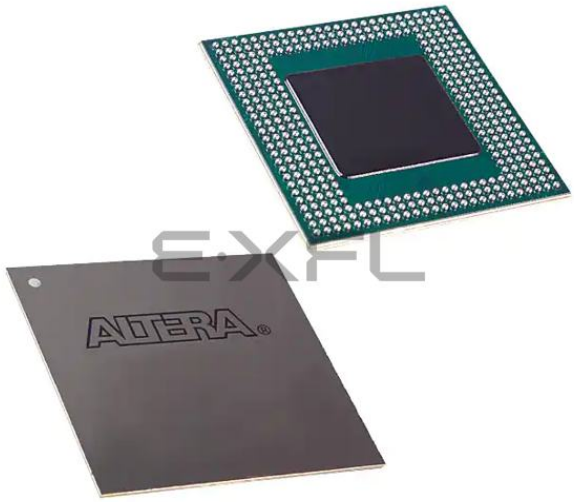
Intel - EPF10K100ABC356-3N Datasheet



Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	624
Number of Logic Elements/Cells	4992
Total RAM Bits	24576
Number of I/O	274
Number of Gates	158000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	356-LBGA
Supplier Device Package	356-BGA (35x35)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k100abc356-3n

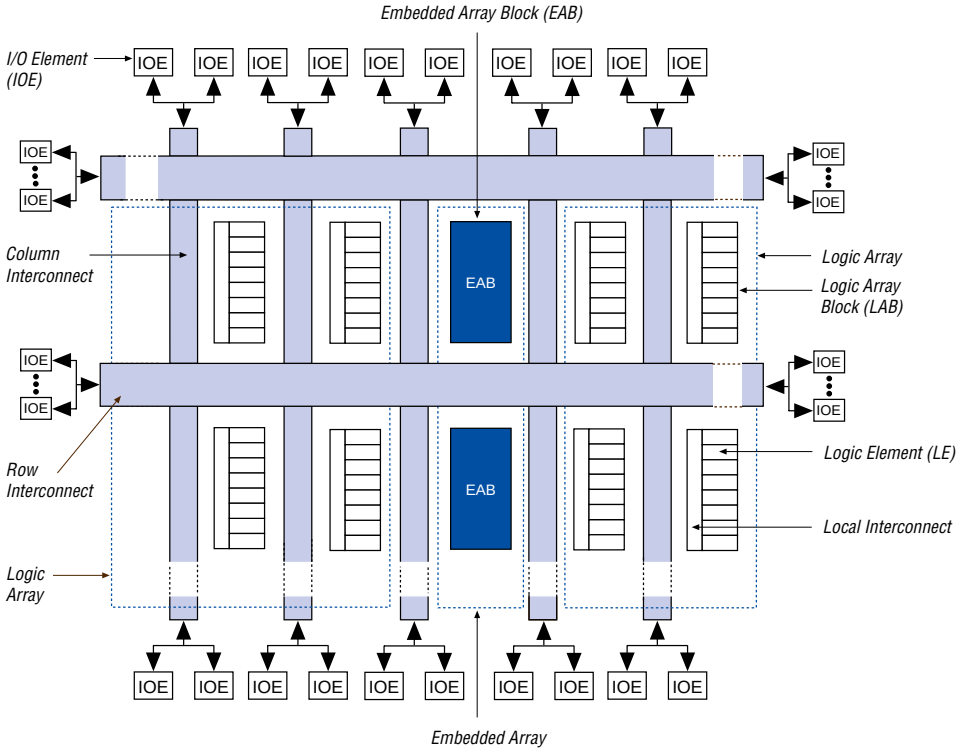
Table 4. FLEX 10K Package Options & I/O Pin Count *Note (1)*

Device	84-Pin PLCC	100-Pin TQFP	144-Pin TQFP	208-Pin PQFP RQFP	240-Pin PQFP RQFP
EPF10K10	59		102	134	
EPF10K10A		66	102	134	
EPF10K20			102	147	189
EPF10K30				147	189
EPF10K30A			102	147	189
EPF10K40				147	189
EPF10K50					189
EPF10K50V					189
EPF10K70					189
EPF10K100					
EPF10K100A					189
EPF10K130V					
EPF10K250A					

Table 5. FLEX 10K Package Options & I/O Pin Count (Continued) *Note (1)*

Device	503-Pin PGA	599-Pin PGA	256-Pin FineLine BGA	356-Pin BGA	484-Pin FineLine BGA	600-Pin BGA	403-Pin PGA
EPF10K10							
EPF10K10A			150		150 *(2)*		
EPF10K20							
EPF10K30				246			
EPF10K30A			191	246	246		
EPF10K40							
EPF10K50				274			310
EPF10K50V				274			
EPF10K70	358						
EPF10K100	406						
EPF10K100A				274	369	406	
EPF10K130V		470				470	
EPF10K250A		470				470	

