Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	216
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	147
Number of Gates	-
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=epf10k30aqc208-1n

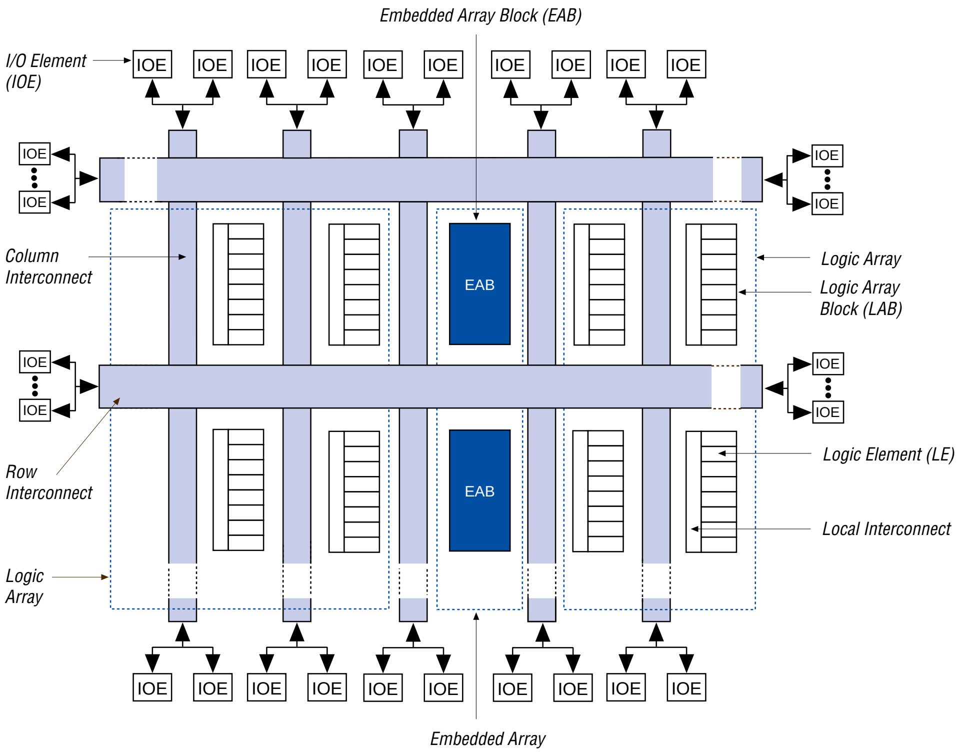
Table 4. FLEX 10K Package Options & I/O Pin Count *Note (1)*

Device	84-Pin PLCC	100-Pin TQFP	144-Pin TQFP	208-Pin PQFP RQFP	240-Pin PQFP RQFP
EPF10K10	59		102	134	
EPF10K10A		66	102	134	
EPF10K20			102	147	189
EPF10K30				147	189
EPF10K30A			102	147	189
EPF10K40				147	189
EPF10K50					189
EPF10K50V					189
EPF10K70					189
EPF10K100					
EPF10K100A					189
EPF10K130V					
EPF10K250A					

Table 5. FLEX 10K Package Options & I/O Pin Count (Continued) *Note (1)*

Device	503-Pin PGA	599-Pin PGA	256-Pin FineLine BGA	356-Pin BGA	484-Pin FineLine BGA	600-Pin BGA	403-Pin PGA
EPF10K10							
EPF10K10A			150		150 *(2)*		
EPF10K20							
EPF10K30				246			
EPF10K30A			191	246	246		
EPF10K40							
EPF10K50				274			310
EPF10K50V				274			
EPF10K70	358						
EPF10K100	406						
EPF10K100A				274	369	406	
EPF10K130V		470				470	
EPF10K250A		470				470	

