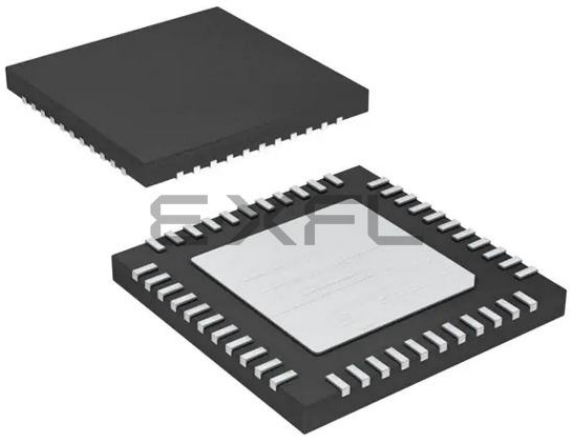

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Peripherals	Brown-out Detect/Reset, Motor Control PWM, QEI, POR, PWM, WDT
Number of I/O	30
Program Memory Size	24KB (8K x 24)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	2.5V ~ 5.5V
Data Converters	A/D 9x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	44-VQFN Exposed Pad
Supplier Device Package	44-QFN (8x8)
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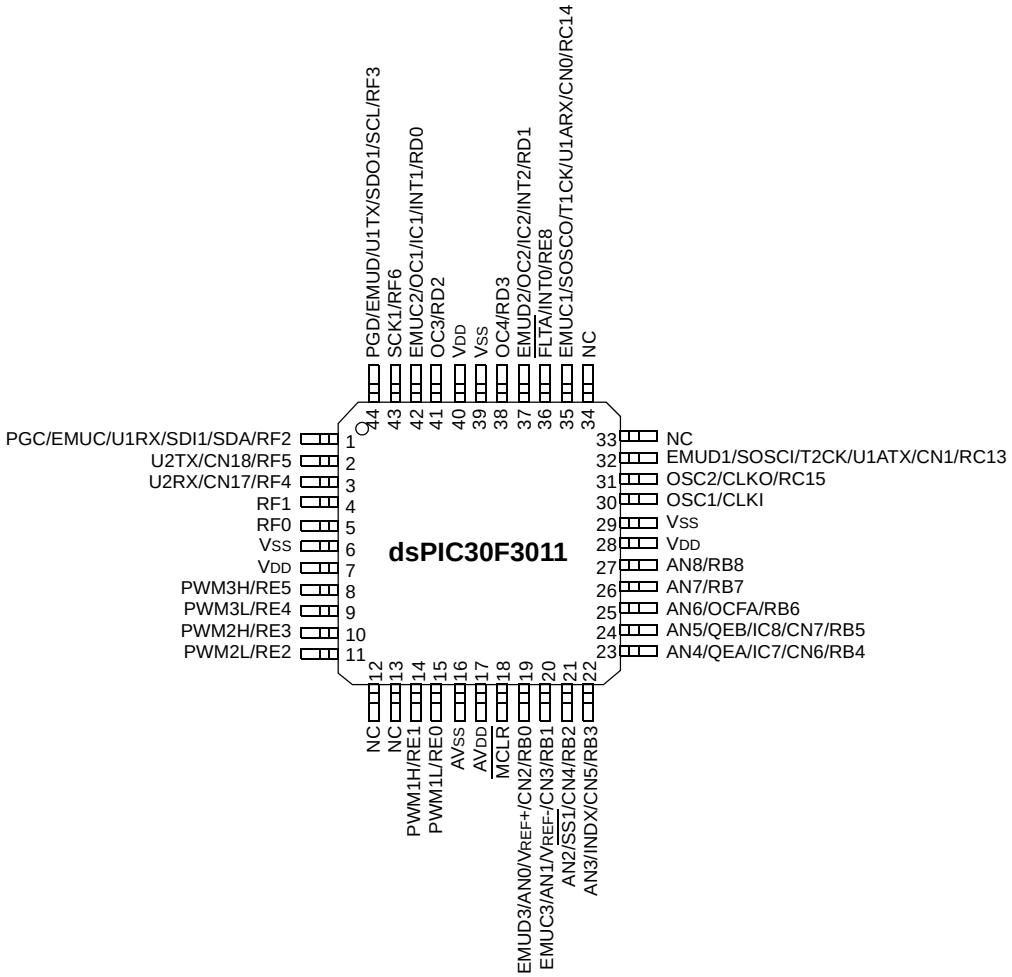
dsPIC30F Motor Control and Power Conversion Family

Device	Pins	Program Mem. Bytes/ Instructions	SRAM Bytes	EEPROM Bytes	Timer 16-Bit	Input Cap	Output Comp/Std PWM	Motor Control PWM	A/D 10-Bit 1 Msps	Quad Enc	UART	SPI	I²C™
dsPIC30F3010	28	24K/8K	1024	1024	5	4	2	6 ch	6 ch	Yes	1	1	1
dsPIC30F3011	40/44	24K/8K	1024	1024	5	4	4	6 ch	9 ch	Yes	2	1	1

Pin Diagrams (Continued)

44-Pin TQFP



NOTES:

NOTES:

3.2.2 DATA SPACES

The X data space is used by all instructions and supports all addressing modes. There are separate read and write data buses. The X read data bus is the return data path for all instructions that view data space as combined X and Y address space. It is also the X address space data path for the dual operand read instructions (MAC class). The X write data bus is the only write path to data space for all instructions.

The X data space also supports Modulo Addressing for all instructions, subject to addressing mode restrictions. Bit-Reversed Addressing is only supported for writes to X data space.

The Y data space is used in concert with the X data space by the MAC class of instructions (CLR, ED, EDAC, MAC, MOVSAC, MPY, MPY.N and MSC) to provide two concurrent data read paths. No writes occur across the Y bus. This class of instructions dedicates two W register pointers, W10 and W11, to always address Y data space, independent of X data space, whereas W8 and W9 always address X data space. Note that during accumulator write back, the data address space is considered a combination of X and Y data spaces, so the write occurs across the X bus. Consequently, the write can be to any address in the entire data space.

The Y data space can only be used for the data prefetch operation associated with the MAC class of instructions. It also supports Modulo Addressing for automated circular buffers. Of course, all other instructions can access the Y data address space through the X data path, as part of the composite linear space.

The boundary between the X and Y data spaces is defined as shown in Figure 3-6 and is not user-programmable. Should an EA point to data outside its own assigned address space, or to a location outside physical memory, an all zero word/byte will be returned. For example, although Y address space is visible by all non-MAC instructions using any addressing mode, an attempt by a MAC instruction to fetch data from that space, using W8 or W9 (X Space Pointers), will return 0x0000.

