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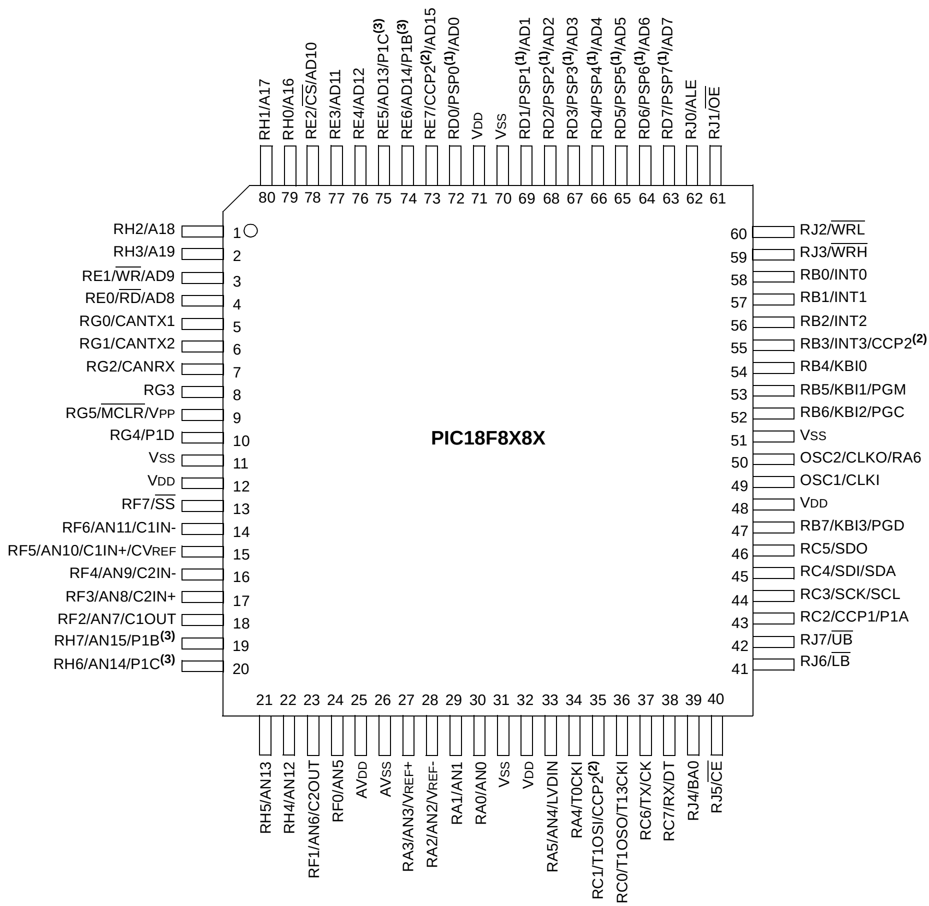
"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	40MHz
Connectivity	CANbus, I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LVD, POR, PWM, WDT
Number of I/O	52
Program Memory Size	48KB (24K x 16)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	3.25K x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 5.5V
Data Converters	A/D 12x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	68-LCC (J-Lead)
Supplier Device Package	68-PLCC (24.23x24.23)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf6585-i-l

Pin Diagrams (Continued)

80-Pin TQFP



Note 1: PSP is available only in Microcontroller mode.

2: CCP2 pin placement depends on CCP2MX and Processor mode settings.

3: P1B and P1C pin placement depends on ECCPMX setting.

Table of Contents

5.0 FLASH PROGRAM MEMORY

The Flash program memory is readable, writable and erasable during normal operation over the entire VDD range.

A read from program memory is executed on one byte at a time. A write to program memory is executed on blocks of 8 bytes at a time. Program memory is erased in blocks of 64 bytes at a time. A bulk erase operation cannot be issued from user code.

Writing or erasing program memory will cease instruction fetches until the operation is complete. The program memory cannot be accessed during the write or erase, therefore, code cannot execute. An internal programming timer terminates program memory writes and erases.

A value written to program memory does not need to be a valid instruction. Executing a program memory location that forms an invalid instruction results in a NOP.

5.1 Table Reads and Table Writes

In order to read and write program memory, there are two operations that allow the processor to move bytes between the program memory space and the data RAM:

- Table Read (TBLRD)
- Table Write (TBLWT)

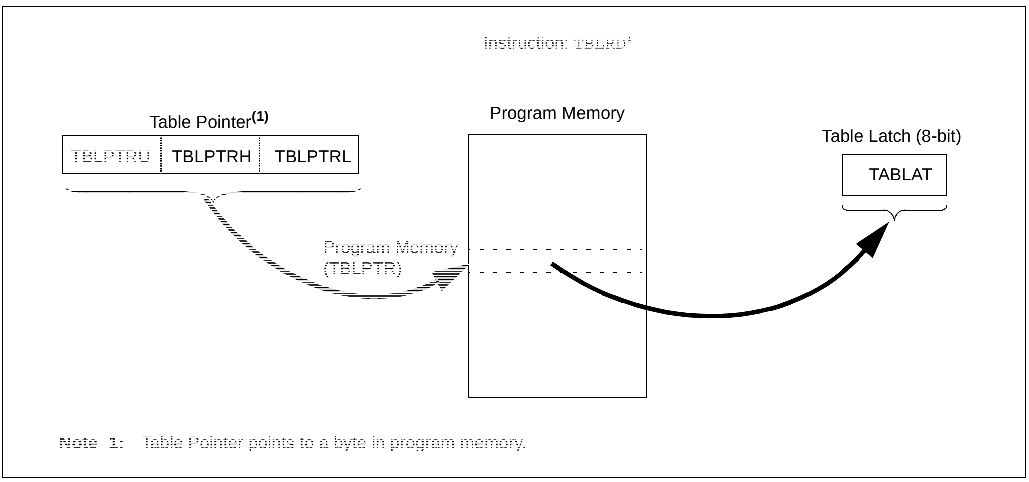
The program memory space is 16 bits wide, while the data RAM space is 8-bits wide. Table reads and table writes move data between these two memory spaces through an 8-bit register (TABLAT).