Figure 1. FLEX 10K Device Block Diagram



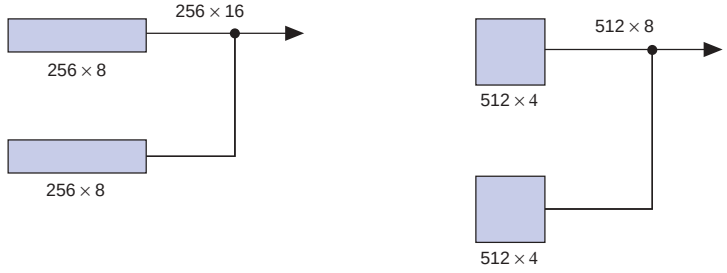
FLEX 10K devices provide six dedicated inputs that drive the flipflops' control inputs to ensure the efficient distribution of high-speed, low-skew (less than 1.5 ns) control signals. These signals use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect. Four of the dedicated inputs drive four global signals. These four global signals can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device.

Embedded Array Block

The EAB is a flexible block of RAM with registers on the input and output ports, and is used to implement common gate array megafunctions. The EAB is also suitable for functions such as multipliers, vector scalars, and error correction circuits, because it is large and flexible. These functions can be combined in applications such as digital filters and microcontrollers.

Larger blocks of RAM are created by combining multiple EABs. For example, two 256 × 8 RAM blocks can be combined to form a 256 × 16 RAM block; two 512 × 4 blocks of RAM can be combined to form a 512 × 8 RAM block. See Figure 3.

Figure 3. Examples of Combining EABs



If necessary, all EABs in a device can be cascaded to form a single RAM block. EABs can be cascaded to form RAM blocks of up to 2,048 words without impacting timing. Altera's software automatically combines EABs to meet a designer's RAM specifications.

EABs provide flexible options for driving and controlling clock signals. Different clocks can be used for the EAB inputs and outputs. Registers can be independently inserted on the data input, EAB output, or the address and `WE` inputs. The global signals and the EAB local interconnect can drive the `WE` signal. The global signals, dedicated clock pins, and EAB local interconnect can drive the EAB clock signals. Because the LEs drive the EAB local interconnect, the LEs can control the `WE` signal or the EAB clock signals.

Each EAB is fed by a row interconnect and can drive out to row and column interconnects. Each EAB output can drive up to two row channels and up to two column channels; the unused row channel can be driven by other LEs. This feature increases the routing resources available for EAB outputs. See Figure 4.

The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock, clear, and preset control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the output of the LE.

The LE has two outputs that drive the interconnect; one drives the local interconnect and the other drives either the row or column FastTrack Interconnect. The two outputs can be controlled independently. For example, the LUT can drive one output while the register drives the other output. This feature, called register packing, can improve LE utilization because the register and the LUT can be used for unrelated functions.

The FLEX 10K architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. The carry chain supports high-speed counters and adders; the cascade chain implements wide-input functions with minimum delay. Carry and cascade chains connect all LEs in an LAB and all LABs in the same row. Intensive use of carry and cascade chains can reduce routing flexibility. Therefore, the use of these chains should be limited to speed-critical portions of a design.

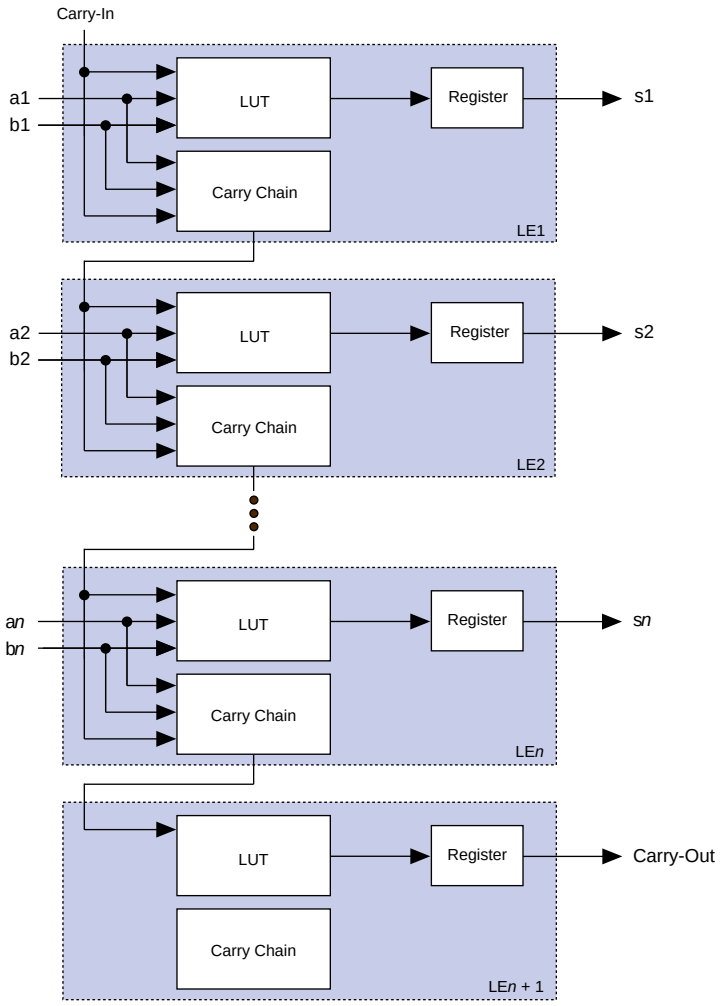
Carry Chain

The carry chain provides a very fast (as low as 0.2 ns) carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the FLEX 10K architecture to implement high-speed counters, adders, and comparators of arbitrary width efficiently. Carry chain logic can be created automatically by the Compiler during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of carry chains.

