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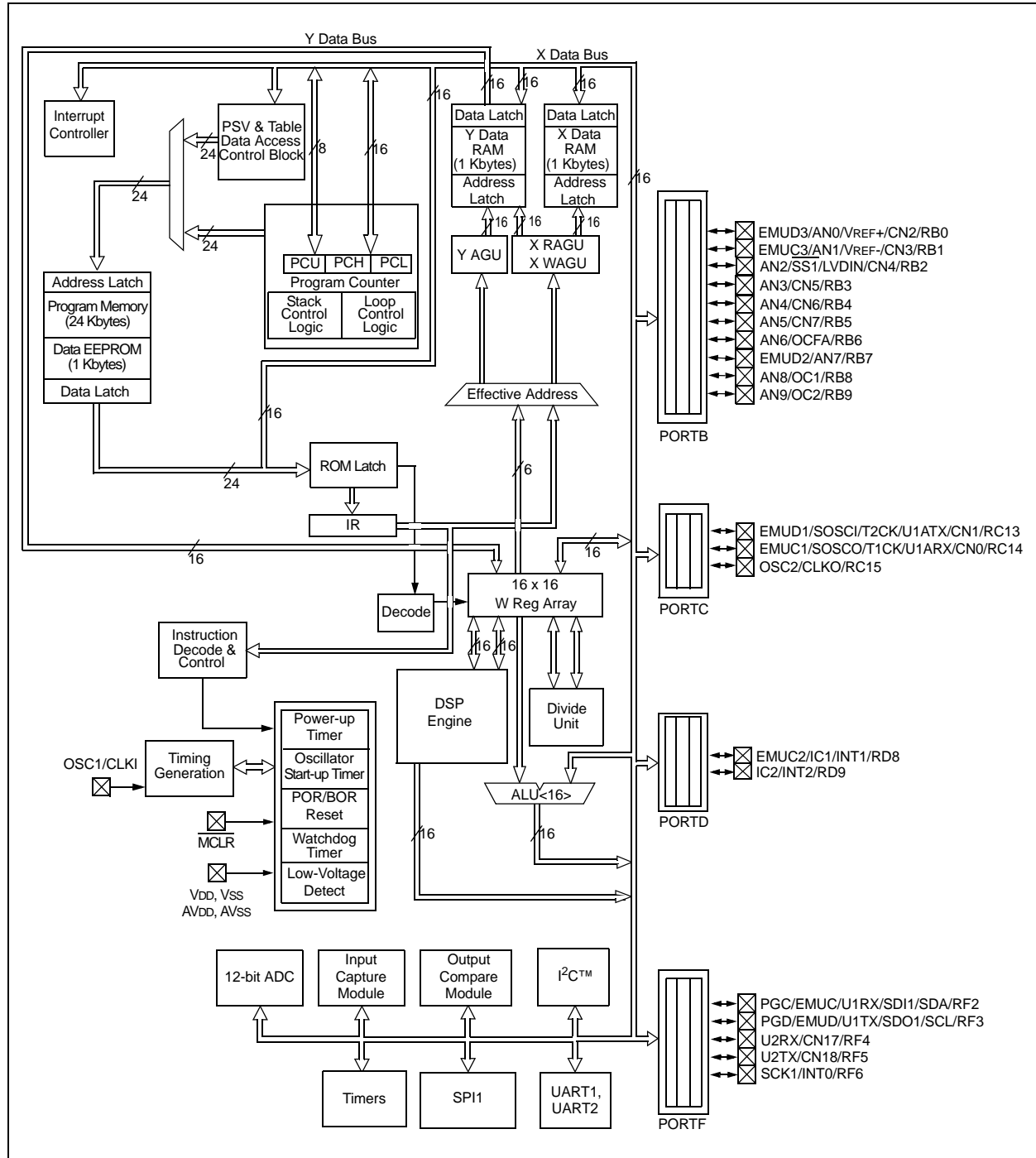
#### Details

Product Status	Active
Core Processor	dsPIC
Core Size	16-Bit
Speed	30 MIPS
Connectivity	I <sup>2</sup> C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	12
Program Memory Size	24KB (8K x 24)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	2.5V ~ 5.5V
Data Converters	A/D 8x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	18-SOIC (0.295", 7.50mm Width)
Supplier Device Package	18-SOIC
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microchip-technology/dspic30f3012-30i-so">https://www.e-xfl.com/product-detail/microchip-technology/dspic30f3012-30i-so</a>

NOTES:

# dsPIC30F2011/2012/3012/3013

**FIGURE 1-4: dsPIC30F3013 BLOCK DIAGRAM**

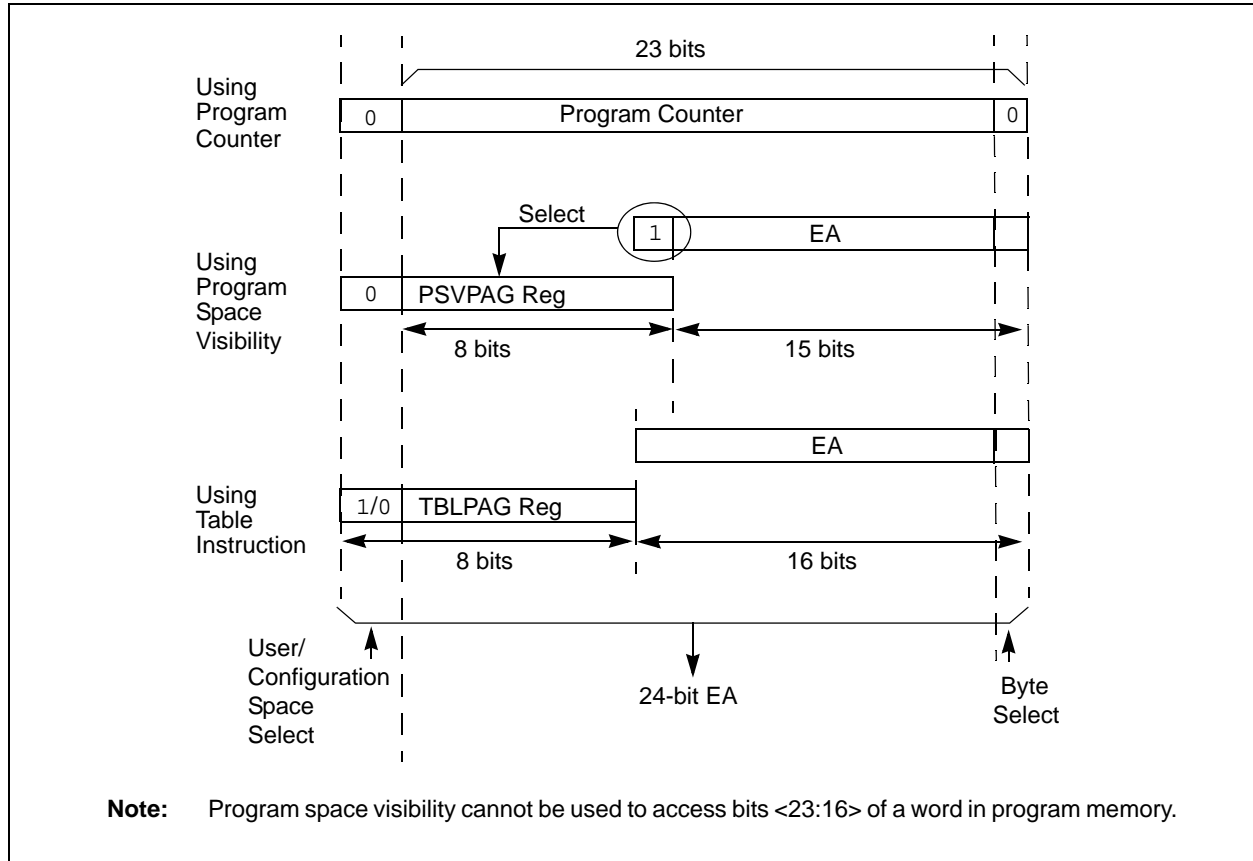


# dsPIC30F2011/2012/3012/3013

**TABLE 3-1: PROGRAM SPACE ADDRESS CONSTRUCTION**

Access Type	Access Space	Program Space Address				
		<23>	<22:16>	<15>	<14:1>	<0>
Instruction Access	User	0	PC<22:1>			0
TBLRD/TBLWT	User (TBLPAG<7> = 0)	TBLPAG<7:0>			Data EA<15:0>	
TBLRD/TBLWT	Configuration (TBLPAG<7> = 1)	TBLPAG<7:0>			Data EA<15:0>	
Program Space Visibility	User	0	PSVPAG<7:0>		Data EA<14:0>	

**FIGURE 3-2: DATA ACCESS FROM PROGRAM SPACE ADDRESS GENERATION**



## 5.4 RTSP Operation

The dsPIC30F Flash program memory is organized into rows and panels. Each row consists of 32 instructions or 96 bytes. Each panel consists of 128 rows or 4K x 24 instructions. RTSP allows the user to erase one row (32 instructions) at a time and to program four instructions at one time. RTSP may be used to program multiple program memory panels, but the table pointer must be changed at each panel boundary.

Each panel of program memory contains write latches that hold 32 instructions of programming data. Prior to the actual programming operation, the write data must be loaded into the panel write latches. The data to be programmed into the panel is loaded in sequential order into the write latches; instruction 0, instruction 1, etc. The instruction words loaded must always be from a 32 address boundary.

