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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Active
Number of LABs/CLBs	72
Number of Logic Elements/Cells	576
Total RAM Bits	-
Number of I/O	102
Number of Gates	-
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=epf10k10atc144-2n

Figure 1. FLEX 10K Device Block Diagram



FLEX 10K devices provide six dedicated inputs that drive the flipflops' control inputs to ensure the efficient distribution of high-speed, low-skew (less than 1.5 ns) control signals. These signals use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect. Four of the dedicated inputs drive four global signals. These four global signals can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device.

Embedded Array Block

The EAB is a flexible block of RAM with registers on the input and output ports, and is used to implement common gate array megafunctions. The EAB is also suitable for functions such as multipliers, vector scalars, and error correction circuits, because it is large and flexible. These functions can be combined in applications such as digital filters and microcontrollers.

Larger blocks of RAM are created by combining multiple EABs. For example, two 256×8 RAM blocks can be combined to form a 256×16 RAM block; two 512×4 blocks of RAM can be combined to form a 512×8 RAM block. See [Figure 3](#).

Figure 3. Examples of Combining EABs



If necessary, all EABs in a device can be cascaded to form a single RAM block. EABs can be cascaded to form RAM blocks of up to 2,048 words without impacting timing. Altera's software automatically combines EABs to meet a designer's RAM specifications.

EABs provide flexible options for driving and controlling clock signals. Different clocks can be used for the EAB inputs and outputs. Registers can be independently inserted on the data input, EAB output, or the address and WE inputs. The global signals and the EAB local interconnect can drive the WE signal. The global signals, dedicated clock pins, and EAB local interconnect can drive the EAB clock signals. Because the LEs drive the EAB local interconnect, the LEs can control the WE signal or the EAB clock signals.

Each EAB is fed by a row interconnect and can drive out to row and column interconnects. Each EAB output can drive up to two row channels and up to two column channels; the unused row channel can be driven by other LEs. This feature increases the routing resources available for EAB outputs. See [Figure 4](#).

The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock, clear, and preset control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the output of the LE.

The LE has two outputs that drive the interconnect; one drives the local interconnect and the other drives either the row or column FastTrack Interconnect. The two outputs can be controlled independently. For example, the LUT can drive one output while the register drives the other output. This feature, called register packing, can improve LE utilization because the register and the LUT can be used for unrelated functions.

The FLEX 10K architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. The carry chain supports high-speed counters and adders; the cascade chain implements wide-input functions with minimum delay. Carry and cascade chains connect all LEs in an LAB and all LABs in the same row. Intensive use of carry and cascade chains can reduce routing flexibility. Therefore, the use of these chains should be limited to speed-critical portions of a design.

Carry Chain

The carry chain provides a very fast (as low as 0.2 ns) carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the FLEX 10K architecture to implement high-speed counters, adders, and comparators of arbitrary width efficiently. Carry chain logic can be created automatically by the Compiler during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of carry chains.

Carry chains longer than eight LEs are automatically implemented by linking LABs together. For enhanced fitting, a long carry chain skips alternate LABs in a row. A carry chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB. For example, the last LE of the first LAB in a row carries to the first LE of the third LAB in the row. The carry chain does not cross the EAB at the middle of the row. For instance, in the EPF10K50 device, the carry chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB.