TABLE 3-2: EFFECT OF INVALID MEMORY ACCESSES

Attempted Operation	Data Returned
EA = an unimplemented address	0x0000
W8 or W9 used to access Y data space in a MAC instruction	0x0000
W10 or W11 used to access X data space in a MAC instruction	0x0000

All effective addresses are 16 bits wide and point to bytes within the data space. Therefore, the data space address range is 64 Kbytes or 32K words.

3.2.3 DATA SPACE WIDTH

The core data width is 16 bits. All internal registers are organized as 16-bit wide words. Data space memory is organized in byte addressable, 16-bit wide blocks.

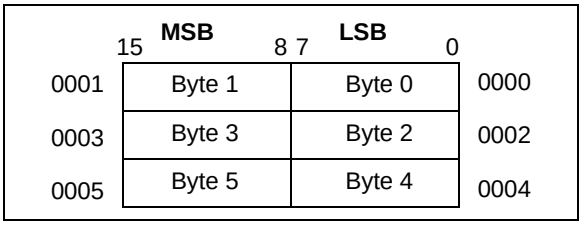
3.2.4 DATA ALIGNMENT

To help maintain backward compatibility with PIC® MCU devices and improve data space memory usage efficiency, the dsPIC30F instruction set supports both word and byte operations. Data is aligned in data memory and registers as words, but all data space EAs resolve to bytes. Data byte reads will read the complete word, which contains the byte, using the LSb of any EA to determine which byte to select. The selected byte is placed onto the LSB of the X data path (no byte accesses are possible from the Y data path as the MAC class of instruction can only fetch words). That is, data memory and registers are organized as two parallel byte-wide entities with shared (word) address decode, but separate write lines. Data byte writes only write to the corresponding side of the array or register which matches the byte address.

As a consequence of this byte accessibility, all effective address calculations (including those generated by the DSP operations, which are restricted to word-sized data) are internally scaled to step through word-aligned memory. For example, the core would recognize that Post-Modified Register Indirect Addressing mode, [Ws++], will result in a value of Ws + 1 for byte operations and Ws + 2 for word operations.

All word accesses must be aligned to an even address. Misaligned word data fetches are not supported, so care must be taken when mixing byte and word operations, or translating from 8-bit MCU code. Should a misaligned read or write be attempted, an address error trap will be generated. If the error occurred on a read, the instruction underway is completed, whereas if it occurred on a write, the instruction will be executed but the write will not occur. In either case, a trap will then be executed, allowing the system and/or user to examine the machine state prior to execution of the address Fault.

FIGURE 3-8: DATA ALIGNMENT



4.0 ADDRESS GENERATOR UNITS

Note: This data sheet summarizes features of this group of dsPIC30F devices and is not intended to be a complete reference source. For more information on the CPU, peripherals, register descriptions and general device functionality, refer to the "*dsPIC30F Family Reference Manual*" (DS70046). For more information on the device instruction set and programming, refer to the "*16-bit MCU and DSC Programmer's Reference Manual*" (DS70157).

The dsPIC DSC core contains two independent address generator units: the X AGU and Y AGU. The Y AGU supports word-sized data reads for the DSP MAC class of instructions only. The dsPIC DSC AGUs support three types of data addressing:

- Linear Addressing
- Modulo (Circular) Addressing
- Bit-Reversed Addressing

Linear and Modulo Data Addressing modes can be applied to data space or program space. Bit-Reversed Addressing is only applicable to data space addresses.

4.1 Instruction Addressing Modes

The addressing modes in Table 4-1 form the basis of the addressing modes optimized to support the specific features of individual instructions. The addressing modes provided in the MAC class of instructions are somewhat different from those in the other instruction types.

4.1.1 FILE REGISTER INSTRUCTIONS

Most file register instructions use a 13-bit address field (f) to directly address data present in the first 8192 bytes of data memory (near data space). Most file register instructions employ a working register, W0, which is denoted as WREG in these instructions. The destination is typically either the same file register, or WREG (with the exception of the MUL instruction), which writes the result to a register or register pair. The MOV instruction allows additional flexibility and can access the entire data space during file register operation.

4.1.2 MCU INSTRUCTIONS

The three-operand MCU instructions are of the form:

```
Operand 3 = Operand 1 <function> Operand 2
```

where Operand 1 is always a working register (i.e., the addressing mode can only be Register Direct), which is referred to as Wb. Operand 2 can be a W register, fetched from data memory, or a 5-bit literal. The result location can be either a W register or an address location. The following addressing modes are supported by MCU instructions:

- Register Direct
- Register Indirect
- Register Indirect Post-Modified
- Register Indirect Pre-Modified
- 5-bit or 10-bit Literal

Note: Not all instructions support all the addressing modes given above. Individual instructions may support different subsets of these addressing modes.

TABLE 4-1: FUNDAMENTAL ADDRESSING MODES SUPPORTED

Addressing Mode	Description
File Register Direct	The address of the file register is specified explicitly.
Register Direct	The contents of a register are accessed directly.
Register Indirect	The contents of Wn forms the EA.
Register Indirect Post-Modified	The contents of Wn forms the EA. Wn is post-modified (incremented or decremented) by a constant value.
Register Indirect Pre-Modified	Wn is pre-modified (incremented or decremented) by a signed constant value to form the EA.
Register Indirect with Register Offset	The sum of Wn and Wb forms the EA.
Register Indirect with Literal Offset	The sum of Wn and a literal forms the EA.

7.2 Erasing Data EEPROM

7.2.1 ERASING A BLOCK OF DATA EEPROM

In order to erase a block of data EEPROM, the NVMADRU and NVMADR registers must initially point to the block of memory to be erased. Configure NVMCON for erasing a block of data EEPROM, and set the ERASE and WREN bits in the NVMCON register. Setting the WR bit initiates the erase, as shown in Example 7-2.

EXAMPLE 7-2: DATA EEPROM BLOCK ERASE

```
; Select data EEPROM block, ERASE, WREN bits
    MOV     #0x4045,W0
    MOV     W0,NVMCON                       ; Initialize NVMCON SFR

; Start erase cycle by setting WR after writing key sequence
    DISI    #5                              ; Block all interrupts with priority <7
                                            ; for next 5 instructions
    MOV     #0x55,W0                        ;
    MOV     W0,NVMKEY                       ; Write the 0x55 key
    MOV     #0xAA,W1                        ;
    MOV     W1,NVMKEY                       ; Write the 0xAA key
    BSET    NVMCON,#WR                      ; Initiate erase sequence
    NOP
    NOP
; Erase cycle will complete in 2mS. CPU is not stalled for the Data Erase Cycle
; User can poll WR bit, use NVMIF or Timer IRQ to determine erasure complete
```

7.2.2 ERASING A WORD OF DATA EEPROM

The TBLPAG and NVMADR registers must point to the block. Select erase a block of data Flash, and set the ERASE and WREN bits in the NVMCON register. Setting the WR bit initiates the erase, as shown in Example 7-3.

