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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	72
Number of Logic Elements/Cells	576
Total RAM Bits	6144
Number of I/O	134
Number of Gates	31000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k10aqc208-3n

Table 4. FLEX 10K Package Options & I/O Pin Count *Note (1)*

Device	84-Pin PLCC	100-Pin TQFP	144-Pin TQFP	208-Pin PQFP RQFP	240-Pin PQFP RQFP
EPF10K10	59		102	134	
EPF10K10A		66	102	134	
EPF10K20			102	147	189
EPF10K30				147	189
EPF10K30A			102	147	189
EPF10K40				147	189
EPF10K50					189
EPF10K50V					189
EPF10K70					189
EPF10K100					
EPF10K100A					189
EPF10K130V					
EPF10K250A					

Table 5. FLEX 10K Package Options & I/O Pin Count (Continued) *Note (1)*

Device	503-Pin PGA	599-Pin PGA	256-Pin FineLine BGA	356-Pin BGA	484-Pin FineLine BGA	600-Pin BGA	403-Pin PGA
EPF10K10							
EPF10K10A			150		150 *(2)*		
EPF10K20							
EPF10K30				246			
EPF10K30A			191	246	246		
EPF10K40							
EPF10K50				274			310
EPF10K50V				274			
EPF10K70	358						
EPF10K100	406						
EPF10K100A				274	369	406	
EPF10K130V		470				470	
EPF10K250A		470				470	

Notes to tables:
(1) FLEX 10K and FLEX 10KA device package types include plastic J-lead chip carrier (PLCC), thin quad flat pack (TQFP), plastic quad flat pack (PQFP), power quad flat pack (RQFP), ball-grid array (BGA), pin-grid array (PGA), and FineLine BGA™ packages.
(2) This option is supported with a 256-pin FineLine BGA package. By using SameFrame pin migration, all FineLine BGA packages are pin compatible. For example, a board can be designed to support both 256-pin and 484-pin FineLine BGA packages. The Altera software automatically avoids conflicting pins when future migration is set.

General Description

Altera's FLEX 10K devices are the industry's first embedded PLDs. Based on reconfigurable CMOS SRAM elements, the Flexible Logic Element MatriX (FLEX) architecture incorporates all features necessary to implement common gate array megafunctions. With up to 250,000 gates, the FLEX 10K family provides the density, speed, and features to integrate entire systems, including multiple 32-bit buses, into a single device.

FLEX 10K devices are reconfigurable, which allows 100% testing prior to shipment. As a result, the designer is not required to generate test vectors for fault coverage purposes. Additionally, the designer does not need to manage inventories of different ASIC designs; FLEX 10K devices can be configured on the board for the specific functionality required.

Table 6 shows FLEX 10K performance for some common designs. All performance values were obtained with Synopsys DesignWare or LPM functions. No special design technique was required to implement the applications; the designer simply inferred or instantiated a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Table 6. FLEX 10K & FLEX 10KA Performance

Application	Resources Used		Performance				Units
	LEs	EABs	-1 Speed Grade	-2 Speed Grade	-3 Speed Grade	-4 Speed Grade	
16-bit loadable counter *(1)*	16	0	204	166	125	95	MHz
16-bit accumulator *(1)*	16	0	204	166	125	95	MHz
16-to-1 multiplexer *(2)*	10	0	4.2	5.8	6.0	7.0	ns
256 × 8 RAM read cycle speed *(3)*	0	1	172	145	108	84	MHz
256 × 8 RAM write cycle speed *(3)*	0	1	106	89	68	63	MHz

Notes:
(1) The speed grade of this application is limited because of clock high and low specifications.
(2) This application uses combinatorial inputs and outputs.
(3) This application uses registered inputs and outputs.

The logic array consists of logic array blocks (LABs). Each LAB contains eight LEs and a local interconnect. An LE consists of a 4-input look-up table (LUT), a programmable flipflop, and dedicated signal paths for carry and cascade functions. The eight LEs can be used to create medium-sized blocks of logic—8-bit counters, address decoders, or state machines—or combined across LABs to create larger logic blocks. Each LAB represents about 96 usable gates of logic.

Signal interconnections within FLEX 10K devices and to and from device pins are provided by the FastTrack Interconnect, a series of fast, continuous row and column channels that run the entire length and width of the device.