Carry chains longer than eight LEs are automatically implemented by linking LABs together. For enhanced fitting, a long carry chain skips alternate LABs in a row. A carry chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB. For example, the last LE of the first LAB in a row carries to the first LE of the third LAB in the row. The carry chain does not cross the EAB at the middle of the row. For instance, in the EPF10K50 device, the carry chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB.

Figure 7 shows how an *n*-bit full adder can be implemented in *n* + 1 LEs with the carry chain. One portion of the LUT generates the sum of two bits using the input signals and the carry-in signal; the sum is routed to the output of the LE. The register can either be bypassed for simple adders or be used for an accumulator function. The carry chain logic generates the carry-out signal, which is routed directly to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it can be used as a general-purpose signal.

Figure 7. Carry Chain Operation (n-bit Full Adder)



LE Operating Modes

The FLEX 10K LE can operate in the following four modes:

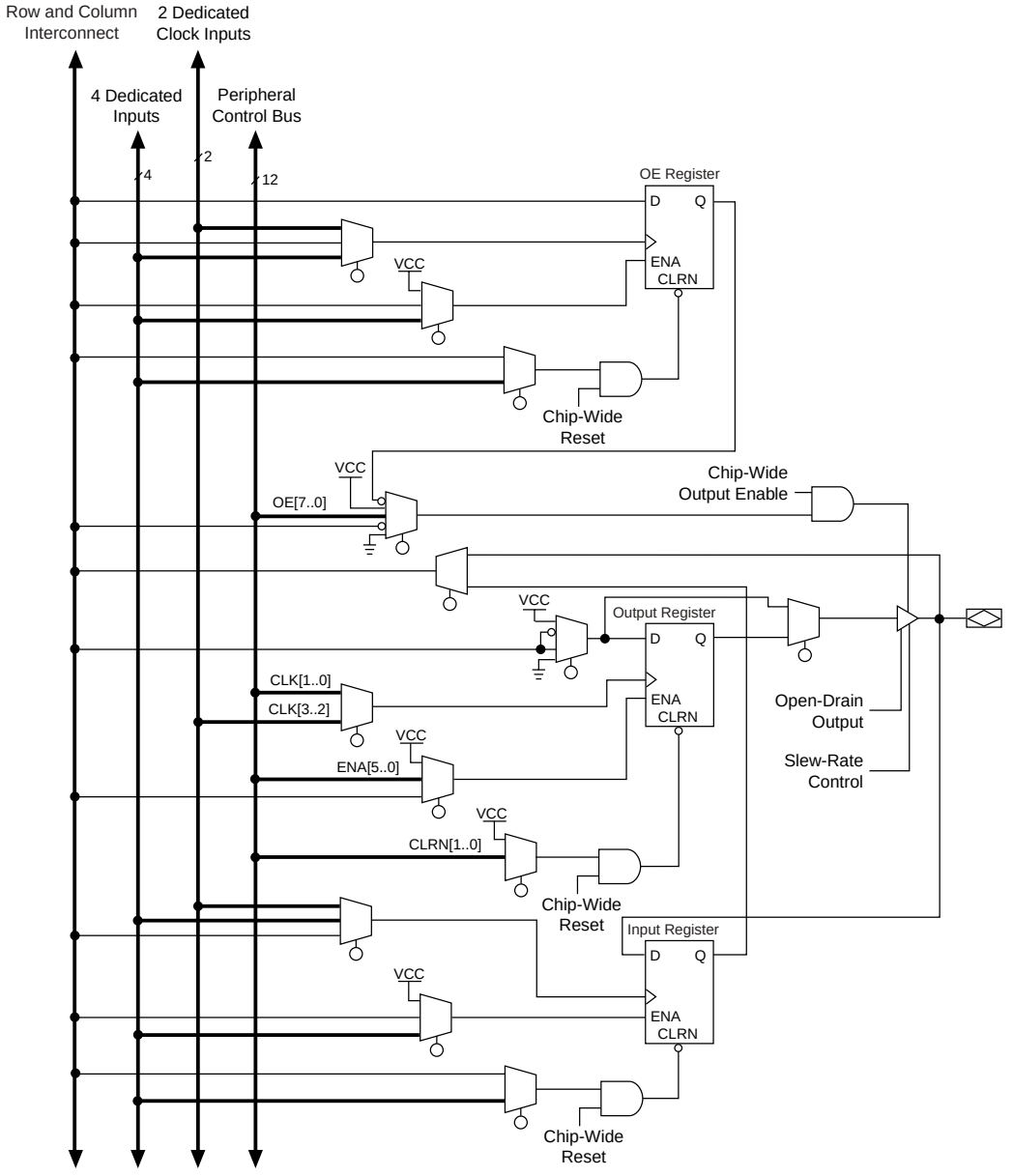
- Normal mode
- Arithmetic mode
- Up/down counter mode
- Clearable counter mode

Each of these modes uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. Three inputs to the LE provide clock, clear, and preset control for the register. The Altera software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions which use a specific LE operating mode for optimal performance.

