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#### Details

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Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I <sup>2</sup> C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	35
Program Memory Size	14KB (8K x 14)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 28x10b; D/A 1x5b, 1x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	40-UFQFN Exposed Pad
Supplier Device Package	40-UQFN (5x5)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1717-e-mv

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# PIC16(L)F1717/8/9





# 5.3 Register Definitions: BOR Control

#### REGISTER 5-1: BORCON: BROWN-OUT RESET CONTROL REGISTER

R/W-1/u	R/W-0/u	U-0	U-0	U-0	U-0	U-0	R-q/u
SBOREN	BORFS <sup>(1)</sup>	—	—	—	—	—	BORRDY
bit 7					-		bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7	SBOREN: Software Brown-out Reset Enable bit         If BOREN <1:0> in Configuration Words ≠ 01:         SBOREN is read/write, but has no effect on the BOR.         If BOREN <1:0> in Configuration Words = 01:         1 = BOR Enabled         0 = BOR Disabled
bit 6	BORFS: Brown-out Reset Fast Start bit <sup>(1)</sup> If BOREN<1:0> = 11 (Always on) or BOREN<1:0> = 00 (Always off) BORFS is Read/Write, but has no effect. If BOREN <1:0> = 10 (Disabled in Sleep) or BOREN<1:0> = 01 (Under software control): 1 = Band gap is forced on always (covers sleep/wake-up/operating cases) 0 = Band gap operates normally, and may turn off
bit 5-1	Unimplemented: Read as '0'
bit 0	<b>BORRDY:</b> Brown-out Reset Circuit Ready Status bit 1 = The Brown-out Reset circuit is active 0 = The Brown-out Reset circuit is inactive

**Note 1:** BOREN<1:0> bits are located in Configuration Words.

![](_page_2_Figure_1.jpeg)

# **10.6 Register Definitions: Flash Program Memory Control**

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			PMDA	AT<7:0>			
bit 7							bit 0
Legend:							
R = Readable I	bit	W = Writable bi	it	U = Unimpler	mented bit, read	l as '0'	
u = Bit is uncha	anged	x = Bit is unkno	wn	-n/n = Value	at POR and BO	R/Value at all c	other Resets
'1' = Bit is set		'0' = Bit is clear	ed				

#### REGISTER 10-1: PMDATL: PROGRAM MEMORY DATA LOW BYTE REGISTER

bit 7-0 PMDAT<7:0>: Read/write value for Least Significant bits of program memory

## REGISTER 10-2: PMDATH: PROGRAM MEMORY DATA HIGH BYTE REGISTER

U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
—	—			PMDA	T<13:8>		
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6 Unimplemented: Read as '0'

bit 5-0 **PMDAT<13:8>**: Read/write value for Most Significant bits of program memory

#### REGISTER 10-3: PMADRL: PROGRAM MEMORY ADDRESS LOW BYTE REGISTER

| R/W-0/0 |
|---------|---------|---------|---------|---------|---------|---------|---------|
|         |         |         | PMAD    | R<7:0>  |         |         |         |
| bit 7   |         |         |         |         |         |         | bit 0   |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **PMADR<7:0>**: Specifies the Least Significant bits for program memory address

# 11.7 PORTD Registers (PIC16(L)F1717/9 only)

## 11.7.1 DATA REGISTER

PORTD is an 8-bit wide bidirectional port. The corresponding data direction register is TRISD (Register 11-26). Setting a TRISD bit (= 1) will make the corresponding PORTD pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISD bit (= 0) will make the corresponding PORTD pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 11-1 shows how to initialize an I/O port.

Reading the PORTD register (Register 11-25) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATD).

## 11.7.2 DIRECTION CONTROL

The TRISD register (Register 11-26) controls the PORTD pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISD register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read '0'.

## 11.7.3 INPUT THRESHOLD CONTROL

The INLVLD register (Register 11-32) controls the input voltage threshold for each of the available PORTD input pins. A selection between the Schmitt Trigger CMOS or the TTL compatible thresholds is available. The input threshold is important in determining the value of a read of the PORTD register and also the level at which an interrupt-on-change occurs, if that feature is enabled. See Table 34-4: I/O Ports for more information on threshold levels.

**Note:** Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

## 11.7.4 OPEN-DRAIN CONTROL

The ODCOND register (Register 11-30) controls the open-drain feature of the port. Open-drain operation is independently selected for each pin. When an ODCOND bit is set, the corresponding port output becomes an open-drain driver capable of sinking current only. When an ODCOND bit is cleared, the corresponding port output pin is the standard push-pull drive capable of sourcing and sinking current.

