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Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	216
Number of Logic Elements/Cells	1728
Total RAM Bits	12288
Number of I/O	191
Number of Gates	69000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	256-BGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k30afc256-2

The FLEX 10K architecture is similar to that of embedded gate arrays, the fastest-growing segment of the gate array market. As with standard gate arrays, embedded gate arrays implement general logic in a conventional “sea-of-gates” architecture. In addition, embedded gate arrays have dedicated die areas for implementing large, specialized functions. By embedding functions in silicon, embedded gate arrays provide reduced die area and increased speed compared to standard gate arrays. However, embedded megafunctions typically cannot be customized, limiting the designer’s options. In contrast, FLEX 10K devices are programmable, providing the designer with full control over embedded megafunctions and general logic while facilitating iterative design changes during debugging.

Each FLEX 10K device contains an embedded array and a logic array. The embedded array is used to implement a variety of memory functions or complex logic functions, such as digital signal processing (DSP), microcontroller, wide-data-path manipulation, and data-transformation functions. The logic array performs the same function as the sea-of-gates in the gate array: it is used to implement general logic, such as counters, adders, state machines, and multiplexers. The combination of embedded and logic arrays provides the high performance and high density of embedded gate arrays, enabling designers to implement an entire system on a single device.

FLEX 10K devices are configured at system power-up with data stored in an Altera serial configuration device or provided by a system controller. Altera offers the EPC1, EPC2, EPC16, and EPC1441 configuration devices, which configure FLEX 10K devices via a serial data stream. Configuration data can also be downloaded from system RAM or from Altera’s BitBlaster™ serial download cable or ByteBlasterMV™ parallel port download cable. After a FLEX 10K device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Because reconfiguration requires less than 320 ms, real-time changes can be made during system operation.

FLEX 10K devices contain an optimized interface that permits microprocessors to configure FLEX 10K devices serially or in parallel, and synchronously or asynchronously. The interface also enables microprocessors to treat a FLEX 10K device as memory and configure the device by writing to a virtual memory location, making it very easy for the designer to reconfigure the device.

Each LAB provides four control signals with programmable inversion that can be used in all eight LEs. Two of these signals can be used as clocks; the other two can be used for clear/preset control. The LAB clocks can be driven by the dedicated clock input pins, global signals, I/O signals, or internal signals via the LAB local interconnect. The LAB preset and clear control signals can be driven by the global signals, I/O signals, or internal signals via the LAB local interconnect. The global control signals are typically used for global clock, clear, or preset signals because they provide asynchronous control with very low skew across the device. If logic is required on a control signal, it can be generated in one or more LEs in any LAB and driven into the local interconnect of the target LAB. In addition, the global control signals can be generated from LE outputs.

Logic Element

The LE, the smallest unit of logic in the FLEX 10K architecture, has a compact size that provides efficient logic utilization. Each LE contains a four-input LUT, which is a function generator that can quickly compute any function of four variables. In addition, each LE contains a programmable flipflop with a synchronous enable, a carry chain, and a cascade chain. Each LE drives both the local and the FastTrack Interconnect. See [Figure 6](#).