The basic sequence for RTSP programming is to set up a Table Pointer, then do a series of TBLWT instructions to load the write latches. Programming is performed by setting the special bits in the NVMCON register. 32 TBLWTL and four TBLWTH instructions are required to load the 32 instructions. If multiple panel programming is required, the Table Pointer needs to be changed and the next set of multiple write latches written.

All of the table write operations are single-word writes (2 instruction cycles), because only the table latches are written. A programming cycle is required for programming each row.

The Flash Program Memory is readable, writable and erasable during normal operation over the entire VDD range.

## 5.5 Control Registers

The four SFRs used to read and write the program Flash memory are:

- NVMCON
- NVMADR
- NVMADRU
- NVMKEY

### 5.5.1 NVMCON REGISTER

The NVMCON register controls which blocks are to be erased, which memory type is to be programmed, and start of the programming cycle.

### 5.5.2 NVMADR REGISTER

The NVMADR register is used to hold the lower two bytes of the Effective Address. The NVMADR register captures the EA<15:0> of the last table instruction that has been executed and selects the row to write.

### 5.5.3 NVMADRU REGISTER

The NVMADRU register is used to hold the upper byte of the Effective Address. The NVMADRU register captures the EA<23:16> of the last table instruction that has been executed.

### 5.5.4 NVMKEY REGISTER

NVMKEY is a write-only register that is used for write protection. To start a programming or an erase sequence, the user must consecutively write 0x55 and 0xAA to the NVMKEY register. Refer to **Section 5.6 “Programming Operations”** for further details.

<b>Note:</b> The user can also directly write to the NVMADR and NVMADRU registers to specify a program memory address for erasing or programming.
---

**TABLE 8-3: dsPIC30F3013 INTERRUPT CONTROLLER REGISTER MAP**

SFR Name	ADR	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset State
INTCON1	0080	NSTDIS	—	—	—	—	OVATE	OVATE	COVTE	—	—	—	MATHERR	ADDRERR	STKERR	OSCFail	—	0000 0000 0000 0000
INTCON2	0082	ALTIVT	DISI	—	—	—	—	—	—	—	—	—	—	—	INT2EP	INT1EP	INT0EP	0000 0000 0000 0000
IFS0	0084	CNIF	MI2CIF	SI2CIF	NVMIF	ADIF	U1TXIF	U1RXIF	SPI1IF	T3IF	T2IF	OC2IF	IC2IF	T1IF	OC1IF	IC1IF	INT0IF	0000 0000 0000 0000
IFS1	0086	—	—	—	—	—	—	U2TXIF	U2RXIF	INT2IF	—	—	—	—	—	—	INT1IF	0000 0000 0000 0000
IFS2	0088	—	—	—	—	—	LVDIF	—	—	—	—	—	—	—	—	—	—	0000 0000 0000 0000
IEC0	008C	CNIE	MI2CIE	SI2CIE	NVMIE	ADIE	U1TXIE	U1RXIE	SPI1IE	T3IE	T2IE	OC2IE	IC2IE	T1IE	OC1IE	IC1IE	INT0IE	0000 0000 0000 0000
IEC1	008E	—	—	—	—	—	—	U2TXIE	U2RXIE	INT2IE	—	—	—	—	—	—	INT1IE	0000 0000 0000 0000
IEC2	0090	—	—	—	—	—	LVDIE	—	—	—	—	—	—	—	—	—	—	0000 0000 0000 0000
IPC0	0094	—	T1IP<2:0>			—	OC1IP<2:0>			—	IC1IP<2:0>			—	INT0IP<2:0>			0100 0100 0100 0100
IPC1	0096	—	T31P<2:0>			—	T2IP<2:0>			—	OC2IP<2:0>			—	IC2IP<2:0>			0100 0100 0100 0100
IPC2	0098	—	ADIP<2:0>			—	U1TXIP<2:0>			—	U1RXIP<2:0>			—	SPI1IP<2:0>			0100 0100 0100 0100
IPC3	009A	—	CNIP<2:0>			—	MI2CIP<2:0>			—	SI2CIP<2:0>			—	NVMIP<2:0>			0100 0100 0100 0100
IPC4	009C	—	—	—	—	—	—	—	—	—	—	—	—	—	INT1IP<2:0>			0000 0000 0000 0100
IPC5	009E	—	INT2IP<2:0>			—	—	—	—	—	—	—	—	—	—	—	—	0100 0000 0000 0000
IPC6	00A0	—	—	—	—	—	—	—	—	—	U2TXIP<2:0>			—	U2RXIP<2:0>			0000 0000 0100 0100
IPC7	00A2	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	0000 0000 0000 0000
IPC8	00A4	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	0000 0000 0000 0000
IPC9	00A6	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—	0000 0000 0000 0000
IPC10	00A8	—	—	—	—	—	LVDIP<2:0>			—	—	—	—	—	—	—	—	0000 0100 0000 0000

**Legend:** u = uninitialized bit; — = unimplemented bit, read as '0'

**Note:** Refer to the "dsPIC30F Family Reference Manual" (DS70046) for descriptions of register bit fields.

**FIGURE 13-1: SPI BLOCK DIAGRAM**

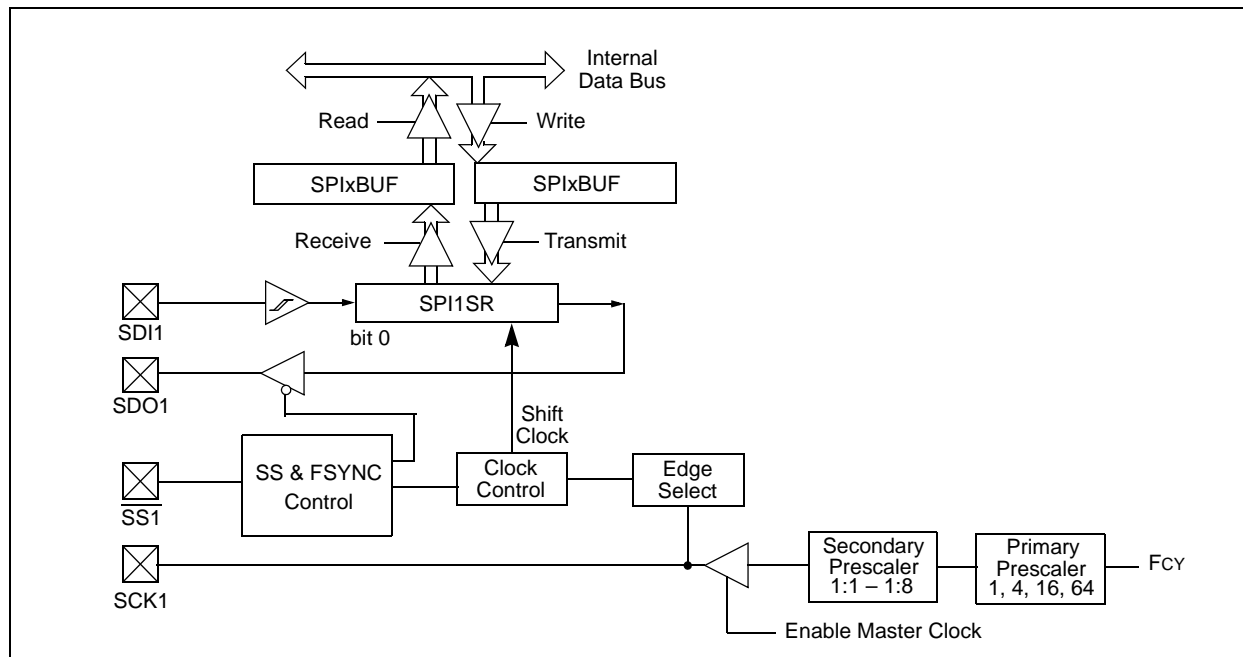


Figure 13-2 depicts the a master/slave connection between two processors. In Master mode, the clock is generated by prescaling the system clock. Data is transmitted as soon as a value is written to SPI1BUF. The interrupt is generated at the middle of the transfer of the last bit.

In Slave mode, data is transmitted and received as external clock pulses appear on SCK. Again, the interrupt is generated when the last bit is latched. If  $\overline{SS1}$  control is enabled, then transmission and reception are enabled only when  $\overline{SS1}$  = low. The SDO1 output will be disabled in  $\overline{SS1}$  mode with  $\overline{SS1}$  high.

The clock provided to the module is ( $F_{osc}/4$ ). This clock is then prescaled by the primary ( $PPRE<1:0>$ ) and the secondary ( $SPRE<2:0>$ ) prescale factors. The CKE bit determines whether transmit occurs on transition from active clock state to Idle clock state, or vice versa. The CKP bit selects the Idle state (high or low) for the clock.

## 13.1.1 WORD AND BYTE COMMUNICATION

A control bit,  $MODE16$  ( $SPI1CON<10>$ ), allows the module to communicate in either 16-bit or 8-bit mode. 16-bit operation is identical to 8-bit operation except that the number of bits transmitted is 16 instead of 8.

The user software must disable the module prior to changing the  $MODE16$  bit. The SPI module is reset when the  $MODE16$  bit is changed by the user.