Table read operations retrieve data from program memory and places it into the data RAM space. Figure 5-1 shows the operation of a table read with program memory and data RAM.

Table write operations store data from the data memory space into holding registers in program memory. The procedure to write the contents of the holding registers into program memory is detailed in **Section 5.5 "Writing to Flash Program Memory"**. Figure 5-2 shows the operation of a table write with program memory and data RAM.

Table operations work with byte entities. A table block containing data, rather than program instructions, is not required to be word aligned. Therefore, a table block can start and end at any byte address. If a table write is being used to write executable code into program memory, program instructions will need to be word aligned.

FIGURE 5-1: TABLE READ OPERATION



10.4 PORTD, TRISD and LATD Registers

PORTD is an 8-bit wide, bidirectional port. The corresponding data direction register is TRISD. Setting a TRISD bit (= `1`) will make the corresponding PORTD pin an input (i.e., put the corresponding output driver in a high-impedance mode). Clearing a TRISD bit (= `0`) will make the corresponding PORTD pin an output (i.e., put the contents of the output latch on the selected pin).

The Data Latch register (LATD) is also memory mapped. Read-modify-write operations on the LATD register read and write the latched output value for PORTD.

PORTD is an 8-bit port with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

Note:	On a Power-on Reset, these pins are configured as digital inputs.

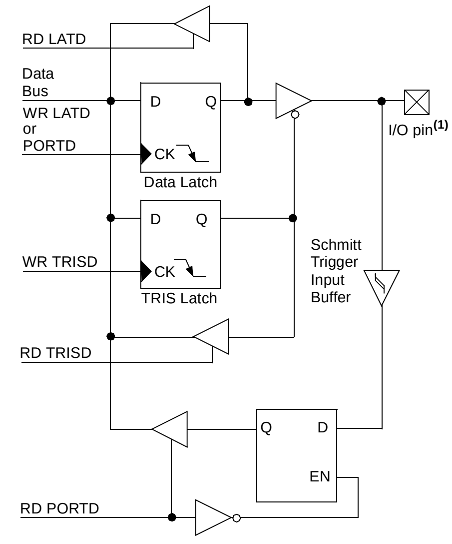
On PIC18F8X8X devices, PORTD is multiplexed with the system bus as the external memory interface; I/O port functions are only available when the system bus is disabled by setting the EBDIS bit in the MEMCOM register (MEMCON<7>). When operating as the external memory interface, PORTD is the low-order byte of the multiplexed address/data bus (AD7:AD0).

PORTD can also be configured as an 8-bit wide microprocessor port (Parallel Slave Port) by setting control bit, PSPMODE (TRISE<4>). In this mode, the input buffers are TTL. See **Section 10.10 "Parallel Slave Port (PSP)"** for additional information.

EXAMPLE 10-4: INITIALIZING PORTD

```
CLRF   PORTD   ; Initialize PORTD by
               ; clearing output
               ; data latches
CLRF   LATD    ; Alternate method
               ; to clear output
               ; data latches
MOVLW  0CFh    ; Value used to
               ; initialize data
               ; direction
MOVWF  TRISD   ; Set RD<3:0> as inputs
               ; RD<5:4> as outputs
               ; RD<7:6> as inputs
```

FIGURE 10-9: PORTD BLOCK DIAGRAM IN I/O PORT MODE



Note 1: I/O pins have diode protection to VDD and VSS.

REGISTER 15-2: CCP2CON REGISTER

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0
bit 7							bit 0

bit 7-6 **Unimplemented:** Read as '0'

bit 5-4 **DC2B1:DC2B0**: PWM Duty Cycle bit 1 and bit 0

Capture mode:
Unused.

Compare mode:
Unused.