Each I/O pin is fed by an I/O element (IOE) located at the end of each row and column of the FastTrack Interconnect. Each IOE contains a bidirectional I/O buffer and a flipflop that can be used as either an output or input register to feed input, output, or bidirectional signals. When used with a dedicated clock pin, these registers provide exceptional performance. As inputs, they provide setup times as low as 1.6 ns and hold times of 0 ns; as outputs, these registers provide clock-to-output times as low as 5.3 ns. IOEs provide a variety of features, such as JTAG BST support, slew-rate control, tri-state buffers, and open-drain outputs.

Figure 1 shows a block diagram of the FLEX 10K architecture. Each group of LEs is combined into an LAB; LABs are arranged into rows and columns. Each row also contains a single EAB. The LABs and EABs are interconnected by the FastTrack Interconnect. IOEs are located at the end of each row and column of the FastTrack Interconnect.

Up/Down Counter Mode

The up/down counter mode offers counter enable, clock enable, synchronous up/down control, and data loading options. These control signals are generated by the data inputs from the LAB local interconnect, the carry-in signal, and output feedback from the programmable register. The Up/down counter mode uses 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading. Data can also be loaded asynchronously with the clear and preset register control signals, without using the LUT resources.

Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Clearable counter mode uses 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is ANDed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-stating without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

In addition to the six clear and preset modes, FLEX 10K devices provide a chip-wide reset pin that can reset all registers in the device. Use of this feature is set during design entry. In any of the clear and preset modes, the chip-wide reset overrides all other signals. Registers with asynchronous presets may be preset when the chip-wide reset is asserted. Inversion can be used to implement the asynchronous preset. Figure 10 shows examples of how to enter a section of a design for the desired functionality.

Signals on the peripheral control bus can also drive the four global signals, referred to as GLOBAL0 through GLOBAL3 in Tables 8 and 9. The internally generated signal can drive the global signal, providing the same low-skew, low-delay characteristics for an internally generated signal as for a signal driven by an input. This feature is ideal for internally generated clear or clock signals with high fan-out. When a global signal is driven by internal logic, the dedicated input pin that drives that global signal cannot be used. The dedicated input pin should be driven to a known logic state (such as ground) and not be allowed to float.

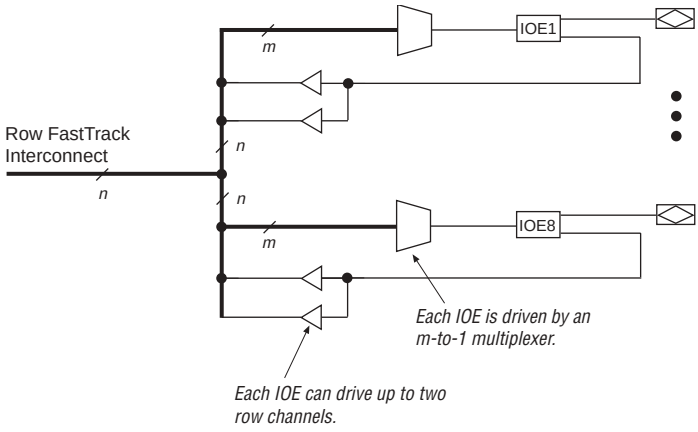
When the chip-wide output enable pin is held low, it will tri-state all pins on the device. This option can be set in the Global Project Device Options menu. Additionally, the registers in the IOE can be reset by holding the chip-wide reset pin low.

Row-to-IOE Connections

When an IOE is used as an input signal, it can drive two separate row channels. The signal is accessible by all LEs within that row. When an IOE is used as an output, the signal is driven by a multiplexer that selects a signal from the row channels. Up to eight IOEs connect to each side of each row channel. See Figure 14.

Figure 14. FLEX 10K Row-to-IOE Connections

The values for m and n are provided in Table 10.

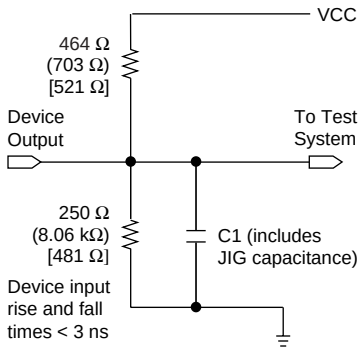


Generic Testing

Each FLEX 10K device is functionally tested. Complete testing of each configurable SRAM bit and all logic functionality ensures 100% yield. AC test measurements for FLEX 10K devices are made under conditions equivalent to those shown in Figure 19. Multiple test patterns can be used to configure devices during all stages of the production flow.

