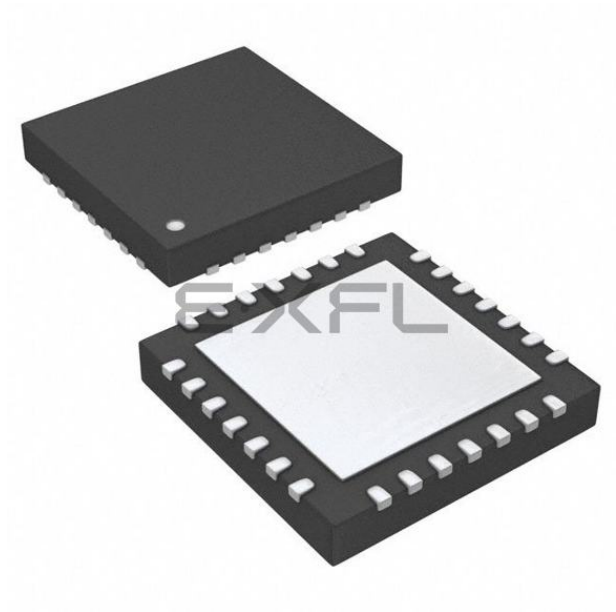
Microchip Technology - PIC16LF648A-I/ML Datasheet



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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	16
Program Memory Size	7KB (4K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 5.5V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-VQFN Exposed Pad
Supplier Device Package	28-QFN (6x6)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf648a-i-ml

4.2.2.1 Status Register

The Status register, shown in Register 4-1, contains the arithmetic status of the ALU; the Reset status and the bank select bits for data memory (SRAM).

The Status register can be the destination for any instruction, like any other register. If the Status register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the $\overline{TO}$ and $\overline{PD}$ bits are non-writable. Therefore, the result of an instruction with the Status register as destination may be different than intended.

For example, `CLRF STATUS` will clear the upper-three bits and set the Z bit. This leaves the Status register as `"000uu1uu"` (where `u` = unchanged).

It is recommended, therefore, that only `BCF`, `BSF`, `SWAPF` and `MOVWF` instructions are used to alter the Status register because these instructions do not affect any Status bit. For other instructions, not affecting any Status bits, see the "Instruction Set Summary".

Note:	The C and DC bits operate as a $\overline{\text{Borrow}}$ and $\overline{\text{Digit Borrow}}$ out bit, respectively, in subtraction. See the `SUBLW` and `SUBWF` instructions for examples.

REGISTER 4-1: STATUS – STATUS REGISTER (ADDRESS: 03h, 83h, 103h, 183h)

R/W-0	R/W-0	R/W-0	R-1	R-1	R/W-x	R/W-x	R/W-x
IRP	RP1	RP0	$\overline{TO}$	$\overline{PD}$	Z	DC	C
bit 7							bit 0

bit 7 **IRP**: Register Bank Select bit (used for indirect addressing)
`1` = Bank 2, 3 (100h-1FFh)
`0` = Bank 0, 1 (00h-FFh)

bit 6-5 **RP<1:0>**: Register Bank Select bits (used for direct addressing)
`00` = Bank 0 (00h-7Fh)
`01` = Bank 1 (80h-FFh)
`10` = Bank 2 (100h-17Fh)
`11` = Bank 3 (180h-1FFh)

bit 4 **$\overline{TO}$**: Time Out bit
`1` = After power-up, `CLRWDT` instruction or `SLEEP` instruction
`0` = A WDT time out occurred

bit 3 **$\overline{PD}$**: Power-down bit
`1` = After power-up or by the `CLRWDT` instruction
`0` = By execution of the `SLEEP` instruction

bit 2 **Z**: Zero bit
`1` = The result of an arithmetic or logic operation is zero
`0` = The result of an arithmetic or logic operation is not zero

bit 1 **DC**: Digit Carry/$\overline{\text{Borrow}}$ bit (`ADDWF`, `ADDLW`, `SUBLW`, `SUBWF` instructions) (for $\overline{\text{Borrow}}$ the polarity is reversed)
`1` = A carry-out from the 4th low order bit of the result occurred
`0` = No carry-out from the 4th low order bit of the result

bit 0 **C**: Carry/$\overline{\text{Borrow}}$ bit (`ADDWF`, `ADDLW`, `SUBLW`, `SUBWF` instructions)
`1` = A carry-out from the Most Significant bit of the result occurred
`0` = No carry-out from the Most Significant bit of the result occurred

Note:	For $\overline{\text{Borrow}}$, the polarity is reversed. A subtraction is executed by adding the two's complement of the second operand. For rotate (`RRF`, `RLF`) instructions, this bit is loaded with either the high or low order bit of the source register.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

4.2.2.3 INTCON Register

The INTCON register is a readable and writable register, which contains the various enable and flag bits for all interrupt sources except the comparator module. See **Section 4.2.2.4 "PIE1 Register"** and **Section 4.2.2.5 "PIR1 Register"** for a description of the comparator enable and flag bits.