The architecture provides a synchronous clock enable to the register in all four modes. The Altera software can set `DATA1` to enable the register synchronously, providing easy implementation of fully synchronous designs.

Figure 9 shows the LE operating modes.

Figure 13. Bidirectional I/O Registers



Slew-Rate Control

The output buffer in each IOE has an adjustable output slew rate that can be configured for low-noise or high-speed performance. A slower slew rate reduces system noise and adds a maximum delay of approximately 2.9 ns. The fast slew rate should be used for speed-critical outputs in systems that are adequately protected against noise. Designers can specify the slew rate on a pin-by-pin basis during design entry or assign a default slew rate to all pins on a device-wide basis. The slow slew rate setting affects only the falling edge of the output.

Open-Drain Output Option

FLEX 10K devices provide an optional open-drain (electrically equivalent to an open-collector) output for each I/O pin. This open-drain output enables the device to provide system-level control signals (e.g., interrupt and write enable signals) that can be asserted by any of several devices. It can also provide an additional wired-OR plane. Additionally, the Altera software can convert tri-state buffers with grounded data inputs to open-drain pins automatically.

Open-drain output pins on FLEX 10K devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a V_{IH} of 3.5 V. When the open-drain pin is active, it will drive low. When the pin is inactive, the trace will be pulled up to 5.0 V by the resistor. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The I_{OL} current specification should be considered when selecting a pull-up resistor.

Output pins on 5.0-V FLEX 10K devices with V_{CCIO} = 3.3 V or 5.0 V (with a pull-up resistor to the 5.0-V supply) can also drive 5.0-V CMOS input pins. In this case, the pull-up transistor will turn off when the pin voltage exceeds 3.3 V. Therefore, the pin does not have to be open-drain.

MultiVolt I/O Interface

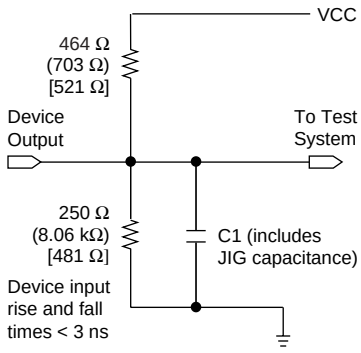
The FLEX 10K device architecture supports the MultiVolt I/O interface feature, which allows FLEX 10K devices to interface with systems of differing supply voltages. These devices have one set of V_{CC} pins for internal operation and input buffers (VCCINT) and another set for I/O output drivers (VCCIO).

Generic Testing

Each FLEX 10K device is functionally tested. Complete testing of each configurable SRAM bit and all logic functionality ensures 100% yield. AC test measurements for FLEX 10K devices are made under conditions equivalent to those shown in Figure 19. Multiple test patterns can be used to configure devices during all stages of the production flow.

Figure 19. FLEX 10K AC Test Conditions

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers without parentheses are for 5.0-V devices or outputs. Numbers in parentheses are for 3.3-V devices or outputs. Numbers in brackets are for 2.5-V devices or outputs.



Operating Conditions

Tables 17 through 21 provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for 5.0-V FLEX 10K devices.

Table 17. FLEX 10K 5.0-V Device Absolute Maximum Ratings *Note (1)*

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CC}	Supply voltage	With respect to ground *(2)*	–2.0	7.0	V
V_I	DC input voltage		–2.0	7.0	V
I_{OUT}	DC output current, per pin		–25	25	mA
T_{STG}	Storage temperature	No bias	–65	150	° C
T_{AMB}	Ambient temperature	Under bias	–65	135	° C
T_J	Junction temperature	Ceramic packages, under bias		150	° C
		PQFP, TQFP, RQFP, and BGA packages, under bias		135	° C

Table 19. FLEX 10K 5.0-V Device DC Operating Conditions *Notes (5), (6)*

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{IH}	High-level input voltage		2.0		V_{CCINT} + 0.5	V
V_{IL}	Low-level input voltage		–0.5		0.8	V
V_{OH}	5.0-V high-level TTL output voltage	I_{OH} = –4 mA DC, V_{CCIO} = 4.75 V *(7)*	2.4			V
	3.3-V high-level TTL output voltage	I_{OH} = –4 mA DC, V_{CCIO} = 3.00 V *(7)*	2.4			V
	3.3-V high-level CMOS output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 3.00 V *(7)*	V_{CCIO} – 0.2			V
V_{OL}	5.0-V low-level TTL output voltage	I_{OL} = 12 mA DC, V_{CCIO} = 4.75 V *(8)*			0.45	V
	3.3-V low-level TTL output voltage	I_{OL} = 12 mA DC, V_{CCIO} = 3.00 V *(8)*			0.45	V
	3.3-V low-level CMOS output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 3.00 V *(8)*			0.2	V
I_I	Input pin leakage current	V_I = V_{CC} or ground *(9)*	–10		10	µA
I_{OZ}	Tri-stated I/O pin leakage current	V_O = V_{CC} or ground *(9)*	–40		40	µA
I_{CC0}	V_{CC} supply current (standby)	V_I = ground, no load		0.5	10	mA