PWM mode:
These bits are the two LSbs of the 10-bit PWM duty cycle. The eight MSbs of the duty cycle are found in CCPR2L.

bit 3-0 **CCP2M3:CCP2M0**: CCP2 Mode Select bits

0000 = Capture/Compare/PWM off (resets CCP2 module)
0001 = Reserved
0010 = Compare mode, toggle output on match
0011 = Reserved
0100 = Capture mode, every falling edge
0101 = Capture mode, every rising edge
0110 = Capture mode, every 4th rising edge
0111 = Capture mode, every 16th rising edge
1000 = Compare mode, initialize CCP pin low, on compare match force CCP pin high
1001 = Compare mode, initialize CCP pin high, on compare match force CCP pin low
1010 = Compare mode, generate software interrupt only, CCP pin is unaffected
1011 = Compare mode, trigger special event, resets TMR1 or TMR3 and starts A/D conversion if A/D module is enabled
11xx = PWM mode

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

15.4 PWM Mode

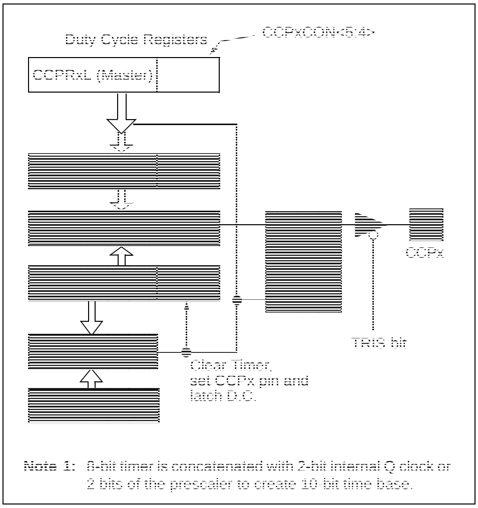
In Pulse Width Modulation (PWM) mode, the CCPx pin produces up to a 10-bit resolution PWM output. For PWM mode to function properly, the TRIS bit for the CCPx pin must be cleared to make it an output.

Note:	Clearing the CCPxCON register will force the CCPx PWM output latch to the default low level. This is not the port data latch.

Figure 15-3 shows a simplified block diagram of the CCP module in PWM mode.

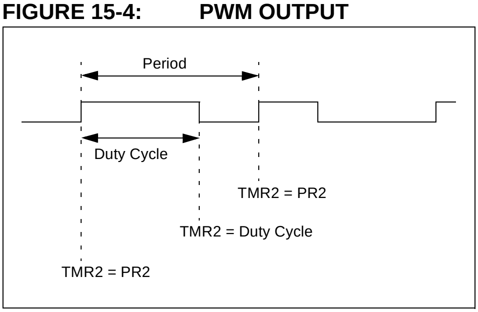
For a step-by-step procedure on how to set up the CCP module for PWM operation, see **Section 15.4.3 "Setup for PWM Operation"**.

FIGURE 15-3: SIMPLIFIED PWM BLOCK DIAGRAM



A PWM output (Figure 15-4) has a time base (period) and a time that the output stays high (duty cycle). The frequency of the PWM is the inverse of the period (1/period).

FIGURE 15-4: PWM OUTPUT



15.4.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following formula.

EQUATION 15-1:

$$\text{PWM Period} = [(PR2) + 1] \cdot 4 \cdot T_{OSC} \cdot (\text{TMR2 Prescale Value})$$

PWM frequency is defined as 1/[PWM period].

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCP1 pin is set (exception: if PWM duty cycle = 0%, the CCP1 pin will not be set)
- The PWM duty cycle is latched from CCPR1L into CCPR1H

Note:	The Timer2 postscaler (see **Section 13.0 "Timer2 Module"**) is not used in the determination of the PWM frequency. The postscaler could be used to have a servo update rate at a different frequency than the PWM output.

15.4.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing to the CCPRxL register and to the CCPxCON<5:4> bits. Up to 10-bit resolution is available. The CCPRxL contains the eight MSbs and the CCPxCON<5:4> contain the two LSbs. This 10-bit value is represented by CCPRxL:CCPxCON<5:4>. The following equation is used to calculate the PWM duty cycle in time.

EQUATION 15-2:

$$\text{PWM Duty Cycle} = (\text{CCPRxL:CCPxCON<5:4>}) \cdot T_{OSC} \cdot (\text{TMR2 Prescale Value})$$

CCPRxL and CCPxCON<5:4> can be written to at any time but the duty cycle value is not latched into CCPRxH until after a match between PR2 and TMR2 occurs (i.e., the period is complete). In PWM mode, CCPRxH is a read-only register.

The CCPRxH register and a 2-bit internal latch are used to double-buffer the PWM duty cycle. This double-buffering is essential for glitchless PWM operation.

When the CCPRxH and 2-bit latch match TMR2, concatenated with an internal 2-bit Q clock or 2 bits of the TMR2 prescaler, the CCPx pin is cleared.

16.2 Enhanced PWM Mode

The Enhanced PWM mode provides additional PWM output options for a broader range of control applications. The module is a backward compatible version of the standard CCP module and offers up to four outputs, designated P1A through P1D. Users are also able to select the polarity of the signal (either active-high or active-low). The module's output mode and polarity are configured by setting the P1M1:P1M0 and CCP1M3:CCP1M0 bits of the CCP1CON register (CCP1CON<7:6> and CCP1CON<3:0>, respectively).

Figure 16-2 shows a simplified block diagram of PWM operation. All control registers are double-buffered and are loaded at the beginning of a new PWM cycle (the period boundary when Timer2 resets) in order to prevent glitches on any of the outputs. The exception is the PWM Delay register, ECCP1DEL, which is loaded at either the duty cycle boundary or the boundary period (whichever comes first). Because of the buffering, the module waits until the assigned timer resets instead of starting immediately. This means that enhanced PWM waveforms do not exactly match the standard PWM waveforms, but are instead offset by one full instruction cycle (4 TOSC).