A basic difference between 8-bit and 16-bit operation is that the data is transmitted out of bit 7 of the SPI1SR for 8-bit operation, and data is transmitted out of bit 15 of the SPI1SR for 16-bit operation. In both modes, data is shifted into bit 0 of the SPI1SR.

## 13.1.2 SDO1 DISABLE

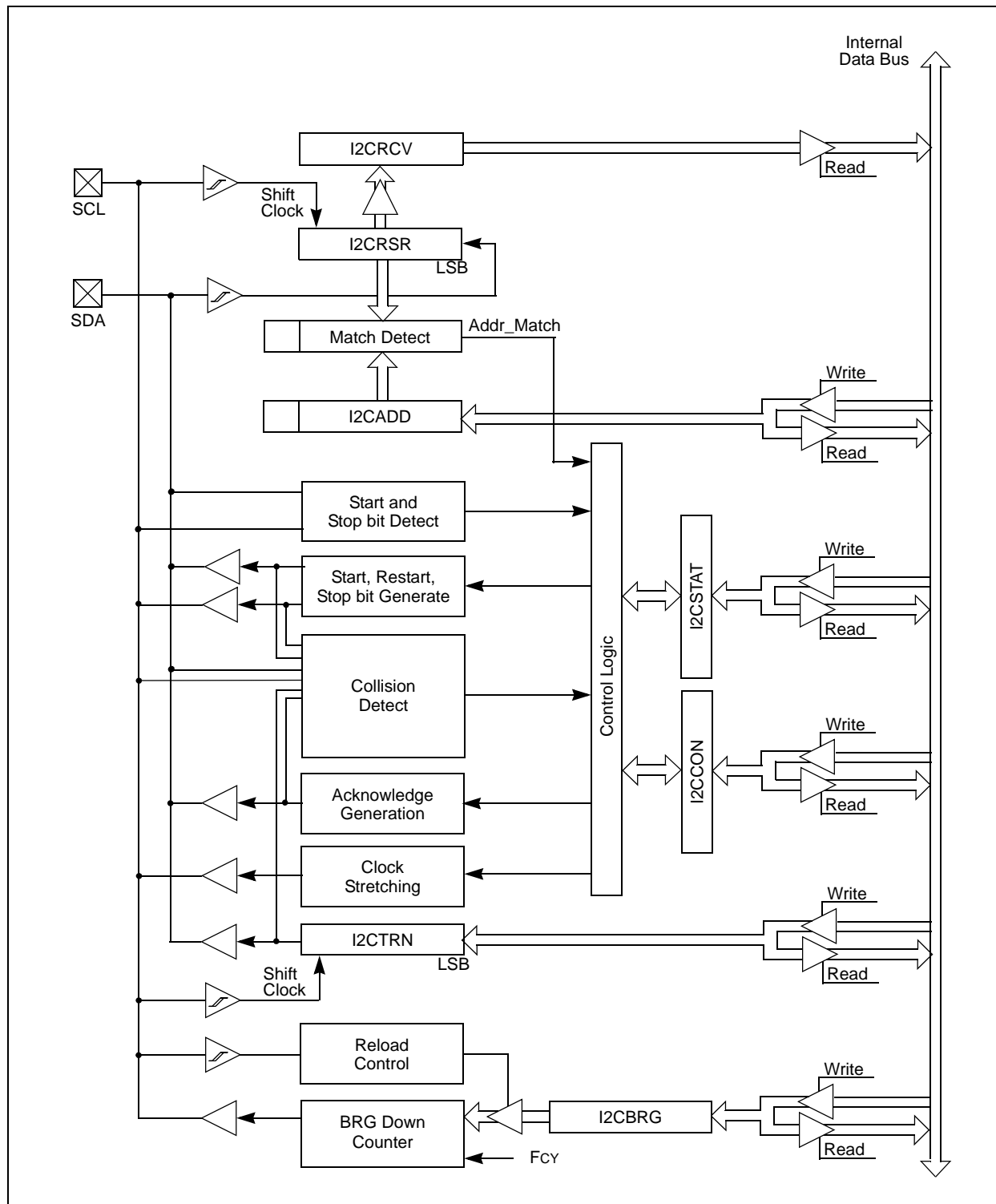
A control bit,  $DISSDO$ , is provided to the SPI1CON register to allow the SDO1 output to be disabled. This will allow the SPI module to be connected in an input only configuration. SDO1 can also be used for general purpose I/O.

## 13.2 Framed SPI Support

The module supports a basic framed SPI protocol in Master or Slave mode. The control bit,  $FRMEN$ , enables framed SPI support and causes the  $\overline{SS1}$  pin to perform the Frame Synchronization Pulse (FSYNC) function. The control bit,  $SPIFSD$ , determines whether the  $\overline{SS1}$  pin is an input or an output (i.e., whether the module receives or generates the Frame Synchronization Pulse). The frame pulse is an active-high pulse for a single SPI clock cycle. When Frame Synchronization is enabled, the data transmission starts only on the subsequent transmit edge of the SPI clock.

# dsPIC30F2011/2012/3012/3013

FIGURE 14-2: I<sup>2</sup>C™ BLOCK DIAGRAM



## 14.4.1 10-BIT MODE SLAVE TRANSMISSION

Once a slave is addressed in this fashion with the full 10-bit address (we will refer to this state as "PRIOR\_ADDR\_MATCH"), the master can begin sending data bytes for a slave reception operation.

## 14.4.2 10-BIT MODE SLAVE RECEPTION

Once addressed, the master can generate a Repeated Start, reset the high byte of the address and set the R\_W bit without generating a Stop bit, thus initiating a slave transmit operation.

## 14.5 Automatic Clock Stretch

In the Slave modes, the module can synchronize buffer reads and write to the master device by clock stretching.

### 14.5.1 TRANSMIT CLOCK STRETCHING

Both 10-bit and 7-bit Transmit modes implement clock stretching by asserting the SCLREL bit after the falling edge of the ninth clock, if the TBF bit is cleared, indicating the buffer is empty.

In Slave Transmit modes, clock stretching is always performed irrespective of the STREN bit.

Clock synchronization takes place following the ninth clock of the transmit sequence. If the device samples an ACK on the falling edge of the ninth clock and if the TBF bit is still clear, then the SCLREL bit is automatically cleared. The SCLREL being cleared to '0' will assert the SCL line low. The user's ISR must set the SCLREL bit before transmission is allowed to continue. By holding the SCL line low, the user has time to service the ISR and load the contents of the I2CTRN before the master device can initiate another transmit sequence.

**Note 1:** If the user loads the contents of I2CTRN, setting the TBF bit before the falling edge of the ninth clock, the SCLREL bit will not be cleared and clock stretching will not occur.

**2:** The SCLREL bit can be set in software, regardless of the state of the TBF bit.

### 14.5.2 RECEIVE CLOCK STRETCHING

The STREN bit in the I2CCON register can be used to enable clock stretching in Slave Receive mode. When the STREN bit is set, the SCL pin will be held low at the end of each data receive sequence.

### 14.5.3 CLOCK STRETCHING DURING 7-BIT ADDRESSING (STREN = 1)

When the STREN bit is set in Slave Receive mode, the SCL line is held low when the buffer register is full. The method for stretching the SCL output is the same for both 7 and 10-bit addressing modes.

Clock stretching takes place following the ninth clock of the receive sequence. On the falling edge of the ninth clock at the end of the ACK sequence, if the RBF bit is set, the SCLREL bit is automatically cleared, forcing the SCL output to be held low. The user's ISR must set the SCLREL bit before reception is allowed to continue. By holding the SCL line low, the user has time to service the ISR and read the contents of the I2CRCV before the master device can initiate another receive sequence. This will prevent buffer overruns from occurring.

**Note 1:** If the user reads the contents of the I2CRCV, clearing the RBF bit before the falling edge of the ninth clock, the SCLREL bit will not be cleared and clock stretching will not occur.

**2:** The SCLREL bit can be set in software regardless of the state of the RBF bit. The user should be careful to clear the RBF bit in the ISR before the next receive sequence in order to prevent an overflow condition.

### 14.5.4 CLOCK STRETCHING DURING 10-BIT ADDRESSING (STREN = 1)

Clock stretching takes place automatically during the addressing sequence. Because this module has a register for the entire address, it is not necessary for the protocol to wait for the address to be updated.

After the address phase is complete, clock stretching will occur on each data receive or transmit sequence as was described earlier.

## 14.6 Software Controlled Clock Stretching (STREN = 1)

When the STREN bit is '1', the SCLREL bit may be cleared by software to allow software to control the clock stretching. The logic will synchronize writes to the SCLREL bit with the SCL clock. Clearing the SCLREL bit will not assert the SCL output until the module detects a falling edge on the SCL output and SCL is sampled low. If the SCLREL bit is cleared by the user while the SCL line has been sampled low, the SCL output will be asserted (held low). The SCL output will remain low until the SCLREL bit is set, and all other devices on the I<sup>2</sup>C bus have de-asserted SCL. This ensures that a write to the SCLREL bit will not violate the minimum high time requirement for SCL.

If the STREN bit is '0', a software write to the SCLREL bit will be disregarded and have no effect on the SCLREL bit.

## 15.5.2 FRAMING ERROR (FERR)

The FERR bit (UxSTA<2>) is set if a '0' is detected instead of a Stop bit. If two Stop bits are selected, both Stop bits must be '1', otherwise FERR will be set. The read-only FERR bit is buffered along with the received data. It is cleared on any Reset.