Figure 19. FLEX 10K AC Test Conditions

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers without parentheses are for 5.0-V devices or outputs. Numbers in parentheses are for 3.3-V devices or outputs. Numbers in brackets are for 2.5-V devices or outputs.



Operating Conditions

Tables 17 through 21 provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for 5.0-V FLEX 10K devices.

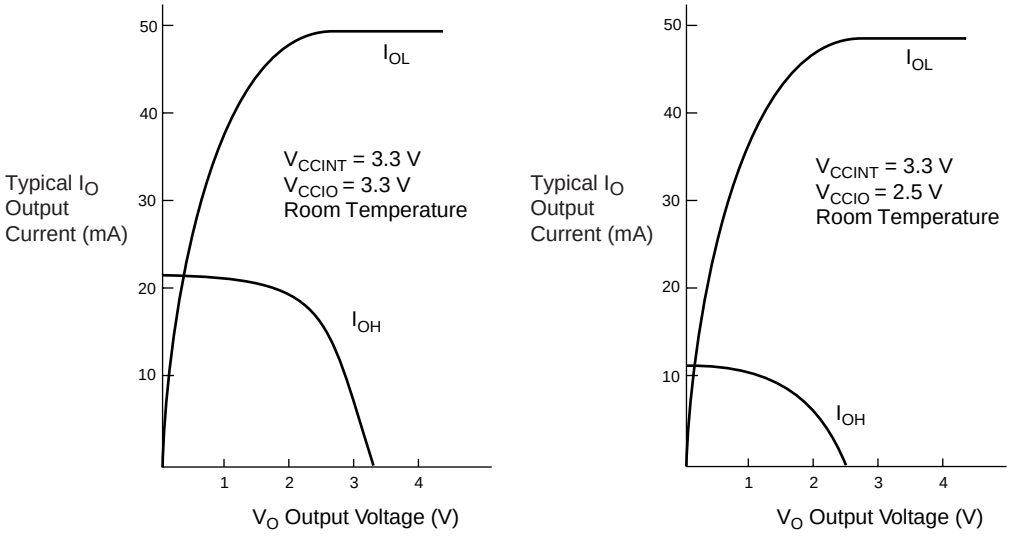
Table 17. FLEX 10K 5.0-V Device Absolute Maximum Ratings *Note (1)*

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CC}	Supply voltage	With respect to ground *(2)*	–2.0	7.0	V
V_I	DC input voltage		–2.0	7.0	V
I_{OUT}	DC output current, per pin		–25	25	mA
T_{STG}	Storage temperature	No bias	–65	150	° C
T_{AMB}	Ambient temperature	Under bias	–65	135	° C
T_J	Junction temperature	Ceramic packages, under bias		150	° C
		PQFP, TQFP, RQFP, and BGA packages, under bias		135	° C

Table 28. FLEX 10KA 3.3-V Device DC Operating Conditions *Notes (6), (7)*

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{IH}	High-level input voltage		1.7 or $0.5 \times V_{CCINT}$, whichever is lower		5.75	V
V_{IL}	Low-level input voltage		–0.5		$0.3 \times V_{CCINT}$	V
V_{OH}	3.3-V high-level TTL output voltage	I_{OH} = –11 mA DC, V_{CCIO} = 3.00 V *(8)*	2.4			V
	3.3-V high-level CMOS output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 3.00 V *(8)*	V_{CCIO} – 0.2			V
	3.3-V high-level PCI output voltage	I_{OH} = –0.5 mA DC, V_{CCIO} = 3.00 to 3.60 V *(8)*	$0.9 \times V_{CCIO}$			V
	2.5-V high-level output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 2.30 V *(8)*	2.1			V
		I_{OH} = –1 mA DC, V_{CCIO} = 2.30 V *(8)*	2.0			V
		I_{OH} = –2 mA DC, V_{CCIO} = 2.30 V *(8)*	1.7			V
V_{OL}	3.3-V low-level TTL output voltage	I_{OL} = 9 mA DC, V_{CCIO} = 3.00 V *(9)*			0.45	V
	3.3-V low-level CMOS output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 3.00 V *(9)*			0.2	V
	3.3-V low-level PCI output voltage	I_{OL} = 1.5 mA DC, V_{CCIO} = 3.00 to 3.60 V *(9)*			$0.1 \times V_{CCIO}$	V
	2.5-V low-level output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 2.30 V *(9)*			0.2	V
		I_{OL} = 1 mA DC, V_{CCIO} = 2.30 V *(9)*			0.4	V
		I_{OL} = 2 mA DC, V_{CCIO} = 2.30 V *(9)*			0.7	V
I_I	Input pin leakage current	V_I = 5.3 V to –0.3 V *(10)*	–10		10	µA
I_{OZ}	Tri-stated I/O pin leakage current	V_O = 5.3 V to –0.3 V *(10)*	–10		10	µA
I_{CC0}	V_{CC} supply current (standby)	V_I = ground, no load		0.3	10	mA
		V_I = ground, no load *(11)*		10		mA