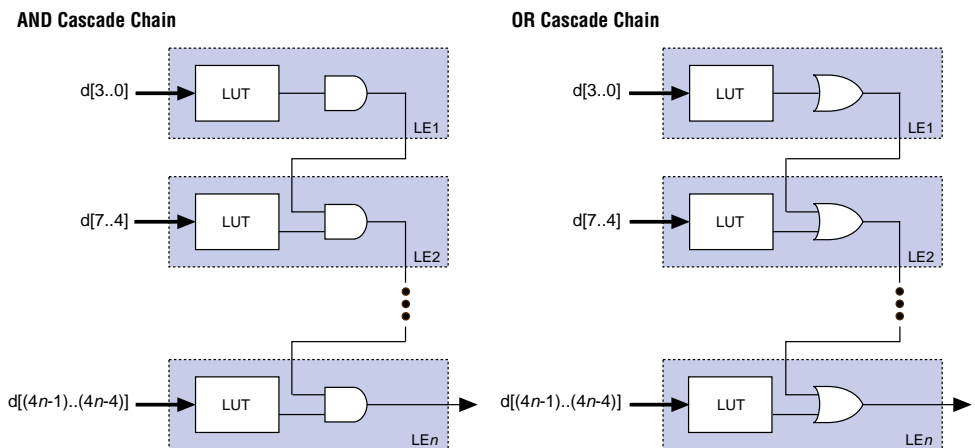
Cascade Chain

With the cascade chain, the FLEX 10K architecture can implement functions that have a very wide fan-in. Adjacent LUTs can be used to compute portions of the function in parallel; the cascade chain serially connects the intermediate values. The cascade chain can use a logical AND or logical OR (via De Morgan's inversion) to connect the outputs of adjacent LEs. Each additional LE provides four more inputs to the effective width of a function, with a delay as low as 0.7 ns per LE. Cascade chain logic can be created automatically by the Compiler during design processing, or manually by the designer during design entry.

Cascade chains longer than eight bits are implemented automatically by linking several LABs together. For easier routing, a long cascade chain skips every other LAB in a row. A cascade chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB (e.g., the last LE of the first LAB in a row cascades to the first LE of the third LAB). The cascade chain does not cross the center of the row (e.g., in the EPF10K50 device, the cascade chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB). This break is due to the EAB's placement in the middle of the row.

Figure 8 shows how the cascade function can connect adjacent LEs to form functions with a wide fan-in. These examples show functions of $4n$ variables implemented with n LEs. The LE delay is as low as 1.6 ns; the cascade chain delay is as low as 0.7 ns. With the cascade chain, 3.7 ns is needed to decode a 16-bit address.

Figure 8. Cascade Chain Operation



Normal Mode

The normal mode is suitable for general logic applications and wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a four-input LUT. The Compiler automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. Either the register or the LUT can be used to drive both the local interconnect and the FastTrack Interconnect at the same time.

The LUT and the register in the LE can be used independently; this feature is known as register packing. To support register packing, the LE has two outputs; one drives the local interconnect and the other drives the FastTrack Interconnect. The DATA4 signal can drive the register directly, allowing the LUT to compute a function that is independent of the registered signal; a three-input function can be computed in the LUT, and a fourth independent signal can be registered. Alternatively, a four-input function can be generated, and one of the inputs to this function can be used to drive the register. The register in a packed LE can still use the clock enable, clear, and preset signals in the LE. In a packed LE, the register can drive the FastTrack Interconnect while the LUT drives the local interconnect, or vice versa.

Arithmetic Mode

The arithmetic mode offers 2 three-input LUTs that are ideal for implementing adders, accumulators, and comparators. One LUT computes a three-input function, and the other generates a carry output. As shown in [Figure 9](#) on page 19, the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, in an adder, this output is the sum of three signals: a, b, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

Up/Down Counter Mode

The up/down counter mode offers counter enable, clock enable, synchronous up/down control, and data loading options. These control signals are generated by the data inputs from the LAB local interconnect, the carry-in signal, and output feedback from the programmable register. The Up/down counter mode uses 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading. Data can also be loaded asynchronously with the clear and preset register control signals, without using the LUT resources.

Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Clearable counter mode uses 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is ANDed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-stating without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

In addition to the six clear and preset modes, FLEX 10K devices provide a chip-wide reset pin that can reset all registers in the device. Use of this feature is set during design entry. In any of the clear and preset modes, the chip-wide reset overrides all other signals. Registers with asynchronous presets may be preset when the chip-wide reset is asserted. Inversion can be used to implement the asynchronous preset. [Figure 10](#) shows examples of how to enter a section of a design for the desired functionality.

Figure 15. FLEX 10K Column-to-IOE Connections

The values for m and n are provided in Table 11.

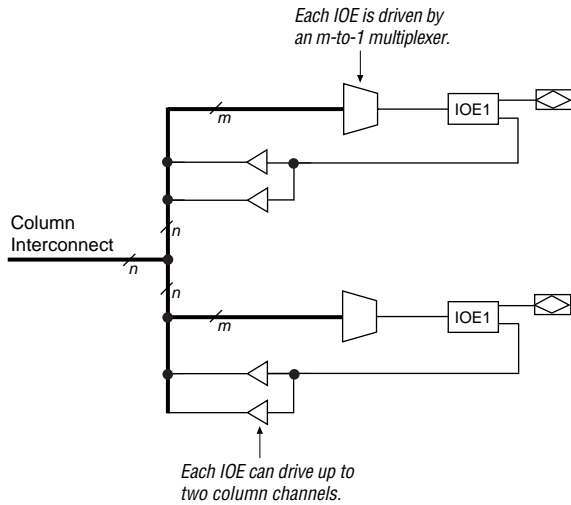


Table 11 lists the FLEX 10K column-to-IOE interconnect resources.