Note:	Interrupt flag bits get set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the global enable bit, GIE (INTCON<7>).

REGISTER 4-3: INTCON – INTERRUPT CONTROL REGISTER (ADDRESS: 0Bh, 8Bh, 10Bh, 18Bh)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-x
GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF
bit 7							bit 0

bit 7 **GIE:** Global Interrupt Enable bit
1 = Enables all un-masked interrupts
0 = Disables all interrupts

bit 6 **PEIE**: Peripheral Interrupt Enable bit
1 = Enables all un-masked peripheral interrupts
0 = Disables all peripheral interrupts

bit 5 **T0IE**: TMR0 Overflow Interrupt Enable bit
1 = Enables the TMR0 interrupt
0 = Disables the TMR0 interrupt

bit 4 **INTE**: RB0/INT External Interrupt Enable bit
1 = Enables the RB0/INT external interrupt
0 = Disables the RB0/INT external interrupt

bit 3 **RBIE**: RB Port Change Interrupt Enable bit
1 = Enables the RB port change interrupt
0 = Disables the RB port change interrupt

bit 2 **T0IF**: TMR0 Overflow Interrupt Flag bit
1 = TMR0 register has overflowed (must be cleared in software)
0 = TMR0 register did not overflow

bit 1 **INTF**: RB0/INT External Interrupt Flag bit
1 = The RB0/INT external interrupt occurred (must be cleared in software)
0 = The RB0/INT external interrupt did not occur

bit 0 **RBIF**: RB Port Change Interrupt Flag bit
1 = When at least one of the RB<7:4> pins changes state (must be cleared in software)
0 = None of the RB<7:4> pins have changed state

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

4.2.2.4 PIE1 Register

This register contains interrupt enable bits.

REGISTER 4-4: PIE1 – PERIPHERAL INTERRUPT ENABLE REGISTER 1 (ADDRESS: 8Ch)

R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0
EEIE	CMIE	RCIE	TXIE	—	CCP1IE	TMR2IE	TMR1IE
bit 7							bit 0

bit 7 **EEIE:** EE Write Complete Interrupt Enable Bit
1 = Enables the EE write complete interrupt
0 = Disables the EE write complete interrupt

bit 6 **CMIE**: Comparator Interrupt Enable bit
1 = Enables the comparator interrupt
0 = Disables the comparator interrupt

bit 5 **RCIE**: USART Receive Interrupt Enable bit
1 = Enables the USART receive interrupt
0 = Disables the USART receive interrupt

bit 4 **TXIE**: USART Transmit Interrupt Enable bit
1 = Enables the USART transmit interrupt
0 = Disables the USART transmit interrupt

bit 3 **Unimplemented**: Read as '0'

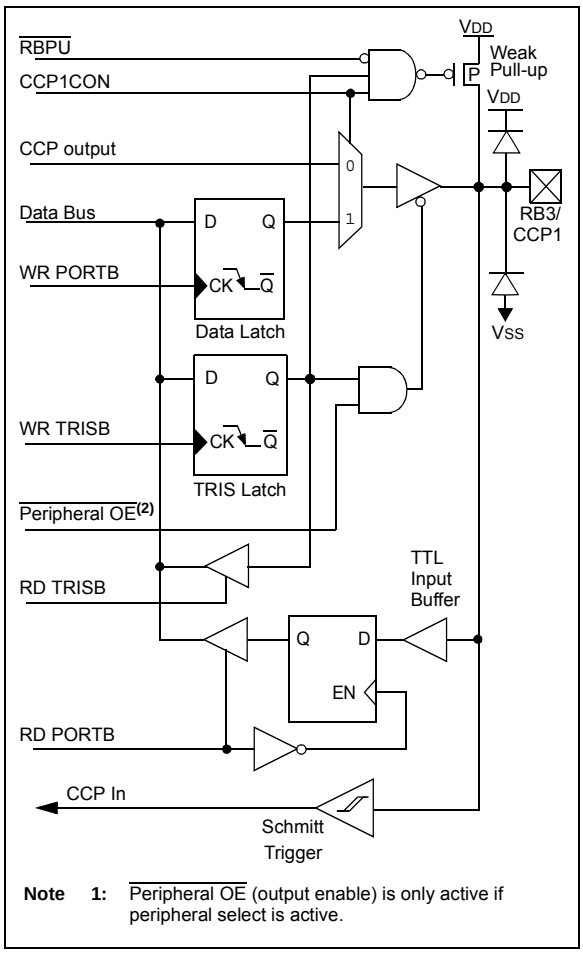
bit 2 **CCP1IE**: CCP1 Interrupt Enable bit
1 = Enables the CCP1 interrupt
0 = Disables the CCP1 interrupt

