Intel - EPF10K30AQI208-3 Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	216
Number of Logic Elements/Cells	1728
Total RAM Bits	12288
Number of I/O	147
Number of Gates	69000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k30aqi208-3

For more information, see the following documents:

- *Configuration Devices for APEX & FLEX Devices Data Sheet*
- *BitBlaster Serial Download Cable Data Sheet*
- *ByteBlasterMV Parallel Port Download Cable Data Sheet*
- *Application Note 116 (Configuring APEX 20K, FLEX 10K & FLEX 6000 Devices)*

FLEX 10K devices are supported by Altera development systems; single, integrated packages that offer schematic, text (including AHDL), and waveform design entry, compilation and logic synthesis, full simulation and worst-case timing analysis, and device configuration. The Altera software provides EDIF 2 0 0 and 3 0 0, LPM, VHDL, Verilog HDL, and other interfaces for additional design entry and simulation support from other industry-standard PC- and UNIX workstation-based EDA tools.

The Altera software works easily with common gate array EDA tools for synthesis and simulation. For example, the Altera software can generate Verilog HDL files for simulation with tools such as Cadence Verilog-XL. Additionally, the Altera software contains EDA libraries that use device-specific features such as carry chains which are used for fast counter and arithmetic functions. For instance, the Synopsys Design Compiler library supplied with the Altera development systems include DesignWare functions that are optimized for the FLEX 10K architecture.

The Altera development systems run on Windows-based PCs and Sun SPARCstation, and HP 9000 Series 700/800 workstations.

See the *MAX+PLUS II Programmable Logic Development System & Software Data Sheet* for more information.

Functional Description

Each FLEX 10K device contains an embedded array to implement memory and specialized logic functions, and a logic array to implement general logic.

The embedded array consists of a series of EABs. When implementing memory functions, each EAB provides 2,048 bits, which can be used to create RAM, ROM, dual-port RAM, or first-in first-out (FIFO) functions. When implementing logic, each EAB can contribute 100 to 600 gates towards complex logic functions, such as multipliers, microcontrollers, state machines, and DSP functions. EABs can be used independently, or multiple EABs can be combined to implement larger functions.

The logic array consists of logic array blocks (LABs). Each LAB contains eight LEs and a local interconnect. An LE consists of a 4-input look-up table (LUT), a programmable flipflop, and dedicated signal paths for carry and cascade functions. The eight LEs can be used to create medium-sized blocks of logic—8-bit counters, address decoders, or state machines—or combined across LABs to create larger logic blocks. Each LAB represents about 96 usable gates of logic.

Signal interconnections within FLEX 10K devices and to and from device pins are provided by the FastTrack Interconnect, a series of fast, continuous row and column channels that run the entire length and width of the device.

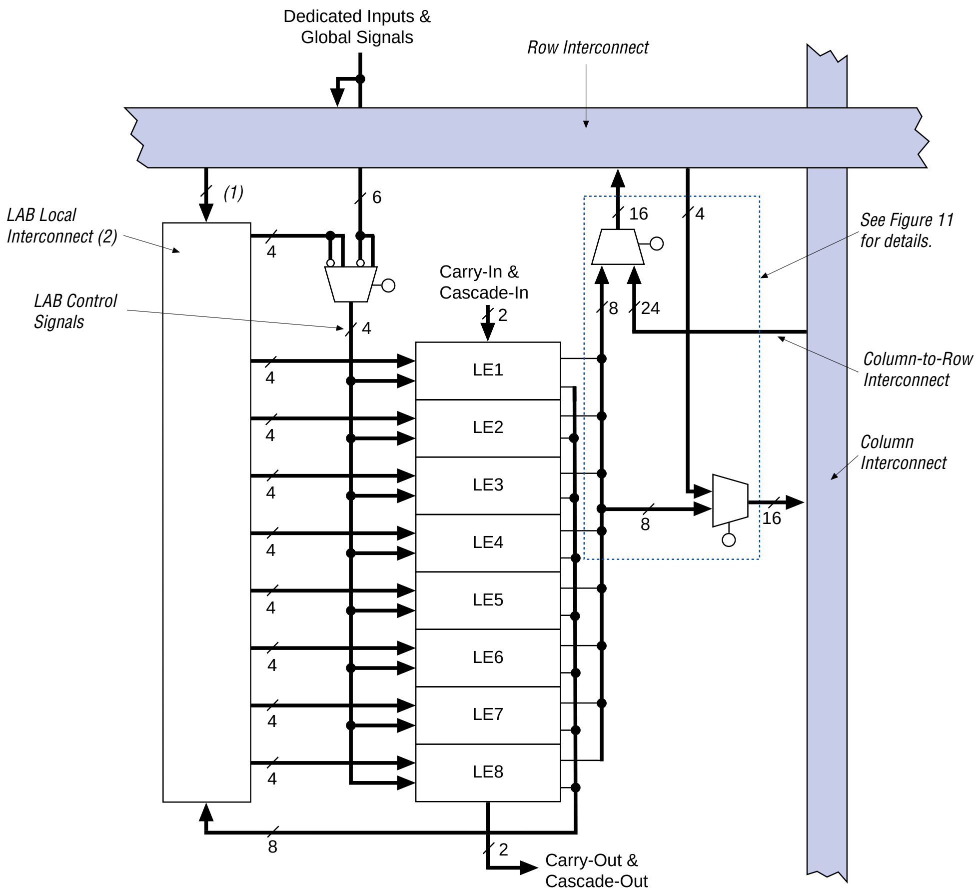
Each I/O pin is fed by an I/O element (IOE) located at the end of each row and column of the FastTrack Interconnect. Each IOE contains a bidirectional I/O buffer and a flipflop that can be used as either an output or input register to feed input, output, or bidirectional signals. When used with a dedicated clock pin, these registers provide exceptional performance. As inputs, they provide setup times as low as 1.6 ns and hold times of 0 ns; as outputs, these registers provide clock-to-output times as low as 5.3 ns. IOEs provide a variety of features, such as JTAG BST support, slew-rate control, tri-state buffers, and open-drain outputs.