As before, the user must manually configure the appropriate TRIS bits for output.

16.2.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following equation.

EQUATION 16-1:

$$\text{PWM Period} = [(\text{PR2}) + 1] \cdot 4 \cdot T_{OSC} \cdot (\text{TMR2 Prescale Value})$$

PWM frequency is defined as 1/[PWM period]. When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCP1 pin is set (if PWM duty cycle = 0%, the CCP1 pin will not be set)
- The PWM duty cycle is copied from CCPR1L into CCPR1H

Note:	The Timer2 postscaler (see **Section 13.0 "Timer2 Module"**) is not used in the determination of the PWM frequency. The postscaler could be used to have a servo update rate at a different frequency than the PWM output.

16.2.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing to the CCPR1L register and to the CCP1CON<5:4> bits. Up to 10-bit resolution is available. The CCPR1L contains the eight MSbs and the CCP1CON<5:4> contains the two LSbs. This 10-bit value is represented by CCPR1L:CCP1CON<5:4>. The PWM duty cycle is calculated by the following equation.

EQUATION 16-2:

$$\text{PWM Duty Cycle} = (\text{CCPR1L:CCP1CON<5:4>}) \cdot T_{OSC} \cdot (\text{TMR2 Prescale Value})$$

CCPR1L and CCP1CON<5:4> can be written to at any time, but the duty cycle value is not copied into CCPR1H until a match between PR2 and TMR2 occurs (i.e., the period is complete). In PWM mode, CCPR1H is a read-only register.

The CCPR1H register and a 2-bit internal latch are used to double-buffer the PWM duty cycle. This double-buffering is essential for glitchless PWM operation. When the CCPR1H and 2-bit latch match TMR2, concatenated with an internal 2-bit Q clock or two bits of the TMR2 prescaler, the CCP1 pin is cleared. The maximum PWM resolution (bits) for a given PWM frequency is given by the following equation:

EQUATION 16-3:

$$\text{PWM Resolution (max)} = \frac{\log\left(\frac{F_{OSC}}{F_{PWM}}\right)}{\log(2)} \text{ bits}$$

Note:	If the PWM duty cycle value is longer than the PWM period, the CCP1 pin will not be cleared.

16.2.3 PWM OUTPUT CONFIGURATIONS

The P1M1:P1M0 bits in the CCP1CON register allow one of four configurations:

- Single Output
- Half-Bridge Output
- Full-Bridge Output, Forward mode
- Full-Bridge Output, Reverse mode

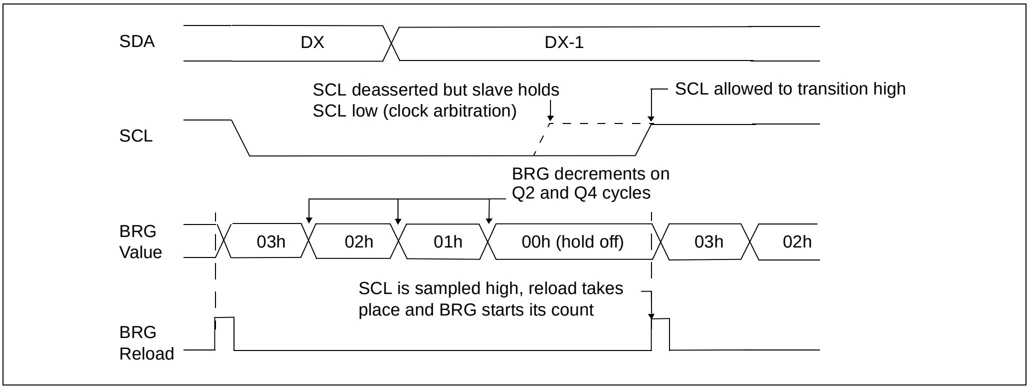
The Single Output mode is the standard PWM mode discussed in **Section 16.2 "Enhanced PWM Mode"**. The Half-Bridge and Full-Bridge Output modes are covered in detail in the sections that follow.

The general relationship of the outputs in all configurations is summarized in Figure 16-3.

17.4.7.1 Clock Arbitration

Clock arbitration occurs when the master, during any receive, transmit or Repeated Start/Stop condition, deasserts the SCL pin (SCL allowed to float high). When the SCL pin is allowed to float high, the Baud Rate Generator (BRG) is suspended from counting until the SCL pin is actually sampled high. When the SCL pin is sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and begins counting. This ensures that the SCL high time will always be at least one BRG rollover count in the event that the clock is held low by an external device (Figure 17-18).

