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Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

Details

E·XFI

Product Status	Active
Core Processor	dsPIC
Core Size	16-Bit
Speed	20 MIPS
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	20
Program Memory Size	12KB (4K x 24)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	2.5V ~ 5.5V
Data Converters	A/D 10x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-SOIC (0.295", 7.50mm Width)
Supplier Device Package	28-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/dspic30f2012-20i-so

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

Pin Diagrams





2.4 DSP Engine

The DSP engine consists of a high-speed 17-bit x 17-bit multiplier, a barrel shifter and a 40-bit adder/subtracter (with two target accumulators, round and saturation logic).

The DSP engine also has the capability to perform inherent accumulator-to-accumulator operations, which require no additional data. These instructions are ADD, SUB and NEG.

The dsPIC30F is a single-cycle instruction flow architecture, therefore, concurrent operation of the DSP engine with MCU instruction flow is not possible. However, some MCU ALU and DSP engine resources may be used concurrently by the same instruction (e.g., ED, EDAC). See Table 2-2.

TABLE 2-2:DSP INSTRUCTIONSUMMARY

Instruction	Algebraic Operation	ACC WB?
CLR	A = 0	Yes
ED	$A = (x - y)^2$	No
EDAC	$A = A + (x - y)^2$	No
MAC	A = A + (x * y)	Yes
MAC	$A = A + x^2$	No
MOVSAC	No change in A	Yes
MPY	$A = x \bullet y$	No
MPY.N	$A = -x \bullet y$	No
MSC	$A = A - x \bullet y$	Yes

The DSP engine has several options selected through various bits in the CPU Core Configuration register (CORCON), which are:

- 1. Fractional or integer DSP multiply (IF).
- 2. Signed or unsigned DSP multiply (US).
- 3. Conventional or convergent rounding (RND).
- 4. Automatic saturation on/off for ACCA (SATA).
- 5. Automatic saturation on/off for ACCB (SATB).
- 6. Automatic saturation on/off for writes to data memory (SATDW).
- 7. Accumulator Saturation mode selection (ACCSAT).

		,						
	Note:	For CC	RCON	layout	, see Tal	ble	3-3.	
Δ	block	diagram	of the	DSP	ongino	ic	shown	ir

A block diagram of the DSP engine is shown in Figure 2-2.

FIGURE 3-1: PROGRAM SPACE MEMORY MAPS



3.2 Data Address Space

The core has two data spaces. The data spaces can be considered either separate (for some DSP instructions), or as one unified linear address range (for MCU instructions). The data spaces are accessed using two Address Generation Units (AGUs) and separate data paths.

3.2.1 DATA SPACE MEMORY MAP

The data space memory is split into two blocks, X and Y data space. A key element of this architecture is that Y space is a subset of X space, and is fully contained within X space. In order to provide an apparent Linear Addressing space, X and Y spaces have contiguous addresses.

When executing any instruction other than one of the MAC class of instructions, the X block consists of the 64 Kbyte data address space (including all Y addresses). When executing one of the MAC class of instructions, the X block consists of the 64 Kbyte data address space, excluding the Y address block (for data reads only). In other words, all other instructions regard the entire data memory as one composite address space. The MAC class instructions extract the Y address space from data space and address it using EAs sourced from W10 and W11. The remaining X data space is addressed using W8 and W9. Both address spaces are concurrently accessed only with the MAC class instructions.

The data space memory map for the dsPIC30F2011 and dsPIC30F2012 is shown in Figure 3-6. The data space memory map for the dsPIC30F3012 and dsPIC30F3013 is shown in Figure 3-7.



FIGURE 3-6: dsPIC30F2011/2012 DATA SPACE MEMORY MAP

4.1.3 MOVE AND ACCUMULATOR INSTRUCTIONS

Move instructions and the DSP accumulator class of instructions provide a greater degree of addressing flexibility than other instructions. In addition to the addressing modes supported by most MCU instructions, move and accumulator instructions also support Register Indirect with Register Offset Addressing mode, also referred to as Register Indexed mode.