Figure 6. FLEX 10K Logic Element

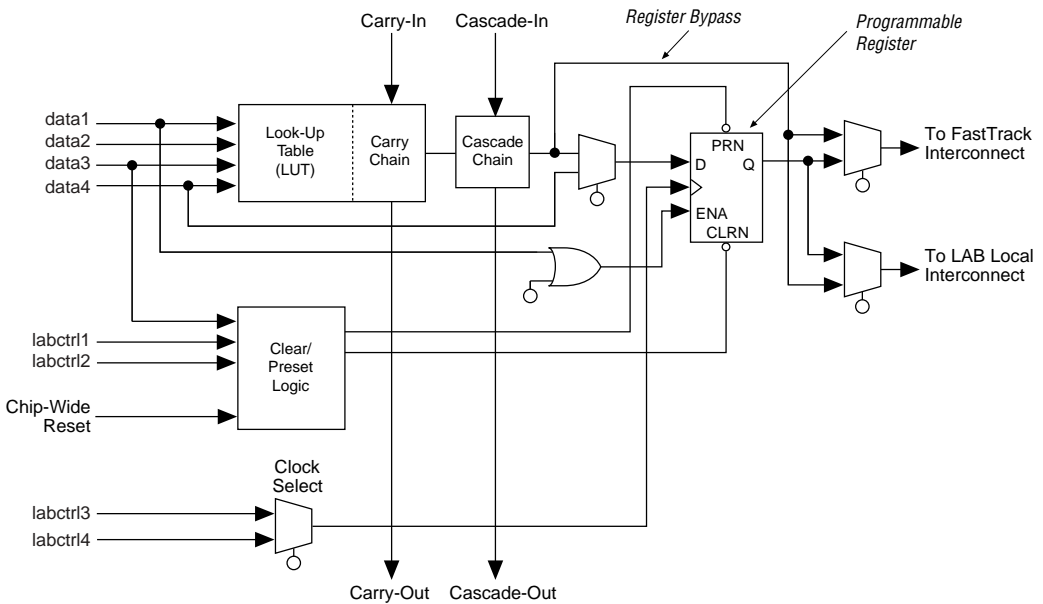


Figure 7 shows how an n -bit full adder can be implemented in $n + 1$ LEs with the carry chain. One portion of the LUT generates the sum of two bits using the input signals and the carry-in signal; the sum is routed to the output of the LE. The register can either be bypassed for simple adders or be used for an accumulator function. The carry chain logic generates the carry-out signal, which is routed directly to the carry-in signal of the next higher-order bit. The final carry-out signal is routed to an LE, where it can be used as a general-purpose signal.

Figure 7. Carry Chain Operation (n -bit Full Adder)

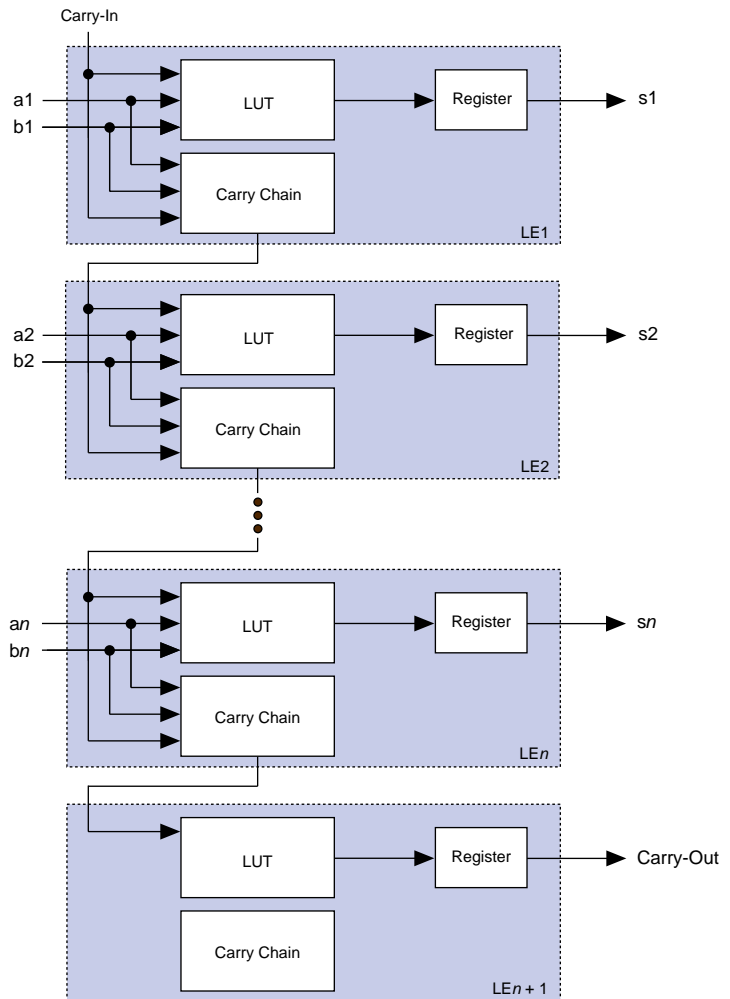
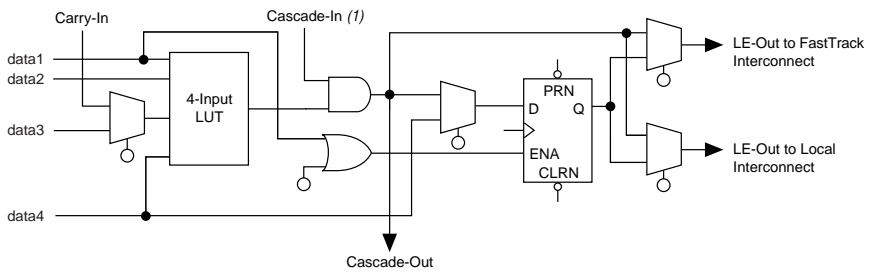
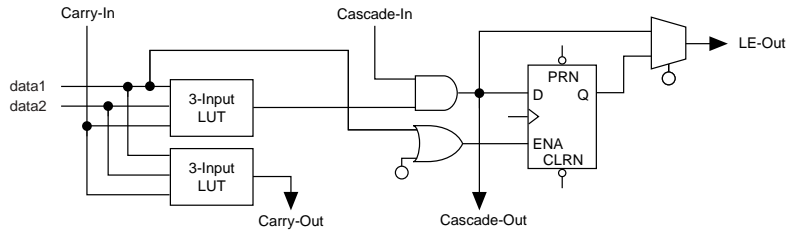