Figure 1 shows a block diagram of the FLEX 10K architecture. Each group of LEs is combined into an LAB; LABs are arranged into rows and columns. Each row also contains a single EAB. The LABs and EABs are interconnected by the FastTrack Interconnect. IOEs are located at the end of each row and column of the FastTrack Interconnect.

Logic Array Block

Each LAB consists of eight LEs, their associated carry and cascade chains, LAB control signals, and the LAB local interconnect. The LAB provides the coarse-grained structure to the FLEX 10K architecture, facilitating efficient routing with optimum device utilization and high performance. See Figure 5.

Figure 5. FLEX 10K LAB



Notes:

(1) EPF10K10, EPF10K10A, EPF10K20, EPF10K30, EPF10K30A, EPF10K40, EPF10K50, and EPF10K50V devices have 22 inputs to the LAB local interconnect channel from the row; EPF10K70, EPF10K100, EPF10K100A, EPF10K130V, and EPF10K250A devices have 26.

(2) EPF10K10, EPF10K10A, EPF10K20, EPF10K30, EPF10K30A, EPF10K40, EPF10K50, and EPF10K50V devices have 30 LAB local interconnect channels; EPF10K70, EPF10K100, EPF10K100A, EPF10K130V, and EPF10K250A devices have 34 LABs.

The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock, clear, and preset control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the output of the LE.

The LE has two outputs that drive the interconnect; one drives the local interconnect and the other drives either the row or column FastTrack Interconnect. The two outputs can be controlled independently. For example, the LUT can drive one output while the register drives the other output. This feature, called register packing, can improve LE utilization because the register and the LUT can be used for unrelated functions.

The FLEX 10K architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. The carry chain supports high-speed counters and adders; the cascade chain implements wide-input functions with minimum delay. Carry and cascade chains connect all LEs in an LAB and all LABs in the same row. Intensive use of carry and cascade chains can reduce routing flexibility. Therefore, the use of these chains should be limited to speed-critical portions of a design.

Carry Chain

The carry chain provides a very fast (as low as 0.2 ns) carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the FLEX 10K architecture to implement high-speed counters, adders, and comparators of arbitrary width efficiently. Carry chain logic can be created automatically by the Compiler during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of carry chains.

Carry chains longer than eight LEs are automatically implemented by linking LABs together. For enhanced fitting, a long carry chain skips alternate LABs in a row. A carry chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB. For example, the last LE of the first LAB in a row carries to the first LE of the third LAB in the row. The carry chain does not cross the EAB at the middle of the row. For instance, in the EPF10K50 device, the carry chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB.

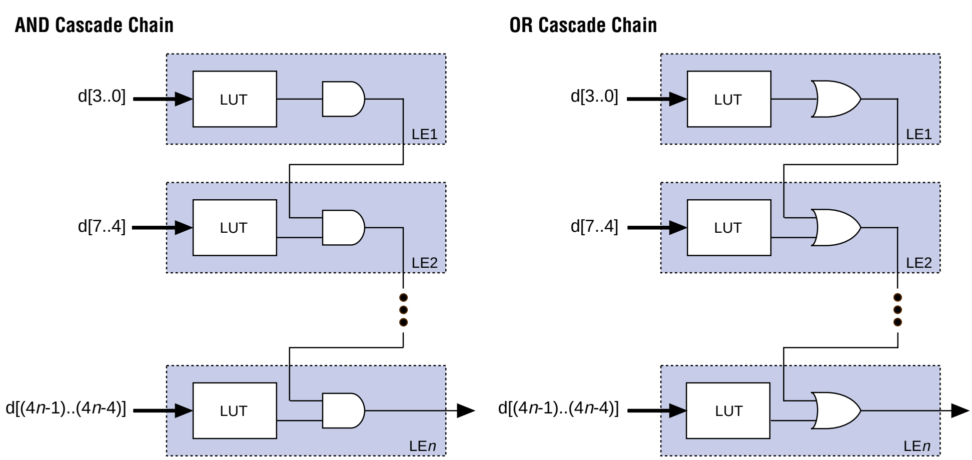
Cascade Chain

With the cascade chain, the FLEX 10K architecture can implement functions that have a very wide fan-in. Adjacent LUTs can be used to compute portions of the function in parallel; the cascade chain serially connects the intermediate values. The cascade chain can use a logical AND or logical OR (via De Morgan's inversion) to connect the outputs of adjacent LEs. Each additional LE provides four more inputs to the effective width of a function, with a delay as low as 0.7 ns per LE. Cascade chain logic can be created automatically by the Compiler during design processing, or manually by the designer during design entry.