Note: For the MOV instructions, the addressing mode specified in the instruction can differ for the source and destination EA. However, the 4-bit Wb (register offset) field is shared between both source and destination (but typically only used by one).

In summary, the following addressing modes are supported by move and accumulator instructions:

- Register Direct
- Register Indirect
- Register Indirect Post-modified
- Register Indirect Pre-modified
- Register Indirect with Register Offset (Indexed)
- Register Indirect with Literal Offset
- 8-bit Literal
- 16-bit Literal
- Note: Not all instructions support all the addressing modes given above. Individual instructions may support different subsets of these addressing modes.

4.1.4 MAC INSTRUCTIONS

The dual source operand DSP instructions (CLR, ED, EDAC, MAC, MPY, MPY.N, MOVSAC and MSC), also referred to as MAC instructions, utilize a simplified set of addressing modes to allow the user to effectively manipulate the data pointers through register indirect tables.

The two source operand prefetch registers must belong to the set {W8, W9, W10, W11}. For data reads, W8 and W9 are always directed to the X RAGU. W10 and W11 are always directed to the Y AGU. The effective addresses generated (before and after modification) must, therefore, be valid addresses within X data space for W8 and W9 and Y data space for W10 and W11.

Note: Register Indirect with Register Offset addressing is only available for W9 (in X space) and W11 (in Y space). In summary, the following addressing modes are supported by the ${\tt MAC}$ class of instructions:

- Register Indirect
- Register Indirect Post-modified by 2
- Register Indirect Post-modified by 4
- Register Indirect Post-modified by 6
- Register Indirect with Register Offset (Indexed)

4.1.5 OTHER INSTRUCTIONS

Besides the various addressing modes outlined above, some instructions use literal constants of various sizes. For example, BRA (branch) instructions use 16-bit signed literals to specify the branch destination directly, whereas the DISI instruction uses a 14-bit unsigned literal field. In some instructions, such as ADD Acc, the source of an operand or result is implied by the opcode itself. Certain operations, such as NOP, do not have any operands.

4.2 Modulo Addressing

Modulo Addressing is a method of providing an automated means to support circular data buffers using hardware. The objective is to remove the need for software to perform data address boundary checks when executing tightly looped code, as is typical in many DSP algorithms.

Modulo Addressing can operate in either data or program space (since the data pointer mechanism is essentially the same for both). One circular buffer can be supported in each of the X (which also provides the pointers into program space) and Y data spaces. Modulo Addressing can operate on any W register pointer. However, it is not advisable to use W14 or W15 for Modulo Addressing since these two registers are used as the Stack Frame Pointer and Stack Pointer, respectively.

In general, any particular circular buffer can only be configured to operate in one direction, as there are certain restrictions on the buffer Start address (for incrementing buffers), or end address (for decrementing buffers) based upon the direction of the buffer.

The only exception to the usage restrictions is for buffers that have a power-of-2 length. As these buffers satisfy the Start and the end address criteria, they can operate in a Bidirectional mode (i.e., address boundary checks are performed on both the lower and upper address boundaries).

NOTES:

TABLE 7-1: PORTB REGISTER MAP FOR dsPIC30F2011/3012

SFR Name	Addr.	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0		Reset	State	
TRISB	02C6	—	—	—	—	—		—	—	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	0000	0000	1111	1111
PORTB	02C8	_	_	_	_	_	_	_	_	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	0000	0000	0000	0000
LATB	02CB	_	_	_	_	_	_		_	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0	0000	0000	0000	0000

Legend: — = unimplemented bit, read as '0'

TABLE 7-2: PORTB REGISTER MAP FOR dsPIC30F2012/3013

SFR Name	Addr.	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset State
TRISB	02C6	_	—	_	—	—	_	TRISB9	TRISB8	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	0000 0011 1111 1111
PORTB	02C8	_	_	_	_	—	_	RB9	RB8	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	0000 0000 0000 0000
LATB	02CB		-			-		LATB9	LATB8	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0	0000 0000 0000 0000

Legend: — = unimplemented bit, read as '0'