FIGURE 17-18: BAUD RATE GENERATOR TIMING WITH CLOCK ARBITRATION



NOTES:

REGISTER 23-51: MSEL3: MASK SELECT REGISTER 3[1]

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
FIL15_1	FIL15_0	FIL14_1	FIL14_0	FIL13_1	FIL13_0	FIL12_1	FIL12_0
bit 7							bit 0

bit 7-6 **FIL15_1:FIL15_0:** Filter 15 Select bits 1 and 0
`11` = No mask
`10` = Filter 15
`01` = Acceptance Mask 1
`00` = Acceptance Mask 0

bit 5-4 **FIL14_1:FIL14_0:** Filter 14 Select bits 1 and 0
`11` = No mask
`10` = Filter 15
`01` = Acceptance Mask 1
`00` = Acceptance Mask 0

bit 3-2 **FIL13_1:FIL13_0:** Filter 13 Select bits 1 and 0
`11` = No mask
`10` = Filter 15
`01` = Acceptance Mask 1
`00` = Acceptance Mask 0

bit 1-0 **FIL12_1:FIL12_0:** Filter 12 Select bits 1 and 0
`11` = No mask
`10` = Filter 15
`01` = Acceptance Mask 1
`00` = Acceptance Mask 0

Note 1: This register is available in Mode 1 and 2 only.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
- n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

TABLE 24-1: CONFIGURATION BITS AND DEVICE IDS

File Name		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Default/ Unprogrammed Value
300001h	CONFIG1H	—	—	$\overline{\text{OSCSEN}}$	—	FOSC3	FOSC2	FOSC1	FOSC0	`--1- 1111`
300002h	CONFIG2L	—	—	—	—	BORV1	BORV0	BODEN	$\overline{\text{PWRTEN}}$	`---- 1111`
300003h	CONFIG2H	—	—	—	WDTPS3	WDTPS2	WDTPS1	WDTPS0	WDTEN	`---1 1111`
300004h(1)	CONFIG3L	WAIT	—	—	—	—	—	PM1	PM0	`1--- --11`
300005h	CONFIG3H	MCLRE	—	—	—	—	—	ECCPMX(4)	CCP2MX	`1--- --11`
300006h	CONFIG4L	$\overline{\text{DEBUG}}$	—	—	—	—	LVP	—	STVREN	`1--- -1-1`
300008h	CONFIG5L	—	—	—	—	CP3(2)	CP2	CP1	CP0	`---- 1111`
300009h	CONFIG5H	CPD	CPB	—	—	—	—	—	—	`11-- ----`
30000Ah	CONFIG6L	—	—	—	—	WRT3(2)	WRT2	WRT1	WRT0	`---- 1111`
30000Bh	CONFIG6H	WRTD	WRTB	WRTC	—	—	—	—	—	`111- ----`
30000Ch	CONFIG7L	—	—	—	—	EBTR3(2)	EBTR2	EBTR1	EBTR0	`---- 1111`
30000Dh	CONFIG7H	—	EBTRB	—	—	—	—	—	—	`-1-- ----`
3FFFFEh	DEVID1	DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0	**(Note 3)**
3FFFFFh	DEVID2	DEV10	DEV9	DEV8	DEV7	DEV6	DEV5	DEV4	DEV3	`0000 1010`

Legend: `x` = unknown, `u` = unchanged, `-` = unimplemented, `q` = value depends on condition.
Shaded cells are unimplemented, read as '`0`'.

Note 1: Unimplemented in PIC18F6X8X devices; maintain this bit set.
2: Unimplemented in PIC18FX585 devices; maintain this bit set.
3: See Register 24-13 for DEVID1 values.
4: Reserved in PIC18F6X8X devices; maintain this bit set.

REGISTER 24-13: DEVICE ID REGISTER 1 FOR PIC18FXX8X DEVICES (ADDRESS 3FFFFEh)

R	R	R	R	R	R	R	R
DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0
bit 7							bit 0

bit 7-5 **DEV2:DEV0:** Device ID bits

`000` = PIC18F8680
`001` = PIC18F6680
`010` = PIC18F8585
`011` = PIC18F6585

bit 4-0 **REV4:REV0:** Revision ID bits

These bits are used to indicate the device revision.

Legend:		
R = Readable bit	P = Programmable bit	U = Unimplemented bit, read as ‘0’
- n = Value when device is unprogrammed		u = Unchanged from programmed state

REGISTER 24-14: DEVICE ID REGISTER 2 FOR PIC18FXX8X DEVICES (ADDRESS 3FFFFFh)

R-0	R-0	R-0	R-0	R-1	R-0	R-1	R-0
DEV10	DEV9	DEV8	DEV7	DEV6	DEV5	DEV4	DEV3
bit 7							bit 0

bit 7-0 **DEV10:DEV3:** Device ID bits

These bits are used with the DEV2:DEV0 bits in the Device ID Register 1 to identify the part number.

`0000 1010` = PIC18F6585/8585/6680/8680

Legend:		
R = Readable bit	P = Programmable bit	U = Unimplemented bit, read as ‘0’
- n = Value when device is unprogrammed		u = Unchanged from programmed state

24.2 Watchdog Timer (WDT)

The Watchdog Timer is a free-running, on-chip RC oscillator which does not require any external components. This RC oscillator is separate from the RC oscillator of the OSC1/CLKI pin. That means that the WDT will run even if the clock on the OSC1/CLKI and OSC2/CLKO/RA6 pins of the device has been stopped, for example, by execution of a `SLEEP` instruction.

During normal operation, a WDT time-out generates a device Reset (Watchdog Timer Reset). If the device is in Sleep mode, a WDT time-out causes the device to wake-up and continue with normal operation (Watchdog Timer wake-up). The $\overline{TO}$ bit in the RCON register will be cleared upon a WDT time-out.