Figure 1. FLEX 10K Device Block Diagram



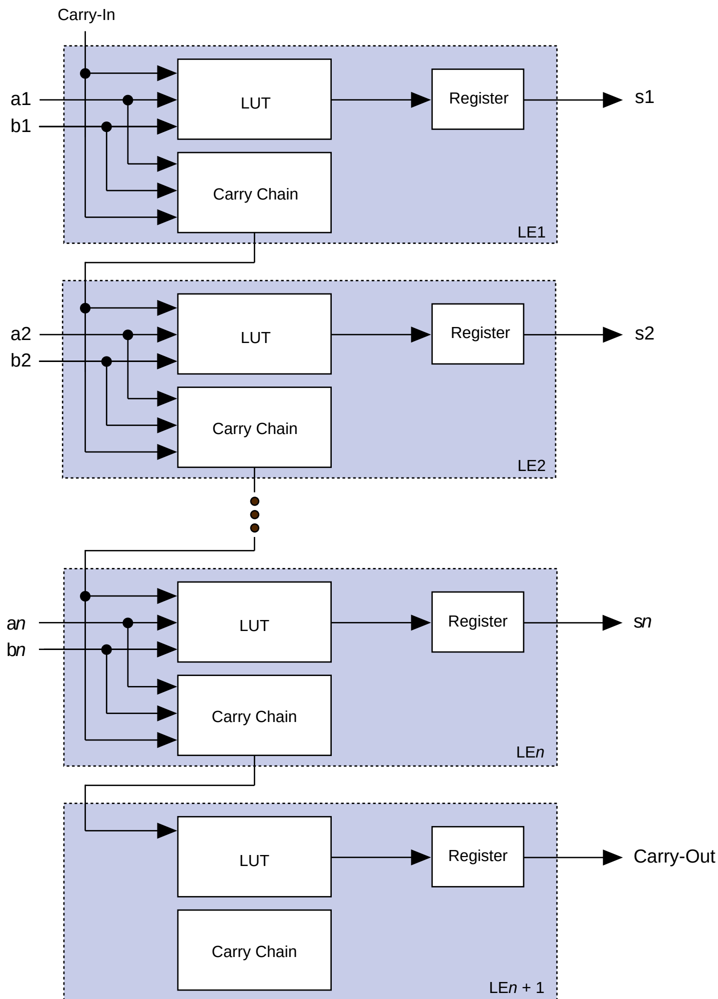
FLEX 10K devices provide six dedicated inputs that drive the flipflops' control inputs to ensure the efficient distribution of high-speed, low-skew (less than 1.5 ns) control signals. These signals use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect. Four of the dedicated inputs drive four global signals. These four global signals can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device.

Embedded Array Block

The EAB is a flexible block of RAM with registers on the input and output ports, and is used to implement common gate array megafunctions. The EAB is also suitable for functions such as multipliers, vector scalars, and error correction circuits, because it is large and flexible. These functions can be combined in applications such as digital filters and microcontrollers.

Figure 7 shows how an n-bit full adder can be implemented in $n + 1$ LEs with the carry chain. One portion of the LUT generates the sum of two bits using the input signals and the carry-in signal; the sum is routed to the output of the LE. The register can either be bypassed for simple adders or be used for an accumulator function. The carry chain logic generates the carry-out signal, which is routed directly to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it can be used as a general-purpose signal.

Figure 7. Carry Chain Operation (n-bit Full Adder)



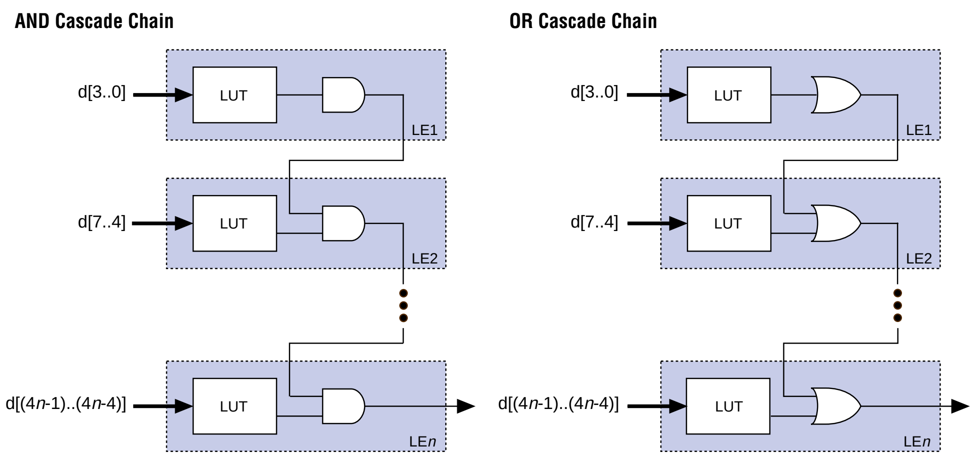
Cascade Chain

With the cascade chain, the FLEX 10K architecture can implement functions that have a very wide fan-in. Adjacent LUTs can be used to compute portions of the function in parallel; the cascade chain serially connects the intermediate values. The cascade chain can use a logical AND or logical OR (via De Morgan's inversion) to connect the outputs of adjacent LEs. Each additional LE provides four more inputs to the effective width of a function, with a delay as low as 0.7 ns per LE. Cascade chain logic can be created automatically by the Compiler during design processing, or manually by the designer during design entry.