EXAMPLE 7-3: DATA EEPROM WORD ERASE

```
; Select data EEPROM word, ERASE, WREN bits
    MOV     #0x4044,W0
    MOV     W0,NVMCON

; Start erase cycle by setting WR after writing key sequence
    DISI    #5                              ; Block all interrupts with priority <7
                                            ; for next 5 instructions
    MOV     #0x55,W0                ;
    MOV     W0,NVMKEY               ; Write the 0x55 key
    MOV     #0xAA,W1                ;
    MOV     W1,NVMKEY               ; Write the 0xAA key
    BSET    NVMCON,#WR              ; Initiate erase sequence
    NOP
    NOP
; Erase cycle will complete in 2mS. CPU is not stalled for the Data Erase Cycle
; User can poll WR bit, use NVMIF or Timer IRQ to determine erasure complete
```

10.0 TIMER2/3 MODULE

Note:	This data sheet summarizes features of this group of dsPIC30F devices and is not intended to be a complete reference source. For more information on the CPU, peripherals, register descriptions and general device functionality, refer to the "*dsPIC30F Family Reference Manual*" (DS70046).

This section describes the 32-bit general purpose timer module (Timer2/3) and associated operational modes. Figure 10-1 depicts the simplified block diagram of the 32-bit Timer2/3 module. Figure 10-2 and Figure 10-3 show Timer2/3 configured as two independent 16-bit timers; Timer2 and Timer3, respectively.

Note:	Timer2 is a 'Type B' timer and Timer3 is a 'Type C' timer. Please refer to appropriate timer type in **Section 23.0 "Electrical Characteristics"**.

The Timer2/3 module is a 32-bit timer, which can be configured as two 16-bit timers, with selectable operating modes. These timers are utilized by other peripheral modules such as:

- Input Capture
- Output Compare/Simple PWM

The following sections provide a detailed description, including setup and control registers, along with associated block diagrams for the operational modes of the timers.

The 32-bit timer has the following modes:

- Two independent 16-bit timers (Timer2 and Timer3) with all 16-bit operating modes (except Asynchronous Counter mode)
- Single 32-bit timer operation
- Single 32-bit synchronous counter

Further, the following operational characteristics are supported:

- ADC Event Trigger
- Timer Gate Operation
- Selectable Prescaler Settings
- Timer Operation during Idle and Sleep modes
- Interrupt on a 32-Bit Period Register Match

These operating modes are determined by setting the appropriate bit(s) in the 16-bit T2CON and T3CON SFRs.

For 32-bit timer/counter operation, Timer2 is the lsw and Timer3 is the msw of the 32-bit timer.

Note:	For 32-bit timer operation, T3CON control bits are ignored. Only T2CON control bits are used for setup and control. Timer2 clock and gate inputs are utilized for the 32-bit timer module, but an interrupt is generated with the Timer3 Interrupt Flag (T3IF) and the interrupt is enabled with the Timer3 Interrupt Enable bit (T3IE).

16-Bit Mode: In the 16-bit mode, Timer2 and Timer3 can be configured as two independent 16-bit timers. Each timer can be set up in either 16-bit Timer mode or 16-bit Synchronous Counter mode. See **Section 9.0 "Timer1 Module"** for details on these two operating modes.

The only functional difference between Timer2 and Timer3 is that Timer2 provides synchronization of the clock prescaler output. This is useful for high-frequency external clock inputs.

32-Bit Timer Mode: In the 32-Bit Timer mode, the timer increments on every instruction cycle up to a match value, preloads into the combined 32-bit Period register, PR3/PR2, then resets to '0' and continues to count.

For synchronous 32-bit reads of the Timer2/Timer3 pair, reading the lsw (TMR2 register) will cause the msw to be read and latched into a 16-bit holding register, termed TMR3HLD.

For synchronous 32-bit writes, the holding register (TMR3HLD) must first be written to. When followed by a write to the TMR2 register, the contents of TMR3HLD will be transferred and latched into the MSB of the 32-bit timer (TMR3).

32-Bit Synchronous Counter Mode: In the 32-Bit Synchronous Counter mode, the timer increments on the rising edge of the applied external clock signal, which is synchronized with the internal phase clocks. The timer counts up to a match value preloaded in the combined 32-bit Period register, PR3/PR2, then resets to '0' and continues.

When the timer is configured for the Synchronous Counter mode of operation and the CPU goes into the Idle mode, the timer will stop incrementing unless the TSIDL (T2CON<13>) bit = 0. If TSIDL = 1, the timer module logic will resume the incrementing sequence upon termination of the CPU Idle mode.

16.2 Framed SPI Support

The module supports a basic framed SPI protocol in Master or Slave mode. The control bit, FRMEN, enables framed SPI support and causes the $\overline{SS1}$ pin to perform the Frame Synchronization (FSYNC) pulse function. The control bit, SPIFSD, determines whether the $\overline{SS1}$ pin is an input or an output (i.e., whether the module receives or generates the frame synchronization pulse). The frame pulse is an active-high pulse for a single SPI clock cycle. When frame synchronization is enabled, the data transmission starts only on the subsequent transmit edge of the SPI clock.