Cascade chains longer than eight bits are implemented automatically by linking several LABs together. For easier routing, a long cascade chain skips every other LAB in a row. A cascade chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB (e.g., the last LE of the first LAB in a row cascades to the first LE of the third LAB). The cascade chain does not cross the center of the row (e.g., in the EPF10K50 device, the cascade chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB). This break is due to the EAB's placement in the middle of the row.

Figure 8 shows how the cascade function can connect adjacent LEs to form functions with a wide fan-in. These examples show functions of 4*n* variables implemented with *n* LEs. The LE delay is as low as 1.6 ns; the cascade chain delay is as low as 0.7 ns. With the cascade chain, 3.7 ns is needed to decode a 16-bit address.

Figure 8. Cascade Chain Operation



Normal Mode

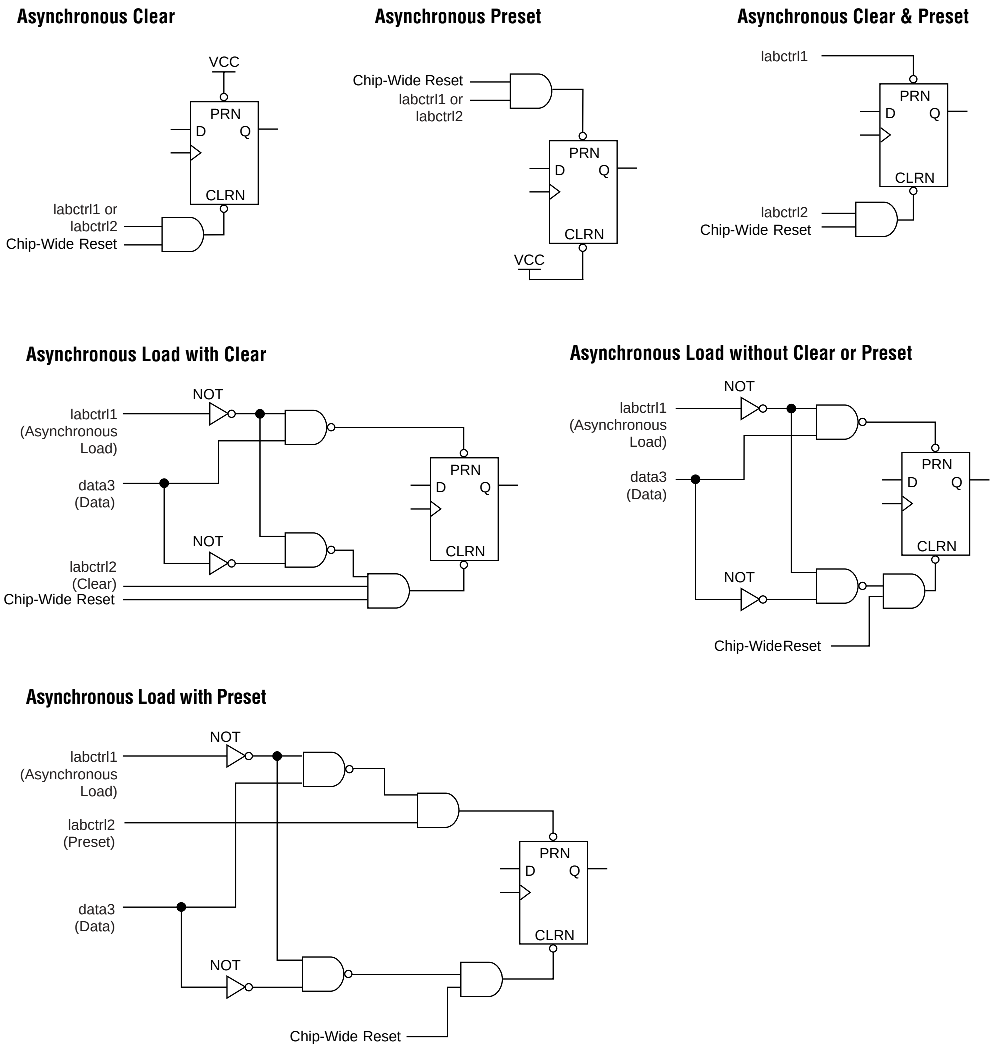
The normal mode is suitable for general logic applications and wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a four-input LUT. The Compiler automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. Either the register or the LUT can be used to drive both the local interconnect and the FastTrack Interconnect at the same time.