The Watchdog Timer is enabled/disabled by a device configuration bit. If the WDT is enabled, software execution may not disable this function. When the WDTEN configuration bit is cleared, the SWDTEN bit enables/disables the operation of the WDT.

The WDT time-out period values may be found in **Section 27.0 "Electrical Characteristics"** under parameter #31. Values for the WDT postscaler may be assigned using the configuration bits.

> **Note 1:** The `CLRWDT` and `SLEEP` instructions clear the WDT and the postscaler if assigned to the WDT and prevent it from timing out and generating a device Reset condition.
>
> **2:** When a `CLRWDT` instruction is executed and the postscaler is assigned to the WDT, the postscaler count will be cleared but the postscaler assignment is not changed.

24.2.1 CONTROL REGISTER

Register 24-15 shows the WDTCON register. This is a readable and writable register which contains a control bit that allows software to override the WDT enable configuration bit, only when the configuration bit has disabled the WDT.

REGISTER 24-15: WDTCON REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	U-0	R/W-0
—	—	—	—	—	—	—	SWDTEN
bit 7							bit 0

bit 7-1 **Unimplemented**: Read as '0'

bit 0 **SWDTEN:** Software Controlled Watchdog Timer Enable bit
1 = Watchdog Timer is on
0 = Watchdog Timer is turned off if the WDTEN configuration bit in the Configuration register = 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

24.4.2 DATA EEPROM CODE PROTECTION

The entire data EEPROM is protected from external reads and writes by two bits: CPD and WRTD. CPD inhibits external reads and writes of data EEPROM. WRTD inhibits external writes to data EEPROM. The CPU can continue to read and write data EEPROM regardless of the protection bit settings.

24.4.3 CONFIGURATION REGISTER PROTECTION

The Configuration registers can be write-protected. The WRTC bit controls protection of the Configuration registers. In User mode, the WRTC bit is readable only. WRTC can only be written via ICSP or an external programmer.

24.5 ID Locations

Eight memory locations (200000h-200007h) are designated as ID locations where the user can store checksum or other code identification numbers. These locations are accessible during normal execution through the `TBLRD` and `TBLWT` instructions or during program/verify. The ID locations can be read when the device is code-protected.

24.6 In-Circuit Serial Programming

PIC18FXX80/XX85 microcontrollers can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data, and three other lines for power, ground and the programming voltage. This allows customers to manufacture boards with unprogrammed devices and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

24.7 In-Circuit Debugger

When the DEBUG bit in Configuration register, CONFIG4L, is programmed to a '0', the in-circuit debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB® IDE. When the microcontroller has this feature enabled, some of the resources are not available for general use. Table 24-4 shows which features are consumed by the background debugger.

TABLE 24-4: DEBUGGER RESOURCES

I/O pins	RB6, RB7
Stack	2 levels
Program Memory	512 bytes
Data Memory	10 bytes

To use the in-circuit debugger function of the microcontroller, the design must implement In-Circuit Serial Programming connections to $\overline{MCLR}$/V_{PP}, V_{DD}, GND, RB7 and RB6. This will interface to the in-circuit debugger module available from Microchip or one of the third party development tool companies.

GOTO — Unconditional Branch

Syntax: [*label*] GOTO k

Operands: $0 \leq k \leq 1048575$

Operation: $k \rightarrow$ PC<20:1>

Status Affected: None

Encoding:

1st word (k<7:0>)	1110	1111	$k_7$kkk	kkkk$_0$
2nd word(k<19:8>)	1111	k_{19}kkk	kkkk	kkkk$_8$

Description: GOTO allows an unconditional branch anywhere within entire 2-Mbyte memory range. The 20-bit value 'k' is loaded into PC<20:1>. GOTO is always a two-cycle instruction.

Words: 2

Cycles: 2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'<7:0>,	No operation	Read literal 'k'<19:8>, Write to PC
No operation	No operation	No operation	No operation

Example: `GOTO THERE`

After Instruction
PC = Address (THERE)

INCF — Increment f

Syntax: [*label*] INCF f [,d [,a]]

Operands: $0 \leq f \leq 255$
$d \in [0,1]$
$a \in [0,1]$

Operation: (f) + 1 $\rightarrow$ dest

Status Affected: C, DC, N, OV, Z

Encoding:

0010	10da	ffff	ffff

Description: The contents of register 'f' are incremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default). If 'a' is '0', the Access Bank will be selected, overriding the BSR value. If 'a' = 1, then the bank will be selected as per the BSR value (default).

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: `INCF CNT, 1, 0`

Before Instruction
CNT = 0xFF
Z = 0
C = ?
DC = ?

After Instruction
CNT = 0x00
Z = 1
C = 1
DC = 1

RLNCF — Rotate Left f (no carry)

Syntax: [*label*] RLNCF f [,d [,a]]

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

Operation: (f<n>) $\rightarrow$ dest<n+1>,
(f<7>) $\rightarrow$ dest<0>

Status Affected: N, Z

Encoding:

0100	01da	ffff	ffff

Description: The contents of register ‘f’ are rotated one bit to the left. If ‘d’ is ‘0,’ the result is placed in W. If ‘d’ is ‘1’, the result is stored back in register ‘f’ (default). If ‘a’ is ‘0’, the Access Bank will be selected, overriding the BSR value. If ‘a’ is ‘1’, then the bank will be selected as per the BSR value (default).