Cascade chains longer than eight bits are implemented automatically by linking several LABs together. For easier routing, a long cascade chain skips every other LAB in a row. A cascade chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB (e.g., the last LE of the first LAB in a row cascades to the first LE of the third LAB). The cascade chain does not cross the center of the row (e.g., in the EPF10K50 device, the cascade chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB). This break is due to the EAB's placement in the middle of the row.

Figure 8 shows how the cascade function can connect adjacent LEs to form functions with a wide fan-in. These examples show functions of 4*n* variables implemented with *n* LEs. The LE delay is as low as 1.6 ns; the cascade chain delay is as low as 0.7 ns. With the cascade chain, 3.7 ns is needed to decode a 16-bit address.

Figure 8. Cascade Chain Operation



Up/Down Counter Mode

The up/down counter mode offers counter enable, clock enable, synchronous up/down control, and data loading options. These control signals are generated by the data inputs from the LAB local interconnect, the carry-in signal, and output feedback from the programmable register. The Up/down counter mode uses 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading. Data can also be loaded asynchronously with the clear and preset register control signals, without using the LUT resources.

Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Clearable counter mode uses 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is ANDed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-stating without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

In addition to the six clear and preset modes, FLEX 10K devices provide a chip-wide reset pin that can reset all registers in the device. Use of this feature is set during design entry. In any of the clear and preset modes, the chip-wide reset overrides all other signals. Registers with asynchronous presets may be preset when the chip-wide reset is asserted. Inversion can be used to implement the asynchronous preset. Figure 10 shows examples of how to enter a section of a design for the desired functionality.

FastTrack Interconnect

In the FLEX 10K architecture, connections between LEs and device I/O pins are provided by the FastTrack Interconnect, which is a series of continuous horizontal and vertical routing channels that traverse the device. This global routing structure provides predictable performance, even in complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.

The FastTrack Interconnect consists of row and column interconnect channels that span the entire device. Each row of LABs is served by a dedicated row interconnect. The row interconnect can drive I/O pins and feed other LABs in the device. The column interconnect routes signals between rows and can drive I/O pins.

A row channel can be driven by an LE or by one of three column channels. These four signals feed dual 4-to-1 multiplexers that connect to two specific row channels. These multiplexers, which are connected to each LE, allow column channels to drive row channels even when all eight LEs in an LAB drive the row interconnect.

Each column of LABs is served by a dedicated column interconnect. The column interconnect can then drive I/O pins or another row's interconnect to route the signals to other LABs in the device. A signal from the column interconnect, which can be either the output of an LE or an input from an I/O pin, must be routed to the row interconnect before it can enter an LAB or EAB. Each row channel that is driven by an IOE or EAB can drive one specific column channel.

Access to row and column channels can be switched between LEs in adjacent pairs of LABs. For example, an LE in one LAB can drive the row and column channels normally driven by a particular LE in the adjacent LAB in the same row, and vice versa. This routing flexibility enables routing resources to be used more efficiently. See Figure 11.

Each IOE selects the clock, clear, clock enable, and output enable controls from a network of I/O control signals called the peripheral control bus. The peripheral control bus uses high-speed drivers to minimize signal skew across devices; it provides up to 12 peripheral control signals that can be allocated as follows:

- Up to eight output enable signals
- Up to six clock enable signals
- Up to two clock signals
- Up to two clear signals

If more than six clock enable or eight output enable signals are required, each IOE on the device can be controlled by clock enable and output enable signals driven by specific LEs. In addition to the two clock signals available on the peripheral control bus, each IOE can use one of two dedicated clock pins. Each peripheral control signal can be driven by any of the dedicated input pins or the first LE of each LAB in a particular row. In addition, an LE in a different row can drive a column interconnect, which causes a row interconnect to drive the peripheral control signal. The chip-wide reset signal will reset all IOE registers, overriding any other control signals.

Tables 8 and 9 list the sources for each peripheral control signal, and the rows that can drive global signals. These tables also show how the output enable, clock enable, clock, and clear signals share 12 peripheral control signals.