## 11.7.5 SLEW RATE CONTROL

The SLRCOND register (Register 11-31) controls the slew rate option for each port pin. Slew rate control is independently selectable for each port pin. When an SLRCOND bit is set, the corresponding port pin drive is slew rate limited. When an SLRCOND bit is cleared, The corresponding port pin drive slews at the maximum rate possible.

## 11.7.6 ANALOG CONTROL

The ANSELD register (Register 11-28) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELD bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELD bits has no effect on digital output functions. A pin with TRIS clear and ANSELD set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note: The ANSELD bits default to the Analog mode after Reset. To use any pins as digital general purpose or peripheral inputs, the corresponding ANSEL bits must be initialized to '0' by user software.

#### 11.7.7 PORTD FUNCTIONS AND OUTPUT PRIORITIES

Each pin defaults to the PORT latch data after Reset. Other functions are selected with the peripheral pin select logic. See **Section 12.0** "**Peripheral Pin Select (PPS) Module**" for more information.

Analog input functions, such as ADC and comparator inputs, are not shown in the peripheral pin select lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELD register. Digital output functions may continue to control the pin when it is in Analog mode.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on page		
RB0PPS	—	—	—			RB0PPS<	4:0>		153		
RB1PPS	—	—	_		RB1PPS<4:0>						
RB2PPS	—	—				RB2PPS<	4:0>		153		
RB3PPS	_	_				RB3PPS<	4:0>		153		
RB4PPS	_	_				RB4PPS<	4:0>		153		
RB5PPS	_	_				RB5PPS<	4:0>		153		
RB6PPS	—	—				RB6PPS<	4:0>		153		
RB7PPS	—	—				RB7PPS<	4:0>		153		
RC0PPS	—	—				RC0PPS<	4:0>		153		
RC1PPS	—	—	_			RC1PPS<	4:0>		153		
RC2PPS	—	—				RC2PPS<	4:0>		153		
RC3PPS	—	—			RC3PPS<4:0>				153		
RC4PPS	_	_				RC4PPS<	4:0>		153		
RC5PPS	_	_				RC5PPS<	4:0>		153		
RC6PPS	_	_				RC6PPS<	4:0>		153		
RC7PPS	_	_				RC7PPS<	4:0>		153		
RD0PPS <sup>(1)</sup>	_	_				RD0PPS<	:4:0>		153		
RD1PPS <sup>(1)</sup>	_	_				RD1PPS<	:4:0>		153		
RD2PPS <sup>(1)</sup>	_	_				RD2PPS<	:4:0>		153		
RD3PPS <sup>(1)</sup>	_	_				RD3PPS<	:4:0>		153		
RD4PPS <sup>(1)</sup>	—	—	_			RD4PPS<	:4:0>		153		
RD5PPS <sup>(1)</sup>	—	—	_			RD5PPS<	:4:0>		153		
RD6PPS <sup>(1)</sup>	—	—	_			RD6PPS<	:4:0>		153		
RD7PPS <sup>(1)</sup>	—	—	—			RD7PPS<	4:0>		153		
RE0PPS <sup>(1)</sup>	—	_	—			RE0PPS<	4:0>		153		
RE1PPS <sup>(1)</sup>	—	—	—			RE1PPS<	4:0>		153		
RE2PPS <sup>(1)</sup>	_	_	_			RE2PPS<	4:0>	RE2PPS<4:0>			

TADIE 12 2.		ASSOCIATED WITH	THE DDS MODILLE	
IADLE IZ-J.	SUMMART OF REGISTERS	ASSOCIATED WITH		

Legend: — = unimplemented, read as '0'. Shaded cells are unused by the PPS module.

Note 1: PIC16(L)F1717/9 only.

## FIGURE 18-7: COG (RISING/FALLING) INPUT BLOCK

![](_page_6_Figure_1.jpeg)

# 18.2 Clock Sources

The COG\_clock is used as the reference clock to the various timers in the peripheral. Timers that use the COG\_clock include:

- · Rising and falling dead-band time
- Rising and falling blanking time
- · Rising and falling event phase delay

Clock sources available for selection include:

- 8 MHz HFINTOSC (active during Sleep)
- Instruction clock (Fosc/4)
- System clock (Fosc)

The clock source is selected with the GxCS<1:0> bits of the COGxCON0 register (Register 18-1).

#### 18.3 Selectable Event Sources

The COG uses any combination of independently selectable event sources to generate the complementary waveform. Sources fall into two categories:

- · Rising event sources
- · Falling event sources

The rising event sources are selected by setting bits in the COGxRIS register (Register 18-3). The falling event sources are selected by setting bits in the COGxFIS register (Register 18-5). All selected sources are 'OR'd together to generate the corresponding event signal. Refer to Figure 18-7.