Table 20. 5.0-V Device Capacitance of EPF10K10, EPF10K20 & EPF10K30 Devices *Note (10)*

Symbol	Parameter	Conditions	Min	Max	Unit
C_{IN}	Input capacitance	V_{IN} = 0 V, f = 1.0 MHz		8	pF
C_{INCLK}	Input capacitance on dedicated clock pin	V_{IN} = 0 V, f = 1.0 MHz		12	pF
C_{OUT}	Output capacitance	V_{OUT} = 0 V, f = 1.0 MHz		8	pF

Table 21. 5.0-V Device Capacitance of EPF10K40, EPF10K50, EPF10K70 & EPF10K100 Devices *Note (10)*

Symbol	Parameter	Conditions	Min	Max	Unit
C_{IN}	Input capacitance	V_{IN} = 0 V, f = 1.0 MHz		10	pF
C_{INCLK}	Input capacitance on dedicated clock pin	V_{IN} = 0 V, f = 1.0 MHz		15	pF
C_{OUT}	Output capacitance	V_{OUT} = 0 V, f = 1.0 MHz		10	pF

Notes to tables:

(1) Microparameters are timing delays contributed by individual architectural elements. These parameters cannot be measured explicitly.

(2) Operating conditions: V_{CCIO} = 5.0 V ± 5% for commercial use in FLEX 10K devices.
V_{CCIO} = 5.0 V ± 10% for industrial use in FLEX 10K devices.
V_{CCIO} = 3.3 V ± 10% for commercial or industrial use in FLEX 10KA devices.

(3) Operating conditions: V_{CCIO} = 3.3 V ± 10% for commercial or industrial use in FLEX 10K devices.
V_{CCIO} = 2.5 V ± 0.2 V for commercial or industrial use in FLEX 10KA devices.

(4) Operating conditions: V_{CCIO} = 2.5 V, 3.3 V, or 5.0 V.

(5) Because the RAM in the EAB is self-timed, this parameter can be ignored when the WE signal is registered.

(6) EAB macroparameters are internal parameters that can simplify predicting the behavior of an EAB at its boundary; these parameters are calculated by summing selected microparameters.

(7) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.

(8) External reference timing parameters are factory-tested, worst-case values specified by Altera. A representative subset of signal paths is tested to approximate typical device applications.

(9) Contact Altera Applications for test circuit specifications and test conditions.

(10) These timing parameters are sample-tested only.

Figures 29 and 30 show the asynchronous and synchronous timing waveforms, respectively, for the EAB macroparameters in Table 34.