Figure 9. FLEX 10K LE Operating Modes

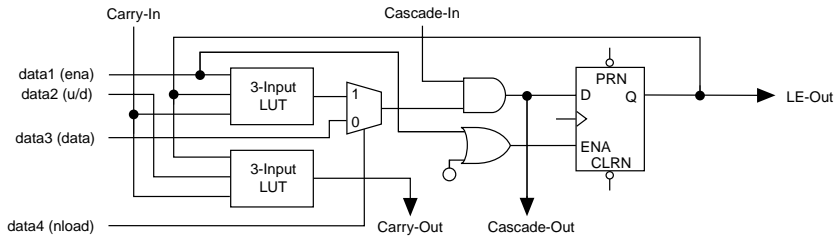
Normal Mode



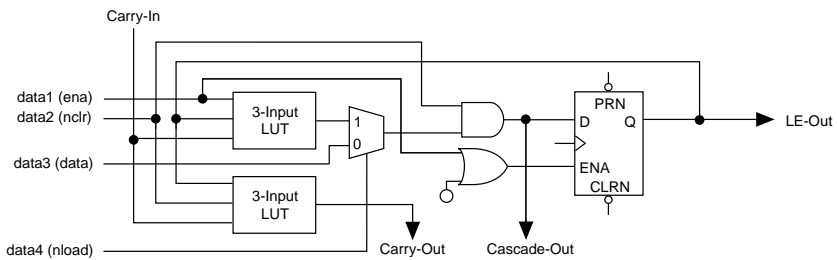
Arithmetic Mode



Up/Down Counter Mode



Clearable Counter Mode



Note:

(1) Packed registers cannot be used with the cascade chain.

SameFrame Pin-Outs

FLEX 10KE devices support the SameFrame pin-out feature for FineLine BGA packages. The SameFrame pin-out feature is the arrangement of balls on FineLine BGA packages such that the lower-ball-count packages form a subset of the higher-ball-count packages. SameFrame pin-outs provide the flexibility to migrate not only from device to device within the same package, but also from one package to another. A given printed circuit board (PCB) layout can support multiple device density/package combinations. For example, a single board layout can support a range of devices from an EPF10K10A device in a 256-pin FineLine BGA package to an EPF10K100A device in a 484-pin FineLine BGA package.

The Altera software provides support to design PCBs with SameFrame pin-out devices. Devices can be defined for present and future use. The Altera software generates pin-outs describing how to lay out a board to take advantage of this migration (see [Figure 16](#)).