## 15.5.3 PARITY ERROR (PERR)

The PERR bit (UxSTA<3>) is set if the parity of the received word is incorrect. This error bit is applicable only if a Parity mode (odd or even) is selected. The read-only PERR bit is buffered along with the received data bytes. It is cleared on any Reset.

## 15.5.4 IDLE STATUS

When the receiver is active (i.e., between the initial detection of the Start bit and the completion of the Stop bit), the RIDLE bit (UxSTA<4>) is '0'. Between the completion of the Stop bit and detection of the next Start bit, the RIDLE bit is '1', indicating that the UART is Idle.

## 15.5.5 RECEIVE BREAK

The receiver will count and expect a certain number of bit times based on the values programmed in the PDSEL (UxMODE<2:1>) and STSEL (UxMODE<0>) bits.

If the break is longer than 13 bit times, the reception is considered complete after the number of bit times specified by PDSEL and STSEL. The URXDA bit is set, FERR is set, zeros are loaded into the receive FIFO, interrupts are generated if appropriate and the RIDLE bit is set.

When the module receives a long break signal and the receiver has detected the Start bit, the data bits and the invalid Stop bit (which sets the FERR), the receiver must wait for a valid Stop bit before looking for the next Start bit. It cannot assume that the break condition on the line is the next Start bit.

Break is regarded as a character containing all '0's with the FERR bit set. The Break character is loaded into the buffer. No further reception can occur until a Stop bit is received. Note that RIDLE goes high when the Stop bit has not yet been received.

## 15.6 Address Detect Mode

Setting the ADDEN bit (UxSTA<5>) enables this special mode in which a 9th bit (URX8) value of '1' identifies the received word as an address, rather than data. This mode is only applicable for 9-bit data communication. The URXISEL control bit does not have any impact on interrupt generation in this mode since an interrupt (if enabled) will be generated every time the received word has the 9th bit set.

## 15.7 Loopback Mode

Setting the LPBACK bit enables this special mode in which the UxTX pin is internally connected to the UxRX pin. When configured for the Loopback mode, the UxRX pin is disconnected from the internal UART receive logic. However, the UxTX pin still functions as in a normal operation.

To select this mode:

- Configure UART for desired mode of operation.
- Set LPBACK = 1 to enable Loopback mode.
- Enable transmission as defined in **Section 15.3 "Transmitting Data"**.

## 15.8 Baud Rate Generator

The UART has a 16-bit Baud Rate Generator to allow maximum flexibility in baud rate generation. The Baud Rate Generator register (UxBRG) is readable and writable. The baud rate is computed as follows:

BRG = 16-bit value held in UxBRG register  
(0 through 65535)

FCY = Instruction Clock Rate (1/Tcy)

The baud rate is given by Equation 15-1.

### EQUATION 15-1: BAUD RATE

$$\text{Baud Rate} = \text{FCY} / (16 * (\text{BRG} + 1))$$

Therefore, the maximum baud rate possible is:

$\text{FCY} / 16$  (if BRG = 0),

and the minimum baud rate possible is:

$\text{FCY} / (16 * 65536)$ .

With a full 16-bit Baud Rate Generator at 30 MIPS operation, the minimum baud rate achievable is 28.5 bps.

## 15.9 Auto-Baud Support

To allow the system to determine baud rates of received characters, the input can be optionally linked to a selected capture input (IC1 for UART1 and IC2 for UART2). To enable this mode, you must program the input capture module to detect the falling and rising edges of the Start bit.

**TABLE 15-1: UART1 REGISTER MAP FOR dsPIC30F2011/2012/3012/3013**

SFR Name	Addr.	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset State
U1MODE	020C	UARTEN	—	USIDL	—	—	ALTIO	—	—	WAKE	LPBACK	ABAUD	—	—	PDSEL1	PDSEL0	STSEL	0000 0000 0000 0000
U1STA	020E	UTXISEL	—	—	—	UTXBRK	UTXEN	UTXBF	TRMT	URXISEL1	URXISEL0	ADDEN	RIDLE	PERR	FERR	OERR	URXDA	0000 0001 0001 0000
U1TXREG	0210	—	—	—	—	—	—	—	UTX8	Transmit Register								0000 000u uuuu uuuu
U1RXREG	0212	—	—	—	—	—	—	—	URX8	Receive Register								0000 0000 0000 0000
U1BRG	0214	Baud Rate Generator Prescaler																0000 0000 0000 0000

**Legend:** u = uninitialized bit; — = unimplemented bit, read as '0'

**TABLE 15-2: UART2 REGISTER MAP FOR dsPIC30F3013<sup>(1)</sup>**

SFR Name	Addr.	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset State
U2MODE	0216	UARTEN	—	USIDL	—	—	—	—	—	WAKE	LPBACK	ABAUD	—	—	PDSEL1	PDSEL0	STSEL	0000 0000 0000 0000
U2STA	0218	UTXISEL	—	—	—	UTXBRK	UTXEN	UTXBF	TRMT	URXISEL1	URXISEL0	ADDEN	RIDLE	PERR	FERR	OERR	URXDA	0000 0001 0001 0000
U2TXREG	021A	—	—	—	—	—	—	—	UTX8	Transmit Register								0000 000u uuuu uuuu
U2RXREG	021C	—	—	—	—	—	—	—	URX8	Receive Register								0000 0000 0000 0000
U2BRG	021E	Baud Rate Generator Prescaler																0000 0000 0000 0000

**Legend:** u = uninitialized bit; — = unimplemented bit, read as '0'

- Note 1:** UART2 is not available on dsPIC30F2011/2012/3012 devices.  
**Note 2:** Refer to the “dsPIC30F Family Reference Manual” (DS70046) for descriptions of register bit fields.

## 16.9 Module Power-Down Modes

The module has two internal power modes.

When the ADON bit is '1', the module is in Active mode; it is fully powered and functional.

When ADON is '0', the module is in Off mode. The digital and analog portions of the circuit are disabled for maximum current savings.

In order to return to the Active mode from Off mode, the user must wait for the ADC circuitry to stabilize.

## 16.10 A/D Operation During CPU Sleep and Idle Modes

### 16.10.1 A/D OPERATION DURING CPU SLEEP MODE

When the device enters Sleep mode, all clock sources to the module are shut down and stay at logic '0'.

If Sleep occurs in the middle of a conversion, the conversion is aborted. The converter will not continue with a partially completed conversion on exit from Sleep mode.

Register contents are not affected by the device entering or leaving Sleep mode.

The ADC module can operate during Sleep mode if the A/D clock source is set to RC (ADRC = 1). When the RC clock source is selected, the ADC module waits one instruction cycle before starting the conversion. This allows the *SLEEP* instruction to be executed which eliminates all digital switching noise from the conversion. When the conversion is complete, the CONV bit will be cleared and the result loaded into the ADCBUF register.

If the A/D interrupt is enabled, the device will wake-up from Sleep. If the A/D interrupt is not enabled, the ADC module will then be turned off, although the ADON bit will remain set.

### 16.10.2 A/D OPERATION DURING CPU IDLE MODE

The ADSIDL bit selects if the module will stop on Idle or continue on Idle. If ADSIDL = 0, the module will continue operation on assertion of Idle mode. If ADSIDL = 1, the module will stop on Idle.

## 16.11 Effects of a Reset

A device Reset forces all registers to their Reset state. This forces the ADC module to be turned off, and any conversion and sampling sequence is aborted. The values that are in the ADCBUF registers are not modified. The A/D Result register will contain unknown data after a Power-on Reset.