Table 11. FLEX 10K Column-to-IOE Interconnect Resources		
Device	Channels per Column (n)	Column Channel per Pin (m)
EPF10K10 EPF10K10A	24	16
EPF10K20	24	16
EPF10K30 EPF10K30A	24	16
EPF10K40	24	16
EPF10K50 EPF10K50V	24	16
EPF10K70	24	16
EPF10K100 EPF10K100A	24	16
EPF10K130V	32	24
EPF10K250A	40	32

Table 27. FLEX 10KA 3.3-V Device Recommended Operating Conditions

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CCINT}	Supply voltage for internal logic and input buffers	(3), (4)	3.00 (3.00)	3.60 (3.60)	V
V_{CCIO}	Supply voltage for output buffers, 3.3-V operation	(3), (4)	3.00 (3.00)	3.60 (3.60)	V
	Supply voltage for output buffers, 2.5-V operation	(3), (4)	2.30 (2.30)	2.70 (2.70)	V
V_I	Input voltage	(5)	-0.5	5.75	V
V_O	Output voltage		0	V_{CCIO}	V
T_A	Ambient temperature	For commercial use	0	70	°C
		For industrial use	-40	85	°C
T_J	Operating temperature	For commercial use	0	85	°C
		For industrial use	-40	100	°C
t_R	Input rise time			40	ns
t_F	Input fall time			40	ns

Figure 22 shows the typical output drive characteristics of EPF10K10A, EPF10K30A, EPF10K100A, and EPF10K250A devices with 3.3-V and 2.5-V V_{CCIO} . The output driver is compliant with the 3.3-V *PCI Local Bus Specification, Revision 2.2* (with 3.3-V V_{CCIO}). Moreover, device analysis shows that the EPF10K10A, EPF10K30A, and EPF 10K100A devices can drive a 5.0-V PCI bus with eight or fewer loads.

Figure 22. Output Drive Characteristics for EPF10K10A, EPF10K30A & EPF10K100A Devices

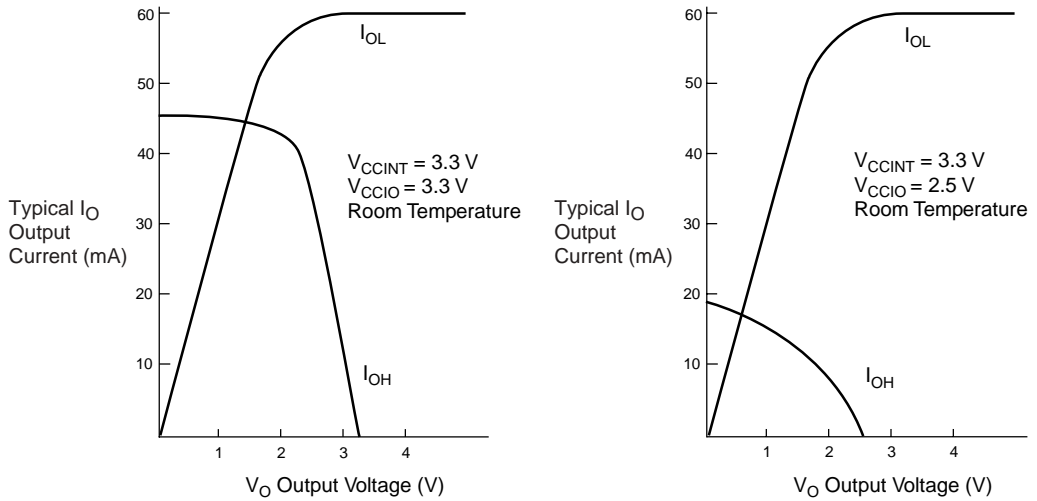


Figure 23 shows the typical output drive characteristics of the EPF10K250A device with 3.3-V and 2.5-V V_{CCIO} .