Figure 23. Output Drive Characteristics for EPF10K250A Device



Timing Model

The continuous, high-performance FastTrack Interconnect routing resources ensure predictable performance and accurate simulation and timing analysis. This predictable performance contrasts with that of FPGAs, which use a segmented connection scheme and therefore have unpredictable performance.

Device performance can be estimated by following the signal path from a source, through the interconnect, to the destination. For example, the registered performance between two LEs on the same row can be calculated by adding the following parameters:

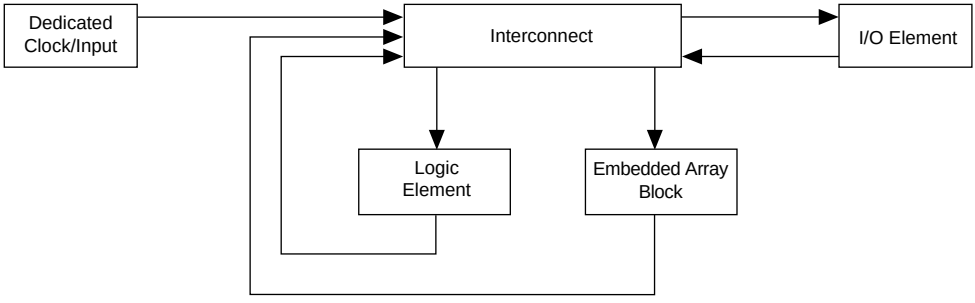
- LE register clock-to-output delay (t_{CO})
- Interconnect delay ($t_{SAMEROW}$)
- LE look-up table delay (t_{LUT})
- LE register setup time (t_{SU})

The routing delay depends on the placement of the source and destination LEs. A more complex registered path may involve multiple combinatorial LEs between the source and destination LEs.

Timing simulation and delay prediction are available with the MAX+PLUS II Simulator and Timing Analyzer, or with industry-standard EDA tools. The Simulator offers both pre-synthesis functional simulation to evaluate logic design accuracy and post-synthesis timing simulation with 0.1-ns resolution. The Timing Analyzer provides point-to-point timing delay information, setup and hold time analysis, and device-wide performance analysis.

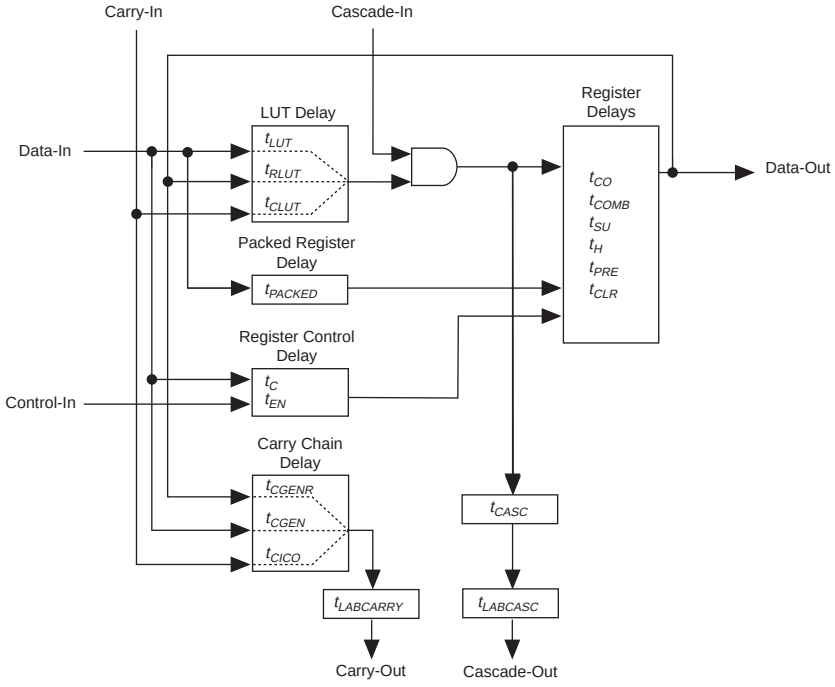
Figure 24 shows the overall timing model, which maps the possible paths to and from the various elements of the FLEX 10K device.