## 18.3.1 EDGE VS. LEVEL SENSING

Event input detection may be selected as level or edge sensitive. The detection mode is individually selectable for every source. Rising source detection modes are selected with the COGxRSIM register (Register 18-4). Falling source detection modes are selected with the COGxFSIM register (Register 18-6). A set bit enables edge detection for the corresponding event source. A cleared bit enables level detection.

In general, events that are driven from a periodic source should be edge detected and events that are derived from voltage thresholds at the target circuit should be level sensitive. Consider the following two examples:

1. The first example is an application in which the period is determined by a 50% duty cycle clock and the COG output duty cycle is determined by a voltage level fed back through a comparator. If the clock input is level sensitive, duty cycles less than 50% will exhibit erratic operation.

2. The second example is similar to the first except that the duty cycle is close to 100%. The feedback comparator high-to-low transition trips the COG drive off, but almost immediately the period source turns the drive back on. If the off cycle is short enough, the comparator input may not reach the low side of the hysteresis band precluding an output change. The comparator output stays low and without a high-to-low transition to trigger the edge sense, the drive of the COG output will be stuck in a constant drive-on condition. See Figure 18-14.

## FIGURE 18-14: EDGE VS LEVEL SENSE

Rising (CCP1)
Falling (C1OUT)
C1IN- hyst I
COGOUT
Edge Sensitive
Rising (CCP1)
Falling (C1OUT)
C1IN- hyst
COGOUT
Level Sensitive

#### 18.3.2 RISING EVENT

The rising event starts the PWM output active duty cycle period. The rising event is the low-to-high transition of the rising\_event output. When the rising event phase delay and dead-band time values are zero, the primary output starts immediately. Otherwise, the primary output is delayed. The rising event source causes all the following actions:

- · Start rising event phase delay counter (if enabled).
- · Clear complementary output after phase delay.
- Start falling event input blanking (if enabled).
- · Start dead-band delay (if enabled).
- · Set primary output after dead-band delay expires.

#### 18.3.3 FALLING EVENT

The falling event terminates the PWM output active duty cycle period. The falling event is the high-to-low transition of the falling\_event output. When the falling event phase delay and dead-band time values are zero, the complementary output starts immediately. Otherwise, the complementary output is delayed. The falling event source causes all the following actions:

- Start falling event phase delay counter (if enabled).
- · Clear primary output.
- · Start rising event input blanking (if enabled).
- · Start falling event dead-band delay (if enabled).
- Set complementary output after dead-band delay expires.

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
GxRIS7	GxRIS6	GxRIS5	GxRIS4	GxRIS3	GxRIS2	GxRIS1	GxRIS0
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'	
u = Bit is unch	anged	x = Bit is unki	nown	-n/n = Value	at POR and BO	R/Value at all o	other Resets
'1' = Bit is set		'0' = Bit is cle	ared	q = Value de	pends on condit	ion	
bit 7	GxRIS7: CO	Gx Rising Ever	nt Input Source	e 7 Enable bit			
	1 = NCO1_0	ut is enabled a	s a rising eve	nt input			
	$0 = NCO1_0$	ut has no effec	t on the rising	event			
bit 6	GxRIS6: CO	Gx Rising Ever	t Input Source	e 6 Enable bit			
	1 = PWM30 0 = PWM3h	as no effect on	d as a rising e				
bit 5	GxRIS5: CO	Gx Rising Ever	t Input Source	e 5 Enable bit			
Sit 0	1 = CCP2 or	itout is enabled	l as a rising ev	vent input			
	0 = CCP2 out	tput has no eff	ect on the risi	ng event			
bit 4	GxRIS4: CO	Gx Rising Ever	t Input Source	e 4 Enable bit			
	1 = CCP1 is	enabled as a r	ising event inp	out			
	0 = CCP1 ha	as no effect on	the rising eve	nt			
bit 3	GxRIS3: CO	Gx Rising Ever	nt Input Source	e 3 Enable bit			
	1 = CLC1 ou	tput is enabled	l as a rising ev	vent input			
hit 0		Cy Dising Ever		ng eveni a 2 Enabla bit			
DIL Z		GX RISING EVER		e 2 Enable bit	out		
	0 = Compara	ator 2 output ha	as no effect or	the rising event in	nt		
bit 1	GxRIS1: CO	Gx Risina Ever	t Input Source	e 1 Enable bit			
	1 = Compara	ator 1 output is	enabled as a	rising event in	out		
	0 = Compara	ator 1 output ha	as no effect or	the rising eve	nt		
bit 0	GxRISO: CO	Gx Rising Ever	t Input Source	e 0 Enable bit			
	1 = Pin selec	ted with COG	PPS control r	egister is enab	oled as rising ev	ent input	
	0 = Pin selec	cted with COG	PPS control h	has no effect of	n the rising ever	nt	