The LUT and the register in the LE can be used independently; this feature is known as register packing. To support register packing, the LE has two outputs; one drives the local interconnect and the other drives the FastTrack Interconnect. The DATA4 signal can drive the register directly, allowing the LUT to compute a function that is independent of the registered signal; a three-input function can be computed in the LUT, and a fourth independent signal can be registered. Alternatively, a four-input function can be generated, and one of the inputs to this function can be used to drive the register. The register in a packed LE can still use the clock enable, clear, and preset signals in the LE. In a packed LE, the register can drive the FastTrack Interconnect while the LUT drives the local interconnect, or vice versa.

Arithmetic Mode

The arithmetic mode offers 2 three-input LUTs that are ideal for implementing adders, accumulators, and comparators. One LUT computes a three-input function, and the other generates a carry output. As shown in Figure 9 on page 19, the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, in an adder, this output is the sum of three signals: a, b, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

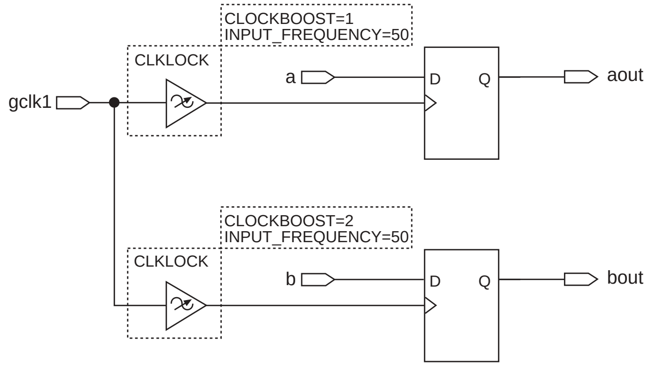
Figure 10. LE Clear & Preset Modes



Asynchronous Clear

The flipflop can be cleared by either `LABCTRL1` or `LABCTRL2`. In this mode, the preset signal is tied to V_{CC} to deactivate it.

Figure 17. Enabling ClockLock & ClockBoost in the Same Design



To use both the ClockLock and ClockBoost circuits in the same design, designers must use Revision C EPF10K100GC503-3DX devices and MAX+PLUS II software versions 7.2 or higher. The die revision is indicated by the third digit of the nine-digit code on the top side of the device.

Output Configuration

This section discusses the peripheral component interconnect (PCI) pull-up clamping diode option, slew-rate control, open-drain output option, MultiVolt I/O interface, and power sequencing for FLEX 10K devices. The PCI pull-up clamping diode, slew-rate control, and open-drain output options are controlled pin-by-pin via Altera logic options. The MultiVolt I/O interface is controlled by connecting V_{CCIO} to a different voltage than V_{CCINT}. Its effect can be simulated in the Altera software via the **Global Project Device Options** dialog box (Assign menu).

PCI Clamping Diodes

The EPF10K10A and EPF10K30A devices have a pull-up clamping diode on every I/O, dedicated input, and dedicated clock pin. PCI clamping diodes clamp the transient overshoot caused by reflected waves to the V_{CCIO} value and are required for 3.3-V PCI compliance. Clamping diodes can also be used to limit overshoot in other systems.

Clamping diodes are controlled on a pin-by-pin basis via a logic option in the Altera software. When V_{CCIO} is 3.3 V, a pin that has the clamping diode turned on can be driven by a 2.5-V or 3.3-V signal, but not a 5.0-V signal. When V_{CCIO} is 2.5 V, a pin that has the clamping diode turned on can be driven by a 2.5-V signal, but not a 3.3-V or 5.0-V signal. However, a clamping diode can be turned on for a subset of pins, which allows devices to bridge between a 3.3-V PCI bus and a 5.0-V device.

Figure 18 shows the timing requirements for the JTAG signals.

Figure 18. JTAG Waveforms

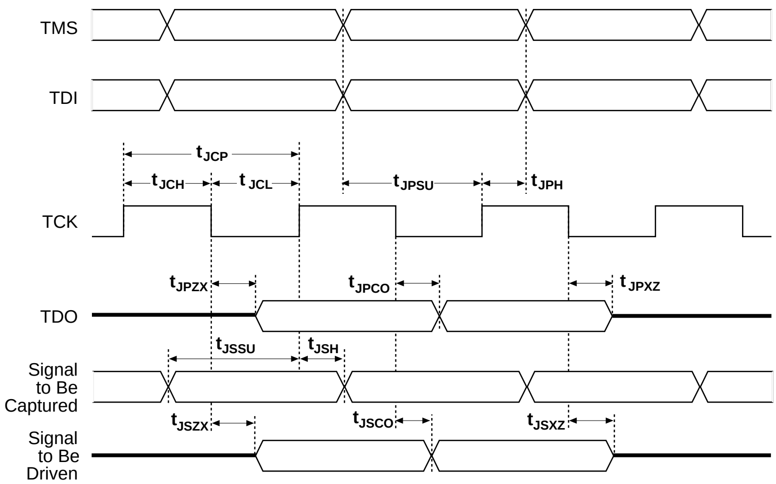


Table 16 shows the timing parameters and values for FLEX 10K devices.