Figure 16. SameFrame Pin-Out Example

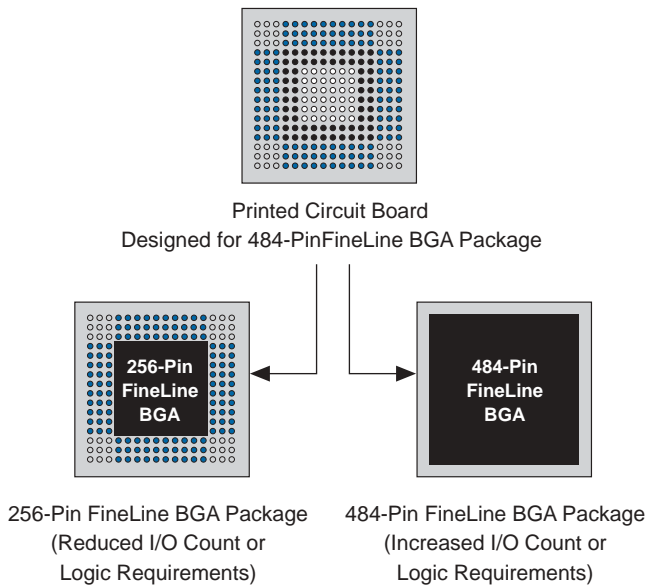
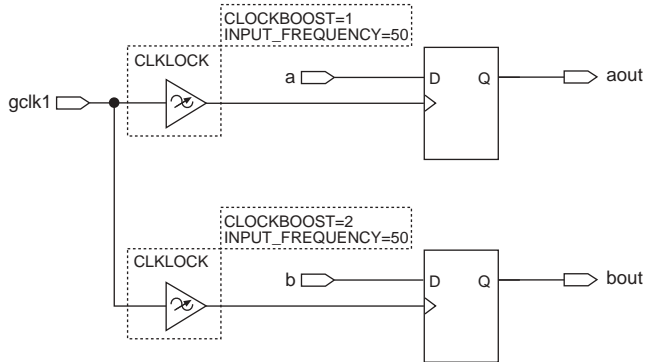


Figure 17. Enabling ClockLock & ClockBoost in the Same Design



To use both the ClockLock and ClockBoost circuits in the same design, designers must use Revision C EPF10K100GC503-3DX devices and MAX+PLUS II software versions 7.2 or higher. The die revision is indicated by the third digit of the nine-digit code on the top side of the device.

Output Configuration

This section discusses the peripheral component interconnect (PCI) pull-up clamping diode option, slew-rate control, open-drain output option, MultiVolt I/O interface, and power sequencing for FLEX 10K devices. The PCI pull-up clamping diode, slew-rate control, and open-drain output options are controlled pin-by-pin via Altera logic options. The MultiVolt I/O interface is controlled by connecting V_{CCIO} to a different voltage than V_{CCINT} . Its effect can be simulated in the Altera software via the **Global Project Device Options** dialog box (Assign menu).

PCI Clamping Diodes

The EPF10K10A and EPF10K30A devices have a pull-up clamping diode on every I/O, dedicated input, and dedicated clock pin. PCI clamping diodes clamp the transient overshoot caused by reflected waves to the V_{CCIO} value and are required for 3.3-V PCI compliance. Clamping diodes can also be used to limit overshoot in other systems.

Clamping diodes are controlled on a pin-by-pin basis via a logic option in the Altera software. When V_{CCIO} is 3.3 V, a pin that has the clamping diode turned on can be driven by a 2.5-V or 3.3-V signal, but not a 5.0-V signal. When V_{CCIO} is 2.5 V, a pin that has the clamping diode turned on can be driven by a 2.5-V signal, but not a 3.3-V or 5.0-V signal. However, a clamping diode can be turned on for a subset of pins, which allows devices to bridge between a 3.3-V PCI bus and a 5.0-V device.

Table 13. FLEX 10K JTAG Instructions

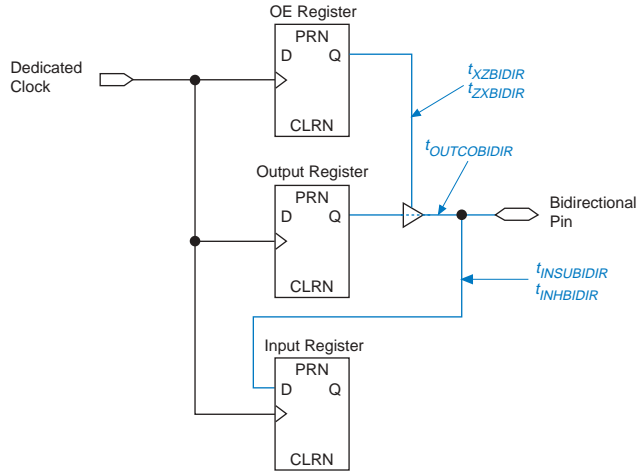
JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern output at the device pins.
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation.
USERCODE	Selects the user electronic signature (USERCODE) register and places it between the TDI and TDO pins, allowing the USERCODE to be serially shifted out of TDO.
IDCODE	Selects the IDCODE register and places it between TDI and TDO, allowing the IDCODE to be serially shifted out of TDO.
ICR Instructions	These instructions are used when configuring a FLEX 10K device via JTAG ports with a BitBlaster, or ByteBlasterMV or MasterBlaster download cable, or using a Jam File (.jam) or Jam Byte-Code File (.jbc) via an embedded processor.