bit 1 **TMR2IE**: TMR2 to PR2 Match Interrupt Enable bit
1 = Enables the TMR2 to PR2 match interrupt
0 = Disables the TMR2 to PR2 match interrupt

bit 0 **TMR1IE**: TMR1 Overflow Interrupt Enable bit
1 = Enables the TMR1 overflow interrupt
0 = Disables the TMR1 overflow interrupt

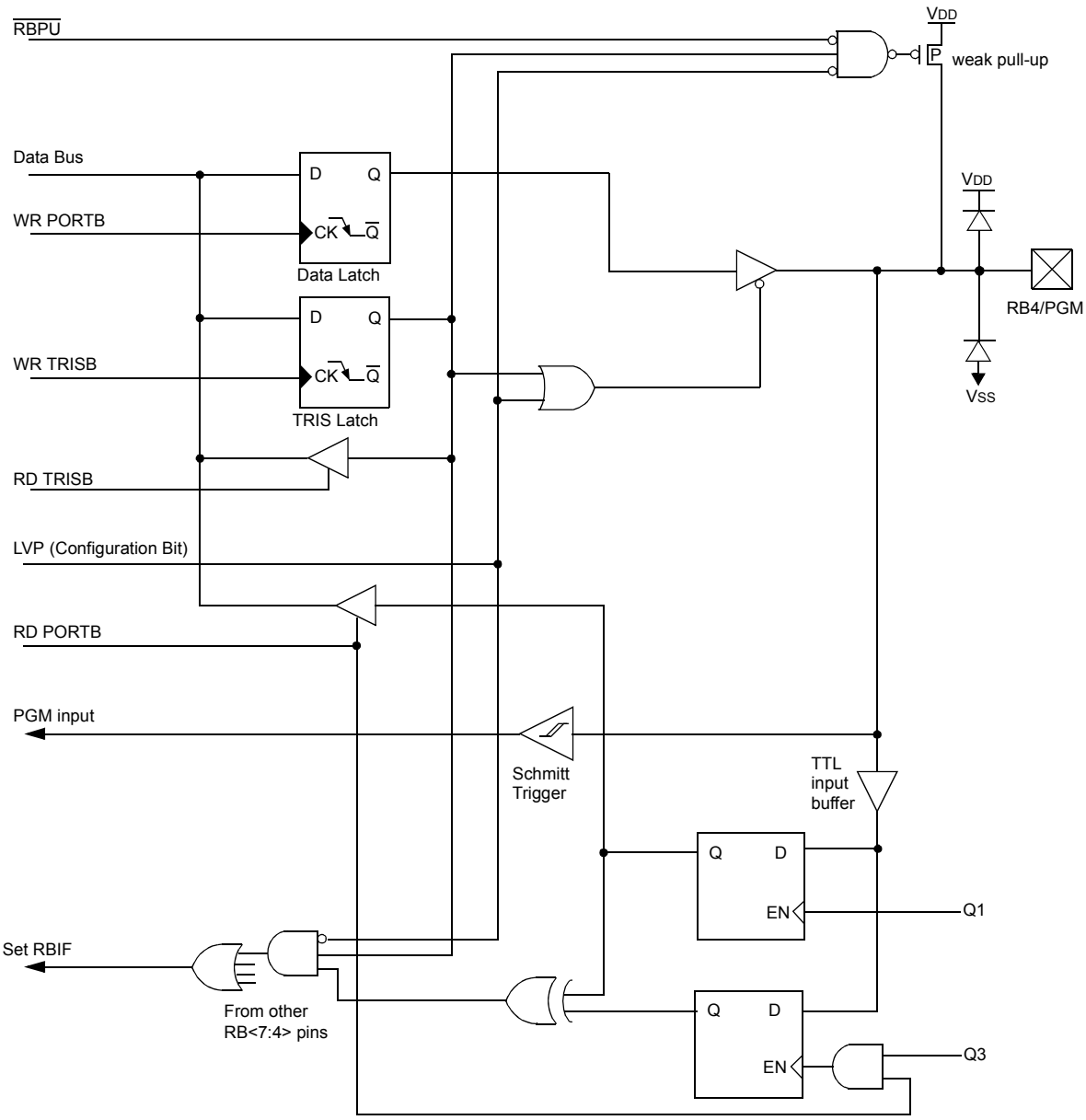
Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