## REGISTER 18-3: COGxRIS: COG RISING EVENT INPUT SELECTION REGISTER

# 19.6 Register Definitions: CLC Control

R/W-0/0	U-0	R-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
LCxEN	_	LCxOUT	LCxINTP	LCxINTN	L	CxMODE<2:0>	
bit 7							bit 0
Legend:							
R = Readable b	bit	W = Writable	bit	U = Unimplen	nented bit, read	l as '0'	
u = Bit is uncha	anged	x = Bit is unkr	nown	-n/n = Value a	at POR and BO	R/Value at all o	ther Resets
'1' = Bit is set		'0' = Bit is clea	ared				
bit 7	LCxEN: Confi	igurable Logic	Cell Enable b	it			
	1 = Configura	able logic cell is	s enabled and	mixing input si	ignals		
	0 = Configura	able logic cell is	s disabled and	has logic zero	output		
bit 6	Unimplement	ted: Read as '	0'				
bit 5	LCxOUT: Cor	nfigurable Logio	c Cell Data Ou	utput bit			
	Read-only: log	gic cell output o	data, after LC	<pre>kPOL; sampled</pre>	from Icx_out w	/ire.	
bit 4	LCxINTP: Co	nfigurable Logi	ic Cell Positive	e Edge Going I	nterrupt Enable	e bit	
	1 = CLCxIF w	vill be set wher	n a rising edge	e occurs on lcx	_out		
	0 = CLCxIF w	vill not be set					
bit 3	LCxINTN: Co	nfigurable Log	ic Cell Negativ	ve Edge Going	Interrupt Enabl	e bit	
	1 = CLCxIF w	vill be set wher	n a falling edg	e occurs on lcx	_out		
<b>h</b> # 0.0					-l		
DIT 2-0		:U>: Configura			de dits		
	111 = Cell is	1-input transpa	arent latch wit	n S and R			
	101 = Cell is	2-input D flip-fl	lop with R				
	100 = Cell is	1-input D flip-fl	op with S and	IR			
	011 = Cell is	S-R latch	•				
	010 = Cell is	4-input AND					
	001 = Cell is	OR-XOR					
		AND-UR					

#### REGISTER 19-1: CLCxCON: CONFIGURABLE LOGIC CELL CONTROL REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
LCxG4D4T	LCxG4D4N	LCxG4D3T	LCxG4D3N	LCxG4D2T	LCxG4D2N	LCxG4D1T	LCxG4D1N
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'	
u = Bit is unch	anged	x = Bit is unkr	nown	-n/n = Value a	at POR and BO	R/Value at all c	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				
bit 7	LCxG4D4T: 0	Gate 4 Data 4 1	rue (non-inve	rted) bit			
	1 = LCxD4T	is gated into LO	CxG4				
	0 = LCxD4T	is not gated int	o LCxG4				
bit 6	LCxG4D4N:	Gate 4 Data 4	Negated (inve	rted) bit			
	1 = LCXD4N 0 = LCYD4N	is pot gated into L					
bit 5		Sate 4 Data 3 1	rue (non-inve	rted) hit			
bit 0	1 = 1  CxD3T	is gated into I (	CxG4				
	0 = LCxD3T	is not gated int	o LCxG4				
bit 4	LCxG4D3N:	Gate 4 Data 3 I	Negated (inve	rted) bit			
	1 = LCxD3N	is gated into L	CxG4				
	0 = LCxD3N	is not gated inf	to LCxG4				
bit 3	LCxG4D2T: (	Gate 4 Data 2 1	True (non-inve	rted) bit			
	1 = LCxD2T	is gated into L0	CXG4				
bit 2		Gate 4 Data 2 I	Vegated (inve	rted) bit			
Dit Z	1 = 1  CxD2N	is dated into L	CxG4	ned) bit			
	0 = LCxD2N	is not gated int	to LCxG4				
bit 1	LCxG4D1T: (	Gate 4 Data 1 1	rue (non-inve	rted) bit			
	1 = LCxD1T	is gated into L0	CxG4				
	0 = LCxD1T	is not gated int	o LCxG4				
bit 0	LCxG4D1N:	Gate 4 Data 1 I	Negated (inve	rted) bit			
	1 = LCxD1N	is gated into L	CxG4				
	0 = LCxD1N	is not gated in	OLCXG4				

## REGISTER 19-10: CLCxGLS3: GATE 4 LOGIC SELECT REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
			NCOXA	CC<15:8>			
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable b	it	U = Unimplen	nented bit, read	l as '0'	
u = Bit is unchanged x = Bit is unknown		own	-n/n = Value a	t POR and BO	R/Value at all o	other Resets	
'1' = Bit is set		'0' = Bit is clea	red				