The instruction register length of FLEX 10K devices is 10 bits. The USERCODE register length in FLEX 10K devices is 32 bits; 7 bits are determined by the user, and 25 bits are predetermined. Tables 14 and 15 show the boundary-scan register length and device IDCODE information for FLEX 10K devices.

Table 14. FLEX 10K Boundary-Scan Register Length

Device	Boundary-Scan Register Length
EPF10K10, EPF10K10A	480
EPF10K20	624
EPF10K30, EPF10K30A	768
EPF10K40	864
EPF10K50, EPF10K50V	960
EPF10K70	1,104
EPF10K100, EPF10K100A	1,248
EPF10K130V	1,440
EPF10K250A	1,440

Figure 28. Synchronous Bidirectional Pin External Timing Model



Tables 32 through 36 describe the FLEX 10K device internal timing parameters. These internal timing parameters are expressed as worst-case values. Using hand calculations, these parameters can be used to estimate design performance. However, before committing designs to silicon, actual worst-case performance should be modeled using timing simulation and analysis. Tables 37 through 38 describe FLEX 10K external timing parameters.

Symbol	Parameter	Conditions
t_{LUT}	LUT delay for data-in	
t_{CLUT}	LUT delay for carry-in	
t_{RLUT}	LUT delay for LE register feedback	
t_{PACKED}	Data-in to packed register delay	
t_{EN}	LE register enable delay	
t_{CICO}	Carry-in to carry-out delay	
t_{CGEN}	Data-in to carry-out delay	
t_{CGENR}	LE register feedback to carry-out delay	
t_{CASC}	Cascade-in to cascade-out delay	
t_C	LE register control signal delay	
t_{CO}	LE register clock-to-output delay	
t_{COMB}	Combinatorial delay	

Table 32. LE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	Parameter	Conditions
t_{SU}	LE register setup time for data and enable signals before clock; LE register recovery time after asynchronous clear, preset, or load	
t_H	LE register hold time for data and enable signals after clock	
t_{PRE}	LE register preset delay	
t_{CLR}	LE register clear delay	
t_{CH}	Minimum clock high time from clock pin	
t_{CL}	Minimum clock low time from clock pin	

Table 33. IOE Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
t_{IOD}	IOE data delay	
t_{IOC}	IOE register control signal delay	
t_{IOCO}	IOE register clock-to-output delay	
t_{IOCOMB}	IOE combinatorial delay	
t_{IOSU}	IOE register setup time for data and enable signals before clock; IOE register recovery time after asynchronous clear	
t_{IOH}	IOE register hold time for data and enable signals after clock	
t_{IOCLR}	IOE register clear time	
t_{OD1}	Output buffer and pad delay, slow slew rate = off, $V_{CCIO} = V_{CCINT}$	C1 = 35 pF (2)
t_{OD2}	Output buffer and pad delay, slow slew rate = off, $V_{CCIO} = \text{low voltage}$	C1 = 35 pF (3)
t_{OD3}	Output buffer and pad delay, slow slew rate = on	C1 = 35 pF (4)
t_{XZ}	IOE output buffer disable delay	
t_{ZX1}	IOE output buffer enable delay, slow slew rate = off, $V_{CCIO} = V_{CCINT}$	C1 = 35 pF (2)
t_{ZX2}	IOE output buffer enable delay, slow slew rate = off, $V_{CCIO} = \text{low voltage}$	C1 = 35 pF (3)
t_{ZX3}	IOE output buffer enable delay, slow slew rate = on	C1 = 35 pF (4)
t_{INREG}	IOE input pad and buffer to IOE register delay	
t_{IOFD}	IOE register feedback delay	
t_{INCOMB}	IOE input pad and buffer to FastTrack Interconnect delay	

Table 40. EPF10K10 & EPF10K20 Device IOE Timing Microparameters *Note (1)*

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
t_{IOD}		1.3		1.6	ns
t_{IOC}		0.5		0.7	ns
t_{IOCO}		0.2		0.2	ns
t_{IOCOMB}		0.0		0.0	ns
t_{IOSU}	2.8		3.2		ns
t_{IOH}	1.0		1.2		ns
t_{IOCLR}		1.0		1.2	ns
t_{OD1}		2.6		3.5	ns
t_{OD2}		4.9		6.4	ns
t_{OD3}		6.3		8.2	ns
t_{XZ}		4.5		5.4	ns
t_{ZX1}		4.5		5.4	ns
t_{ZX2}		6.8		8.3	ns
t_{ZX3}		8.2		10.1	ns
t_{INREG}		6.0		7.5	ns
t_{IOFD}		3.1		3.5	ns
t_{INCOMB}		3.1		3.5	ns

