall Length	D	1.980	–	2.095
Tip to Seating Plane	L	.115	–	.200
Lead Thickness	c	.008	–	.015
Upper Lead Width	b1	.030	–	.070
Lower Lead Width	b	.014	–	.023
Overall Row Spacing §	eB	–	–	.700

Notes:

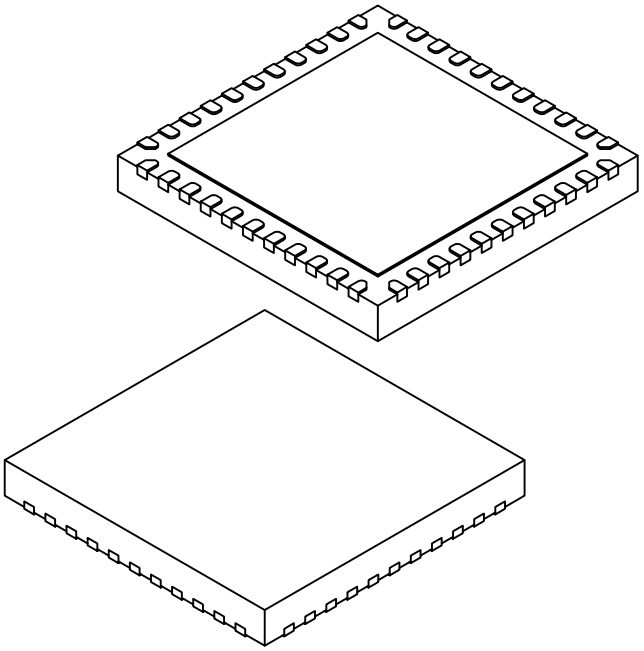
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. § Significant Characteristic.
3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-016B

44-Lead Plastic Quad Flat, No Lead Package (ML) - 8x8 mm Body [QFN or VQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		44	
Pitch	e		0.65 BSC	
Overall Height	A	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Terminal Thickness	A3		0.20 REF	
Overall Width	E		8.00 BSC	
Exposed Pad Width	E2	6.25	6.45	6.60
Overall Length	D		8.00 BSC	
Exposed Pad Length	D2	6.25	6.45	6.60
Terminal Width	b	0.20	0.30	0.35
Terminal Length	L	0.30	0.40	0.50
Terminal-to-Exposed-Pad	K	0.20	-	-

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Package is saw singulated
3. Dimensioning and tolerancing per ASME Y14.5M

   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

   REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-103D Sheet 2 of 2