Figure 29. EAB Asynchronous Timing Waveforms

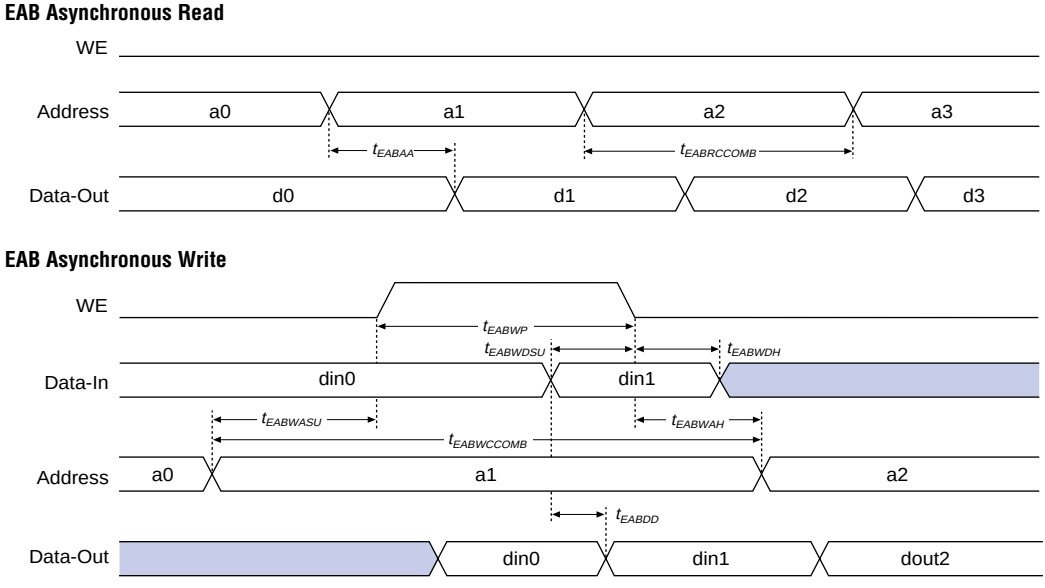


Table 41. EPF10K10 & EPF10K20 Device EAB Internal Microparameters *Note (1)*

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
$t_{EABDATA1}$		1.5		1.9	ns
$t_{EABDATA2}$		4.8		6.0	ns
t_{EABWE1}		1.0		1.2	ns
t_{EABWE2}		5.0		6.2	ns
t_{EABCLK}		1.0		2.2	ns
t_{EABCO}		0.5		0.6	ns
$t_{EABBYPASS}$		1.5		1.9	ns
t_{EABSU}	1.5		1.8		ns
t_{EABH}	2.0		2.5		ns
t_{AA}		8.7		10.7	ns
t_{WP}	5.8		7.2		ns
t_{WDSU}	1.6		2.0		ns
t_{WDH}	0.3		0.4		ns
t_{WASU}	0.5		0.6		ns
t_{WAH}	1.0		1.2		ns
t_{WO}		5.0		6.2	ns
t_{DD}		5.0		6.2	ns
t_{EABOUT}		0.5		0.6	ns
t_{EABCH}	4.0		4.0		ns
t_{EABCL}	5.8		7.2		ns

Table 54. EPF10K50 Device Interconnect Timing Microparameters *Note (1)*

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
$t_{DIN2IOE}$		8.4		10.2	ns
t_{DIN2LE}		3.6		4.8	ns
$t_{DIN2DATA}$		5.5		7.2	ns
$t_{DCLK2IOE}$		4.6		6.2	ns
$t_{DCLK2LE}$		3.6		4.8	ns
$t_{SAMELAB}$		0.3		0.3	ns
$t_{SAMEROW}$		3.3		3.7	ns
$t_{SAMECOLUMN}$		3.9		4.1	ns
$t_{DIFFROW}$		7.2		7.8	ns
$t_{TWOROWS}$		10.5		11.5	ns
$t_{LEPERIPH}$		7.5		8.2	ns
$t_{LABCARRY}$		0.4		0.6	ns
$t_{LABCASC}$		2.4		3.0	ns

Table 55. EPF10K30, EPF10K40 & EPF10K50 Device External Timing Parameters *Note (1)*

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
t_{DRR}		17.2		21.1	ns
t_{INSU} *(2)*, *(3)*	5.7		6.4		ns
t_{INH} *(3)*	0.0		0.0		ns
t_{OUTCO} *(3)*	2.0	8.8	2.0	11.2	ns

Table 56. EPF10K30, EPF10K40 & EPF10K50 Device External Bidirectional Timing Parameters *Note (1)*

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
$t_{INSUBIDIR}$	4.1		4.6		ns
$t_{INHBIDIR}$	0.0		0.0		ns
$t_{OUTCOBIDIR}$	2.0	8.8	2.0	11.2	ns
$t_{XZBIDIR}$		12.3		15.0	ns
$t_{ZXBIDIR}$		12.3		15.0	ns

Table 68. EPF10K100 Device Interconnect Timing Microparameters *Note (1)*

Symbol	-3DX Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		10.3		10.3		12.2	ns
t_{DIN2LE}		4.8		4.8		6.0	ns
$t_{DIN2DATA}$		7.3		7.3		11.0	ns
$t_{DCLK2IOE}$ without ClockLock or ClockBoost circuitry		6.2		6.2		7.7	ns
$t_{DCLK2IOE}$ with ClockLock or ClockBoost circuitry		2.3		–		–	ns
$t_{DCLK2LE}$ without ClockLock or ClockBoost circuitry		4.8		4.8		6.0	ns
$t_{DCLK2LE}$ with ClockLock or ClockBoost circuitry		2.3		–		–	ns
$t_{SAMELAB}$		0.4		0.4		0.5	ns
$t_{SAMEROW}$		4.9		4.9		5.5	ns
$t_{SAMECOLUMN}$		5.1		5.1		5.4	ns
$t_{DIFFROW}$		10.0		10.0		10.9	ns
$t_{TWOROWS}$		14.9		14.9		16.4	ns
$t_{LEPERIPH}$		6.9		6.9		8.1	ns
$t_{LABCARRY}$		0.9		0.9		1.1	ns
$t_{LABCASC}$		3.0		3.0		3.2	ns

Table 89. EPF10K10A Device Interconnect Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		4.2		5.0		6.5	ns
t_{DIN2LE}		2.2		2.6		3.4	ns
$t_{DIN2DATA}$		4.3		5.2		7.1	ns
$t_{DCLK2IOE}$		4.2		4.9		6.6	ns
$t_{DCLK2LE}$		2.2		2.6		3.4	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		2.2		2.4		2.9	ns
$t_{SAMECOLUMN}$		0.8		1.0		1.4	ns
$t_{DIFFROW}$		3.0		3.4		4.3	ns
$t_{TWOROWS}$		5.2		5.8		7.2	ns
$t_{LEPERIPH}$		1.8		2.2		2.8	ns
$t_{LABCARRY}$		0.5		0.5		0.7	ns
$t_{LABCASC}$		0.9		1.0		1.5	ns