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
t_{DRR}		16.1		20.0	ns
t_{INSU} (2), (3)	5.5		6.0		ns
t_{INH} (3)	0.0		0.0		ns
t_{OUTCO} (3)	2.0	6.7	2.0	8.4	ns

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
$t_{\text{INSUBIDIR}}$	4.5		5.6		ns
t_{INHBIDIR}	0.0		0.0		ns
$t_{\text{OUTCOBIDIR}}$	2.0	6.7	2.0	8.4	ns
t_{XZBIDIR}		10.5		13.4	ns
t_{ZXBIDIR}		10.5		13.4	ns

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
$t_{\text{INSUBIDIR}}$	4.6		5.7		ns
t_{INHBIDIR}	0.0		0.0		ns
$t_{\text{OUTCOBIDIR}}$	2.0	6.7	2.0	8.4	ns
t_{XZBIDIR}		10.5		13.4	ns
t_{ZXBIDIR}		10.5		13.4	ns

Notes to tables:

- (1) All timing parameters are described in Tables 32 through 38 in this data sheet.
- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.

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Tables 64 through 70 show EPF10K100 device internal and external timing parameters.

Symbol	-3DX Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		1.5		1.5		2.0	ns
t_{CLUT}		0.4		0.4		0.5	ns
t_{RLUT}		1.6		1.6		2.0	ns
t_{PACKED}		0.9		0.9		1.3	ns
t_{EN}		0.9		0.9		1.2	ns
t_{CICO}		0.2		0.2		0.3	ns
t_{CGEN}		1.1		1.1		1.4	ns
t_{CGENR}		1.2		1.2		1.5	ns
t_{CASC}		1.1		1.1		1.3	ns
t_C		0.8		0.8		1.0	ns
t_{CO}		1.0		1.0		1.4	ns
t_{COMB}		0.5		0.5		0.7	ns
t_{SU}	2.1		2.1		2.6		ns
t_H	2.3		2.3		3.1		ns
t_{PRE}		1.0		1.0		1.4	ns
t_{CLR}		1.0		1.0		1.4	ns
t_{CH}	4.0		4.0		4.0		ns
t_{CL}	4.0		4.0		4.0		ns

Table 67. EPF10K100 Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-3DX Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		13.7		13.7		17.0	ns
$t_{EABRCCOMB}$	13.7		13.7		17.0		ns
$t_{EABRCREG}$	9.7		9.7		11.9		ns
t_{EABWP}	5.8		5.8		7.2		ns
$t_{EABWCCOMB}$	7.3		7.3		9.0		ns
$t_{EABWCREG}$	13.0		13.0		16.0		ns
t_{EABDD}		10.0		10.0		12.5	ns
$t_{EABDATACO}$		2.0		2.0		3.4	ns
$t_{EABDATASU}$	5.3		5.3		5.6		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	5.5		5.5		5.8		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	5.5		5.5		5.8		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	2.1		2.1		2.7		ns
t_{EABWAH}	0.0		0.0		0.0		ns
t_{EABWO}		9.5		9.5		11.8	ns

Table 68. EPF10K100 Device Interconnect Timing Microparameters *Note (1)*

Symbol	-3DX Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		10.3		10.3		12.2	ns
t_{DIN2LE}		4.8		4.8		6.0	ns
$t_{DIN2DATA}$		7.3		7.3		11.0	ns
$t_{DCLK2IOE}$ without ClockLock or ClockBoost circuitry		6.2		6.2		7.7	ns
$t_{DCLK2IOE}$ with ClockLock or ClockBoost circuitry		2.3		–		–	ns
$t_{DCLK2LE}$ without ClockLock or ClockBoost circuitry		4.8		4.8		6.0	ns
$t_{DCLK2LE}$ with ClockLock or ClockBoost circuitry		2.3		–		–	ns
$t_{SAMELAB}$		0.4		0.4		0.5	ns
$t_{SAMEROW}$		4.9		4.9		5.5	ns
$t_{SAMECOLUMN}$		5.1		5.1		5.4	ns
$t_{DIFFROW}$		10.0		10.0		10.9	ns
$t_{TWOROWS}$		14.9		14.9		16.4	ns
$t_{LEPERIPH}$		6.9		6.9		8.1	ns
$t_{LABCARRY}$		0.9		0.9		1.1	ns
$t_{LABCASC}$		3.0		3.0		3.2	ns