Table 15. 32-Bit FLEX 10K Device IDCODE *Note (1)*

Device	IDCODE (32 Bits)			
	Version (4 Bits)	Part Number (16 Bits)	Manufacturer's Identity (11 Bits)	1 (1 Bit) *(2)*
EPF10K10, EPF10K10A	0000	0001 0000 0001 0000	00001101110	1
EPF10K20	0000	0001 0000 0010 0000	00001101110	1
EPF10K30, EPF10K30A	0000	0001 0000 0011 0000	00001101110	1
EPF10K40	0000	0001 0000 0100 0000	00001101110	1
EPF10K50, EPF10K50V	0000	0001 0000 0101 0000	00001101110	1
EPF10K70	0000	0001 0000 0111 0000	00001101110	1
EPF10K100, EPF10K100A	0000	0000 0001 0000 0000	00001101110	1
EPF10K130V	0000	0000 0001 0011 0000	00001101110	1
EPF10K250A	0000	0000 0010 0101 0000	00001101110	1

Notes:

(1) The most significant bit (MSB) is on the left.

(2) The least significant bit (LSB) for all JTAG IDCODEs is 1.

FLEX 10K devices include weak pull-ups on JTAG pins.

For more information, see the following documents:

- *Application Note 39 (IEEE 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices)*
- *BitBlaster Serial Download Cable Data Sheet*
- *ByteBlasterMV Parallel Port Download Cable Data Sheet*
- *Jam Programming & Test Language Specification*

Timing simulation and delay prediction are available with the MAX+PLUS II Simulator and Timing Analyzer, or with industry-standard EDA tools. The Simulator offers both pre-synthesis functional simulation to evaluate logic design accuracy and post-synthesis timing simulation with 0.1-ns resolution. The Timing Analyzer provides point-to-point timing delay information, setup and hold time analysis, and device-wide performance analysis.

Figure 24 shows the overall timing model, which maps the possible paths to and from the various elements of the FLEX 10K device.

Figure 24. FLEX 10K Device Timing Model

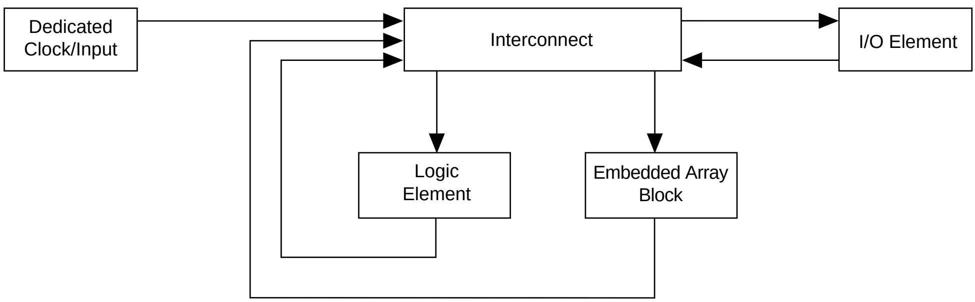


Figure 26. FLEX 10K Device IOE Timing Model

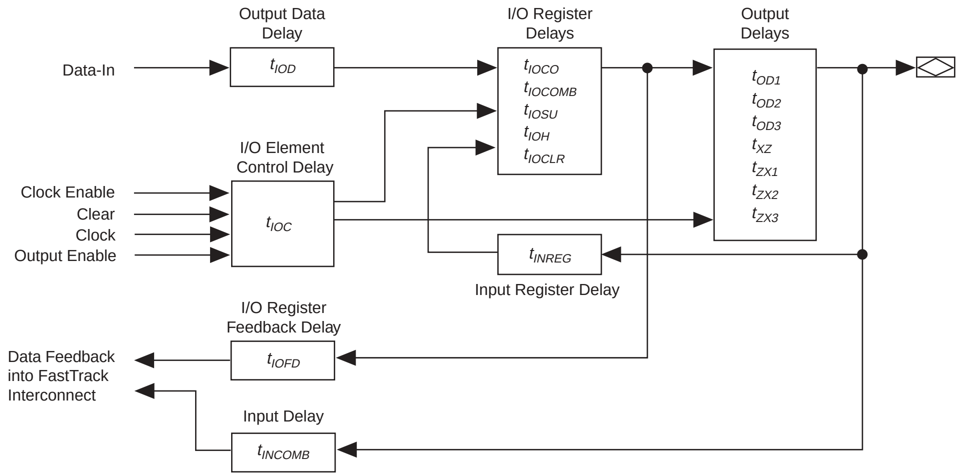
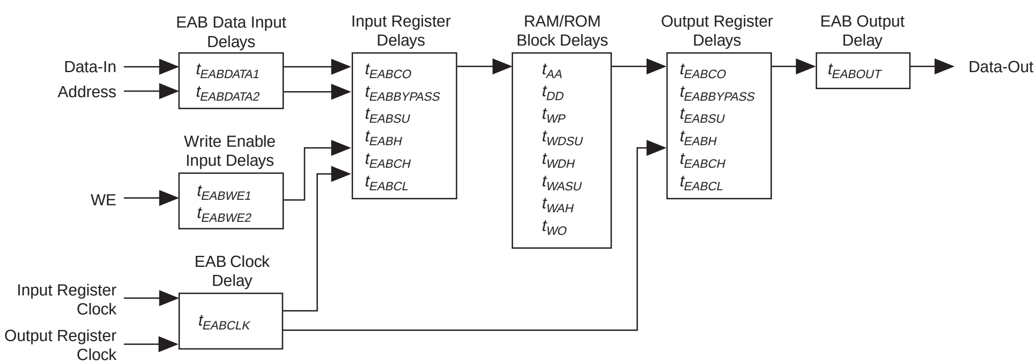