#### **REGISTER 20-4:** NCOXACCH: NCOX ACCUMULATOR REGISTER – HIGH BYTE

bit 7-0 NCOxACC<15:8>: NCOx Accumulator, High Byte

#### REGISTER 20-5: NCOxACCU: NCOx ACCUMULATOR REGISTER – UPPER BYTE

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
						0.40.40	
	—				NCOXAC	3<19:16>	
bit 7 bi					bit 0		
Legend:							
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'							
u = Bit is unch	Bit is unchanged x = Bit is unknown		-n/n = Value at POR and BOR/Value at all other Rese			other Resets	

bit 7-4	Unimplemented: Read as '0	ı'
---------	---------------------------	----

'1' = Bit is set

bit 3-0 NCOxACC<19:16>: NCOx Accumulator, Upper Byte

'0' = Bit is cleared

# **REGISTER 20-6:** NCOxINCL: NCOx INCREMENT REGISTER – LOW BYTE<sup>(1)</sup>

R/W-0/0	R/W-1/1						
			NCOxIN	C<7:0>			
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

#### bit 7-0 NCOxINC<7:0>: NCOx Increment, Low Byte

Note 1: Write the NCOxINCH register first, then the NCOxINCL register. See Section 20.1.4 "Increment Registers" for more information.

# 22.1 OPA Module Performance

Common AC and DC performance specifications for the OPA module:

- Common Mode Voltage Range
- Leakage Current
- Input Offset Voltage
- Open Loop Gain
- Gain Bandwidth Product

**Common mode voltage range** is the specified voltage range for the OPA+ and OPA- inputs, for which the OPA module will perform to within its specifications. The OPA module is designed to operate with input voltages between Vss and VDD. Behavior for Common mode voltages greater than VDD, or below Vss, are not guaranteed.

**Leakage current** is a measure of the small source or sink currents on the OPA+ and OPA- inputs. To minimize the effect of leakage currents, the effective impedances connected to the OPA+ and OPA- inputs should be kept as small as possible and equal.

**Input offset voltage** is a measure of the voltage difference between the OPA+ and OPA- inputs in a closed loop circuit with the OPA in its linear region. The offset voltage will appear as a DC offset in the output equal to the input offset voltage, multiplied by the gain of the circuit. The input offset voltage is also affected by the Common mode voltage. The OPA is factory calibrated to minimize the input offset voltage of the module.

**Open loop gain** is the ratio of the output voltage to the differential input voltage, (OPA+) - (OPA-). The gain is greatest at DC and falls off with frequency.

**Gain Bandwidth Product** or GBWP is the frequency at which the open loop gain falls off to 0 dB.

## 22.1.1 **OPA Module Control**

The OPA module is enabled by setting the OPAxEN bit of the OPAxCON register. When enabled, the OPA forces the output driver of OPAxOUT pin into tri-state to prevent contention between the driver and the OPA output.

Note: When the OPA module is enabled, the OPAxOUT pin is driven by the op amp output, not by the PORT digital driver. Refer to Table 34-17: Operational Amplifier (OPA) for the op amp output drive capability.

## 22.1.2 UNITY GAIN MODE

The OPAxUG bit of the OPAxCON register selects the Unity Gain mode. When unity gain is selected, the OPA output is connected to the inverting input and the OPAxIN pin is relinquished, releasing the pin for general purpose input and output.

## 22.2 Effects of Reset

A device Reset forces all registers to their Reset state. This disables the OPA module.

![](_page_13_Figure_1.jpeg)

![](_page_13_Figure_2.jpeg)

#### 30.2.1 SPI MODE REGISTERS

The MSSP module has five registers for SPI mode operation. These are:

- MSSP STATUS register (SSP1STAT)
- MSSP Control register 1 (SSP1CON1)
- MSSP Control register 3 (SSP1CON3)
- MSSP Data Buffer register (SSP1BUF)
- MSSP Address register (SSP1ADD)
- MSSP Shift register (SSP1SR)
- (Not directly accessible)

SSP1CON1 and SSP1STAT are the control and STA-TUS registers in SPI mode operation. The SSP1CON1 register is readable and writable. The lower six bits of the SSP1STAT are read-only. The upper two bits of the SSP1STAT are read/write.

In one SPI master mode, SSP1ADD can be loaded with a value used in the Baud Rate Generator. More information on the Baud Rate Generator is available in **Section 30.7 "Baud Rate Generator"**.

SSP1SR is the shift register used for shifting data in and out. SSP1BUF provides indirect access to the SSP1SR register. SSP1BUF is the buffer register to which data bytes are written, and from which data bytes are read.

In receive operations, SSP1SR and SSP1BUF together create a buffered receiver. When SSP1SR receives a complete byte, it is transferred to SSP1BUF and the SSP1IF interrupt is set.