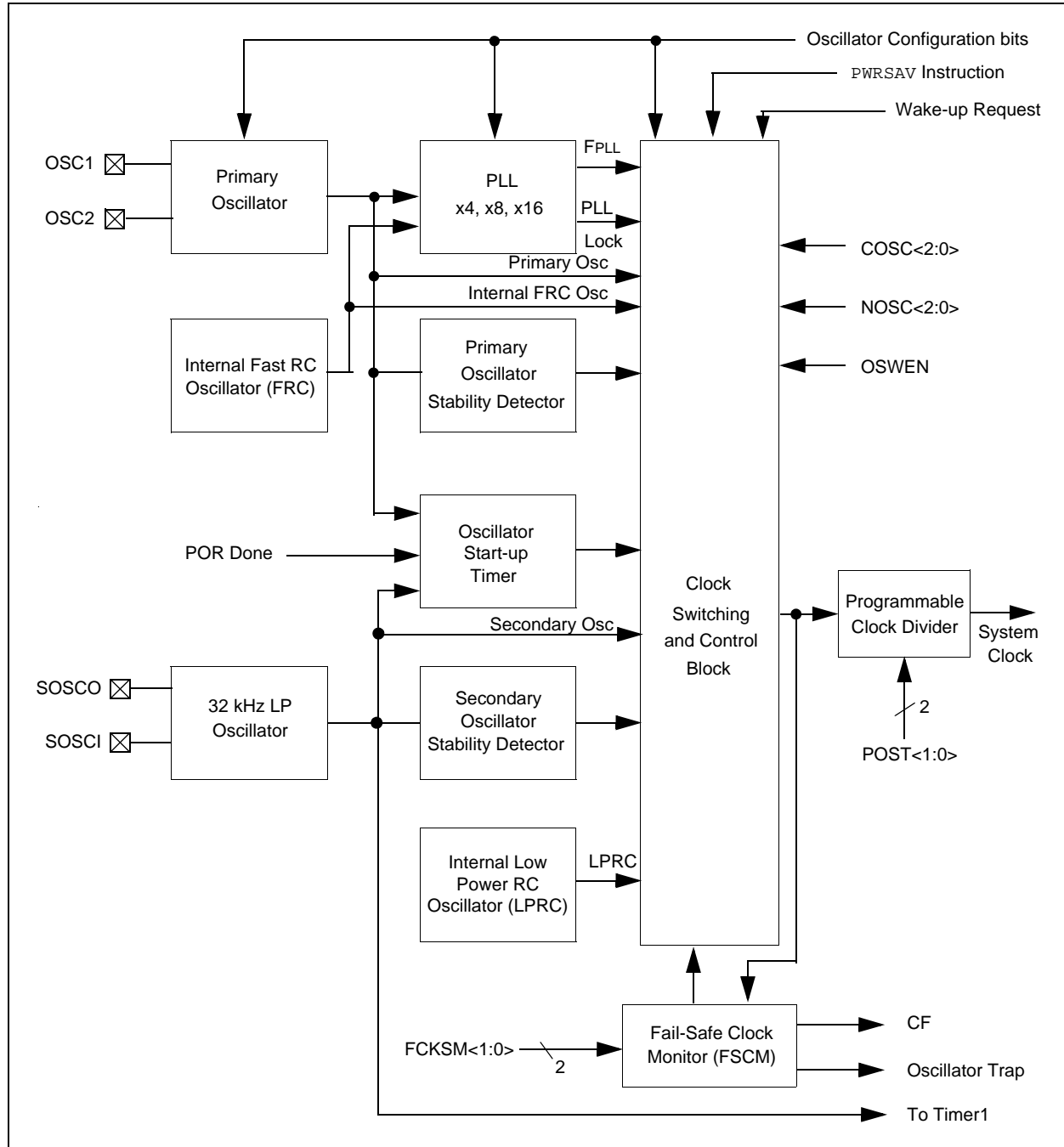
## 16.12 Output Formats

The A/D result is 12 bits wide. The data buffer RAM is also 12 bits wide. The 12-bit data can be read in one of four different formats. The FORM<1:0> bits select the format. Each of the output formats translates to a 16-bit result on the data bus.

**FIGURE 16-4: A/D OUTPUT DATA FORMATS**

RAM Contents:						d11	d10	d09	d08	d07	d06	d05	d04	d03	d02	d01	d00				
Read to Bus:																					
Signed Fractional						$\overline{d11}$	d10	d09	d08	d07	d06	d05	d04	d03	d02	d01	d00	0	0	0	0
Fractional						d11	d10	d09	d08	d07	d06	d05	d04	d03	d02	d01	d00	0	0	0	0
Signed Integer						$\overline{d11}$	$\overline{d11}$	$\overline{d11}$	$\overline{d11}$	$\overline{d11}$	d10	d09	d08	d07	d06	d05	d04	d03	d02	d01	d00
Integer						0	0	0	0	d11	d10	d09	d08	d07	d06	d05	d04	d03	d02	d01	d00

**FIGURE 17-1: OSCILLATOR SYSTEM BLOCK DIAGRAM**



## 17.3.1.1 POR with Long Crystal Start-up Time (with FSCM Enabled)

The oscillator start-up circuitry is not linked to the POR circuitry. Some crystal circuits (especially low frequency crystals) will have a relatively long start-up time. Therefore, one or more of the following conditions is possible after the POR timer and the PWRT have expired:

- The oscillator circuit has not begun to oscillate.
- The Oscillator Start-up Timer has not expired (if a crystal oscillator is used).
- The PLL has not achieved a LOCK (if PLL is used).

If the FSCM is enabled and one of the above conditions is true, then a clock failure trap will occur. The device will automatically switch to the FRC oscillator and the user can switch to the desired crystal oscillator in the trap ISR.

## 17.3.1.2 Operating without FSCM and PWRT

If the FSCM is disabled and the Power-up Timer (PWRT) is also disabled, then the device will exit rapidly from Reset on power-up. If the clock source is FRC, LPRC, ERC or EC, it will be active immediately.

If the FSCM is disabled and the system clock has not started, the device will be in a frozen state at the Reset vector until the system clock starts. From the user's perspective, the device will appear to be in Reset until a system clock is available.

## 17.3.2 BOR: PROGRAMMABLE BROWN-OUT RESET

The BOR (Brown-out Reset) module is based on an internal voltage reference circuit. The main purpose of the BOR module is to generate a device Reset when a brown-out condition occurs. Brown-out conditions are generally caused by glitches on the AC mains (i.e., missing portions of the AC cycle waveform due to bad power transmission lines, or voltage sags due to excessive current draw when a large inductive load is turned on).

The BOR module allows selection of one of the following voltage trip points (see Table 20-11):

- 2.6V-2.71V
- 4.1V-4.4V
- 4.58V-4.73V

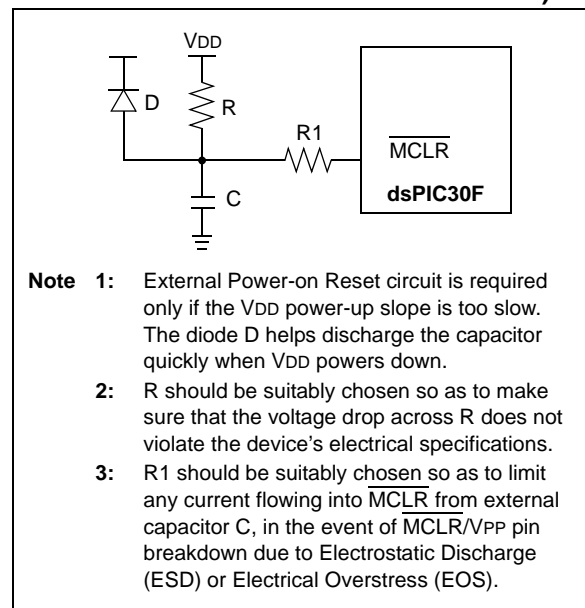
**Note:** The BOR voltage trip points indicated here are nominal values provided for design guidance only. Refer to the Electrical Specifications in the specific device data sheet for BOR voltage limit specifications.

A BOR will generate a Reset pulse which will reset the device. The BOR will select the clock source based on the device Configuration bit values (FOS<2:0> and FPR<4:0>). Furthermore, if an Oscillator mode is selected, the BOR will activate the Oscillator Start-up Timer (OST). The system clock is held until OST expires. If the PLL is used, then the clock will be held until the LOCK bit (OSCCON<5>) is '1'.

Concurrently, the POR time-out (TPOR) and the PWRT time-out (TPWRT) will be applied before the internal Reset is released. If TPWRT = 0 and a crystal oscillator is being used, then a nominal delay of TFSCM = 100  $\mu$ s is applied. The total delay in this case is (TPOR + TFSCM).

The BOR Status bit (RCON<1>) will be set to indicate that a BOR has occurred. The BOR circuit, if enabled, will continue to operate while in Sleep or Idle modes and will reset the device should VDD fall below the BOR threshold voltage.

**FIGURE 17-6: EXTERNAL POWER-ON RESET CIRCUIT (FOR SLOW VDD POWER-UP)**



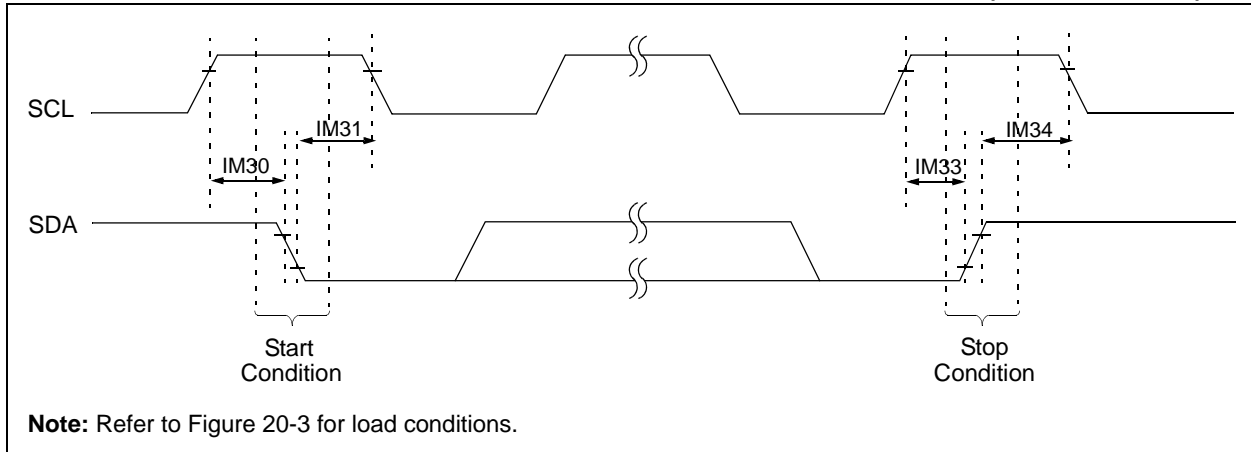
**Note:** Dedicated supervisory devices, such as the MCP1XX and MCP8XX, may also be used as an external Power-on Reset circuit.