Figure 24. FLEX 10K Device Timing Model



Figures 25 through 27 show the delays that correspond to various paths and functions within the LE, IOE, and EAB timing models.

Figure 25. FLEX 10K Device LE Timing Model



Notes to tables:

(1) Microparameters are timing delays contributed by individual architectural elements. These parameters cannot be measured explicitly.

(2) Operating conditions: V_{CCIO} = 5.0 V ± 5% for commercial use in FLEX 10K devices.
V_{CCIO} = 5.0 V ± 10% for industrial use in FLEX 10K devices.
V_{CCIO} = 3.3 V ± 10% for commercial or industrial use in FLEX 10KA devices.

(3) Operating conditions: V_{CCIO} = 3.3 V ± 10% for commercial or industrial use in FLEX 10K devices.
V_{CCIO} = 2.5 V ± 0.2 V for commercial or industrial use in FLEX 10KA devices.

(4) Operating conditions: V_{CCIO} = 2.5 V, 3.3 V, or 5.0 V.

(5) Because the RAM in the EAB is self-timed, this parameter can be ignored when the WE signal is registered.

(6) EAB macroparameters are internal parameters that can simplify predicting the behavior of an EAB at its boundary; these parameters are calculated by summing selected microparameters.

(7) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.

(8) External reference timing parameters are factory-tested, worst-case values specified by Altera. A representative subset of signal paths is tested to approximate typical device applications.

(9) Contact Altera Applications for test circuit specifications and test conditions.

(10) These timing parameters are sample-tested only.

Figures 29 and 30 show the asynchronous and synchronous timing waveforms, respectively, for the EAB macroparameters in Table 34.