FIGURE 16-1: SPI BLOCK DIAGRAM

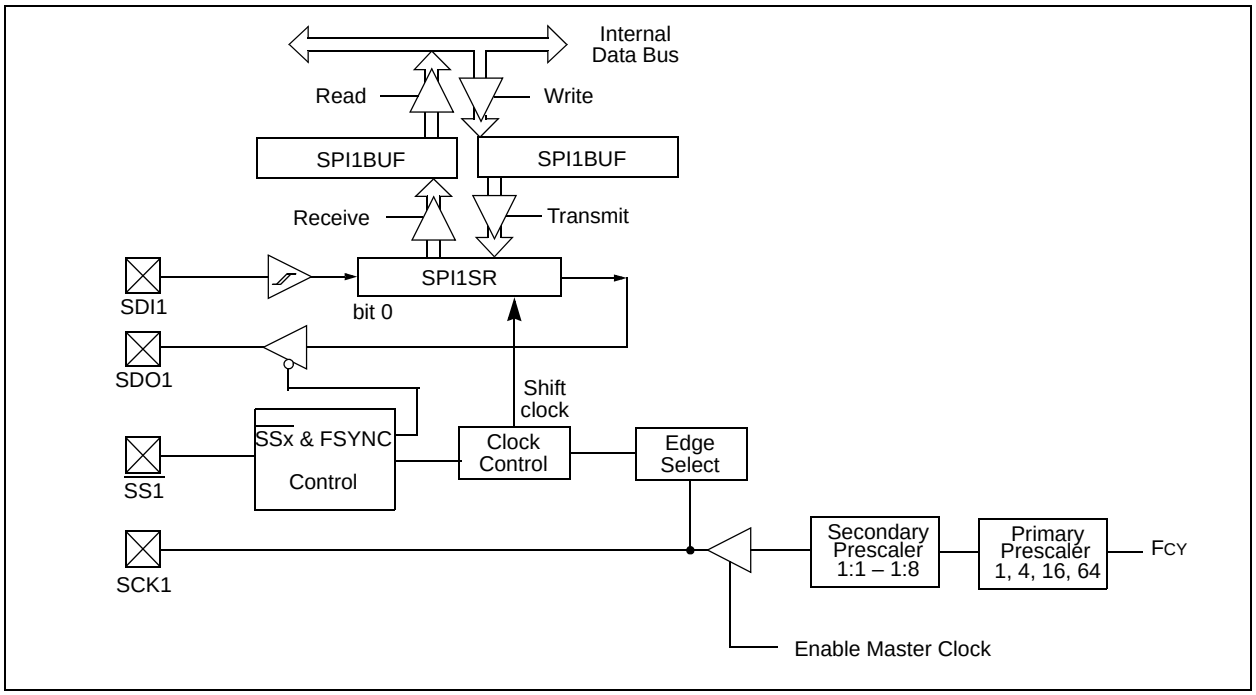
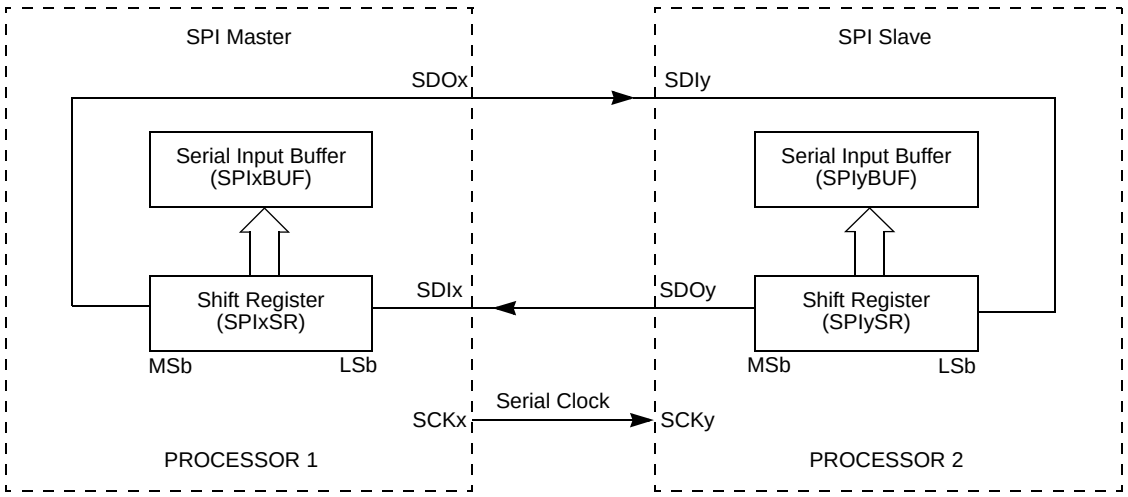


FIGURE 16-2: SPI MASTER/SLAVE CONNECTION



Note: x = 1 or 2, y = 1 or 2.

19.1 ADC Result Buffer

The module contains a 16-word, dual port, read-only buffer, called ADCBUF0...ADCBUFF, to buffer the ADC results. The RAM is 10 bits wide, but is read into different format 16-bit words. The contents of the sixteen ADC Conversion Result Buffer registers, ADCBUF0 through ADCBUFF, cannot be written by user software.

19.2 Conversion Operation

After the ADC module has been configured, the sample acquisition is started by setting the SAMP bit. Various sources, such as a programmable bit, timer time-outs and external events, will terminate acquisition and start a conversion. When the A/D conversion is complete, the result is loaded into ADCBUF0...ADCBUFF, and the A/D Interrupt Flag, ADIF, and the DONE bit are set after the number of samples specified by the SMPI bit.

The following steps should be followed for doing an A/D conversion:

- Configure the ADC module:
  - Configure analog pins, voltage reference and digital I/O
  - Select A/D input channels
  - Select A/D conversion clock
  - Select A/D conversion trigger
  - Turn on A/D module
- Configure A/D interrupt (if required):
  - Clear ADIF bit
  - Select A/D interrupt priority
- Start sampling
- Wait the required acquisition time
- Trigger acquisition end; start conversion
- Wait for A/D conversion to complete, by either:
  - Waiting for the A/D interrupt
  - Waiting for the DONE bit to be set
- Read A/D result buffer; clear ADIF if required

19.3 Selecting the Conversion Sequence

Several groups of control bits select the sequence in which the A/D connects inputs to the sample/hold channels, converts channels, writes the buffer memory and generates interrupts. The sequence is controlled by the sampling clocks.

The SIMSAM bit controls the acquire/convert sequence for multiple channels. If the SIMSAM bit is '0', the two or four selected channels are acquired and converted sequentially, with two or four sample clocks. If the SIMSAM bit is '1', two or four selected channels are acquired simultaneously, with one sample clock. The channels are then converted sequentially. Obviously, if there is only 1 channel selected, the SIMSAM bit is not applicable.

The CHPS bits select how many channels are sampled. This can vary from 1, 2 or 4 channels. If the CHPS bits select 1 channel, the CH0 channel will be sampled at the sample clock and converted. The result is stored in the buffer. If the CHPS bits select 2 channels, the CH0 and CH1 channels will be sampled and converted. If the CHPS bits select 4 channels, the CH0, CH1, CH2 and CH3 channels will be sampled and converted.

The SMPI bits select the number of acquisition/conversion sequences that would be performed before an interrupt occurs. This can vary from 1 sample per interrupt to 16 samples per interrupt.

The user cannot program a combination of CHPS and SMPI bits that specifies more than 16 conversions per interrupt, or 8 conversions per interrupt, depending on the BUFM bit. The BUFM bit, when set, will split the 16-word results buffer (ADCBUF0...ADCBUFF) into two 8-word groups. Writing to the 8-word buffers will be alternated on each interrupt event. Use of the BUFM bit will depend on how much time is available for moving data out of the buffers after the interrupt, as determined by the application.

If the processor can quickly unload a full buffer within the time it takes to acquire and convert one channel, the BUFM bit can be '0' and up to 16 conversions may be done per interrupt. The processor will have one sample and conversion time to move the sixteen conversions.

If the processor cannot unload the buffer within the acquisition and conversion time, the BUFM bit should be '1'. For example, if SMPI<3:0> (ADCON2<5:2>) = 0111, then eight conversions will be loaded into 1/2 of the buffer, following which an interrupt occurs. The next eight conversions will be loaded into the other 1/2 of the buffer. The processor will have the entire time between interrupts to move the eight conversions.