Table 81. EPF10K130V Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		11.2		14.2		14.2	ns
$t_{EABRCCOMB}$	11.1		14.2		14.2		ns
$t_{EABRCREG}$	8.5		10.8		10.8		ns
t_{EABWP}	3.7		4.7		4.7		ns
$t_{EABWCCOMB}$	7.6		9.7		9.7		ns
$t_{EABWCREG}$	14.0		17.8		17.8		ns
t_{EABDD}		11.1		14.2		14.2	ns
$t_{EABDATACO}$		3.6		4.6		4.6	ns
$t_{EABDATASU}$	4.4		5.6		5.6		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	4.4		5.6		5.6		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	4.6		5.9		5.9		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	3.9		5.0		5.0		ns
t_{EABWAH}	0.0		0.0		0.0		ns
t_{EABWO}		11.1		14.2		14.2	ns

Table 87. EPF10K10A Device EAB Internal Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		3.3		3.9		5.2	ns
$t_{EABDATA2}$		1.0		1.3		1.7	ns
t_{EABWE1}		2.6		3.1		4.1	ns
t_{EABWE2}		2.7		3.2		4.3	ns
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		1.2		1.4		1.8	ns
$t_{EABBYPASS}$		0.1		0.2		0.2	ns
t_{EABSU}	1.4		1.7		2.2		ns
t_{EABH}	0.1		0.1		0.1		ns
t_{AA}		4.5		5.4		7.3	ns
t_{WP}	2.0		2.4		3.2		ns
t_{WDSU}	0.7		0.8		1.1		ns
t_{WDH}	0.5		0.6		0.7		ns
t_{WASU}	0.6		0.7		0.9		ns
t_{WAH}	0.9		1.1		1.5		ns
t_{WO}		3.3		3.9		5.2	ns
t_{DD}		3.3		3.9		5.2	ns
t_{EABOUT}		0.1		0.1		0.2	ns
t_{EABCH}	3.0		3.5		4.0		ns
t_{EABCL}	3.03		3.5		4.0		ns

Notes to tables:

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- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.

ClockLock & ClockBoost Timing Parameters

For the ClockLock and ClockBoost circuitry to function properly, the incoming clock must meet certain requirements. If these specifications are not met, the circuitry may not lock onto the incoming clock, which generates an erroneous clock within the device. The clock generated by the ClockLock and ClockBoost circuitry must also meet certain specifications. If the incoming clock meets these requirements during configuration, the ClockLock and ClockBoost circuitry will lock onto the clock during configuration. The circuit will be ready for use immediately after configuration. Figure 31 illustrates the incoming and generated clock specifications.

Figure 31. Specifications for the Incoming & Generated Clocks

The t_I parameter refers to the nominal input clock period; the t_O parameter refers to the nominal output clock period.

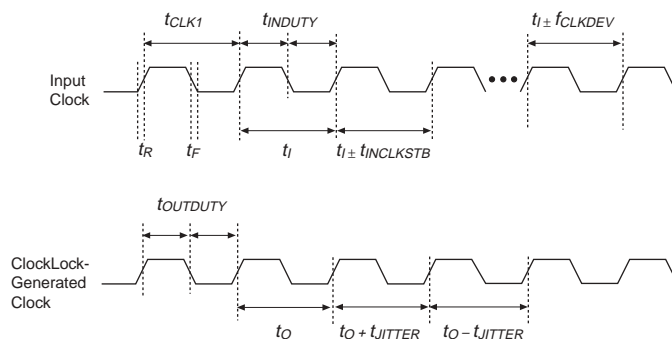
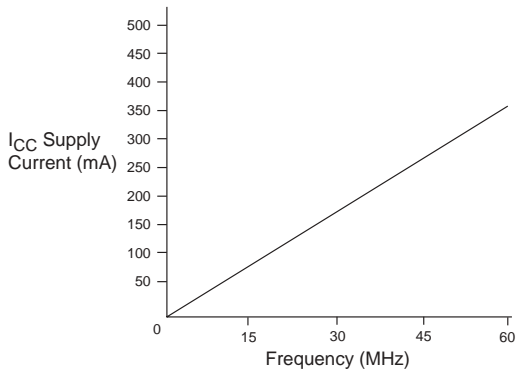


Table 113 summarizes the ClockLock and ClockBoost parameters.