Figure 27. FLEX 10K Device EAB Timing Model



Figures 28 shows the timing model for bidirectional I/O pin timing.

Table 32. LE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	Parameter	Conditions
t_{SU}	LE register setup time for data and enable signals before clock; LE register recovery time after asynchronous clear, preset, or load	
t_{H}	LE register hold time for data and enable signals after clock	
t_{PRE}	LE register preset delay	
t_{CLR}	LE register clear delay	
t_{CH}	Minimum clock high time from clock pin	
t_{CL}	Minimum clock low time from clock pin	

Table 33. IOE Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
t_{IOD}	IOE data delay	
t_{IOC}	IOE register control signal delay	
t_{IOCO}	IOE register clock-to-output delay	
t_{IOCOMB}	IOE combinatorial delay	
t_{IOSU}	IOE register setup time for data and enable signals before clock; IOE register recovery time after asynchronous clear	
t_{IOH}	IOE register hold time for data and enable signals after clock	
t_{IOCLR}	IOE register clear time	
t_{OD1}	Output buffer and pad delay, slow slew rate = off, $V_{CCIO} = V_{CCINT}$	C1 = 35 pF *(2)*
t_{OD2}	Output buffer and pad delay, slow slew rate = off, V_{CCIO} = low voltage	C1 = 35 pF *(3)*
t_{OD3}	Output buffer and pad delay, slow slew rate = on	C1 = 35 pF *(4)*
t_{XZ}	IOE output buffer disable delay	
t_{ZX1}	IOE output buffer enable delay, slow slew rate = off, $V_{CCIO} = V_{CCINT}$	C1 = 35 pF *(2)*
t_{ZX2}	IOE output buffer enable delay, slow slew rate = off, V_{CCIO} = low voltage	C1 = 35 pF *(3)*
t_{ZX3}	IOE output buffer enable delay, slow slew rate = on	C1 = 35 pF *(4)*
t_{INREG}	IOE input pad and buffer to IOE register delay	
t_{IOFD}	IOE register feedback delay	
t_{INCOMB}	IOE input pad and buffer to FastTrack Interconnect delay	

Notes to tables:

(1) Microparameters are timing delays contributed by individual architectural elements. These parameters cannot be measured explicitly.

(2) Operating conditions: V_{CCIO} = 5.0 V ± 5% for commercial use in FLEX 10K devices.
V_{CCIO} = 5.0 V ± 10% for industrial use in FLEX 10K devices.
V_{CCIO} = 3.3 V ± 10% for commercial or industrial use in FLEX 10KA devices.

(3) Operating conditions: V_{CCIO} = 3.3 V ± 10% for commercial or industrial use in FLEX 10K devices.
V_{CCIO} = 2.5 V ± 0.2 V for commercial or industrial use in FLEX 10KA devices.

(4) Operating conditions: V_{CCIO} = 2.5 V, 3.3 V, or 5.0 V.

(5) Because the RAM in the EAB is self-timed, this parameter can be ignored when the `WE` signal is registered.

(6) EAB macroparameters are internal parameters that can simplify predicting the behavior of an EAB at its boundary; these parameters are calculated by summing selected microparameters.

(7) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.

(8) External reference timing parameters are factory-tested, worst-case values specified by Altera. A representative subset of signal paths is tested to approximate typical device applications.

(9) Contact Altera Applications for test circuit specifications and test conditions.

(10) These timing parameters are sample-tested only.