Table 16. JTAG Timing Parameters & Values

Symbol	Parameter	Min	Max	Unit
t_{JCP}	TCK clock period	100		ns
t_{JCH}	TCK clock high time	50		ns
t_{JCL}	TCK clock low time	50		ns
t_{JPSU}	JTAG port setup time	20		ns
t_{JPH}	JTAG port hold time	45		ns
t_{JPCO}	JTAG port clock to output		25	ns
t_{JPZX}	JTAG port high impedance to valid output		25	ns
t_{JPXZ}	JTAG port valid output to high impedance		25	ns
t_{JSSU}	Capture register setup time	20		ns
t_{JSH}	Capture register hold time	45		ns
t_{JSCO}	Update register clock to output		35	ns
t_{JSZX}	Update register high-impedance to valid output		35	ns
t_{JSXZ}	Update register valid output to high impedance		35	ns

Table 28. FLEX 10KA 3.3-V Device DC Operating Conditions *Notes (6), (7)*						
Symbol	**Parameter**	**Conditions**	**Min**	**Typ**	**Max**	**Unit**
V_{IH}	High-level input voltage		1.7 or $0.5 \times V_{CCINT}$, whichever is lower		5.75	V
V_{IL}	Low-level input voltage		–0.5		$0.3 \times V_{CCINT}$	V
V_{OH}	3.3-V high-level TTL output voltage	I_{OH} = –11 mA DC, V_{CCIO} = 3.00 V *(8)*	2.4			V
	3.3-V high-level CMOS output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 3.00 V *(8)*	$V_{CCIO} - 0.2$			V
	3.3-V high-level PCI output voltage	I_{OH} = –0.5 mA DC, V_{CCIO} = 3.00 to 3.60 V *(8)*	$0.9 \times V_{CCIO}$			V
	2.5-V high-level output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 2.30 V *(8)*	2.1			V
		I_{OH} = –1 mA DC, V_{CCIO} = 2.30 V *(8)*	2.0			V
		I_{OH} = –2 mA DC, V_{CCIO} = 2.30 V *(8)*	1.7			V
V_{OL}	3.3-V low-level TTL output voltage	I_{OL} = 9 mA DC, V_{CCIO} = 3.00 V *(9)*			0.45	V
	3.3-V low-level CMOS output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 3.00 V *(9)*			0.2	V
	3.3-V low-level PCI output voltage	I_{OL} = 1.5 mA DC, V_{CCIO} = 3.00 to 3.60 V *(9)*			$0.1 \times V_{CCIO}$	V
	2.5-V low-level output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 2.30 V *(9)*			0.2	V
		I_{OL} = 1 mA DC, V_{CCIO} = 2.30 V *(9)*			0.4	V
		I_{OL} = 2 mA DC, V_{CCIO} = 2.30 V *(9)*			0.7	V
I_I	Input pin leakage current	V_I = 5.3 V to –0.3 V *(10)*	–10		10	μA
I_{OZ}	Tri-stated I/O pin leakage current	V_O = 5.3 V to –0.3 V *(10)*	–10		10	μA
I_{CC0}	V_{CC} supply current (standby)	V_I = ground, no load		0.3	10	mA
		V_I = ground, no load *(11)*		10		mA

Figure 26. FLEX 10K Device IOE Timing Model

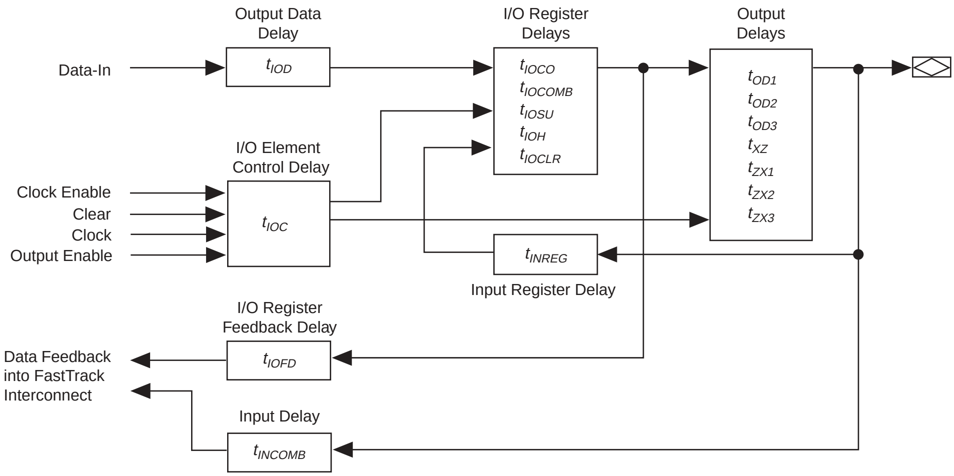
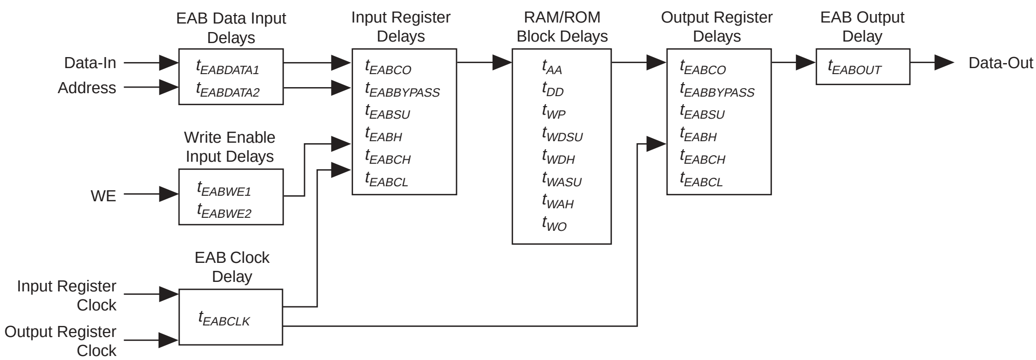