Table 34. EAB Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
$t_{EABDATA1}$	Data or address delay to EAB for combinatorial input	
$t_{EABDATA2}$	Data or address delay to EAB for registered input	
t_{EABWE1}	Write enable delay to EAB for combinatorial input	
t_{EABWE2}	Write enable delay to EAB for registered input	
t_{EABCLK}	EAB register clock delay	
t_{EABCO}	EAB register clock-to-output delay	
$t_{EABYPASS}$	Bypass register delay	
t_{EABSU}	EAB register setup time before clock	
t_{EABH}	EAB register hold time after clock	
t_{AA}	Address access delay	
t_{WP}	Write pulse width	
t_{WDSU}	Data setup time before falling edge of write pulse	(5)
t_{WDH}	Data hold time after falling edge of write pulse	(5)
t_{WASU}	Address setup time before rising edge of write pulse	(5)
t_{WAH}	Address hold time after falling edge of write pulse	(5)
t_{WO}	Write enable to data output valid delay	
t_{DD}	Data-in to data-out valid delay	
t_{EABOUT}	Data-out delay	
t_{EABCH}	Clock high time	
t_{EABCL}	Clock low time	

Table 35. EAB Timing Macroparameters <i>Notes (1), (6)</i>		
Symbol	Parameter	Conditions
t_{EABAA}	EAB address access delay	
$t_{EABRCCOMB}$	EAB asynchronous read cycle time	
$t_{EABRCREG}$	EAB synchronous read cycle time	
t_{EABWP}	EAB write pulse width	
$t_{EABWCCOMB}$	EAB asynchronous write cycle time	
$t_{EABWCREG}$	EAB synchronous write cycle time	
t_{EABDD}	EAB data-in to data-out valid delay	
$t_{EABDATACO}$	EAB clock-to-output delay when using output registers	
$t_{EABDATASU}$	EAB data/address setup time before clock when using input register	
$t_{EABDATAH}$	EAB data/address hold time after clock when using input register	
$t_{EABWESU}$	EAB \overline{WE} setup time before clock when using input register	
t_{EABWEH}	EAB \overline{WE} hold time after clock when using input register	
$t_{EABWDSU}$	EAB data setup time before falling edge of write pulse when not using input registers	
t_{EABWDH}	EAB data hold time after falling edge of write pulse when not using input registers	
$t_{EABWASU}$	EAB address setup time before rising edge of write pulse when not using input registers	
t_{EABWAH}	EAB address hold time after falling edge of write pulse when not using input registers	
t_{EABWO}	EAB write enable to data output valid delay	

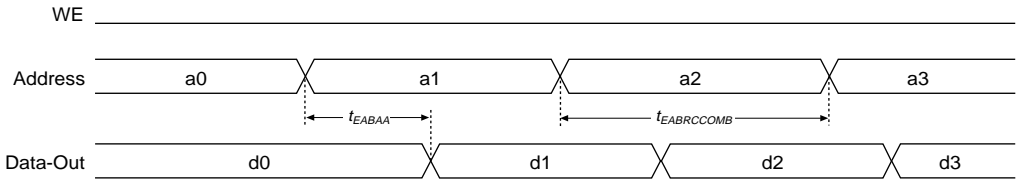
Notes to tables:

- (1) Microparameters are timing delays contributed by individual architectural elements. These parameters cannot be measured explicitly.
- (2) Operating conditions: $V_{CCIO} = 5.0\text{ V} \pm 5\%$ for commercial use in FLEX 10K devices.
 $V_{CCIO} = 5.0\text{ V} \pm 10\%$ for industrial use in FLEX 10K devices.
 $V_{CCIO} = 3.3\text{ V} \pm 10\%$ for commercial or industrial use in FLEX 10KA devices.
- (3) Operating conditions: $V_{CCIO} = 3.3\text{ V} \pm 10\%$ for commercial or industrial use in FLEX 10K devices.
 $V_{CCIO} = 2.5\text{ V} \pm 0.2\text{ V}$ for commercial or industrial use in FLEX 10KA devices.
- (4) Operating conditions: $V_{CCIO} = 2.5\text{ V}, 3.3\text{ V}, \text{ or } 5.0\text{ V}$.
- (5) Because the RAM in the EAB is self-timed, this parameter can be ignored when the WE signal is registered.
- (6) EAB macroparameters are internal parameters that can simplify predicting the behavior of an EAB at its boundary; these parameters are calculated by summing selected microparameters.
- (7) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.
- (8) External reference timing parameters are factory-tested, worst-case values specified by Altera. A representative subset of signal paths is tested to approximate typical device applications.
- (9) Contact Altera Applications for test circuit specifications and test conditions.
- (10) These timing parameters are sample-tested only.

Figures 29 and 30 show the asynchronous and synchronous timing waveforms, respectively, for the EAB macroparameters in Table 34.