Figure 29. EAB Asynchronous Timing Waveforms

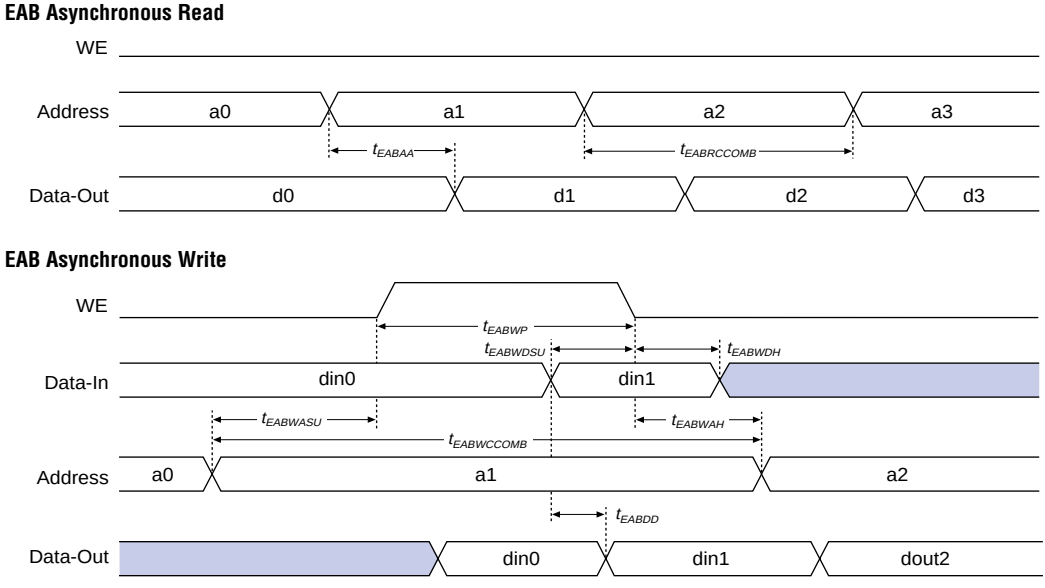


Table 43. EPF10K10 Device Interconnect Timing Microparameters *Note (1)*					
Symbol	**-3 Speed Grade**		**-4 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	
$t_{DIN2IOE}$		4.8		6.2	ns
t_{DIN2LE}		2.6		3.8	ns
$t_{DIN2DATA}$		4.3		5.2	ns
$t_{DCLK2IOE}$		3.4		4.0	ns
$t_{DCLK2LE}$		2.6		3.8	ns
$t_{SAMELAB}$		0.6		0.6	ns
$t_{SAMEROW}$		3.6		3.8	ns
$t_{SAMECOLUMN}$		0.9		1.1	ns
$t_{DIFFROW}$		4.5		4.9	ns
$t_{TWOROWS}$		8.1		8.7	ns
$t_{LEPERIPH}$		3.3		3.9	ns
$t_{LABCARRY}$		0.5		0.8	ns
$t_{LABCASC}$		2.7		3.0	ns

Table 44. EPF10K20 Device Interconnect Timing Microparameters *Note (1)*					
Symbol	**-3 Speed Grade**		**-4 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	
$t_{DIN2IOE}$		5.2		6.6	ns
t_{DIN2LE}		2.6		3.8	ns
$t_{DIN2DATA}$		4.3		5.2	ns
$t_{DCLK2IOE}$		4.3		4.0	ns
$t_{DCLK2LE}$		2.6		3.8	ns
$t_{SAMELAB}$		0.6		0.6	ns
$t_{SAMEROW}$		3.7		3.9	ns
$t_{SAMECOLUMN}$		1.4		1.6	ns
$t_{DIFFROW}$		5.1		5.5	ns
$t_{TWOROWS}$		8.8		9.4	ns
$t_{LEPERIPH}$		4.7		5.6	ns
$t_{LABCARRY}$		0.5		0.8	ns
$t_{LABCASC}$		2.7		3.0	ns

Table 51. EPF10K30, EPF10K40 & EPF10K50 Device EAB Internal Timing Macroparameters *Note (1)*					
Symbol	**-3 Speed Grade**		**-4 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	
t_{EABAA}		13.7		17.0	ns
$t_{EABRCCOMB}$	13.7		17.0		ns
$t_{EABRCREG}$	9.7		11.9		ns
t_{EABWP}	5.8		7.2		ns
$t_{EABWCCOMB}$	7.3		9.0		ns
$t_{EABWCREG}$	13.0		16.0		ns
t_{EABDD}		10.0		12.5	ns
$t_{EABDATACO}$		2.0		3.4	ns
$t_{EABDATASU}$	5.3		5.6		ns
$t_{EABDATAH}$	0.0		0.0		ns
$t_{EABWESU}$	5.5		5.8		ns
t_{EABWEH}	0.0		0.0		ns
$t_{EABWDSU}$	5.5		5.8		ns
t_{EABWDH}	0.0		0.0		ns
$t_{EABWASU}$	2.1		2.7		ns
t_{EABWAH}	0.0		0.0		ns
t_{EABWO}		9.5		11.8	ns

Table 58. EPF10K70 Device IOE Timing Microparameters *Note (1)*

Symbol	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		0.0		0.0		0.0	ns
t_{IOC}		0.4		0.5		0.7	ns
t_{IOCO}		0.4		0.4		0.9	ns
t_{IOCOMB}		0.0		0.0		0.0	ns
t_{IOSU}	4.5		5.0		6.2		ns
t_{IOH}	0.4		0.5		0.7		ns
t_{IOCLR}		0.6		0.7		1.6	ns
t_{OD1}		3.6		4.0		5.0	ns
t_{OD2}		5.6		6.3		7.3	ns
t_{OD3}		6.9		7.7		8.7	ns
t_{XZ}		5.5		6.2		6.8	ns
t_{ZX1}		5.5		6.2		6.8	ns
t_{ZX2}		7.5		8.5		9.1	ns
t_{ZX3}		8.8		9.9		10.5	ns
t_{INREG}		8.0		9.0		10.2	ns
t_{IOFD}		7.2		8.1		10.3	ns
t_{INCOMB}		7.2		8.1		10.3	ns