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register ‘f’	Process Data	Write to destination

Example: `RLNCF REG, 1, 0`

Before Instruction
REG = `1010 1011`

After Instruction
REG = `0101 0111`

RRCF — Rotate Right f through Carry

Syntax: [*label*] RRCF f [,d [,a]]

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

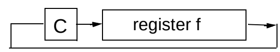
Operation: (f<n>) $\rightarrow$ dest<n-1>,
(f<0>) $\rightarrow$ C,
(C) $\rightarrow$ dest<7>

Status Affected: C, N, Z

Encoding:

0011	00da	ffff	ffff

Description: The contents of register ‘f’ are rotated one bit to the right through the Carry flag. If ‘d’ is ‘0’, the result is placed in W. If ‘d’ is ‘1’, the result is placed back in register ‘f’ (default). If ‘a’ is ‘0’, the Access Bank will be selected, overriding the BSR value. If ‘a’ is ‘1’, then the bank will be selected as per the BSR value (default).



Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register ‘f’	Process Data	Write to destination

Example: `RRCF REG, 0, 0`

Before Instruction
REG = `1110 0110`
C = 0

After Instruction
REG = `1110 0110`
W = `0111 0011`
C = 0

26.9 MPLAB ICE 2000 High-Performance Universal In-Circuit Emulator

The MPLAB ICE 2000 universal in-circuit emulator is intended to provide the product development engineer with a complete microcontroller design tool set for PIC microcontrollers. Software control of the MPLAB ICE 2000 in-circuit emulator is advanced by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICE 2000 is a full-featured emulator system with enhanced trace, trigger and data monitoring features. Interchangeable processor modules allow the system to be easily reconfigured for emulation of different processors. The universal architecture of the MPLAB ICE in-circuit emulator allows expansion to support new PIC microcontrollers.

The MPLAB ICE 2000 in-circuit emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft® Windows 32-bit operating system were chosen to best make these features available in a simple, unified application.

26.10 MPLAB ICE 4000 High-Performance Universal In-Circuit Emulator

The MPLAB ICE 4000 universal in-circuit emulator is intended to provide the product development engineer with a complete microcontroller design tool set for high-end PIC microcontrollers. Software control of the MPLAB ICE in-circuit emulator is provided by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICD 4000 is a premium emulator system, providing the features of MPLAB ICE 2000, but with increased emulation memory and high-speed performance for dsPIC30F and PIC18XXXX devices. Its advanced emulator features include complex triggering and timing, up to 2 Mb of emulation memory and the ability to view variables in real-time.

The MPLAB ICE 4000 in-circuit emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft Windows 32-bit operating system were chosen to best make these features available in a simple, unified application.

26.11 MPLAB ICD 2 In-Circuit Debugger

Microchip's In-Circuit Debugger, MPLAB ICD 2, is a powerful, low-cost, run-time development tool, connecting to the host PC via an RS-232 or high-speed USB interface. This tool is based on the Flash PIC MCUs and can be used to develop for these and other PIC microcontrollers. The MPLAB ICD 2 utilizes the in-circuit debugging capability built into the Flash devices. This feature, along with Microchip's In-Circuit Serial Programming™ (ICSP™) protocol, offers cost effective in-circuit Flash debugging from the graphical user interface of the MPLAB Integrated Development Environment. This enables a designer to develop and debug source code by setting breakpoints, single-stepping and watching variables, CPU status and peripheral registers. Running at full speed enables testing hardware and applications in real-time. MPLAB ICD 2 also serves as a development programmer for selected PIC devices.

26.12 PRO MATE II Universal Device Programmer

The PRO MATE II is a universal, CE compliant device programmer with programmable voltage verification at V_{DDMIN} and V_{DDMAX} for maximum reliability. It features an LCD display for instructions and error messages and a modular detachable socket assembly to support various package types. In Stand-Alone mode, the PRO MATE II device programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode.

26.13 MPLAB PM3 Device Programmer

The MPLAB PM3 is a universal, CE compliant device programmer with programmable voltage verification at V_{DDMIN} and V_{DDMAX} for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages and a modular detachable socket assembly to support various package types. The ICSP™ cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 device programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. MPLAB PM3 connects to the host PC via an RS-232 or USB cable. MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices and incorporates an SD/MMC card for file storage and secure data applications.

27.2 DC Characteristics: Power-down and Supply Current PIC18FXX8X (Industrial, Extended) PIC18LFXX8X (Industrial) (Continued)