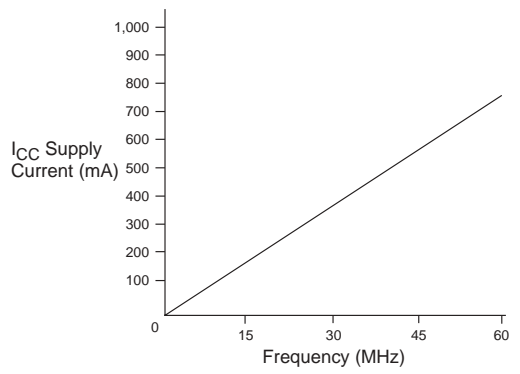
Symbol	Parameter	Min	Typ	Max	Unit
t_R	Input rise time			2	ns
t_F	Input fall time			2	ns
t_{INDUTY}	Input duty cycle	45		55	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)	30		80	MHz
t_{CLK1}	Input clock period (ClockBoost clock multiplication factor equals 1)	12.5		33.3	ns
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)	16		50	MHz
t_{CLK2}	Input clock period (ClockBoost clock multiplication factor equals 2)	20		62.5	ns

Figure 32. $I_{CCACTIVE}$ vs. Operating Frequency (Part 1 of 3)

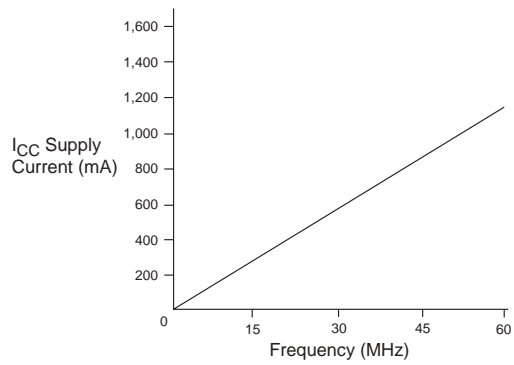
EPF10K10



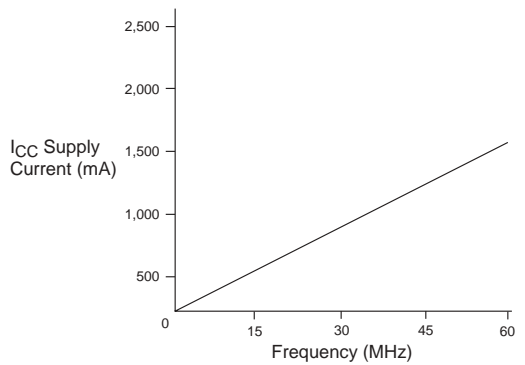
EPF10K20



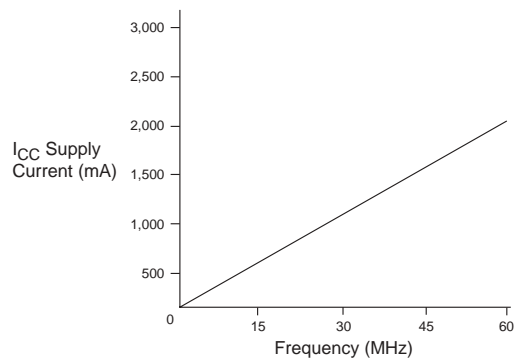
EPF10K30



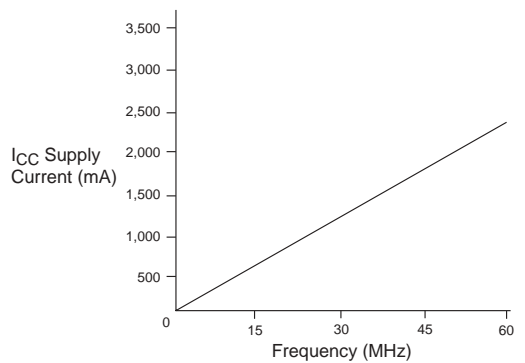
EPF10K40



EPF10K50



EPF10K70





Notes:



101 Innovation Drive
San Jose, CA 95134
(408) 544-7000
<http://www.altera.com>
Applications Hotline:
(800) 800-EPLD
Customer Marketing:
(408) 544-7104
Literature Services:
lit_req@altera.com

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