Table 61. EPF10K70 Device Interconnect Timing Microparameters *Note (1)*

Symbol	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		6.6		7.3		8.8	ns
t_{DIN2LE}		4.2		4.8		6.0	ns
$t_{DIN2DATA}$		6.5		7.1		10.8	ns
$t_{DCLK2IOE}$		5.5		6.2		7.7	ns
$t_{DCLK2LE}$		4.2		4.8		6.0	ns
$t_{SAMELAB}$		0.4		0.4		0.5	ns
$t_{SAMEROW}$		4.8		4.9		5.5	ns
$t_{SAMECOLUMN}$		3.3		3.4		3.7	ns
$t_{DIFFROW}$		8.1		8.3		9.2	ns
$t_{TWOROWS}$		12.9		13.2		14.7	ns
$t_{LEPERIPH}$		5.5		5.7		6.5	ns
$t_{LABCARRY}$		0.8		0.9		1.1	ns
$t_{LABCASC}$		2.7		3.0		3.2	ns

Table 62. EPF10K70 Device External Timing Parameters *Note (1)*

Symbol	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		17.2		19.1		24.2	ns
t_{INSU} *(2)*, *(3)*	6.6		7.3		8.0		ns
t_{INH} *(3)*	0.0		0.0		0.0		ns
t_{OUTCO} *(3)*	2.0	9.9	2.0	11.1	2.0	14.3	ns

Table 63. EPF10K70 Device External Bidirectional Timing Parameters *Note (1)*

Symbol	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{INSUBIDIR}$	7.4		8.1		10.4		ns
$t_{INHBIDIR}$	0.0		0.0		0.0		ns
$t_{OUTCOBIDIR}$	2.0	9.9	2.0	11.1	2.0	14.3	ns
$t_{XZBIDIR}$		13.7		15.4		18.5	ns
$t_{ZXBIDIR}$		13.7		15.4		18.5	ns

Table 67. EPF10K100 Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-3DX Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		13.7		13.7		17.0	ns
$t_{EABRCCOMB}$	13.7		13.7		17.0		ns
$t_{EABRCREG}$	9.7		9.7		11.9		ns
t_{EABWP}	5.8		5.8		7.2		ns
$t_{EABWCCOMB}$	7.3		7.3		9.0		ns
$t_{EABWCREG}$	13.0		13.0		16.0		ns
t_{EABDD}		10.0		10.0		12.5	ns
$t_{EABDATACO}$		2.0		2.0		3.4	ns
$t_{EABDATASU}$	5.3		5.3		5.6		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	5.5		5.5		5.8		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	5.5		5.5		5.8		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	2.1		2.1		2.7		ns
t_{EABWAH}	0.0		0.0		0.0		ns
t_{EABWO}		9.5		9.5		11.8	ns

Table 69. EPF10K100 Device External Timing Parameters *Note (1)*

Symbol	-3DX Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		19.1		19.1		24.2	ns
t_{INSU} *(2)*, *(3)*, *(4)*	7.8		7.8		8.5		ns
t_{OUTCO} *(3)*, *(4)*	2.0	11.1	2.0	11.1	2.0	14.3	ns
t_{INH} *(3)*	0.0		0.0		0.0		ns
t_{INSU} *(2)*, *(3)*, *(5)*	6.2		–		–		ns
t_{OUTCO} *(3)*, *(5)*	2.0	6.7		–		–	ns

Table 70. EPF10K100 Device External Bidirectional Timing Parameters *Note (1)*

Symbol	-3DX Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{INSUBIDIR}$ *(4)*	8.1		8.1		10.4		ns
$t_{INHBIDIR}$ *(4)*	0.0		0.0		0.0		ns
$t_{OUTCOBIDIR}$ *(4)*	2.0	11.1	2.0	11.1	2.0	14.3	ns
$t_{XZBIDIR}$ *(4)*		15.3		15.3		18.4	ns
$t_{ZXBIDIR}$ *(4)*		15.3		15.3		18.4	ns
$t_{INSUBIDIR}$ *(5)*	9.1		–		–		ns
$t_{INHBIDIR}$ *(5)*	0.0		–		–		ns
$t_{OUTCOBIDIR}$ *(5)*	2.0	7.2	–	–	–	–	ns
$t_{XZBIDIR}$ *(5)*		14.3		–		–	ns
$t_{ZXBIDIR}$ *(5)*		14.3		–		–	ns

Notes to tables:

(1) All timing parameters are described in Tables 32 through 38 in this data sheet.
(2) Using an LE to register the signal may provide a lower setup time.
(3) This parameter is specified by characterization.
(4) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
(5) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Notes to tables:
(1) All timing parameters are described in Tables 32 through 38 in this data sheet.
(2) Using an LE to register the signal may provide a lower setup time.
(3) This parameter is specified by characterization.