PIC18LFXX8X (Industrial)		Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial					
PIC18FXX8X (Industrial, Extended)		**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended					
Param. No.	**Device**	**Typ**	**Max**	**Units**	**Conditions**		
	Supply Current (IDD)(2,3)						
	PIC18FXX8X	13	27	mA	-40°C	VDD = 4.2V	FOSC = 25 MHz, EC oscillator
		15	27	mA	+25°C		
		19	29	mA	+85°C		
	PIC18FXX8X	17	31	mA	-40°C	VDD = 5.0V	
		21	31	mA	+25°C		
		23	34	mA	+85°C		
	PIC18FXX8X	20	34	mA	-40°C	VDD = 4.2V	FOSC = 40 MHz, EC oscillator
		24	34	mA	+25°C		
		29	44	mA	+85°C		
	PIC18FXX8X	28	46	mA	-40°C	VDD = 5.0V	
		33	46	mA	+25°C		
		40	51	mA	+85°C		
D014	PIC18LFXX8X	27	45	μA	-10°C	VDD = 2.0V	FOSC = 32 kHz, Timer1 as clock
		30	50	μA	+25°C		
		32	54	μA	+70°C		
	PIC18LFXX8X	33	55	μA	-10°C	VDD = 3.0V	
		36	60	μA	+25°C		
		39	65	μA	+70°C		
	All devices	75	125	μA	-10°C	VDD = 5.0V	
		90	150	μA	+25°C		
		113	188	μA	+70°C		

Legend: Shading of rows is to assist in readability of the table.

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or VSS and all features that add delta current disabled (such as WDT, Timer1 Oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.
The test conditions for all IDD measurements in active operation mode are:
OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;
MCLR = VDD; WDT enabled/disabled as specified.

3: For RC oscillator configurations, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/2REXT (mA) with REXT in kΩ.

FIGURE 27-18: EXAMPLE SPI SLAVE MODE TIMING (CKE = 1)

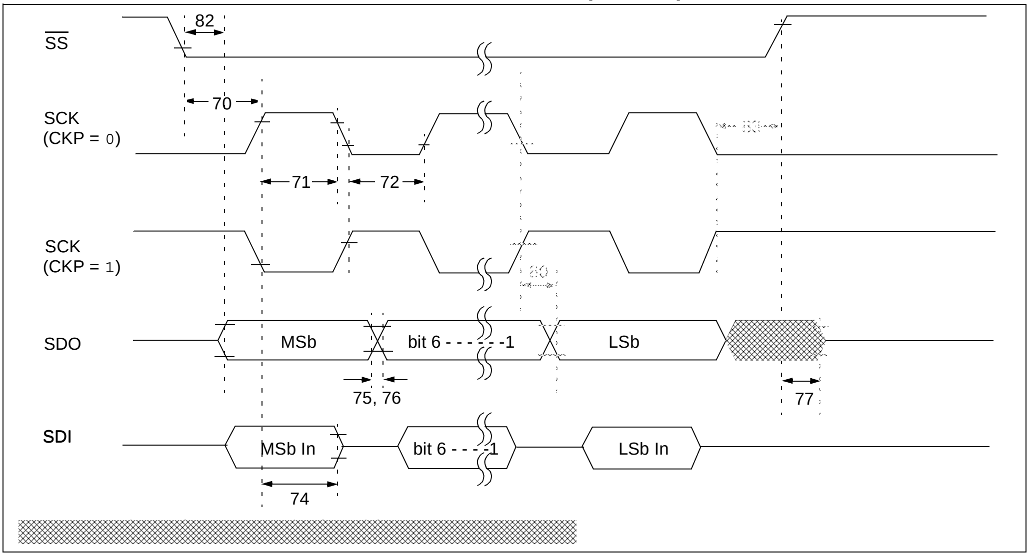


TABLE 27-18: EXAMPLE SPI SLAVE MODE REQUIREMENTS (CKE = 1)

Param. No.	Symbol	Characteristic		Min	Max	Units	Conditions
70	TSSL2SCH, TSSL2SCL	$\overline{SS}$ ↓ to SCK ↓ or SCK ↑ Input		TCY	—	ns	
71	TSCH	SCK Input High Time (Slave mode)	Continuous	1.25 TCY + 30	—	ns	
71A			Single Byte	40	—	ns	**(Note 1)**
72	TSCL	SCK Input Low Time (Slave mode)	Continuous	1.25 TCY + 30	—	ns	
72A			Single Byte	40	—	ns	**(Note 1)**
73A	TB2B	Last Clock Edge of Byte 1 to the First Clock Edge of Byte 2		1.5 TCY + 40	—	ns	**(Note 2)**
74	TSCH2DIL, TSCL2DIL	Hold Time of SDI Data Input to SCK Edge		100	—	ns	
75	TDOR	SDO Data Output Rise Time	PIC18FXX8X	—	25	ns	
			PIC18LFXX8X		45	ns	
76	TDOF	SDO Data Output Fall Time		—	25	ns	
77	TSSH2DOZ	$\overline{SS}$ ↑ to SDO Output High-Impedance		10	50	ns	
78	TSCR	SCK Output Rise Time (Master mode)	PIC18FXX8X	—	25	ns	
			PIC18LFXX8X	—	45	ns	
79	TSCF	SCK Output Fall Time (Master mode)		—	25	ns	
80	TSCH2DOV, TSCL2DOV	SDO Data Output Valid after SCK Edge	PIC18FXX8X	—	50	ns	
			PIC18LFXX8X	—	100	ns	
82	TSSL2DOV	SDO Data Output Valid after $\overline{SS}$ ↓ Edge	PIC18FXX8X	—	50	ns	
			PIC18LFXX8X	—	100	ns	
83	TSCH2SSH, TSCL2SSH	$\overline{SS}$ ↑ after SCK Edge		1.5 TCY + 40	—	ns	

Note 1: Requires the use of Parameter #73A.

2: Only if Parameter #71A and #72A are used.

Q

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