The ALTS bit can be used to alternate the inputs selected during the sampling sequence. The input multiplexer has two sets of sample inputs: MUX A and MUX B. If the ALTS bit is '0', only the MUX A inputs are selected for sampling. If the ALTS bit is '1' and SMPI<3:0> = 0000, on the first sample/convert sequence, the MUX A inputs are selected, and on the next acquire/convert sequence, the MUX B inputs are selected.

The CSCNA bit (ADCON2<10>) will allow the CH0 channel inputs to be alternately scanned across a selected number of analog inputs for the MUX A group. The inputs are selected by the ADCSSL register. If a particular bit in the ADCSSL register is '1', the corresponding input is selected. The inputs are always scanned from lower to higher numbered inputs, starting after each interrupt. If the number of inputs selected is greater than the number of samples taken per interrupt, the higher numbered inputs are unused.

FIGURE 20-1: OSCILLATOR SYSTEM BLOCK DIAGRAM

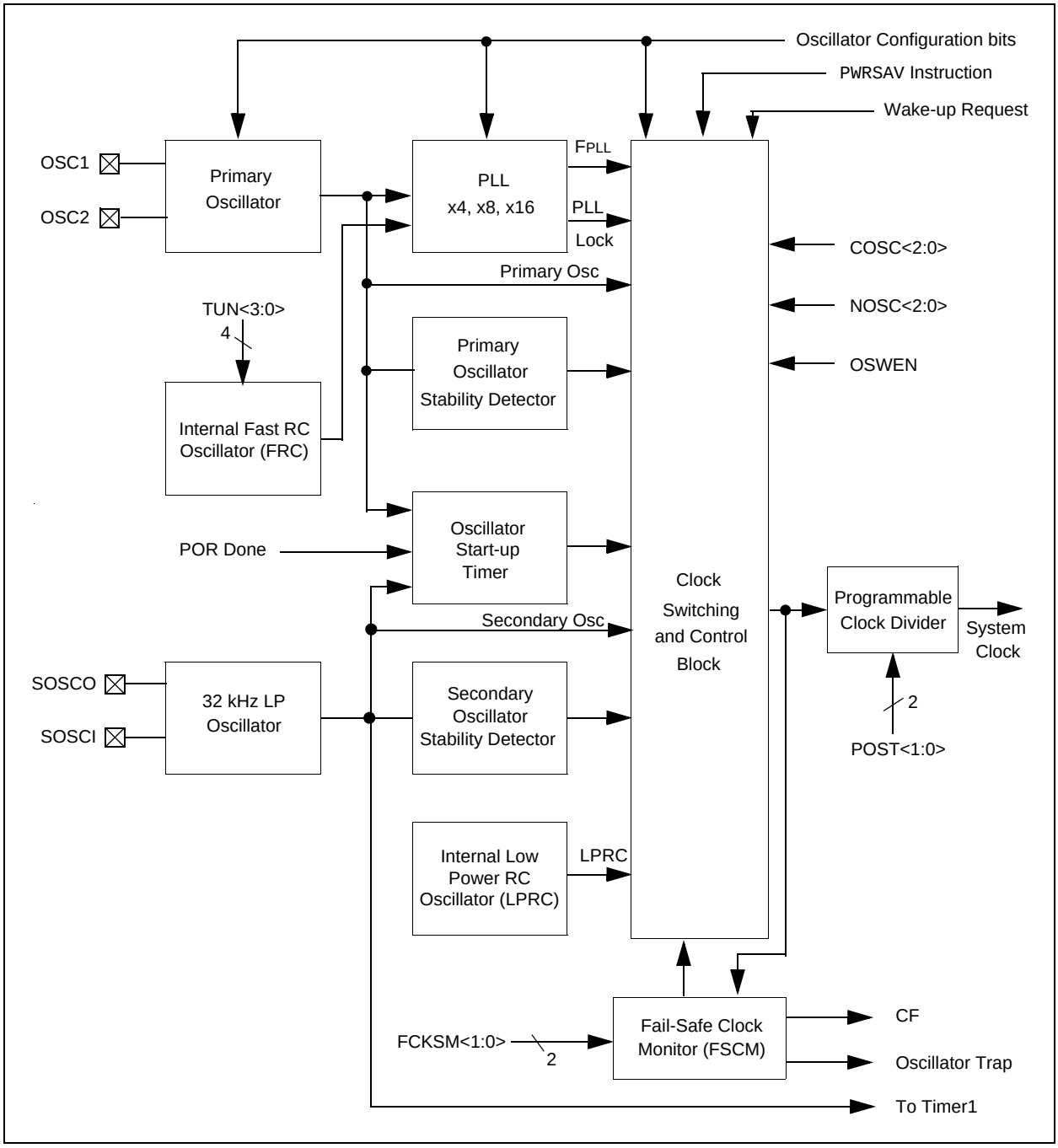


TABLE 21-1: SYMBOLS USED IN OPCODE DESCRIPTIONS (CONTINUED)

Field	Description
Wb	Base W register ∈ {W0..W15}
Wd	Destination W register ∈ { Wd, [Wd], [Wd++], [Wd--], [++Wd], [--Wd] }
Wdo	Destination W register ∈ { Wnd, [Wnd], [Wnd++], [Wnd--], [++Wnd], [--Wnd], [Wnd+Wb] }
Wm,Wn	Dividend, Divisor Working register pair (direct addressing)
Wm*Wm	Multiplicand and Multiplier working register pair for Square instructions ∈ {W4*W4,W5*W5,W6*W6,W7*W7}
Wm*Wn	Multiplicand and Multiplier working register pair for DSP instructions ∈ {W4*W5,W4*W6,W4*W7,W5*W6,W5*W7,W6*W7}
Wn	One of 16 working registers ∈ {W0..W15}
Wnd	One of 16 destination working registers ∈ {W0..W15}
Wns	One of 16 source working registers ∈ {W0..W15}
WREG	W0 (working register used in File register instructions)
Ws	Source W register ∈ { Ws, [Ws], [Ws++], [Ws--], [++Ws], [--Ws] }
Wso	Source W register ∈ { Wns, [Wns], [Wns++], [Wns--], [++Wns], [--Wns], [Wns+Wb] }
Wx	X data space Prefetch Address register for DSP instructions ∈ {[W8]+=6, [W8]+=4, [W8]+=2, [W8], [W8]-=6, [W8]-=4, [W8]-=2, [W9]+=6, [W9]+=4, [W9]+=2, [W9], [W9]-=6, [W9]-=4, [W9]-=2, [W9+W12],none}
Wxd	X data space Prefetch Destination register for DSP instructions ∈ {W4..W7}
Wy	Y data space Prefetch Address register for DSP instructions ∈ {[W10]+=6, [W10]+=4, [W10]+=2, [W10], [W10]-=6, [W10]-=4, [W10]-=2, [W11]+=6, [W11]+=4, [W11]+=2, [W11], [W11]-=6, [W11]-=4, [W11]-=2, [W11+W12], none}
Wyd	Y data space Prefetch Destination register for DSP instructions ∈ {W4..W7}