Tables 85 through 91 show EPF10K10A device internal and external timing parameters.

Table 85. EPF10K10A Device LE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.9		1.2		1.6	ns
t_{CLUT}		1.2		1.4		1.9	ns
t_{RLUT}		1.9		2.3		3.0	ns
t_{PACKED}		0.6		0.7		0.9	ns
t_{EN}		0.5		0.6		0.8	ns
t_{CICO}		02		0.3		0.4	ns
t_{CGEN}		0.7		0.9		1.1	ns
t_{CGENR}		0.7		0.9		1.1	ns
t_{CASC}		1.0		1.2		1.7	ns
t_{C}		1.2		1.4		1.9	ns
t_{CO}		0.5		0.6		0.8	ns
t_{COMB}		0.5		0.6		0.8	ns
t_{SU}	1.1		1.3		1.7		ns
t_{H}	0.6		0.7		0.9		ns
t_{PRE}		0.5		0.6		0.9	ns
t_{CLR}		0.5		0.6		0.9	ns
t_{CH}	3.0		3.5		4.0		ns
t_{CL}	3.0		3.5		4.0		ns

Table 86. EPF10K10A Device IOE Timing Microparameters *Note (1)* **(Part 1 of 2)**

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
		1.3		1.5		2.0	ns
t_{IOC}		0.2		0.3		0.3	ns
t_{IOCO}		0.2		0.3		0.4	ns
t_{IOCOMB}		0.6		0.7		0.9	ns
t_{IOSU}	0.8		1.0		1.3		ns

Notes to tables:
(1) All timing parameters are described in Tables 32 through 37 in this data sheet.
(2) Using an LE to register the signal may provide a lower setup time.
(3) This parameter is specified by characterization.

ClockLock & ClockBoost Timing Parameters

For the ClockLock and ClockBoost circuitry to function properly, the incoming clock must meet certain requirements. If these specifications are not met, the circuitry may not lock onto the incoming clock, which generates an erroneous clock within the device. The clock generated by the ClockLock and ClockBoost circuitry must also meet certain specifications. If the incoming clock meets these requirements during configuration, the ClockLock and ClockBoost circuitry will lock onto the clock during configuration. The circuit will be ready for use immediately after configuration. Figure 31 illustrates the incoming and generated clock specifications.

Figure 31. Specifications for the Incoming & Generated Clocks

The t_I parameter refers to the nominal input clock period; the t_O parameter refers to the nominal output clock period.

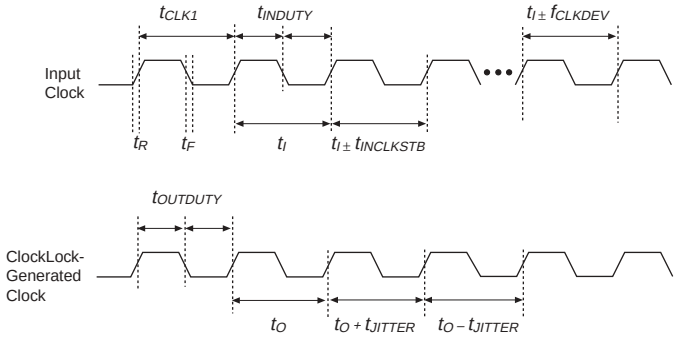


Table 113 summarizes the ClockLock and ClockBoost parameters.

Table 113. ClockLock & ClockBoost Parameters (Part 1 of 2)

Symbol	Parameter	Min	Typ	Max	Unit
t_R	Input rise time			2	ns
t_F	Input fall time			2	ns
t_{INDUTY}	Input duty cycle	45		55	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)	30		80	MHz
t_{CLK1}	Input clock period (ClockBoost clock multiplication factor equals 1)	12.5		33.3	ns
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)	16		50	MHz
t_{CLK2}	Input clock period (ClockBoost clock multiplication factor equals 2)	20		62.5	ns