# dsPIC30F2011/2012/3012/3013

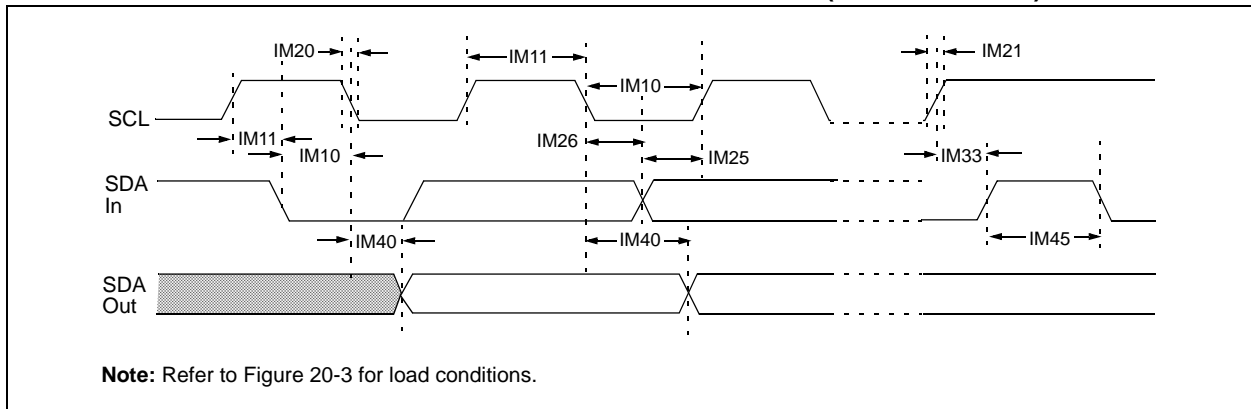
**TABLE 18-2: INSTRUCTION SET OVERVIEW (CONTINUED)**

Base Instr #	Assembly Mnemonic	Assembly Syntax	Description	# of Words	# of Cycles	Status Flags Affected
66	RRNC	RRNC <i>f</i>	<i>f</i> = Rotate Right (No Carry) <i>f</i>	1	1	N,Z
		RRNC <i>f</i> , WREG	WREG = Rotate Right (No Carry) <i>f</i>	1	1	N,Z
		RRNC <i>Ws</i> , <i>Wd</i>	<i>Wd</i> = Rotate Right (No Carry) <i>Ws</i>	1	1	N,Z
67	SAC	SAC <i>Acc</i> , #Slit4, <i>Wdo</i>	Store Accumulator	1	1	None
		SAC.R <i>Acc</i> , #Slit4, <i>Wdo</i>	Store Rounded Accumulator	1	1	None
68	SE	SE <i>Ws</i> , <i>Wnd</i>	<i>Wnd</i> = sign-extended <i>Ws</i>	1	1	C,N,Z
69	SETM	SETM <i>f</i>	<i>f</i> = 0xFFFF	1	1	None
		SETM WREG	WREG = 0xFFFF	1	1	None
		SETM <i>Ws</i>	<i>Ws</i> = 0xFFFF	1	1	None
70	SFTAC	SFTAC <i>Acc</i> , <i>Wn</i>	Arithmetic Shift Accumulator by ( <i>Wn</i> )	1	1	OA,OB,OAB,SA,SB,SAB
		SFTAC <i>Acc</i> , #Slit6	Arithmetic Shift Accumulator by Slit6	1	1	OA,OB,OAB,SA,SB,SAB
71	SL	SL <i>f</i>	<i>f</i> = Left Shift <i>f</i>	1	1	C,N,OV,Z
		SL <i>f</i> , WREG	WREG = Left Shift <i>f</i>	1	1	C,N,OV,Z
		SL <i>Ws</i> , <i>Wd</i>	<i>Wd</i> = Left Shift <i>Ws</i>	1	1	C,N,OV,Z
		SL <i>Wb</i> , <i>Wns</i> , <i>Wnd</i>	<i>Wnd</i> = Left Shift <i>Wb</i> by <i>Wns</i>	1	1	N,Z
		SL <i>Wb</i> , #lit5, <i>Wnd</i>	<i>Wnd</i> = Left Shift <i>Wb</i> by lit5	1	1	N,Z
72	SUB	SUB <i>Acc</i>	Subtract Accumulators	1	1	OA,OB,OAB,SA,SB,SAB
		SUB <i>f</i>	<i>f</i> = <i>f</i> - WREG	1	1	C,DC,N,OV,Z
		SUB <i>f</i> , WREG	WREG = <i>f</i> - WREG	1	1	C,DC,N,OV,Z
		SUB #lit10, <i>Wn</i>	<i>Wn</i> = <i>Wn</i> - lit10	1	1	C,DC,N,OV,Z
		SUB <i>Wb</i> , <i>Ws</i> , <i>Wd</i>	<i>Wd</i> = <i>Wb</i> - <i>Ws</i>	1	1	C,DC,N,OV,Z
		SUB <i>Wb</i> , #lit5, <i>Wd</i>	<i>Wd</i> = <i>Wb</i> - lit5	1	1	C,DC,N,OV,Z
73	SUBB	SUBB <i>f</i>	<i>f</i> = <i>f</i> - WREG - ( $\overline{C}$ )	1	1	C,DC,N,OV,Z
		SUBB <i>f</i> , WREG	WREG = <i>f</i> - WREG - ( $\overline{C}$ )	1	1	C,DC,N,OV,Z
		SUBB #lit10, <i>Wn</i>	<i>Wn</i> = <i>Wn</i> - lit10 - ( $\overline{C}$ )	1	1	C,DC,N,OV,Z
		SUBB <i>Wb</i> , <i>Ws</i> , <i>Wd</i>	<i>Wd</i> = <i>Wb</i> - <i>Ws</i> - ( $\overline{C}$ )	1	1	C,DC,N,OV,Z
		SUBB <i>Wb</i> , #lit5, <i>Wd</i>	<i>Wd</i> = <i>Wb</i> - lit5 - ( $\overline{C}$ )	1	1	C,DC,N,OV,Z
74	SUBR	SUBR <i>f</i>	<i>f</i> = WREG - <i>f</i>	1	1	C,DC,N,OV,Z
		SUBR <i>f</i> , WREG	WREG = WREG - <i>f</i>	1	1	C,DC,N,OV,Z
		SUBR <i>Wb</i> , <i>Ws</i> , <i>Wd</i>	<i>Wd</i> = <i>Ws</i> - <i>Wb</i>	1	1	C,DC,N,OV,Z
		SUBR <i>Wb</i> , #lit5, <i>Wd</i>	<i>Wd</i> = lit5 - <i>Wb</i>	1	1	C,DC,N,OV,Z
75	SUBBR	SUBBR <i>f</i>	<i>f</i> = WREG - <i>f</i> - ( $\overline{C}$ )	1	1	C,DC,N,OV,Z
		SUBBR <i>f</i> , WREG	WREG = WREG - <i>f</i> - ( $\overline{C}$ )	1	1	C,DC,N,OV,Z
		SUBBR <i>Wb</i> , <i>Ws</i> , <i>Wd</i>	<i>Wd</i> = <i>Ws</i> - <i>Wb</i> - ( $\overline{C}$ )	1	1	C,DC,N,OV,Z
		SUBBR <i>Wb</i> , #lit5, <i>Wd</i>	<i>Wd</i> = lit5 - <i>Wb</i> - ( $\overline{C}$ )	1	1	C,DC,N,OV,Z
76	SWAP	SWAP.b <i>Wn</i>	<i>Wn</i> = nibble swap <i>Wn</i>	1	1	None
		SWAP <i>Wn</i>	<i>Wn</i> = byte swap <i>Wn</i>	1	1	None
77	TBLRDH	TBLRDH <i>Ws</i> , <i>Wd</i>	Read Prog<23:16> to <i>Wd</i> <7:0>	1	2	None
78	TBLRDL	TBLRDL <i>Ws</i> , <i>Wd</i>	Read Prog<15:0> to <i>Wd</i>	1	2	None
79	TBLWTH	TBLWTH <i>Ws</i> , <i>Wd</i>	Write <i>Ws</i> <7:0> to Prog<23:16>	1	2	None
80	TBLWTL	TBLWTL <i>Ws</i> , <i>Wd</i>	Write <i>Ws</i> to Prog<15:0>	1	2	None
81	ULNK	ULNK	Unlink frame pointer	1	1	None
82	XOR	XOR <i>f</i>	<i>f</i> = <i>f</i> .XOR. WREG	1	1	N,Z
		XOR <i>f</i> , WREG	WREG = <i>f</i> .XOR. WREG	1	1	N,Z
		XOR #lit10, <i>Wn</i>	<i>Wd</i> = lit10 .XOR. <i>Wd</i>	1	1	N,Z
		XOR <i>Wb</i> , <i>Ws</i> , <i>Wd</i>	<i>Wd</i> = <i>Wb</i> .XOR. <i>Ws</i>	1	1	N,Z
		XOR <i>Wb</i> , #lit5, <i>Wd</i>	<i>Wd</i> = <i>Wb</i> .XOR. lit5	1	1	N,Z
83	ZE	ZE <i>Ws</i> , <i>Wnd</i>	<i>Wnd</i> = Zero-extend <i>Ws</i>	1	1	C,Z,N