Table 90. EPF10K10A External Reference Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		10.0		12.0		16.0	ns
t_{INSU} *(2)*, *(3)*	1.6		2.1		2.8		ns
t_{INH} *(3)*	0.0		0.0		0.0		ns
t_{OUTCO} *(3)*	2.0	5.8	2.0	6.9	2.0	9.2	ns

Table 91. EPF10K10A Device External Bidirectional Timing Parameters *Note (1)*

Symbol	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{INSUBIDIR}$	2.4		3.3		4.5		ns
$t_{INHBIDIR}$	0.0		0.0		0.0		ns
$t_{OUTCOBIDIR}$	2.0	5.8	2.0	6.9	2.0	9.2	ns
$t_{XZBIDIR}$		6.3		7.5		9.9	ns
$t_{ZXBIDIR}$		6.3		7.5		9.9	ns

Table 95. EPF10K30A Device EAB Internal Timing Macroparameters *Note (1)*							
Symbol	**-1 Speed Grade**		**-2 Speed Grade**		**-3 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	**Min**	**Max**	
t_{EABAA}		9.7		11.6		16.2	ns
$t_{EABRCCOMB}$	9.7		11.6		16.2		ns
$t_{EABRCREG}$	5.9		7.1		9.7		ns
t_{EABWP}	3.8		4.5		5.9		ns
$t_{EABWCCOMB}$	4.0		4.7		6.3		ns
$t_{EABWCREG}$	9.8		11.6		16.6		ns
t_{EABDD}		9.2		11.0		16.1	ns
$t_{EABDATACO}$		1.7		2.1		3.4	ns
$t_{EABDATASU}$	2.3		2.7		3.5		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	3.3		3.9		4.9		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	3.2		3.8		5.0		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	3.7		4.4		5.1		ns
t_{EABWAH}	0.0		0.0		0.0		ns
t_{EABWO}		6.1		7.3		11.3	ns

Table 100. EPF10K100A Device IOE Timing Microparameters *Note (1)*							
Symbol	**-1 Speed Grade**		**-2 Speed Grade**		**-3 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	**Min**	**Max**	
t_{IOD}		2.5		2.9		3.4	ns
t_{IOC}		0.3		0.3		0.4	ns
t_{IOCO}		0.2		0.2		0.3	ns
t_{IOCOMB}		0.5		0.6		0.7	ns
t_{IOSU}	1.3		1.7		1.8		ns
t_{IOH}	0.2		0.2		0.3		ns
t_{IOCLR}		1.0		1.2		1.4	ns
t_{OD1}		2.2		2.6		3.0	ns
t_{OD2}		4.5		5.3		6.1	ns
t_{OD3}		6.8		7.9		9.3	ns
t_{XZ}		2.7		3.1		3.7	ns
t_{ZX1}		2.7		3.1		3.7	ns
t_{ZX2}		5.0		5.8		6.8	ns
t_{ZX3}		7.3		8.4		10.0	ns
t_{INREG}		5.3		6.1		7.2	ns
t_{IOFD}		4.7		5.5		6.4	ns
t_{INCOMB}		4.7		5.5		6.4	ns

Table 113. ClockLock & ClockBoost Parameters (Part 2 of 2)

Symbol	Parameter	Min	Typ	Max	Unit
$f_{CLKDEV1}$	Input deviation from user specification in MAX+PLUS II (ClockBoost clock multiplication factor equals 1) *(1)*			±1	MHz
$f_{CLKDEV2}$	Input deviation from user specification in MAX+PLUS II (ClockBoost clock multiplication factor equals 2) *(1)*			±0.5	MHz
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)			100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock *(2)*			10	µs
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock *(3)*			1	ns
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock	40	50	60	%

Notes:

(1) To implement the ClockLock and ClockBoost circuitry with the MAX+PLUS II software, designers must specify the input frequency. The MAX+PLUS II software tunes the PLL in the ClockLock and ClockBoost circuitry to this frequency. The f_{CLKDEV} parameter specifies how much the incoming clock can differ from the specified frequency during device operation. Simulation does not reflect this parameter.

(2) During device configuration, the ClockLock and ClockBoost circuitry is configured before the rest of the device. If the incoming clock is supplied during configuration, the ClockLock and ClockBoost circuitry locks during configuration, because the t_{LOCK} value is less than the time required for configuration.

(3) The t_{JITTER} specification is measured under long-term observation.

Power Consumption

The supply power (P) for FLEX 10K devices can be calculated with the following equation:

$$P = P_{INT} + P_{IO} = (I_{CCSTANDBY} + I_{CCACTIVE}) \times V_{CC} + P_{IO}$$

Typical $I_{CCSTANDBY}$ values are shown as I_{CC0} in the FLEX 10K device DC operating conditions tables on pages 46, 49, and 52 of this data sheet. The $I_{CCACTIVE}$ value depends on the switching frequency and the application logic. This value is calculated based on the amount of current that each LE typically consumes. The P_{IO} value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

Compared to the rest of the device, the embedded array consumes a negligible amount of power. Therefore, the embedded array can be ignored when calculating supply current.