TABLE 7-3: PORTC REGISTER MAP FOR dsPIC30F2011/2012/3012/3013

SFR Nam	e Addr.	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset State
TRISC	02CC	TRISC15	TRISC14	TRISC13	-	-	-	_	-	-	-	-	—	_	—	—	_	1110 0000 0000 0000
PORTO	02CE	RC15	RC14	RC13	_	—	—	_	—	_	-	—	_	_	_	_	_	0000 0000 0000 0000
LATC	02D0	LATC15	LATC14	LATC13	-	_	_	_	—	_	-	—	_	_	_	_	_	0000 0000 0000 0000

Legend: — = unimplemented bit, read as '0'

TABLE 7-4: PORTD REGISTER MAP FOR dsPIC30F2011/3012

SFR Name	Addr.	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset State
TRISD	02D2	_	-	—	_	-	_	—	_	-		_		_	_		TRISD0	0000 0000 0000 0000
PORTD	02D4	_	_	_	—	—	_	_	—	—	—	—	_	—	_	_	RD0	0000 0000 0000 0000
LATD	02D6	_	_	_	_	_	_	_	_	_	_	_	_	_	_	_	LATD0	0000 0000 0000 0000

Legend: — = unimplemented bit, read as '0'

16.4 Programming the Start of Conversion Trigger

The conversion trigger will terminate acquisition and start the requested conversions.

The SSRC<2:0> bits select the source of the conversion trigger. The SSRC bits provide for up to four alternate sources of conversion trigger.

When SSRC<2:0> = 000, the conversion trigger is under software control. Clearing the SAMP bit will cause the conversion trigger.

When SSRC<2:0> = 111 (Auto-Start mode), the conversion trigger is under A/D clock control. The SAMC bits select the number of A/D clocks between the start of acquisition and the start of conversion. This provides the fastest conversion rates on multiple channels. SAMC must always be at least one clock cycle.

Other trigger sources can come from timer modules or external interrupts.

16.5 Aborting a Conversion

Clearing the ADON bit during a conversion will abort the current conversion and stop the sampling sequencing until the next sampling trigger. The ADCBUF will not be updated with the partially completed A/D conversion sample. That is, the ADCBUF will continue to contain the value of the last completed conversion (or the last value written to the ADCBUF register).

If the clearing of the ADON bit coincides with an auto-start, the clearing has a higher priority and a new conversion will not start.

After the A/D conversion is aborted, a 2 TAD wait is required before the next sampling may be started by setting the SAMP bit.

16.6 Selecting the ADC Conversion Clock

The ADC conversion requires 14 TAD. The source of the ADC conversion clock is software selected, using a 6-bit counter. There are 64 possible options for TAD.

EQUATION 16-1: ADC CONVERSION CLOCK

TAD = TCY * (0.5*(ADCS < 5:0 > + 1))

The internal RC oscillator is selected by setting the ADRC bit.

For correct ADC conversions, the ADC conversion clock (TAD) must be selected to ensure a minimum TAD time of 334 nsec (for VDD = 5V). Refer to **Section 20.0 "Electrical Characteristics"** for minimum TAD under other operating conditions.

Example 16-1 shows a sample calculation for the ADCS<5:0> bits, assuming a device operating speed of 30 MIPS.

EXAMPLE 16-1: ADC CONVERSION CLOCK AND SAMPLING RATE CALCULATION

Minimum TAD = 334 nsec TCY = 33 .33 nsec (30 MIPS) $ADCS < 5:0 > = 2 \frac{TAD}{TCY} - 1$ $= 2 \cdot \frac{334 \text{ nsec}}{33.33 \text{ nsec}} - 1$ = 19.04Therefore. Set ADCS<5:0> = 19 Actual TAD = $\frac{\text{TCY}}{2}$ (ADCS<5:0>+1) $=\frac{33.33 \text{ nsec}}{2}$ (19 + 1) = 334 nsec If SSRC<2:0> = '111' and SAMC<4:0> = '00001' Since, Sampling Time = Acquisition Time + Conversion Time = 1 TAD + 14 TAD= 15 x 334 nsec Therefore, Sampling Rate = (15 x 334 nsec)

 $= \sim 200 \text{ kHz}$

NOTES:

17.4 Watchdog Timer (WDT)

17.4.1 WATCHDOG TIMER OPERATION

The primary function of the Watchdog Timer (WDT) is to reset the processor in the event of a software malfunction. The WDT is a free-running timer which runs off an on-chip RC oscillator, requiring no external component. Therefore, the WDT timer will continue to operate even if the main processor clock (e.g., the crystal oscillator) fails.