During transmission, the SSP1BUF is not buffered. A write to SSP1BUF will write to both SSP1BUF and SSP1SR.

## 30.2.2 SPI MODE OPERATION

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits (SSP1CON1<5:0> and SSP1STAT<7:6>). These control bits allow the following to be specified:

- Master mode (SCK is the clock output)
- Slave mode (SCK is the clock input)
- Clock Polarity (Idle state of SCK)
- Data Input Sample Phase (middle or end of data output time)
- Clock Edge (output data on rising/falling edge of SCK)
- Clock Rate (Master mode only)
- · Slave Select mode (Slave mode only)

To enable the serial port, SSP Enable bit, SSPEN of the SSP1CON1 register, must be set. To reset or reconfigure SPI mode, clear the SSPEN bit, re-initialize the SSP1CONx registers and then set the <u>SSPEN</u> bit. This configures the SDI, SDO, SCK and <u>SS</u> pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the TRIS register) appropriately programmed as follows:

- · SDI must have corresponding TRIS bit set
- SDO must have corresponding TRIS bit cleared
  SCK (Master mode) must have corresponding
- TRIS bit cleared
- SCK (Slave mode) must have corresponding <u>TRIS</u> bit set
- SS must have corresponding TRIS bit set

Any serial port function that is not desired may be overridden by programming the corresponding data direction (TRIS) register to the opposite value.

## 30.4.9 ACKNOWLEDGE SEQUENCE

The 9th SCL pulse for any transferred byte in  $I^2C$  is dedicated as an Acknowledge. It allows receiving devices to respond back to the transmitter by pulling the SDA line low. The transmitter must release control of the line during this time to shift in the response. The Acknowledge (ACK) is an active-low signal, pulling the SDA line low indicates to the transmitter that the device has received the transmitted data and is ready to receive more.

The result of an  $\overline{ACK}$  is placed in the ACKSTAT bit of the SSP1CON2 register.

Slave software, when the AHEN and DHEN bits are set, allow the user to set the ACK value sent back to the transmitter. The ACKDT bit of the SSP1CON2 register is set/cleared to determine the response.

Slave hardware will generate an ACK response if the AHEN and DHEN bits of the SSP1CON3 register are clear.

There are certain conditions where an  $\overline{ACK}$  will not be sent by the slave. If the BF bit of the SSP1STAT register or the SSPOV bit of the SSP1CON1 register are set when a byte is received.

When the module is addressed, after the eighth falling edge of SCL on the bus, the ACKTIM bit of the SSP1CON3 register is set. The ACKTIM bit indicates the acknowledge time of the active bus. The ACKTIM Status bit is only active when the AHEN bit or DHEN bit is enabled.

## 30.5 I<sup>2</sup>C SLAVE MODE OPERATION

The MSSP Slave mode operates in one of four modes selected by the SSPM bits of SSP1CON1 register. The modes can be divided into 7-bit and 10-bit Addressing mode. 10-bit Addressing modes operate the same as 7-bit with some additional overhead for handling the larger addresses.

Modes with Start and Stop bit interrupts operate the same as the other modes with SSP1IF additionally getting set upon detection of a Start, Restart, or Stop condition.

#### 30.5.1 SLAVE MODE ADDRESSES

The SSP1ADD register (Register 30-6) contains the Slave mode address. The first byte received after a Start or Restart condition is compared against the value stored in this register. If the byte matches, the value is loaded into the SSP1BUF register and an interrupt is generated. If the value does not match, the module goes idle and no indication is given to the software that anything happened.

The SSP Mask register (Register 30-5) affects the address matching process. See **Section 30.5.9** "**SSP Mask Register**" for more information.

#### 30.5.1.1 I<sup>2</sup>C Slave 7-Bit Addressing Mode

In 7-bit Addressing mode, the LSb of the received data byte is ignored when determining if there is an address match.

#### 30.5.1.2 I<sup>2</sup>C Slave 10-Bit Addressing Mode

In 10-bit Addressing mode, the first received byte is compared to the binary value of '1 1 1 1 0 A9 A8 0'. A9 and A8 are the two MSb's of the 10-bit address and stored in bits 2 and 1 of the SSP1ADD register.

After the acknowledge of the high byte the UA bit is set and SCL is held low until the user updates SSP1ADD with the low address. The low address byte is clocked in and all eight bits are compared to the low address value in SSP1ADD. Even if there is not an address match; SSP1IF and UA are set, and SCL is held low until SSP1ADD is updated to receive a high byte again. When SSP1ADD is updated the UA bit is cleared. This ensures the module is ready to receive the high address byte on the next communication.