Figure 27. FLEX 10K Device EAB Timing Model



Figures 28 shows the timing model for bidirectional I/O pin timing.

Table 34. EAB Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
$t_{EABDATA1}$	Data or address delay to EAB for combinatorial input	
$t_{EABDATA2}$	Data or address delay to EAB for registered input	
t_{EABWE1}	Write enable delay to EAB for combinatorial input	
t_{EABWE2}	Write enable delay to EAB for registered input	
t_{EABCLK}	EAB register clock delay	
t_{EABCO}	EAB register clock-to-output delay	
$t_{EABBYPASS}$	Bypass register delay	
t_{EABSU}	EAB register setup time before clock	
t_{EABH}	EAB register hold time after clock	
t_{AA}	Address access delay	
t_{WP}	Write pulse width	
t_{WDSU}	Data setup time before falling edge of write pulse	*(5)*
t_{WDH}	Data hold time after falling edge of write pulse	*(5)*
t_{WASU}	Address setup time before rising edge of write pulse	*(5)*
t_{WAH}	Address hold time after falling edge of write pulse	*(5)*
t_{WO}	Write enable to data output valid delay	
t_{DD}	Data-in to data-out valid delay	
t_{EABOUT}	Data-out delay	
t_{EABCH}	Clock high time	
t_{EABCL}	Clock low time	

Table 49. EPF10K30, EPF10K40 & EPF10K50 Device IOE Timing Microparameters *Note (1)*					
Symbol	**-3 Speed Grade**		**-4 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	
t_{IOD}		0.4		0.6	ns
t_{IOC}		0.5		0.9	ns
t_{IOCO}		0.4		0.5	ns
t_{IOCOMB}		0.0		0.0	ns
t_{IOSU}	3.1		3.5		ns
t_{IOH}	1.0		1.9		ns
t_{IOCLR}		1.0		1.2	ns
t_{OD1}		3.3		3.6	ns
t_{OD2}		5.6		6.5	ns
t_{OD3}		7.0		8.3	ns
t_{XZ}		5.2		5.5	ns
t_{ZX1}		5.2		5.5	ns
t_{ZX2}		7.5		8.4	ns
t_{ZX3}		8.9		10.2	ns
t_{INREG}		7.7		10.0	ns
t_{IOFD}		3.3		4.0	ns
t_{INCOMB}		3.3		4.0	ns

Table 51. EPF10K30, EPF10K40 & EPF10K50 Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
t_{EABAA}		13.7		17.0	ns
$t_{EABRCCOMB}$	13.7		17.0		ns
$t_{EABRCREG}$	9.7		11.9		ns
t_{EABWP}	5.8		7.2		ns
$t_{EABWCCOMB}$	7.3		9.0		ns
$t_{EABWCREG}$	13.0		16.0		ns
t_{EABDD}		10.0		12.5	ns
$t_{EABDATACO}$		2.0		3.4	ns
$t_{EABDATASU}$	5.3		5.6		ns
$t_{EABDATAH}$	0.0		0.0		ns
$t_{EABWESU}$	5.5		5.8		ns
t_{EABWEH}	0.0		0.0		ns
$t_{EABWDSU}$	5.5		5.8		ns
t_{EABWDH}	0.0		0.0		ns
$t_{EABWASU}$	2.1		2.7		ns
t_{EABWAH}	0.0		0.0		ns
t_{EABWO}		9.5		11.8	ns

Table 52. EPF10K30 Device Interconnect Timing Microparameters *Note (1)*

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
$t_{DIN2IOE}$		6.9		8.7	ns
t_{DIN2LE}		3.6		4.8	ns
$t_{DIN2DATA}$		5.5		7.2	ns
$t_{DCLK2IOE}$		4.6		6.2	ns
$t_{DCLK2LE}$		3.6		4.8	ns
$t_{SAMELAB}$		0.3		0.3	ns
$t_{SAMEROW}$		3.3		3.7	ns
$t_{SAMECOLUMN}$		2.5		2.7	ns
$t_{DIFFROW}$		5.8		6.4	ns
$t_{TWOROWS}$		9.1		10.1	ns
$t_{LEPERIPH}$		6.2		7.1	ns
$t_{LABCARRY}$		0.4		0.6	ns
$t_{LABCASC}$		2.4		3.0	ns