TABLE 21-2: INSTRUCTION SET OVERVIEW (CONTINUED)

Base Instr #	Assembly Mnemonic	Assembly Syntax	Description	# of words	# of cycles	Status Flags Affected
46	MOV	MOV f,Wn	Move f to Wn	1	1	None
		MOV f	Move f to f	1	1	N,Z
		MOV f,WREG	Move f to WREG	1	1	N,Z
		MOV #lit16,Wn	Move 16-bit Literal to Wn	1	1	None
		MOV.b #lit8,Wn	Move 8-bit Literal to Wn	1	1	None
		MOV Wn,f	Move Wn to f	1	1	None
		MOV Wso,Wdo	Move Ws to Wd	1	1	None
		MOV WREG,f	Move WREG to f	1	1	N,Z
		MOV.D Wns,Wd	Move Double from W(ns):W(ns + 1) to Wd	1	2	None
		MOV.D Ws,Wnd	Move Double from Ws to W(nd + 1):W(nd)	1	2	None
47	MOVSAC	MOVSAC Acc,Wx,Wxd,Wy,Wyd,AWB	Prefetch and Store Accumulator	1	1	None
48	MPY	MPY Wm*Wn,Acc,Wx,Wxd,Wy,Wyd	Multiply Wm by Wn to Accumulator	1	1	OA,OB,OAB, SA,SB,SAB
		MPY Wm*Wm,Acc,Wx,Wxd,Wy,Wyd	Square Wm to Accumulator	1	1	OA,OB,OAB, SA,SB,SAB
49	MPY.N	MPY.N Wm*Wn,Acc,Wx,Wxd,Wy,Wyd	-(Multiply Wm by Wn) to Accumulator	1	1	None
50	MSC	MSC Wm*Wm,Acc,Wx,Wxd,Wy,Wyd, AWB	Multiply and Subtract from Accumulator	1	1	OA,OB,OAB, SA,SB,SAB
51	MUL	MUL.SS Wb,Ws,Wnd	{Wnd+1, Wnd} = signed(Wb) * signed(Ws)	1	1	None
		MUL.SU Wb,Ws,Wnd	{Wnd+1, Wnd} = signed(Wb) * unsigned(Ws)	1	1	None
		MUL.US Wb,Ws,Wnd	{Wnd+1, Wnd} = unsigned(Wb) * signed(Ws)	1	1	None
		MUL.UU Wb,Ws,Wnd	{Wnd+1, Wnd} = unsigned(Wb) * unsigned(Ws)	1	1	None
		MUL.SU Wb,#lit5,Wnd	{Wnd+1, Wnd} = signed(Wb) * unsigned(lit5)	1	1	None
		MUL.UU Wb,#lit5,Wnd	{Wnd+1, Wnd} = unsigned(Wb) * unsigned(lit5)	1	1	None
		MUL f	W3:W2 = f * WREG	1	1	None
52	NEG	NEG Acc	Negate Accumulator	1	1	OA,OB,OAB, SA,SB,SAB
		NEG f	f = $\bar{f}$ + 1	1	1	C,DC,N,OV,Z
		NEG f,WREG	WREG = $\bar{f}$ + 1	1	1	C,DC,N,OV,Z
		NEG Ws,Wd	Wd = $\overline{Ws}$ + 1	1	1	C,DC,N,OV,Z
53	NOP	NOP	No Operation	1	1	None
		NOPR	No Operation	1	1	None
54	POP	POP f	Pop f from Top-of-Stack (TOS)	1	1	None
		POP Wdo	Pop from Top-of-Stack (TOS) to Wdo	1	1	None
		POP.D Wnd	Pop from Top-of-Stack (TOS) to W(nd):W(nd + 1)	1	2	None
		POP.S	Pop Shadow Registers	1	1	All
55	PUSH	PUSH f	Push f to Top-of-Stack (TOS)	1	1	None
		PUSH Wso	Push Wso to Top-of-Stack (TOS)	1	1	None
		PUSH.D Wns	Push W(ns):W(ns + 1) to Top-of-Stack (TOS)	1	2	None
		PUSH.S	Push Shadow Registers	1	1	None
56	PWRSAV	PWRSAV #lit1	Go into Sleep or Idle mode	1	1	WDTO,Sleep
57	RCALL	RCALL Expr	Relative Call	1	2	None
		RCALL Wn	Computed Call	1	2	None
58	REPEAT	REPEAT #lit14	Repeat Next Instruction lit14 + 1 Times	1	1	None
		REPEAT Wn	Repeat Next Instruction (Wn) + 1 Times	1	1	None
59	RESET	RESET	Software Device Reset	1	1	None
60	RETFIE	RETFIE	Return from Interrupt	1	3 (2)	None
61	RETLW	RETLW #lit10,Wn	Return with Literal in Wn	1	3 (2)	None
62	RETURN	RETURN	Return from Subroutine	1	3 (2)	None

TABLE 23-8: DC CHARACTERISTICS: I/O PIN INPUT SPECIFICATIONS

DC CHARACTERISTICS			Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for Industrial -40°C ≤ TA ≤ +125°C for Extended				
Param No.	**Symbol**	**Characteristic**	**Min**	**Typ(1)**	**Max**	**Units**	**Conditions**
	VIL	**Input Low Voltage(2)**					
DI10		I/O Pins: with Schmitt Trigger Buffer	VSS	—	0.2 VDD	V	
DI15		$\overline{\text{MCLR}}$	VSS	—	0.2 VDD	V	
DI16		OSC1 (in XT, HS and LP modes)	VSS	—	0.2 VDD	V	
DI17		OSC1 (in RC mode)(3)	VSS	—	0.3 VDD	V	
DI18		SDA, SCL	VSS	—	0.3 VDD	V	SMbus disabled
DI19		SDA, SCL	VSS	—	0.8	V	SMbus enabled
	VIH	**Input High Voltage(2)**					
DI20		I/O Pins: with Schmitt Trigger Buffer	0.8 VDD	—	VDD	V	
DI25		$\overline{\text{MCLR}}$	0.8 VDD	—	VDD	V	
DI26		OSC1 (in XT, HS and LP modes)	0.7 VDD	—	VDD	V	
DI27		OSC1 (in RC mode)(3)	0.9 VDD	—	VDD	V	
DI28		SDA, SCL	0.7 VDD	—	VDD	V	SMbus disabled
DI29		SDA, SCL	2.1	—	VDD	V	SMbus enabled
	ICNPU	**CNxx Pull-up Current(2)**					
DI30			50	250	400	μA	VDD = 5V, VPIN = VSS
	IIL	**Input Leakage Current(2,4,5)**					
DI50		I/O Ports	—	0.01	±1	μA	VSS ≤ VPIN ≤ VDD, Pin at high-impedance
DI51		Analog Input Pins	—	0.50	—	μA	VSS ≤ VPIN ≤ VDD, Pin at high-impedance
DI55		$\overline{\text{MCLR}}$	—	0.05	±5	μA	VSS ≤ VPIN ≤ VDD
DI56		OSC1	—	0.05	±5	μA	VSS ≤ VPIN ≤ VDD, XT, HS and LP Oscillator mode

Note 1: Data in "Typ" column is at 5V, 25°C unless otherwise stated. Parameters are for design guidance only and are not tested.