The $I_{CCACTIVE}$ value is calculated with the following equation:

$$I_{CCACTIVE} = K \times \mathbf{f_{MAX}} \times N \times \mathbf{tog_{LC}} \times \frac{\mu A}{MHz \times LE}$$

The parameters in this equation are shown below:

$\mathbf{f_{MAX}}$ = Maximum operating frequency in MHz
N = Total number of logic cells used in the device
$\mathbf{tog_{LC}}$ = Average percent of logic cells toggling at each clock (typically 12.5%)
K = Constant, shown in Tables 114 and 115

Table 114. FLEX 10K K Constant Values

Device	K Value
EPF10K10	82
EPF10K20	89
EPF10K30	88
EPF10K40	92
EPF10K50	95
EPF10K70	85
EPF10K100	88

Table 115. FLEX 10KA K Constant Values

Device	K Value
EPF10K10A	17
EPF10K30A	17
EPF10K50V	19
EPF10K100A	19
EPF10K130V	22
EPF10K250A	23

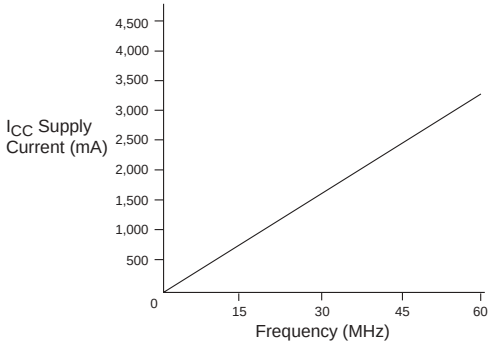
This calculation provides an I_{CC} estimate based on typical conditions with no output load. The actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

To better reflect actual designs, the power model (and the constant *K* in the power calculation equations) for continuous interconnect FLEX devices assumes that logic cells drive FastTrack Interconnect channels. In contrast, the power model of segmented FPGAs assumes that all logic cells drive only one short interconnect segment. This assumption may lead to inaccurate results, compared to measured power consumption for an actual design in a segmented interconnect FPGA.

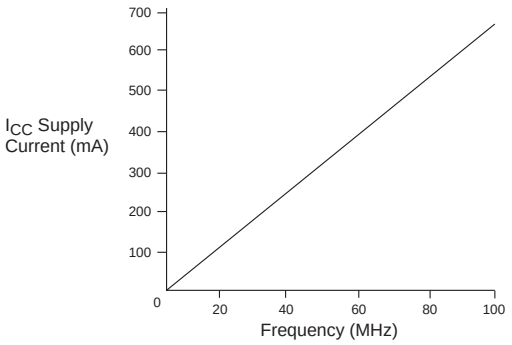
Figure 32 shows the relationship between the current and operating frequency of FLEX 10K devices.

Figure 32. $I_{CCACTIVE}$ vs. Operating Frequency (Part 2 of 3)

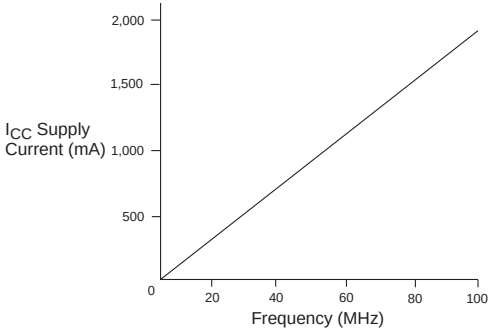
EPF10K100



EPF10K50V



EPF10K130V



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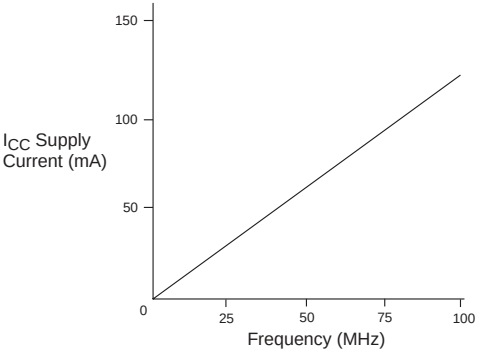
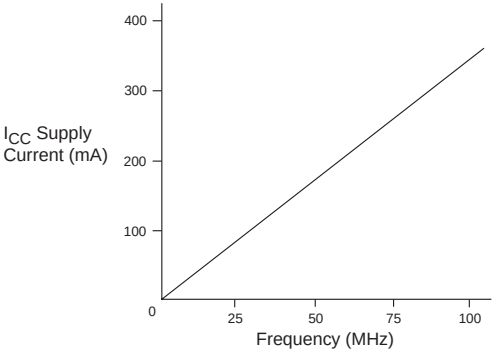


EPF10K30A



EPF10K100A