Table 53. EPF10K40 Device Interconnect Timing Microparameters *Note (1)*

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
$t_{DIN2IOE}$		7.6		9.4	ns
t_{DIN2LE}		3.6		4.8	ns
$t_{DIN2DATA}$		5.5		7.2	ns
$t_{DCLK2IOE}$		4.6		6.2	ns
$t_{DCLK2LE}$		3.6		4.8	ns
$t_{SAMELAB}$		0.3		0.3	ns
$t_{SAMEROW}$		3.3		3.7	ns
$t_{SAMECOLUMN}$		3.1		3.2	ns
$t_{DIFFROW}$		6.4		6.4	ns
$t_{TWOROWS}$		9.7		10.6	ns
$t_{LEPERIPH}$		6.4		7.1	ns
$t_{LABCARRY}$		0.4		0.6	ns
$t_{LABCASC}$		2.4		3.0	ns

Table 59. EPF10K70 Device EAB Internal Microparameters *Note (1)*

Symbol	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.3		1.5		1.9	ns
$t_{EABDATA2}$		4.3		4.8		6.0	ns
t_{EABWE1}		0.9		1.0		1.2	ns
t_{EABWE2}		4.5		5.0		6.2	ns
t_{EABCLK}		0.9		1.0		2.2	ns
t_{EABCO}		0.4		0.5		0.6	ns
$t_{EABBYPASS}$		1.3		1.5		1.9	ns
t_{EABSU}	1.3		1.5		1.8		ns
t_{EABH}	1.8		2.0		2.5		ns
t_{AA}		7.8		8.7		10.7	ns
t_{WP}	5.2		5.8		7.2		ns
t_{WDSU}	1.4		1.6		2.0		ns
t_{WDH}	0.3		0.3		0.4		ns
t_{WASU}	0.4		0.5		0.6		ns
t_{WAH}	0.9		1.0		1.2		ns
t_{WO}		4.5		5.0		6.2	ns
t_{DD}		4.5		5.0		6.2	ns
t_{EABOUT}		0.4		0.5		0.6	ns
t_{EABCH}	4.0		4.0		4.0		ns
t_{EABCL}	5.2		5.8		7.2		ns

Table 67. EPF10K100 Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-3DX Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		13.7		13.7		17.0	ns
$t_{EABRCCOMB}$	13.7		13.7		17.0		ns
$t_{EABRCREG}$	9.7		9.7		11.9		ns
t_{EABWP}	5.8		5.8		7.2		ns
$t_{EABWCCOMB}$	7.3		7.3		9.0		ns
$t_{EABWCREG}$	13.0		13.0		16.0		ns
t_{EABDD}		10.0		10.0		12.5	ns
$t_{EABDATACO}$		2.0		2.0		3.4	ns
$t_{EABDATASU}$	5.3		5.3		5.6		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	5.5		5.5		5.8		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	5.5		5.5		5.8		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	2.1		2.1		2.7		ns
t_{EABWAH}	0.0		0.0		0.0		ns
t_{EABWO}		9.5		9.5		11.8	ns

Table 72. EPF10K50V Device IOE Timing Microparameters Note (1)									
Symbol	**-1 Speed Grade**		**-2 Speed Grade**		**-3 Speed Grade**		**-4 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	**Min**	**Max**	**Min**	**Max**	
t_{IOD}		1.2		1.6		1.9		2.1	ns
t_{IOC}		0.3		0.4		0.5		0.5	ns
t_{IOCO}		0.3		0.3		0.4		0.4	ns
t_{IOCOMB}		0.0		0.0		0.0		0.0	ns
t_{IOSU}	2.8		2.8		3.4		3.9		ns
t_{IOH}	0.7		0.8		1.0		1.4		ns
t_{IOCLR}		0.5		0.6		0.7		0.7	ns
t_{OD1}		2.8		3.2		3.9		4.7	ns
t_{OD2}		–		–		–		–	ns
t_{OD3}		6.5		6.9		7.6		8.4	ns
t_{XZ}		2.8		3.1		3.8		4.6	ns
t_{ZX1}		2.8		3.1		3.8		4.6	ns
t_{ZX2}		–		–		–		–	ns
t_{ZX3}		6.5		6.8		7.5		8.3	ns
t_{INREG}		5.0		5.7		7.0		9.0	ns
t_{IOFD}		1.5		1.9		2.3		2.7	ns
t_{INCOMB}		1.5		1.9		2.3		2.7	ns

Notes to tables:
(1) All timing parameters are described in Tables 32 through 37 in this data sheet.
(2) Using an LE to register the signal may provide a lower setup time.
(3) This parameter is specified by characterization.

ClockLock & ClockBoost Timing Parameters

For the ClockLock and ClockBoost circuitry to function properly, the incoming clock must meet certain requirements. If these specifications are not met, the circuitry may not lock onto the incoming clock, which generates an erroneous clock within the device. The clock generated by the ClockLock and ClockBoost circuitry must also meet certain specifications. If the incoming clock meets these requirements during configuration, the ClockLock and ClockBoost circuitry will lock onto the clock during configuration. The circuit will be ready for use immediately after configuration. Figure 31 illustrates the incoming and generated clock specifications.