A high and low address match as a write request is required at the start of all 10-bit addressing communication. A transmission can be initiated by issuing a Restart once the slave is addressed, and clocking in the high address with the R/W bit set. The slave hardware will then acknowledge the read request and prepare to clock out data. This is only valid for a slave after it has received a complete high and low address byte match.

## 30.6.6 I<sup>2</sup>C MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address is accomplished by simply writing a value to the SSP1BUF register. This action will set the Buffer Full flag bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDA pin after the falling edge of SCL is asserted. SCL is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCL is released high. When the SCL pin is released high, it is held that way for TBRG. The data on the SDA pin must remain stable for that duration and some hold time after the next falling edge of SCL. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDA. This allows the slave device being addressed to respond with an  $\overline{ACK}$  bit during the ninth bit time if an address match occurred, or if data was received properly. The status of ACK is written into the ACKSTAT bit on the rising edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared. If not, the bit is set. After the ninth clock, the SSP1IF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSP1BUF, leaving SCL low and SDA unchanged (Figure 30-28).

After the write to the SSP1BUF, each bit of the address will be shifted out on the falling edge of SCL until all seven address bits and the R/W bit are completed. On the falling edge of the eighth clock, the master will release the SDA pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDA pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT Status bit of the SSP1CON2 register. Following the falling edge of the ninth clock transmission of the address, the SSP1IF is set, the BF flag is cleared and the Baud Rate Generator is turned off until another write to the SSP1BUF takes place, holding SCL low and allowing SDA to float.

## 30.6.6.1 BF Status Flag

In Transmit mode, the BF bit of the SSP1STAT register is set when the CPU writes to SSP1BUF and is cleared when all eight bits are shifted out.

## 30.6.6.2 WCOL Status Flag

If the user writes the SSP1BUF when a transmit is already in progress (i.e., SSP1SR is still shifting out a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).

WCOL must be cleared by software before the next transmission.

## 30.6.6.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit of the SSP1CON2 register is cleared when the slave has sent an Acknowledge ( $\overline{ACK} = 0$ ) and is set when the slave does not Acknowledge ( $\overline{ACK} = 1$ ). A slave sends an Acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

30.6.6.4 Typical Transmit Sequence

- 1. The user generates a Start condition by setting the SEN bit of the SSP1CON2 register.
- 2. SSP1IF is set by hardware on completion of the Start.
- 3. SSP1IF is cleared by software.
- 4. The MSSP module will wait the required start time before any other operation takes place.
- 5. The user loads the SSP1BUF with the slave address to transmit.
- 6. Address is shifted out the SDA pin until all eight bits are transmitted. Transmission begins as soon as SSP1BUF is written to.
- 7. The MSSP module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSP1CON2 register.
- 8. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSP1IF bit.
- 9. The user loads the SSP1BUF with eight bits of data.
- 10. Data is shifted out the SDA pin until all eight bits are transmitted.
- 11. The MSSP module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSP1CON2 register.
- 12. Steps 8-11 are repeated for all transmitted data bytes.
- 13. The user generates a Stop or Restart condition by setting the PEN or RSEN bits of the SSP1CON2 register. Interrupt is generated once the Stop/Restart condition is complete.

# 31.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

The Enhanced Universal Synchronous Asynchronous Receiver Transmitter (EUSART) module is a serial I/O communications peripheral. It contains all the clock generators, shift registers and data buffers necessary to perform an input or output serial data transfer independent of device program execution. The EUSART, also known as a Serial Communications Interface (SCI), can be configured as a full-duplex asynchronous system or half-duplex synchronous system. Full-Duplex mode is useful for communications with peripheral systems, such as CRT terminals and personal computers. Half-Duplex Synchronous mode is intended for communications with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs or other microcontrollers. These devices typically do not have internal clocks for baud rate generation and require the external clock signal provided by a master synchronous device.

The EUSART module includes the following capabilities:

- · Full-duplex asynchronous transmit and receive
- Two-character input buffer
- · One-character output buffer
- Programmable 8-bit or 9-bit character length
- Address detection in 9-bit mode
- Input buffer overrun error detection
- Received character framing error detection
- Half-duplex synchronous master
- Half-duplex synchronous slave
- Programmable clock polarity in synchronous modes
- Sleep operation

The EUSART module implements the following additional features, making it ideally suited for use in Local Interconnect Network (LIN) bus systems:

- Automatic detection and calibration of the baud rate
- Wake-up on Break reception
- 13-bit Break character transmit

Block diagrams of the EUSART transmitter and receiver are shown in Figure 31-1 and Figure 31-2.