17.4.2 ENABLING AND DISABLING THE WDT

The Watchdog Timer can be "Enabled" or "Disabled" only through a Configuration bit (FWDTEN) in the Configuration register, FWDT.

Setting FWDTEN = 1 enables the Watchdog Timer. The enabling is done when programming the device. By default, after chip erase, FWDTEN bit = 1. Any device programmer capable of programming dsPIC30F devices allows programming of this and other Configuration bits.

If enabled, the WDT will increment until it overflows or "times out". A WDT time-out will force a device Reset (except during Sleep). To prevent a WDT time-out, the user must clear the Watchdog Timer using a CLRWDT instruction.

If a WDT times out during Sleep, the device will wake-up. The WDTO bit in the RCON register will be cleared to indicate a wake-up resulting from a WDT time-out.

Setting FWDTEN = 0 allows user software to enable/disable the Watchdog Timer via the SWDTEN (RCON<5>) control bit.

17.5 Low-Voltage Detect

The Low-Voltage Detect (LVD) module is used to detect when the VDD of the device drops below a threshold value, VLVD, which is determined by the LVDL<3:0> bits (RCON<11:8>) and is thus user programmable. The internal voltage reference circuitry requires a nominal amount of time to stabilize, and the BGST bit (RCON<13>) indicates when the voltage reference has stabilized.

In some devices, the LVD threshold voltage may be applied externally on the LVDIN pin.

The LVD module is enabled by setting the LVDEN bit (RCON<12>).

17.6 Power-Saving Modes

There are two power-saving states that can be entered through the execution of a special instruction, PWRSAV; these are Sleep and Idle.

The format of the PWRSAV instruction is as follows:

PWRSAV <parameter>, where 'parameter' defines Idle or Sleep mode.

17.6.1 SLEEP MODE

In Sleep mode, the clock to the CPU and peripherals is shut down. If an on-chip oscillator is being used, it is shut down.

The Fail-Safe Clock Monitor is not functional during Sleep since there is no clock to monitor. However, LPRC clock remains active if WDT is operational during Sleep.

The brown-out protection circuit and the Low-Voltage Detect circuit, if enabled, will remain functional during Sleep.

The processor wakes up from Sleep if at least one of the following conditions has occurred:

- any interrupt that is individually enabled and meets the required priority level
- any Reset (POR, BOR and MCLR)
- WDT time-out

On waking up from Sleep mode, the processor will restart the same clock that was active prior to entry into Sleep mode. When clock switching is enabled, bits COSC<2:0> will determine the oscillator source that will be used on wake-up. If clock switch is disabled, then there is only one system clock.

Note: If a POR or BOR occurred, the selection of the oscillator is based on the FOS<2:0> and FPR<4:0> Configuration bits.

If the clock source is an oscillator, the clock to the device will be held off until OST times out (indicating a stable oscillator). If PLL is used, the system clock is held off until LOCK = 1 (indicating that the PLL is stable). In either case, TPOR, TLOCK and TPWRT delays are applied.

If EC, FRC, LPRC or ERC oscillators are used, then a delay of TPOR (~ 10 μ s) is applied. This is the smallest delay possible on wake-up from Sleep.

Moreover, if LP oscillator was active during Sleep and LP is the oscillator used on wake-up, then the start-up delay will be equal to TPOR. PWRT delay and OST timer delay are not applied. In order to have the smallest possible start-up delay when waking up from Sleep, one of these faster wake-up options should be selected before entering Sleep.