Figure 31. Specifications for the Incoming & Generated Clocks

The t_I parameter refers to the nominal input clock period; the t_O parameter refers to the nominal output clock period.

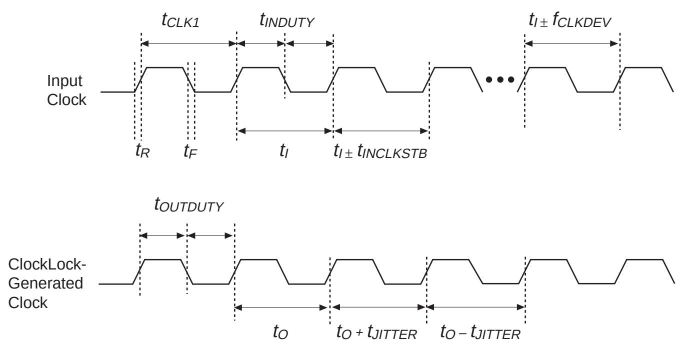


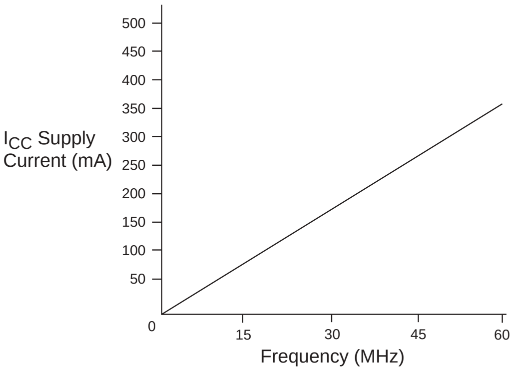
Table 113 summarizes the ClockLock and ClockBoost parameters.

Table 113. ClockLock & ClockBoost Parameters (Part 1 of 2)

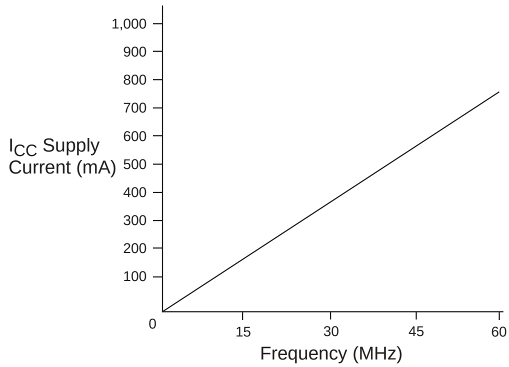
Symbol	Parameter	Min	Typ	Max	Unit
t_R	Input rise time			2	ns
t_F	Input fall time			2	ns
t_{INDUTY}	Input duty cycle	45		55	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)	30		80	MHz
t_{CLK1}	Input clock period (ClockBoost clock multiplication factor equals 1)	12.5		33.3	ns
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)	16		50	MHz
t_{CLK2}	Input clock period (ClockBoost clock multiplication factor equals 2)	20		62.5	ns

Figure 32. $I_{CCACTIVE}$ vs. Operating Frequency (Part 1 of 3)

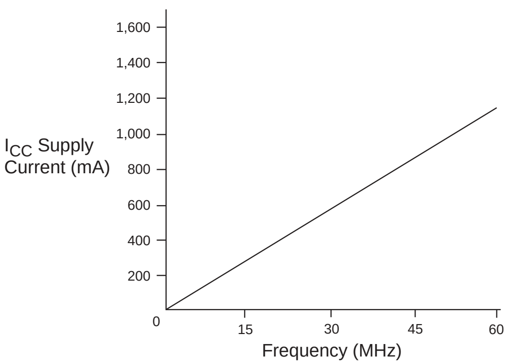
EPF10K10



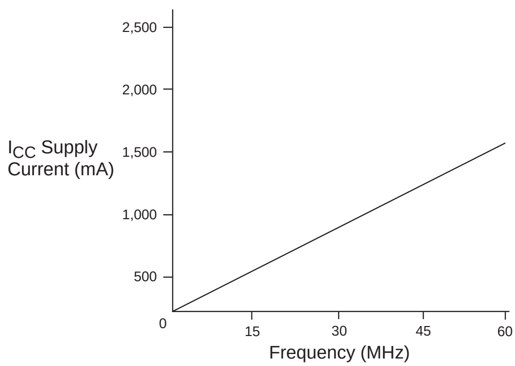
EPF10K20



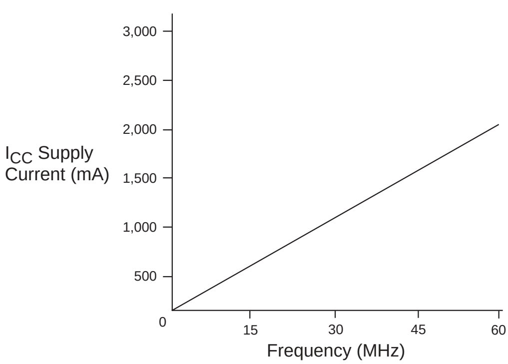
EPF10K30



EPF10K40



EPF10K50



EPF10K70