2: These parameters are characterized but not tested in manufacturing.

3: In RC oscillator configuration, the OSC1/CLKI pin is a Schmitt Trigger input. It is not recommended that the dsPIC30F device be driven with an external clock while in RC mode.

4: The leakage current on the $\overline{\text{MCLR}}$ pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

5: Negative current is defined as current sourced by the pin.

TABLE 23-13: EXTERNAL CLOCK TIMING REQUIREMENTS

AC CHARACTERISTICS			Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for Industrial -40°C ≤ TA ≤ +125°C for Extended				
Param No.	**Symbol**	**Characteristic**	**Min**	**Typ(1)**	**Max**	**Units**	**Conditions**
OS10	FOSC	External CLKI Frequency(2) (External clocks allowed only in EC mode)	DC 4 4 4	— — — —	40 10 10 7.5	MHz MHz MHz MHz	EC EC with 4x PLL EC with 8x PLL EC with 16x PLL
		Oscillator Frequency(2)	DC 0.4 4 4 4 4 10 31 — —	— — — — — — — — 7.37 512	4 4 10 10 10 7.5 25 33 — —	MHz MHz MHz MHz MHz MHz MHz kHz MHz kHz	RC XTL XT XT with 4x PLL XT with 8x PLL XT with 16x PLL HS LP FRC internal LPRC internal
OS20	TOSC	TOSC = 1/FOSC	—	—	—	—	See parameter OS10 for FOSC value
OS25	TCY	Instruction Cycle Time(2,3)	33	—	DC	ns	See Table 23-16
OS30	TosL, TosH	External Clock in (OSC1) High or Low Time(2)	.45 x TOSC	—	—	ns	EC
OS31	TosR, TosF	External Clock in (OSC1) Rise or Fall Time(2)	—	—	20	ns	EC
OS40	TckR	CLKO Rise Time(2,4)	—	—	—	ns	See parameter DO31
OS41	TckF	CLKO Fall Time(2,4)	—	—	—	ns	See parameter DO32

Note 1: Data in "Typ" column is at 5V, 25°C unless otherwise stated. Parameters are for design guidance only and are not tested.

2: These parameters are characterized but not tested in manufacturing.

3: Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "Min." values with an external clock applied to the OSC1/CLKI pin. When an external clock input is used, the "Max." cycle time limit is "DC" (no clock) for all devices.

4: Measurements are taken in EC or ERC modes. The CLKO signal is measured on the OSC2 pin. CLKO is low for the Q1-Q2 period (1/2 TCY) and high for the Q3-Q4 period (1/2 TCY).

FIGURE 23-8: TIMERQ (QEI MODULE) EXTERNAL CLOCK TIMING CHARACTERISTICS

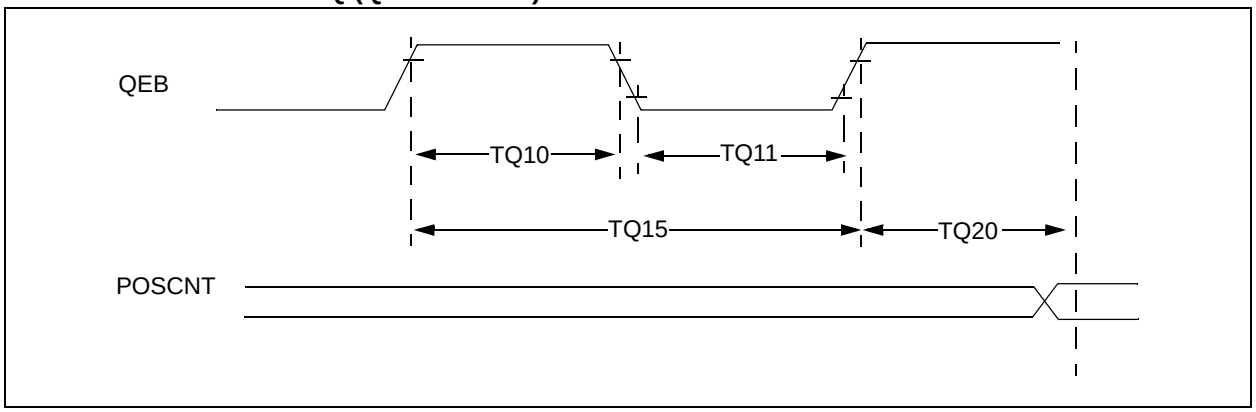


TABLE 23-25: QEI MODULE EXTERNAL CLOCK TIMING REQUIREMENTS

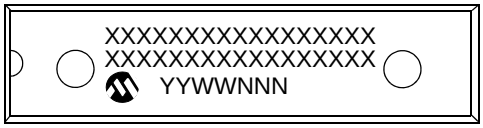
AC CHARACTERISTICS				Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for Industrial -40°C ≤ TA ≤ +125°C for Extended				
Param No.	**Symbol**	**Characteristic(1)**		**Min**	**Typ**	**Max**	**Units**	**Conditions**
TQ10	TtQH	TQCK High Time	Synchronous, with prescaler	TCY + 20	—	—	ns	Must also meet parameter TQ15
TQ11	TtQL	TQCK Low Time	Synchronous, with prescaler	TCY + 20	—	—	ns	Must also meet parameter TQ15
TQ15	TtQP	TQCP Input Period	Synchronous, with prescaler	2 * TCY + 40	—	—	ns	
TQ20	TCKEXTMRL	Delay from External TQCK Clock Edge to Timer Increment		0.5 TCY	—	1.5 TCY	—	

Note 1: These parameters are characterized but not tested in manufacturing.