Figure 29. EAB Asynchronous Timing Waveforms

EAB Asynchronous Read



EAB Asynchronous Write

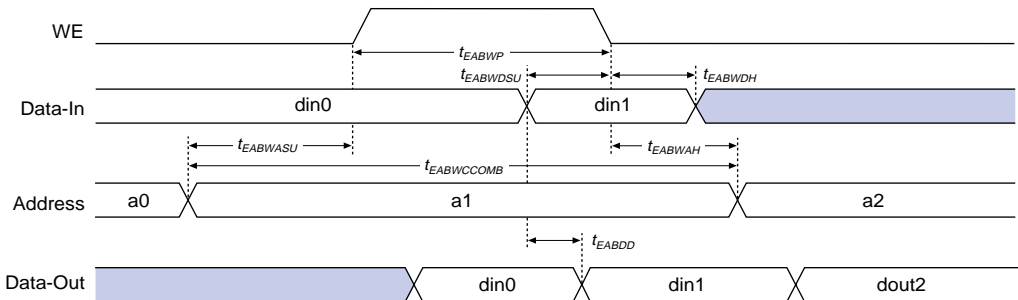


Table 58. EPF10K70 Device IOE Timing Microparameters *Note (1)*

Symbol	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		0.0		0.0		0.0	ns
t_{IOC}		0.4		0.5		0.7	ns
t_{IOCO}		0.4		0.4		0.9	ns
t_{IOCOMB}		0.0		0.0		0.0	ns
t_{IOSU}	4.5		5.0		6.2		ns
t_{IOH}	0.4		0.5		0.7		ns
t_{IOCLR}		0.6		0.7		1.6	ns
t_{OD1}		3.6		4.0		5.0	ns
t_{OD2}		5.6		6.3		7.3	ns
t_{OD3}		6.9		7.7		8.7	ns
t_{XZ}		5.5		6.2		6.8	ns
t_{ZX1}		5.5		6.2		6.8	ns
t_{ZX2}		7.5		8.5		9.1	ns
t_{ZX3}		8.8		9.9		10.5	ns
t_{INREG}		8.0		9.0		10.2	ns
t_{IOFD}		7.2		8.1		10.3	ns
t_{INCOMB}		7.2		8.1		10.3	ns

Table 87. EPF10K10A Device EAB Internal Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		3.3		3.9		5.2	ns
$t_{EABDATA2}$		1.0		1.3		1.7	ns
t_{EABWE1}		2.6		3.1		4.1	ns
t_{EABWE2}		2.7		3.2		4.3	ns
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		1.2		1.4		1.8	ns
$t_{EABBYPASS}$		0.1		0.2		0.2	ns
t_{EABSU}	1.4		1.7		2.2		ns
t_{EABH}	0.1		0.1		0.1		ns
t_{AA}		4.5		5.4		7.3	ns
t_{WP}	2.0		2.4		3.2		ns
t_{WDSU}	0.7		0.8		1.1		ns
t_{WDH}	0.5		0.6		0.7		ns
t_{WASU}	0.6		0.7		0.9		ns
t_{WAH}	0.9		1.1		1.5		ns
t_{WO}		3.3		3.9		5.2	ns
t_{DD}		3.3		3.9		5.2	ns
t_{EABOUT}		0.1		0.1		0.2	ns
t_{EABCH}	3.0		3.5		4.0		ns
t_{EABCL}	3.03		3.5		4.0		ns

Table 95. EPF10K30A Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		9.7		11.6		16.2	ns
$t_{EABRCCOMB}$	9.7		11.6		16.2		ns
$t_{EABRCREG}$	5.9		7.1		9.7		ns
t_{EABWP}	3.8		4.5		5.9		ns
$t_{EABWCCOMB}$	4.0		4.7		6.3		ns
$t_{EABWCREG}$	9.8		11.6		16.6		ns
t_{EABDD}		9.2		11.0		16.1	ns
$t_{EABDATACO}$		1.7		2.1		3.4	ns
$t_{EABDATASU}$	2.3		2.7		3.5		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	3.3		3.9		4.9		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	3.2		3.8		5.0		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	3.7		4.4		5.1		ns
t_{EABWAH}	0.0		0.0		0.0		ns
t_{EABWO}		6.1		7.3		11.3	ns