Field	Description
Wb	Base W register ∈ {W0W15}
Wd	Destination W register ∈ { Wd, [Wd], [Wd++], [Wd], [++Wd], [Wd] }
Wdo	Destination W register ∈ { Wnd, [Wnd], [Wnd++], [Wnd], [++Wnd], [Wnd], [Wnd+Wb] }
Wm,Wn	Dividend, Divisor working register pair (direct addressing)
Wm*Wm	Multiplicand and Multiplier working register pair for Square instructions ∈ {W4*W4,W5*W5,W6*W6,W7*W7}
Wm*Wn	Multiplicand and Multiplier working register pair for DSP instructions ∈ {W4*W5,W4*W6,W4*W7,W5*W6,W5*W7,W6*W7}
Wn	One of 16 working registers ∈ {W0W15}
Wnd	One of 16 destination working registers ∈ {W0W15}
Wns	One of 16 source working registers ∈ {W0W15}
WREG	W0 (working register used in file register instructions)
Ws	Source W register ∈ { Ws, [Ws], [Ws++], [Ws], [++Ws], [Ws] }
Wso	Source W register ∈ { Wns, [Wns], [Wns++], [Wns], [++Wns], [Wns], [Wns+Wb] }
Wx	X data space prefetch address register for DSP instructions ∈ {[W8]+=6, [W8]+=4, [W8]+=2, [W8], [W8]-=6, [W8]-=4, [W8]-=2, [W9]+=6, [W9]+=4, [W9]+=2, [W9], [W9]-=6, [W9]-=4, [W9]-=2, [W9+W12],none}
Wxd	X data space prefetch destination register for DSP instructions ∈ {W4W7}
Wy	Y data space prefetch address register for DSP instructions ∈ {[W10]+=6, [W10]+=4, [W10]+=2, [W10], [W10]-=6, [W10]-=4, [W10]-=2, [W11]+=6, [W11]+=4, [W11]+=2, [W11], [W11]-=6, [W11]-=4, [W11]-=2, [W11+W12], none}
Wyd	Y data space prefetch destination register for DSP instructions ∈ {W4W7}

TABLE 18-1: SYMBOLS USED IN OPCODE DESCRIPTIONS (CONTINUED)

19.0 DEVELOPMENT SUPPORT

The PIC[®] microcontrollers and dsPIC[®] digital signal controllers are supported with a full range of software and hardware development tools:

- Integrated Development Environment
- MPLAB[®] IDE Software
- Compilers/Assemblers/Linkers
 - MPLAB C Compiler for Various Device Families
 - HI-TECH C for Various Device Families
 - MPASM[™] Assembler
 - MPLINK[™] Object Linker/ MPLIB[™] Object Librarian
 - MPLAB Assembler/Linker/Librarian for Various Device Families
- Simulators
 - MPLAB SIM Software Simulator
- Emulators
 - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debuggers
 - MPLAB ICD 3
 - PICkit[™] 3 Debug Express
- Device Programmers
 - PICkit[™] 2 Programmer
 - MPLAB PM3 Device Programmer
- Low-Cost Demonstration/Development Boards, Evaluation Kits, and Starter Kits

19.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16/32-bit microcontroller market. The MPLAB IDE is a Windows[®] operating system-based application that contains:

- A single graphical interface to all debugging tools
 - Simulator
 - Programmer (sold separately)
 - In-Circuit Emulator (sold separately)
 - In-Circuit Debugger (sold separately)
- A full-featured editor with color-coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High-level source code debugging
- Mouse over variable inspection
- Drag and drop variables from source to watch windows
- Extensive on-line help
- Integration of select third party tools, such as IAR C Compilers

The MPLAB IDE allows you to:

- Edit your source files (either C or assembly)
- One-touch compile or assemble, and download to emulator and simulator tools (automatically updates all project information)
- Debug using:
 - Source files (C or assembly)
 - Mixed C and assembly
 - Machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost-effective simulators, through low-cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increased flexibility and power.