FIGURE 5-11: BLOCK DIAGRAM OF RB3/CCP1 PIN



Note 1: Peripheral OE (output enable) is only active if peripheral select is active.

FIGURE 5-12: BLOCK DIAGRAM OF RB4/PGM PIN



Note: The low-voltage programming disables the interrupt-on-change and the weak pull-ups on RB4.

5.3 I/O Programming Considerations

5.3.1 BIDIRECTIONAL I/O PORTS

Any instruction that writes operates internally as a read followed by a write operation. The BCF and BSF instructions, for example, read the register into the CPU, execute the bit operation and write the result back to the register. Caution must be used when these instructions are applied to a port with both inputs and outputs defined. For example, a BSF operation on bit 5 of PORTB will cause all eight bits of PORTB to be read into the CPU. Then the BSF operation takes place on bit 5 and PORTB is written to the output latches. If another bit of PORTB is used as a bidirectional I/O pin (e.g., bit 0) and is defined as an input at this time, the input signal present on the pin itself would be read into the CPU and rewritten to the data latch of this particular pin, overwriting the previous content. As long as the pin stays in the Input mode, no problem occurs. However, if bit 0 is switched into Output mode later on, the content of the data latch may now be unknown.

Reading a port register reads the values of the port pins. Writing to the port register writes the value to the port latch. When using read-modify-write instructions (ex. BCF, BSF, etc.) on a port, the value of the port pins is read, the desired operation is done to this value, and this value is then written to the port latch.

Example 5-2 shows the effect of two sequential read-modify-write instructions (ex., BCF, BSF, etc.) on an I/O port.

A pin actively outputting a Low or High should not be driven from external devices at the same time in order to change the level on this pin ("wired-OR", "wired-AND"). The resulting high output currents may damage the chip.

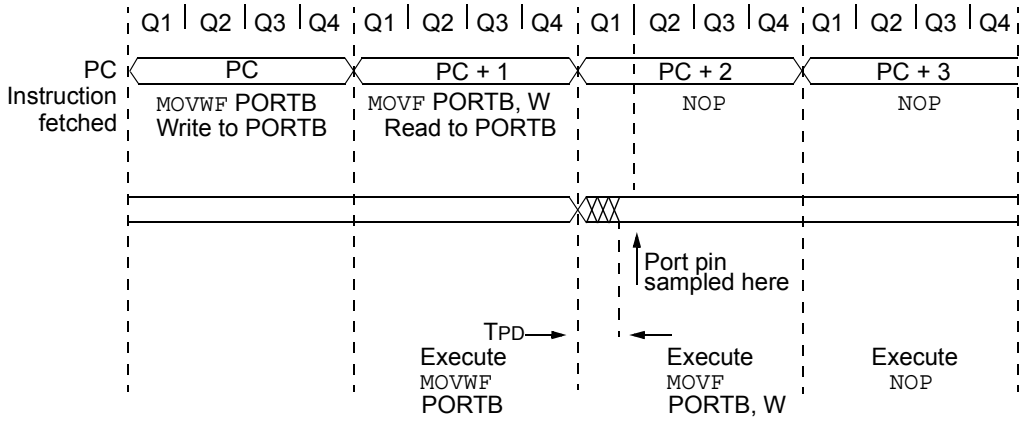
EXAMPLE 5-2: READ-MODIFY-WRITE INSTRUCTIONS ON AN I/O PORT

```
;Initial PORT settings:PORTB<7:4> Inputs
;                      PORTB<3:0> Outputs
;PORTB<7:6> have external pull-up and are
;not connected to other circuitry
;
;                      PORT latchPORT Pins
                       ---------- ----------
    BCF STATUS, RP0    ;
    BCF PORTB, 7       ;01pp pppp 11pp pppp
    BSF STATUS, RP0    ;
    BCF TRISB, 7       ;10pp pppp 11pp pppp
    BCF TRISB, 6       ;10pp pppp 10pp pppp
;
;Note that the user may have expected the
;pin values to be 00pp pppp. The 2nd BCF
;caused RB7 to be latched as the pin value
;(High).
```