Figures 29 and 30 show the asynchronous and synchronous timing waveforms, respectively, for the EAB macroparameters in Table 34.

Figure 29. EAB Asynchronous Timing Waveforms

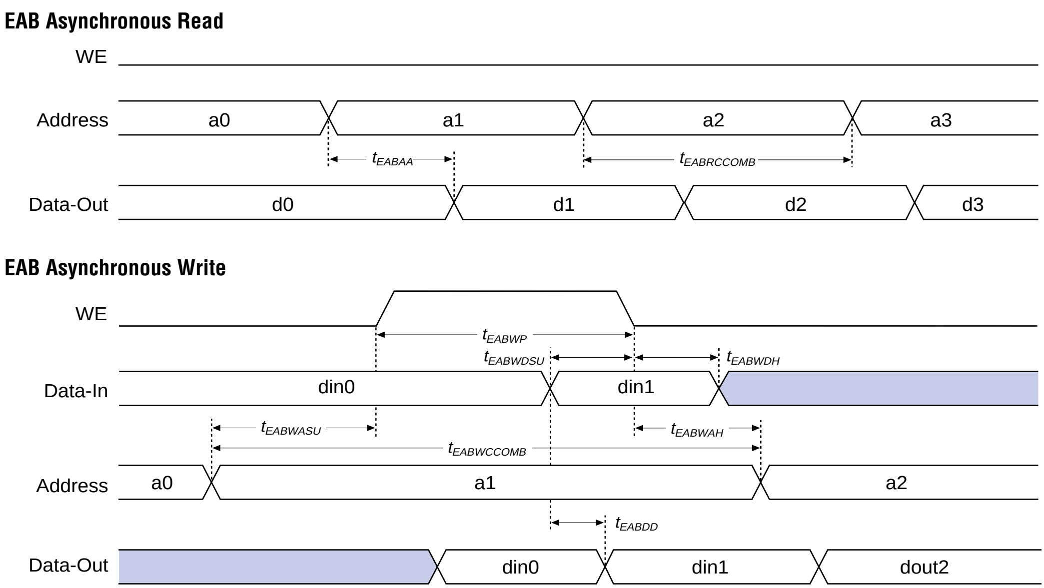


Table 59. EPF10K70 Device EAB Internal Microparameters *Note (1)*

Symbol	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.3		1.5		1.9	ns
$t_{EABDATA2}$		4.3		4.8		6.0	ns
t_{EABWE1}		0.9		1.0		1.2	ns
t_{EABWE2}		4.5		5.0		6.2	ns
t_{EABCLK}		0.9		1.0		2.2	ns
t_{EABCO}		0.4		0.5		0.6	ns
$t_{EABBYPASS}$		1.3		1.5		1.9	ns
t_{EABSU}	1.3		1.5		1.8		ns
t_{EABH}	1.8		2.0		2.5		ns
t_{AA}		7.8		8.7		10.7	ns
t_{WP}	5.2		5.8		7.2		ns
t_{WDSU}	1.4		1.6		2.0		ns
t_{WDH}	0.3		0.3		0.4		ns
t_{WASU}	0.4		0.5		0.6		ns
t_{WAH}	0.9		1.0		1.2		ns
t_{WO}		4.5		5.0		6.2	ns
t_{DD}		4.5		5.0		6.2	ns
t_{EABOUT}		0.4		0.5		0.6	ns
t_{EABCH}	4.0		4.0		4.0		ns
t_{EABCL}	5.2		5.8		7.2		ns

Table 69. EPF10K100 Device External Timing Parameters Note (1)

Symbol	-3DX Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		19.1		19.1		24.2	ns
t_{INSU} (2), (3), (4)	7.8		7.8		8.5		ns
t_{OUTCO} (3), (4)	2.0	11.1	2.0	11.1	2.0	14.3	ns
t_{INH} (3)	0.0		0.0		0.0		ns
t_{INSU} (2), (3), (5)	6.2		–		–		ns
t_{OUTCO} (3), (5)	2.0	6.7		–		–	ns

Table 70. EPF10K100 Device External Bidirectional Timing Parameters Note (1)