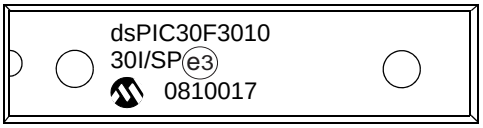
24.0 PACKAGING INFORMATION

24.1 Package Marking Information

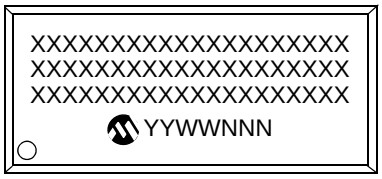
28-Lead PDIP (Skinny DIP)



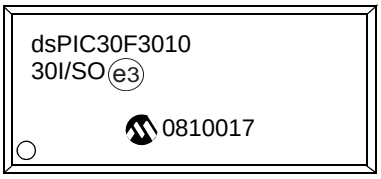
Example



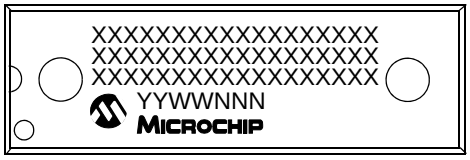
28-Lead SOIC



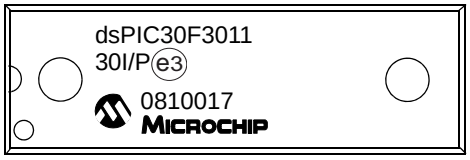
Example



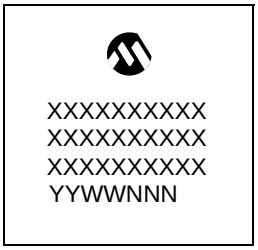
40-Lead PDIP



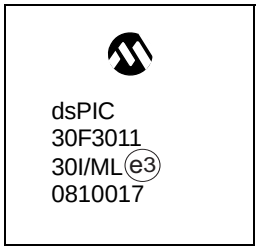
Example



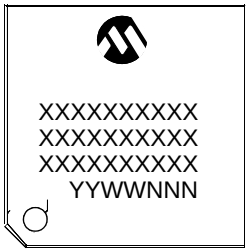
44-Lead QFN



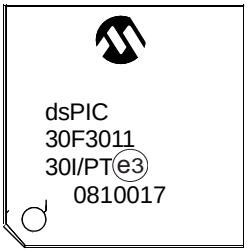
Example



44-Lead TQFP



Example

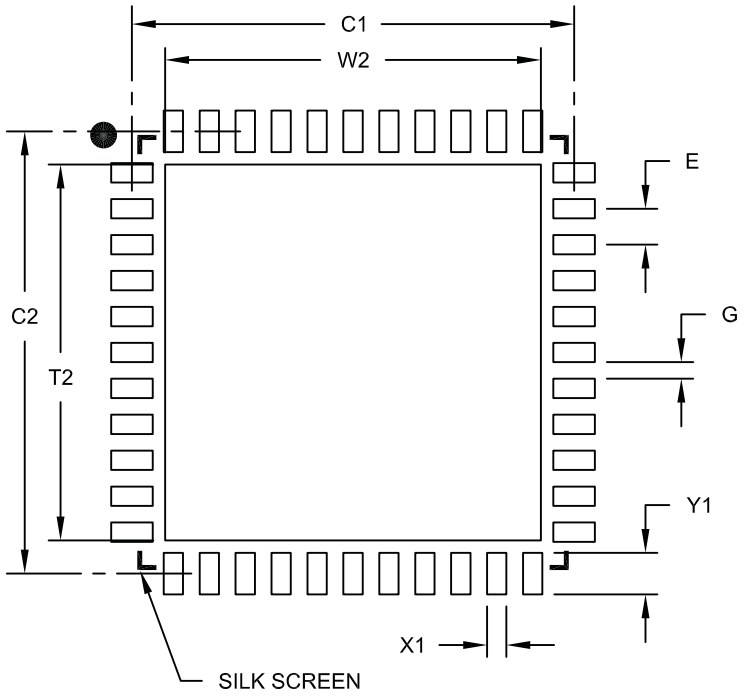


Legend:	XX...X	Customer-specific information
	Y	Year code (last digit of calendar year)
	YY	Year code (last 2 digits of calendar year)
	WW	Week code (week of January 1 is week '01')
	NNN	Alphanumeric traceability code
	(e3)	Pb-free JEDEC designator for Matte Tin (Sn)
	*	This package is Pb-free. The Pb-free JEDEC designator ((e3)) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

44-Lead Plastic Quad Flat, No Lead Package (ML) – 8x8 mm Body [QFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		0.65 BSC	
Optional Center Pad Width	W2			6.80
Optional Center Pad Length	T2			6.80
Contact Pad Spacing	C1		8.00	
Contact Pad Spacing	C2		8.00	
Contact Pad Width (X44)	X1			0.35
Contact Pad Length (X44)	Y1			0.80
Distance Between Pads	G	0.25		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2103A

APPENDIX A: REVISION HISTORY

Revision B (May 2006)

Previous versions of this data sheet contained Advance or Preliminary Information. They were distributed with incomplete characterization data.

This revision reflects these updates:

- Supported I^2C Slave Addresses (see Table 17-1)
- ADC Conversion Clock selection to allow 1 Msps operation (see **Section 19.0 "10-bit High-Speed Analog-to-Digital Converter (ADC) Module"**)
- Operating Current (I_{DD}) Specifications (see Table 23-5)
- Power-Down Current (I_{PD}) (see Table 23-7)
- I/O pin Input Specifications (see Table 23-8)
- BOR voltage limits (see Table 23-10)
- Watchdog Timer time-out limits (see Table 23-20)

Revision C (September 2006)

Updates made to **Section 23.0 "Electrical Characteristics"**.

Revision D (January 2007)

This revision includes updates to the packaging diagrams.

Revision E (April 2008)

This revision reflects these updates:

- Added OSCTUN register information and updated the OSCCON register information (removed TUN bits) in System Integration Register Map (see Table 20-7)
- Changed the location of the input reference in the 10-Bit High-Speed ADC Functional Block Diagram (see Figure 19-1)
- Added Fuse Configuration Register (FICD) details (see **Section 20.6 "Device Configuration Registers"** and Table 20-8)
- Added Note 2 in Device Configuration Registers table (Table 20-8)
- Updated FOSC register bit definition in Device Configuration Registers table (Table 20-8)
- Electrical Specifications:
  - Updated values for parameters DO10, DO16, DO20, and DO26 (see Table 23-9)
  - 10-Bit High-Speed ADC t_{PDU} timing parameter (time to stabilize) has been updated from 20 µs typical to 20 µs maximum (see Table 23-39)
  - Parameter OS65 (Internal RC Accuracy) has been expanded to reflect multiple Min and Max values for different temperatures (see Table 23-18)
  - Parameter DC12 (RAM Data Retention Voltage) has been updated to include a Min value (see Table 23-4)
  - Parameter D134 (Erase/Write Cycle Time) has been updated to include Min and Max values and the Typ value has been removed (see Table 23-11)
  - Removed parameters OS62 (Internal FRC Jitter) and OS64 (Internal FRC Drift) and Note 2 from AC Characteristics (see Table 23-17)
  - Parameter OS63 (Internal FRC Accuracy) has been expanded to reflect multiple Min and Max values for different temperatures (see Table 23-17)
  - Updated Min and Max values and Conditions for parameter SY11 and updated Min, Typ, and Max values and Conditions for parameter SY20 (see Table 23-20)
- Additional minor corrections throughout the document