TABLE 20-7: DC CHARACTERISTICS: POWER-DOWN CURRENT (IPD)

DC CHARACT	ERISTICS		Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended								
Parameter No.	Typical ⁽¹⁾	Мах	Units			Conditions					
Power-Down	Current (IPD) ⁽	2)									
DC60a	0.3	_	μΑ	25°C							
DC60b	1.3	30.0	μΑ	85°C	3.3V						
DC60c	16.0	60.0	μA	125°C		Rass Rower Down Current(3)					
DC60e	0.5	—	μA	25°C		Base Power-Down Currenter					
DC60f	3.7	45.0	μA	85°C	5V						
DC60g	25.0	90.0	μA	125°C							
DC61a	6.0	9.0	μA	25°C							
DC61b	6.0	9.0	μA	85°C	3.3V						
DC61c	6.0	9.0	μA	125°C		Watabdag Timor Current: Alwor(3)					
DC61e	13.0	20.0	μA	25°C							
DC61f	12.0	20.0	μΑ	85°C	5V						
DC61g	12.0	20.0	μΑ	125°C							
DC62a	4.0	10.0	μA	25°C		-					
DC62b	5.0	10.0	μΑ	85°C	3.3V						
DC62c	4.0	10.0	μA	125°C		Timor 1 w/22 kHz Crystol: Alt 22(3)					
DC62e	4.0	15.0	μA	25°C							
DC62f	6.0	15.0	μΑ	85°C	5V						
DC62g	5.0	15.0	μΑ	125°C							
DC63a	33.0	53.0	μA	25°C							
DC63b	35.0	53.0	μΑ	85°C	3.3V						
DC63c	19.0	53.0	μA	125°C		BOB On: Albon(3)					
DC63e	38.0	62.0	μA	25°C		BOR OII. ABOR 7					
DC63f	41.0	62.0	μΑ	85°C	5V						
DC63g	41.0	62.0	μA	125°C							
DC66a	21.0	40.0	μA	25°C							
DC66b	26.0	40.0	μA	85°C	3.3V						
DC66c	27.0	40.0	μA	125°C		Low-Voltage Detect: ∆IL∨D ⁽³⁾					
DC66e	25.0	44.0	μA	25°C							
DC66f	27.0	44.0	μA	85°C	5V						
DC66g	29.0	44.0	μA	125°C							

Note 1: Data in the Typical column is at 5V, 25°C unless otherwise stated. Parameters are for design guidance only and are not tested.

2: Base IPD is measured with all peripherals and clocks shut down. All I/Os are configured as inputs and pulled high. LVD, BOR, WDT, etc. are all switched off.

3: The Δ current is the additional current consumed when the module is enabled. This current should be added to the base IPD current.

FIGURE 20-11: OC/PWM MODULE TIMING CHARACTERISTICS



TABLE 20-28: SIMPLE OC/PWM MODE TIMING REQUIREMENTS

АС СНА	RACTER	ISTICS		Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended						
Param No.	Symbol	Characteristic ⁽¹⁾	Min	Typ ⁽²⁾	Max	Units	Conditions			
OC15	Tfd	Fault Input to PWM I/O Change			50	ns				
OC20	TFLT	Fault Input Pulse Width	50		_	ns				

Note 1: These parameters are characterized but not tested in manufacturing.

2: Data in "Typ" column is at 5V, 25°C unless otherwise stated. Parameters are for design guidance only and are not tested.







TABLE 20-34: I²C[™] BUS DATA TIMING REQUIREMENTS (SLAVE MODE)