Symbol	-3DX Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{INSUBIDIR}$ (4)	8.1		8.1		10.4		ns
$t_{INHBIDIR}$ (4)	0.0		0.0		0.0		ns
$t_{OUTCOBIDIR}$ (4)	2.0	11.1	2.0	11.1	2.0	14.3	ns
$t_{XZBIDIR}$ (4)		15.3		15.3		18.4	ns
$t_{ZXBIDIR}$ (4)		15.3		15.3		18.4	ns
$t_{INSUBIDIR}$ (5)	9.1		–		–		ns
$t_{INHBIDIR}$ (5)	0.0		–		–		ns
$t_{OUTCOBIDIR}$ (5)	2.0	7.2	–	–	–	–	ns
$t_{XZBIDIR}$ (5)		14.3		–		–	ns
$t_{ZXBIDIR}$ (5)		14.3		–		–	ns

Notes to tables:

(1) All timing parameters are described in Tables 32 through 38 in this data sheet.
(2) Using an LE to register the signal may provide a lower setup time.
(3) This parameter is specified by characterization.
(4) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
(5) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Notes to tables:

(1) All timing parameters are described in Tables 32 through 38 in this data sheet.
(2) Using an LE to register the signal may provide a lower setup time.
(3) This parameter is specified by characterization.

Tables 78 through 84 show EPF10K130V device internal and external timing parameters.

Table 78. EPF10K130V Device LE Timing Microparameters *Note (1)*

Symbol	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		1.3		1.8		2.3	ns
t_{CLUT}		0.5		0.7		0.9	ns
t_{RLUT}		1.2		1.7		2.2	ns
t_{PACKED}		0.5		0.6		0.7	ns
t_{EN}		0.6		0.8		1.0	ns
t_{CICO}		0.2		0.3		0.4	ns
t_{CGEN}		0.3		0.4		0.5	ns
t_{CGENR}		0.7		1.0		1.3	ns
t_{CASC}		0.9		1.2		1.5	ns
t_{C}		1.9		2.4		3.0	ns
t_{CO}		0.6		0.9		1.1	ns
t_{COMB}		0.5		0.7		0.9	ns
t_{SU}	0.2		0.2		0.3		ns
t_{H}	0.0		0.0		0.0		ns
t_{PRE}		2.4		3.1		3.9	ns
t_{CLR}		2.4		3.1		3.9	ns
t_{CH}	4.0		4.0		4.0		ns
t_{CL}	4.0		4.0		4.0		ns

Table 96. EPF10K30A Device Interconnect Timing Microparameters *Note (1)*							
Symbol	**-1 Speed Grade**		**-2 Speed Grade**		**-3 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	**Min**	**Max**	
$t_{DIN2IOE}$		3.9		4.4		5.1	ns
t_{DIN2LE}		1.2		1.5		1.9	ns
$t_{DIN2DATA}$		3.2		3.6		4.5	ns
$t_{DCLK2IOE}$		3.0		3.5		4.6	ns
$t_{DCLK2LE}$		1.2		1.5		1.9	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		2.3		2.4		2.7	ns
$t_{SAMECOLUMN}$		1.3		1.4		1.9	ns
$t_{DIFFROW}$		3.6		3.8		4.6	ns
$t_{TWOROWS}$		5.9		6.2		7.3	ns
$t_{LEPERIPH}$		3.5		3.8		4.1	ns
$t_{LABCARRY}$		0.3		0.4		0.5	ns
$t_{LABCASC}$		0.9		1.1		1.4	ns

Table 97. EPF10K30A External Reference Timing Parameters *Note (1)*							
Symbol	**-1 Speed Grade**		**-2 Speed Grade**		**-3 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	**Min**	**Max**	
t_{DRR}		11.0		13.0		17.0	ns
t_{INSU} *(2)*, *(3)*	2.5		3.1		3.9		ns
t_{INH} *(3)*	0.0		0.0		0.0		ns
t_{OUTCO} *(3)*	2.0	5.4	2.0	6.2	2.0	8.3	ns

Table 98. EPF10K30A Device External Bidirectional Timing Parameters *Note (1)*							
Symbol	**-1 Speed Grade**		**-2 Speed Grade**		**-3 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	**Min**	**Max**	
$t_{INSUBIDIR}$	4.2		4.9		6.8		ns
$t_{INHBIDIR}$	0.0		0.0		0.0		ns
$t_{OUTCOBIDIR}$	2.0	5.4	2.0	6.2	2.0	8.3	ns
$t_{XZBIDIR}$		6.2		7.5		9.8	ns
$t_{ZXBIDIR}$		6.2		7.5		9.8	ns