Table 96. EPF10K30A Device Interconnect Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		3.9		4.4		5.1	ns
t_{DIN2LE}		1.2		1.5		1.9	ns
$t_{DIN2DATA}$		3.2		3.6		4.5	ns
$t_{DCLK2IOE}$		3.0		3.5		4.6	ns
$t_{DCLK2LE}$		1.2		1.5		1.9	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		2.3		2.4		2.7	ns
$t_{SAMECOLUMN}$		1.3		1.4		1.9	ns
$t_{DIFFROW}$		3.6		3.8		4.6	ns
$t_{TROWROWS}$		5.9		6.2		7.3	ns
$t_{LEPERIPH}$		3.5		3.8		4.1	ns
$t_{LABCARRY}$		0.3		0.4		0.5	ns
$t_{LABCASC}$		0.9		1.1		1.4	ns

Table 97. EPF10K30A External Reference Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		11.0		13.0		17.0	ns
t_{INSU} (2), (3)	2.5		3.1		3.9		ns
t_{INH} (3)	0.0		0.0		0.0		ns
t_{OUTCO} (3)	2.0	5.4	2.0	6.2	2.0	8.3	ns

Table 98. EPF10K30A Device External Bidirectional Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{INSUBIDIR}$	4.2		4.9		6.8		ns
$t_{INHBIDIR}$	0.0		0.0		0.0		ns
$t_{OUTCOBIDIR}$	2.0	5.4	2.0	6.2	2.0	8.3	ns
$t_{XZBIDIR}$		6.2		7.5		9.8	ns
$t_{ZXBIDIR}$		6.2		7.5		9.8	ns

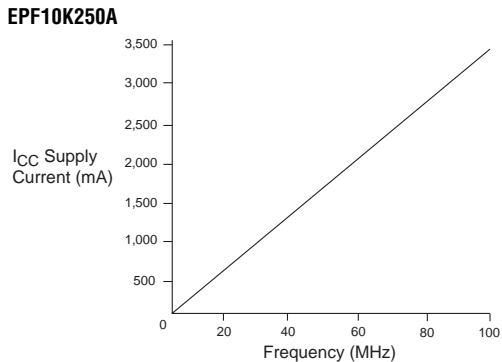
Notes to tables:

- (1) All timing parameters are described in Tables 32 through 38 in this data sheet.
- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.

Tables 99 through 105 show EPF10K100A device internal and external timing parameters.

Table 99. EPF10K100A Device LE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		1.0		1.2		1.4	ns
t_{CLUT}		0.8		0.9		1.1	ns
t_{RLUT}		1.4		1.6		1.9	ns
t_{PACKED}		0.4		0.5		0.5	ns
t_{EN}		0.6		0.7		0.8	ns
t_{CICO}		0.2		0.2		0.3	ns
t_{CGEN}		0.4		0.4		0.6	ns
t_{CGENR}		0.6		0.7		0.8	ns
t_{CASC}		0.7		0.9		1.0	ns
t_C		0.9		1.0		1.2	ns
t_{CO}		0.2		0.3		0.3	ns
t_{COMB}		0.6		0.7		0.8	ns
t_{SU}	0.8		1.0		1.2		ns
t_H	0.3		0.5		0.5		ns
t_{PRE}		0.3		0.3		0.4	ns
t_{CLR}		0.3		0.3		0.4	ns
t_{CH}	2.5		3.5		4.0		ns
t_{CL}	2.5		3.5		4.0		ns

Figure 32. $I_{CCACTIVE}$ vs. Operating Frequency (Part 3 of 3)

Configuration & Operation



The FLEX 10K architecture supports several configuration schemes. This section summarizes the device operating modes and available device configuration schemes.

See *Application Note 116 (Configuring APEX 20K, FLEX 10K & FLEX 6000 Devices)* for detailed descriptions of device configuration options, device configuration pins, and for information on configuring FLEX 10K devices, including sample schematics, timing diagrams, and configuration parameters.

Operating Modes

The FLEX 10K architecture uses SRAM configuration elements that require configuration data to be loaded every time the circuit powers up. The process of physically loading the SRAM data into the device is called *configuration*. Before configuration, as VCC rises, the device initiates a Power-On Reset (POR). This POR event clears the device and prepares it for configuration. The FLEX 10K POR time does not exceed 50 μ s.

During initialization, which occurs immediately after configuration, the device resets registers, enables I/O pins, and begins to operate as a logic device. The I/O pins are tri-stated during power-up, and before and during configuration. Together, the configuration and initialization processes are called *command mode*; normal device operation is called *user mode*.



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