**FIGURE 20-16: I<sup>2</sup>C™ BUS START/STOP BITS TIMING CHARACTERISTICS (MASTER MODE)**



**FIGURE 20-17: I<sup>2</sup>C™ BUS DATA TIMING CHARACTERISTICS (MASTER MODE)**



# dsPIC30F2011/2012/3012/3013

**TABLE 20-33: I<sup>2</sup>C™ BUS DATA TIMING REQUIREMENTS (MASTER MODE)**

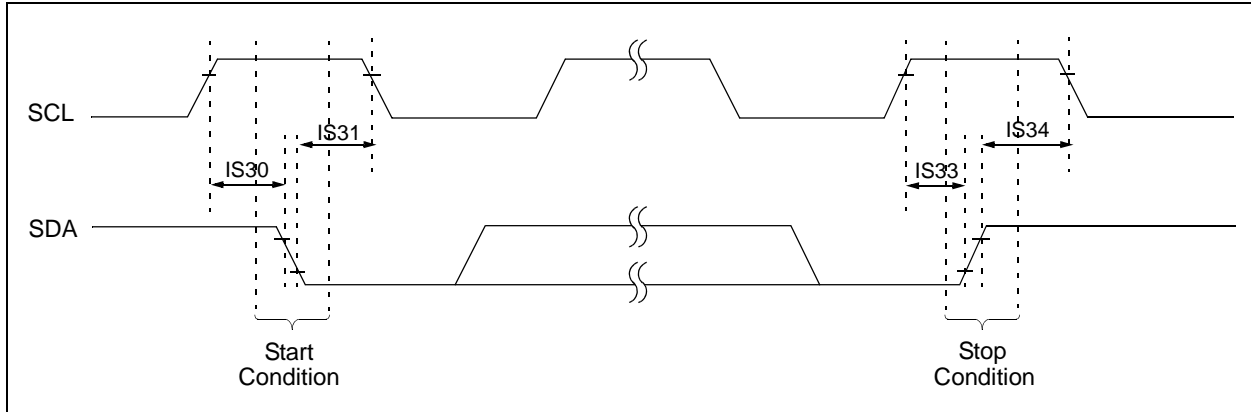
AC CHARACTERISTICS				Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for Industrial -40°C ≤ TA ≤ +125°C for Extended			
Param No.	Symbol	Characteristic		Min <sup>(1)</sup>	Max	Units	Conditions
IM10	TLO:SCL	Clock Low Time	100 kHz mode	Tcy/2 (BRG + 1)	—	μs	
			400 kHz mode	Tcy/2 (BRG + 1)	—	μs	
			1 MHz mode <sup>(2)</sup>	Tcy/2 (BRG + 1)	—	μs	
IM11	THI:SCL	Clock High Time	100 kHz mode	Tcy/2 (BRG + 1)	—	μs	
			400 kHz mode	Tcy/2 (BRG + 1)	—	μs	
			1 MHz mode <sup>(2)</sup>	Tcy/2 (BRG + 1)	—	μs	
IM20	TF:SCL	SDA and SCL Fall Time	100 kHz mode	—	300	ns	Cb is specified to be from 10 to 400 pF
			400 kHz mode	20 + 0.1 Cb	300	ns	
			1 MHz mode <sup>(2)</sup>	—	100	ns	
IM21	TR:SCL	SDA and SCL Rise Time	100 kHz mode	—	1000	ns	Cb is specified to be from 10 to 400 pF
			400 kHz mode	20 + 0.1 Cb	300	ns	
			1 MHz mode <sup>(2)</sup>	—	300	ns	
IM25	TSU:DAT	Data Input Setup Time	100 kHz mode	250	—	ns	
			400 kHz mode	100	—	ns	
			1 MHz mode <sup>(2)</sup>	—	—	ns	
IM26	THD:DAT	Data Input Hold Time	100 kHz mode	0	—	ns	
			400 kHz mode	0	0.9	μs	
			1 MHz mode <sup>(2)</sup>	—	—	ns	
IM30	TSU:STA	Start Condition Setup Time	100 kHz mode	Tcy/2 (BRG + 1)	—	μs	Only relevant for Repeated Start condition
			400 kHz mode	Tcy/2 (BRG + 1)	—	μs	
			1 MHz mode <sup>(2)</sup>	Tcy/2 (BRG + 1)	—	μs	
IM31	THD:STA	Start Condition Hold Time	100 kHz mode	Tcy/2 (BRG + 1)	—	μs	After this period the first clock pulse is generated
			400 kHz mode	Tcy/2 (BRG + 1)	—	μs	
			1 MHz mode <sup>(2)</sup>	Tcy/2 (BRG + 1)	—	μs	
IM33	TSU:STO	Stop Condition Setup Time	100 kHz mode	Tcy/2 (BRG + 1)	—	μs	
			400 kHz mode	Tcy/2 (BRG + 1)	—	μs	
			1 MHz mode <sup>(2)</sup>	Tcy/2 (BRG + 1)	—	μs	
IM34	THD:STO	Stop Condition Hold Time	100 kHz mode	Tcy/2 (BRG + 1)	—	ns	
			400 kHz mode	Tcy/2 (BRG + 1)	—	ns	
			1 MHz mode <sup>(2)</sup>	Tcy/2 (BRG + 1)	—	ns	
IM40	TAA:SCL	Output Valid From Clock	100 kHz mode	—	3500	ns	
			400 kHz mode	—	1000	ns	
			1 MHz mode <sup>(2)</sup>	—	—	ns	
IM45	TBF:SDA	Bus Free Time	100 kHz mode	4.7	—	μs	Time the bus must be free before a new transmission can start
			400 kHz mode	1.3	—	μs	
			1 MHz mode <sup>(2)</sup>	—	—	μs	
IM50	CB	Bus Capacitive Loading		—	400	pF	

**Note 1:** BRG is the value of the I<sup>2</sup>C Baud Rate Generator. Refer to **Section 21. “Inter-Integrated Circuit™ (I<sup>2</sup>C)”** (DS70068) in the *dsPIC30F Family Reference Manual* (DS70046).

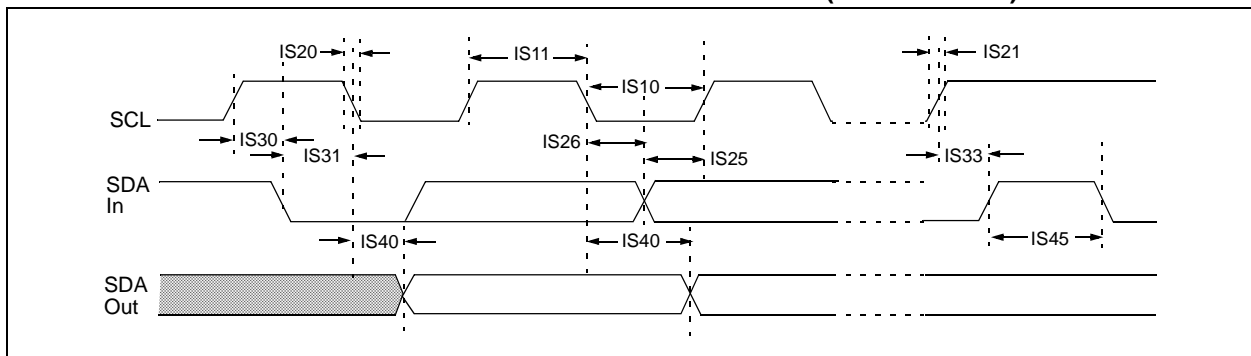
**2:** Maximum pin capacitance = 10 pF for all I<sup>2</sup>C™ pins (for 1 MHz mode only).