Table 100. EPF10K100A Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		2.5		2.9		3.4	ns
t_{IOC}		0.3		0.3		0.4	ns
t_{IOCO}		0.2		0.2		0.3	ns
t_{IOCOMB}		0.5		0.6		0.7	ns
t_{IOSU}	1.3		1.7		1.8		ns
t_{IOH}	0.2		0.2		0.3		ns
t_{IOCLR}		1.0		1.2		1.4	ns
t_{OD1}		2.2		2.6		3.0	ns
t_{OD2}		4.5		5.3		6.1	ns
t_{OD3}		6.8		7.9		9.3	ns
t_{XZ}		2.7		3.1		3.7	ns
t_{ZX1}		2.7		3.1		3.7	ns
t_{ZX2}		5.0		5.8		6.8	ns
t_{ZX3}		7.3		8.4		10.0	ns
t_{INREG}		5.3		6.1		7.2	ns
t_{IOFD}		4.7		5.5		6.4	ns
t_{INCOMB}		4.7		5.5		6.4	ns

Notes to tables:
(1) All timing parameters are described in Tables 32 through 38 in this data sheet.
(2) Using an LE to register the signal may provide a lower setup time.
(3) This parameter is specified by characterization.

Tables 106 through 112 show EPF10K250A device internal and external timing parameters.

Table 106. EPF10K250A Device LE Timing Microparameters Note (1)

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.9		1.0		1.4	ns
t_{CLUT}		1.2		1.3		1.6	ns
t_{RLUT}		2.0		2.3		2.7	ns
t_{PACKED}		0.4		0.4		0.5	ns
t_{EN}		1.4		1.6		1.9	ns
t_{CICO}		0.2		0.3		0.3	ns
t_{CGEN}		0.4		0.6		0.6	ns
t_{CGENR}		0.8		1.0		1.1	ns
t_{CASC}		0.7		0.8		1.0	ns
t_C		1.2		1.3		1.6	ns
t_{CO}		0.6		0.7		0.9	ns
t_{COMB}		0.5		0.6		0.7	ns
t_{SU}	1.2		1.4		1.7		ns
t_H	1.2		1.3		1.6		ns
t_{PRE}		0.7		0.8		0.9	ns
t_{CLR}		0.7		0.8		0.9	ns
t_{CH}	2.5		3.0		3.5		ns
t_{CL}	2.5		3.0		3.5		ns

Notes to tables:
(1) All timing parameters are described in Tables 32 through 37 in this data sheet.
(2) Using an LE to register the signal may provide a lower setup time.
(3) This parameter is specified by characterization.

ClockLock & ClockBoost Timing Parameters

For the ClockLock and ClockBoost circuitry to function properly, the incoming clock must meet certain requirements. If these specifications are not met, the circuitry may not lock onto the incoming clock, which generates an erroneous clock within the device. The clock generated by the ClockLock and ClockBoost circuitry must also meet certain specifications. If the incoming clock meets these requirements during configuration, the ClockLock and ClockBoost circuitry will lock onto the clock during configuration. The circuit will be ready for use immediately after configuration. Figure 31 illustrates the incoming and generated clock specifications.

Figure 31. Specifications for the Incoming & Generated Clocks

The t_I parameter refers to the nominal input clock period; the t_O parameter refers to the nominal output clock period.

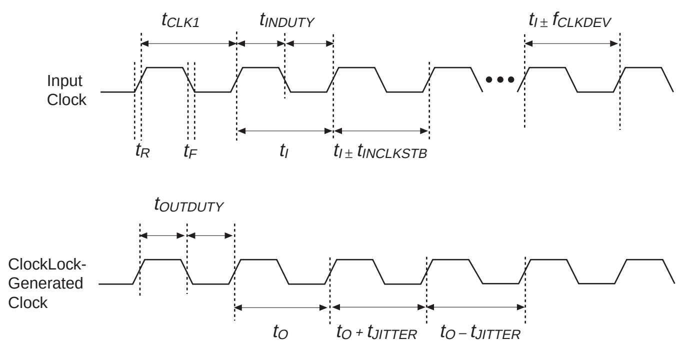


Table 113 summarizes the ClockLock and ClockBoost parameters.

Table 113. ClockLock & ClockBoost Parameters (Part 1 of 2)

Symbol	Parameter	Min	Typ	Max	Unit
t_R	Input rise time			2	ns
t_F	Input fall time			2	ns
t_{INDUTY}	Input duty cycle	45		55	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)	30		80	MHz
t_{CLK1}	Input clock period (ClockBoost clock multiplication factor equals 1)	12.5		33.3	ns
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)	16		50	MHz
t_{CLK2}	Input clock period (ClockBoost clock multiplication factor equals 2)	20		62.5	ns