The EUSART transmit output (TX\_out) is available to the TX/CK pin and internally to the following peripherals:

• Configurable Logic Cell (CLC)

#### FIGURE 31-1: EUSART TRANSMIT BLOCK DIAGRAM

![](_page_16_Figure_23.jpeg)

## 31.1.2.4 Receive Framing Error

Each character in the receive FIFO buffer has a corresponding framing error Status bit. A framing error indicates that a Stop bit was not seen at the expected time. The framing error status is accessed via the FERR bit of the RC1STA register. The FERR bit represents the status of the top unread character in the receive FIFO. Therefore, the FERR bit must be read before reading the RC1REG.

The FERR bit is read-only and only applies to the top unread character in the receive FIFO. A framing error (FERR = 1) does not preclude reception of additional characters. It is not necessary to clear the FERR bit. Reading the next character from the FIFO buffer will advance the FIFO to the next character and the next corresponding framing error.

The FERR bit can be forced clear by clearing the SPEN bit of the RC1STA register which resets the EUSART. Clearing the CREN bit of the RC1STA register does not affect the FERR bit. A framing error by itself does not generate an interrupt.

Note:	If all receive characters in the receive
	FIFO have framing errors, repeated reads
	of the RC1REG will not clear the FERR
	bit.

#### 31.1.2.5 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before the FIFO is accessed. When this happens the OERR bit of the RC1STA register is set. The characters already in the FIFO buffer can be read but no additional characters will be received until the error is cleared. The error must be cleared by either clearing the CREN bit of the RC1STA register or by resetting the EUSART by clearing the SPEN bit of the RC1STA register.

#### 31.1.2.6 Receiving 9-Bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RC1STA register is set the EUSART will shift nine bits into the RSR for each character received. The RX9D bit of the RC1STA register is the ninth and Most Significant data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RC1REG.

#### 31.1.2.7 Address Detection

A special Address Detection mode is available for use when multiple receivers share the same transmission line, such as in RS-485 systems. Address detection is enabled by setting the ADDEN bit of the RC1STA register.

Address detection requires 9-bit character reception. When address detection is enabled, only characters with the ninth data bit set will be transferred to the receive FIFO buffer, thereby setting the RCIF interrupt bit. All other characters will be ignored.

Upon receiving an address character, user software determines if the address matches its own. Upon address match, user software must disable address detection by clearing the ADDEN bit before the next Stop bit occurs. When user software detects the end of the message, determined by the message protocol used, software places the receiver back into the Address Detection mode by setting the ADDEN bit.

# PIC16(L)F1717/8/9

![](_page_18_Figure_1.jpeg)

![](_page_18_Figure_2.jpeg)

![](_page_18_Figure_3.jpeg)

# PIC16(L)F1717/8/9

Note: Unless otherwise noted, VIN = 5V, Fosc = 500 kHz, CIN = 0.1  $\mu$ F, TA = 25°C.

![](_page_19_Figure_2.jpeg)

**FIGURE 35-31:** IDD, HS Oscillator, 32 MHz (8 MHz + 4x PLL), PIC16LF1717/8/9 Only.

![](_page_19_Figure_4.jpeg)

FIGURE 35-32: IDD, HS Oscillator, 32 MHz (8 MHz + 4x PLL), PIC16F1717/8/9 Only.

![](_page_19_Figure_6.jpeg)

FIGURE 35-33: IPD Base, LP Sleep Mode, PIC16LF1717/8/9 Only.

![](_page_19_Figure_8.jpeg)

**FIGURE 35-34:** IPD Base, LP Sleep Mode (VREGPM = 1), PIC16F1717/8/9 Only.

![](_page_19_Figure_10.jpeg)

FIGURE 35-35: IPD, Watchdog Timer (WDT), PIC16LF1717/8/9 Only.

![](_page_19_Figure_12.jpeg)

FIGURE 35-36: IPD, Watchdog Timer (WDT), PIC16F1717/8/9 Only.

## 37.2 Package Details

The following sections give the technical details of the packages.

## 28-Lead Skinny Plastic Dual In-Line (SP) – 300 mil Body [SPDIP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

![](_page_20_Figure_5.jpeg)

	Units		INCHES	
	Dimension Limits	MIN	NOM	MAX
Number of Pins	N		28	
Pitch	е		.100 BSC	
Top to Seating Plane	A	-	-	.200
Molded Package Thickness	A2	.120	.135	.150
Base to Seating Plane	A1	.015	-	-
Shoulder to Shoulder Width	E	.290	.310	.335
Molded Package Width	E1	.240	.285	.295
Overall Length	D	1.345	1.365	1.400
Tip to Seating Plane	L	.110	.130	.150
Lead Thickness	С	.008	.010	.015
Upper Lead Width	b1	.040	.050	.070
Lower Lead Width	b	.014	.018	.022
Overall Row Spacing §	eB	_	_	.430

#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic.
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-070B