AC CHA	RACTERIS	STICS	Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended								
Param No.	Symbol	Charact	teristic	Min	Max	Units	Conditions				
IS10	TLO:SCL	Clock Low Time	100 kHz mode	4.7	_	μs	Device must operate at a minimum of 1.5 MHz				
			400 kHz mode	1.3	_	μs	Device must operate at a minimum of 10 MHz.				
			1 MHz mode ⁽¹⁾	0.5	_	μs					
IS11	THI:SCL	Clock High Time	100 kHz mode	4.0		μs	Device must operate at a minimum of 1.5 MHz				
			400 kHz mode	0.6	_	μs	Device must operate at a minimum of 10 MHz				
			1 MHz mode ⁽¹⁾	0.5		μs					
IS20	TF:SCL	SDA and SCL	100 kHz mode	—	300	ns	CB is specified to be from				
		Fall Time	400 kHz mode	20 + 0.1 Св	300	ns	10 to 400 pF				
			1 MHz mode ⁽¹⁾	—	100	ns					
IS21	IS21 TR:SCL SDA and SCL		100 kHz mode	_	1000	ns	CB is specified to be from				
		Rise Time	400 kHz mode	20 + 0.1 Св	300	ns	10 to 400 pF				
			1 MHz mode ⁽¹⁾	_	300	ns					

Note 1: Maximum pin capacitance = 10 pF for all I^2C^{TM} pins (for 1 MHz mode only).

AC CH	ARACTERI	STICS	Standard Operating Conditions: 2.7V to 5.5V (unless otherwise stated) Operating temperature-40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended							
Param No.	Symbol	Characteristic	Min.	Тур	Max.	Units	Conditions			
		Cloc	k Parame	ters						
AD50	TAD	A/D Clock Period	334	_		ns	VDD = 3-5.5V (Note 1)			
AD51	tRC	A/D Internal RC Oscillator Period	1.2	1.5	1.8	μs				
		Con	version R	ate						
AD55	tCONV	Conversion Time	—	14 Tad		ns				
AD56a	FCNV	Throughput Rate	_	200		ksps	VDD = VREF = 5V, Industrial temperature			
AD56b	FCNV	Throughput Rate	_	100		ksps	VDD = VREF = 5V, Extended temperature			
AD57	TSAMP	Sampling Time	1 Tad			ns	VDD = 3-5.5V source resistance Rs = 0-2.5 k Ω			
		Timin	ig Parame	eters						
AD60	tPCS	Conversion Start from Sample Trigger	—	1 Tad		ns				
AD61	tPSS	Sample Start from Setting Sample (SAMP) Bit	0.5 Tad	—	1.5 Tad	ns				
AD62	tCSS	Conversion Completion to Sample Start (ASAM = 1)	_	0.5 TAD		ns				
AD63	tdpu (2)	Time to Stabilize Analog Stage from A/D Off to A/D On		—	20	μs				

TABLE 20-37: 12-BIT A/D CONVERSION TIMING REQUIREMENTS

Note 1: Because the sample caps will eventually lose charge, clock rates below 10 kHz can affect linearity performance, especially at elevated temperatures.

2: tDPU is the time required for the ADC module to stabilize when it is turned on (ADCON1<ADON> = 1). During this time the ADC result is indeterminate.

18-Lead Plastic Dual In-Line (P) – 300 mil Body [PDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units		INCHES	
Dimensi	on Limits	MIN	NOM	MAX
Number of Pins	Ν		18	
Pitch	е		.100 BSC	
Top to Seating Plane	Α	-	-	.210
Molded Package Thickness	A2	.115	.130	.195
Base to Seating Plane	A1	.015	-	-
Shoulder to Shoulder Width	E	.300	.310	.325
Molded Package Width	E1	.240	.250	.280
Overall Length	D	.880	.900	.920
Tip to Seating Plane	L	.115	.130	.150
Lead Thickness	С	.008	.010	.014
Upper Lead Width	b1	.045	.060	.070
Lower Lead Width	b	.014	.018	.022
Overall Row Spacing §	eB	-	-	.430

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. § Significant Characteristic.

3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.

4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-007B

44-Lead Plastic Quad Flat, No Lead Package (ML) – 8x8 mm Body [QFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	ts MILLIMETERS		ETERS
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E	0.65 BSC		
Optional Center Pad Width	W2			6.80
Optional Center Pad Length	T2			6.80
Contact Pad Spacing	C1		8.00	
Contact Pad Spacing	C2		8.00	
Contact Pad Width (X44)	X1			0.35
Contact Pad Length (X44)	Y1			0.80
Distance Between Pads	G	0.25		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

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