# dsPIC30F2011/2012/3012/3013

**FIGURE 20-18: I<sup>2</sup>C™ BUS START/STOP BITS TIMING CHARACTERISTICS (SLAVE MODE)**



**FIGURE 20-19: I<sup>2</sup>C™ BUS DATA TIMING CHARACTERISTICS (SLAVE MODE)**



**TABLE 20-34: I<sup>2</sup>C™ BUS DATA TIMING REQUIREMENTS (SLAVE MODE)**

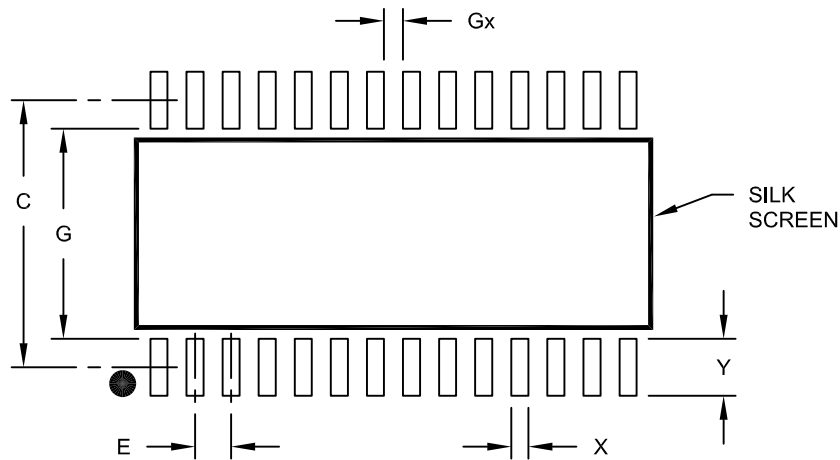
AC CHARACTERISTICS				Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤ T <sub>A</sub> ≤ +85°C for Industrial -40°C ≤ T <sub>A</sub> ≤ +125°C for Extended			
Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
IS10	TLO:SCL	Clock Low Time	100 kHz mode	4.7	—	μs	Device must operate at a minimum of 1.5 MHz
			400 kHz mode	1.3	—	μs	Device must operate at a minimum of 10 MHz.
			1 MHz mode <sup>(1)</sup>	0.5	—	μs	
IS11	THI:SCL	Clock High Time	100 kHz mode	4.0	—	μs	Device must operate at a minimum of 1.5 MHz
			400 kHz mode	0.6	—	μs	Device must operate at a minimum of 10 MHz
			1 MHz mode <sup>(1)</sup>	0.5	—	μs	
IS20	TF:SCL	SDA and SCL Fall Time	100 kHz mode	—	300	ns	Cb is specified to be from 10 to 400 pF
			400 kHz mode	20 + 0.1 Cb	300	ns	
			1 MHz mode <sup>(1)</sup>	—	100	ns	
IS21	TR:SCL	SDA and SCL Rise Time	100 kHz mode	—	1000	ns	Cb is specified to be from 10 to 400 pF
			400 kHz mode	20 + 0.1 Cb	300	ns	
			1 MHz mode <sup>(1)</sup>	—	300	ns	

**Note 1:** Maximum pin capacitance = 10 pF for all I<sup>2</sup>C™ pins (for 1 MHz mode only).

# dsPIC30F2011/2012/3012/3013

28-Lead Plastic Small Outline (SO) - Wide, 7.50 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



## RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		1.27 BSC	
Contact Pad Spacing	C		9.40	
Contact Pad Width (X28)	X			0.60
Contact Pad Length (X28)	Y			2.00
Distance Between Pads	Gx	0.67		
Distance Between Pads	G	7.40		

### Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

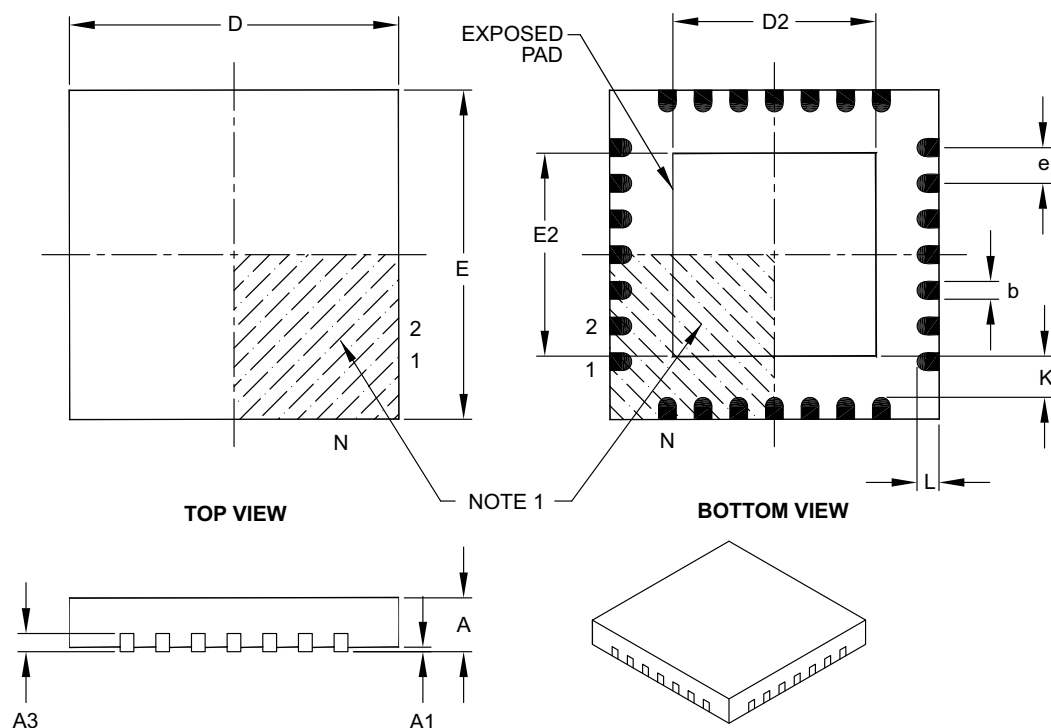
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2052A

# dsPIC30F2011/2012/3012/3013

## 28-Lead Plastic Quad Flat, No Lead Package (MM) – 6x6x0.9 mm Body [QFN-S] with 0.40 mm Contact Length

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N	28		
Pitch	e	0.65 BSC		
Overall Height	A	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3	0.20 REF		
Overall Width	E	6.00 BSC		
Exposed Pad Width	E2	3.65	3.70	4.70
Overall Length	D	6.00 BSC		
Exposed Pad Length	D2	3.65	3.70	4.70
Contact Width	b	0.23	0.38	0.43
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	K	0.20	—	—

### Notes:

- Pin 1 visual index feature may vary, but must be located within the hatched area.
- Package is saw singulated.
- Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
  - REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-124B

# dsPIC30F2011/2012/3012/3013

## INDEX

### Numerics

12-bit Analog-to-Digital Converter (A/D) Module .....	113
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### A

A/D .....	113
Aborting a Conversion .....	115
ADCHS Register .....	113
ADCON1 Register .....	113
ADCON2 Register .....	113
ADCON3 Register .....	113
ADCSSL Register .....	113
ADPCFG Register .....	113
Configuring Analog Port Pins .....	60, 119
Connection Considerations .....	119
Conversion Operation .....	114
Effects of a Reset .....	118
Operation During CPU Idle Mode .....	118
Operation During CPU Sleep Mode .....	118
Output Formats .....	118
Power-Down Modes .....	118
Programming the Sample Trigger .....	115
Register Map .....	121
Result Buffer .....	114
Sampling Requirements .....	117
Selecting the Conversion Sequence .....	114
AC Characteristics .....	160
Load Conditions .....	160
AC Temperature and Voltage Specifications .....	160
ADC .....	
Selecting the Conversion Clock .....	115
ADC Conversion Speeds .....	116
Address Generator Units .....	43
Alternate Vector Table .....	69
Analog-to-Digital Converter. <i>See</i> ADC.	
Assembler .....	
MPASM Assembler .....	146
Automatic Clock Stretch .....	100
During 10-bit Addressing (STREN = 1) .....	100
During 7-bit Addressing (STREN = 1) .....	100
Receive Mode .....	100
Transmit Mode .....	100

### B

Bandgap Start-up Time .....	
Requirements .....	166
Timing Characteristics .....	166
Barrel Shifter .....	27
Bit-Reversed Addressing .....	46
Example .....	47
Implementation .....	46
Modifier Values Table .....	47
Sequence Table (16-Entry) .....	47
Block Diagrams .....	
12-bit ADC Functional .....	113
16-bit Timer1 Module .....	73
16-bit Timer2 .....	79
16-bit Timer3 .....	79
32-bit Timer2/3 .....	78
DSP Engine .....	24
dsPIC30F2011 .....	12
dsPIC30F2012 .....	13
dsPIC30F3013 .....	15
External Power-on Reset Circuit .....	131

I <sup>2</sup> C .....	98
Input Capture Mode .....	83
Oscillator System .....	125
Output Compare Mode .....	87
Reset System .....	129
Shared Port Structure .....	59
SPI .....	94
SPI Master/Slave Connection .....	95
UART Receiver .....	106
UART Transmitter .....	105
BOR Characteristics .....	158
BOR. <i>See</i> Brown-out Reset.	
Brown-out Reset .....	
Characteristics .....	158
Timing Requirements .....	165

### C

C Compilers .....	
MPLAB C18 .....	146
CAN Module .....	
I/O Timing Characteristics .....	181
I/O Timing Requirements .....	181
CLKOUT and I/O Timing .....	
Characteristics .....	164
Requirements .....	164
Code Examples .....	
Data EEPROM Block Erase .....	56
Data EEPROM Block Write .....	58
Data EEPROM Read .....	55
Data EEPROM Word Erase .....	56
Data EEPROM Word Write .....	57
Erasing a Row of Program Memory .....	51
Initiating a Programming Sequence .....	52
Loading Write Latches .....	52
Code Protection .....	123
Control Registers .....	50
NVMADR .....	50
NVMADRU .....	50
NVMCON .....	50
NVMKEY .....	50
Core Architecture .....	
Overview .....	19
CPU Architecture Overview .....	19
Customer Change Notification Service .....	205
Customer Notification Service .....	205
Customer Support .....	205

### D

Data Accumulators and Adder/Subtractor .....	25
Data Space Write Saturation .....	27
Overflow and Saturation .....	25
Round Logic .....	26
Write-Back .....	26
Data Address Space .....	35
Alignment .....	38
Alignment (Figure) .....	38
Effect of Invalid Memory Accesses (Table) .....	38
MCU and DSP (MAC Class) Instructions Example ....	37
Memory Map .....	35, 36
Near Data Space .....	39
Software Stack .....	39
Spaces .....	38
Width .....	38
Data EEPROM Memory .....	55
Erasing .....	56
Erasing, Block .....	56