5.3.2 SUCCESSIVE OPERATIONS ON I/O PORTS

The actual write to an I/O port happens at the end of an instruction cycle, whereas for reading, the data must be valid at the beginning of the instruction cycle (Figure 5-16). Therefore, care must be exercised if a write followed by a read operation is carried out on the same I/O port. The sequence of instructions should be such to allow the pin voltage to stabilize (load dependent) before the next instruction, which causes that file to be read into the CPU, is executed. Otherwise, the previous state of that pin may be read into the CPU rather than the new state. When in doubt, it is better to separate these instructions with a NOP or another instruction not accessing this I/O port.

FIGURE 5-16: SUCCESSIVE I/O OPERATION



Note 1: This example shows write to PORTB followed by a read from PORTB.

2: Data setup time = (0.25 TCY - TPD) where TCY = instruction cycle and TPD = propagation delay of Q1 cycle to output valid. Therefore, at higher clock frequencies, a write followed by a read may be problematic.

6.3.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control (i.e., it can be changed “on-the-fly” during program execution). Use the instruction sequences shown in Example 6-1 when changing the prescaler assignment from Timer0 to WDT, to avoid an unintended device Reset.

EXAMPLE 6-1: CHANGING PRESCALER (TIMER0 → WDT)

```
BCF     STATUS, RP0     ;Skip if already in
                        ;Bank 0
CLRWDT                  ;Clear WDT
CLRF    TMR0            ;Clear TMR0 and
                        ;Prescaler
BSF     STATUS, RP0     ;Bank 1
MOVLW   '00101111'b     ;These 3 lines
                        ;(5, 6, 7)
MOVWF   OPTION_REG      ;are required only
                        ;if desired PS<2:0>
                        ;are
CLRWDT                  ;000 or 001
MOVLW   '00101xxx'b     ;Set Postscaler to
MOVWF   OPTION_REG      ;desired WDT rate
BCF     STATUS, RP0     ;Return to Bank 0
```

To change prescaler from the WDT to the Timer0 module, use the sequence shown in Example 6-2. This precaution must be taken even if the WDT is disabled.

EXAMPLE 6-2: CHANGING PRESCALER (WDT → TIMER0)

```
CLRWDT                  ;Clear WDT and
                        ;prescaler
BSF     STATUS, RP0
MOVLW   b'xxxx0xxx'     ;Select TMR0, new
                        ;prescale value and
                        ;clock source
MOVWF   OPTION_REG
BCF     STATUS, RP0
```

TABLE 6-1: REGISTERS ASSOCIATED WITH TIMER0

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR	Value on All Other Resets
01h, 101h	TMR0	Timer0 Module Register								xxxx xxxx	uuuu uuuu
0Bh, 8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000u
81h, 181h	OPTION(2)	$\overline{\text{RBPU}}$	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
85h	TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111

Legend: - = Unimplemented locations, read as ‘0’, u = unchanged, x = unknown. Shaded cells are not used for Timer0.

Note 1: Option is referred by OPTION_REG in MPLAB® IDE Software.

8.0 TIMER2 MODULE

Timer2 is an 8-bit timer with a prescaler and a postscaler. It can be used as the PWM time base for PWM mode of the CCP module. The TMR2 register is readable and writable, and is cleared on any device Reset.

The input clock (FOSC/4) has a prescale option of 1:1, 1:4 or 1:16, selected by control bits T2CKPS<1:0> (T2CON<1:0>).

The Timer2 module has an 8-bit period register PR2. The TMR2 register value increments from 00h until it matches the PR2 register value and then resets to 00h on the next increment cycle. The PR2 register is a readable and writable register. The PR2 register is initialized to FFh upon Reset.

The match output of Timer2 goes through a 4-bit postscaler (which gives a 1:1 to 1:16 scaling inclusive) to generate a Timer2 interrupt (latched in flag bit TMR2IF, (PIR1<1>)).

Timer2 can be shut off by clearing control bit TMR2ON (T2CON<2>) to minimize power consumption.

Register 8-1 shows the Timer2 control register.

8.1 Timer2 Prescaler and Postscaler

The prescaler and postscaler counters are cleared when any of the following occurs:

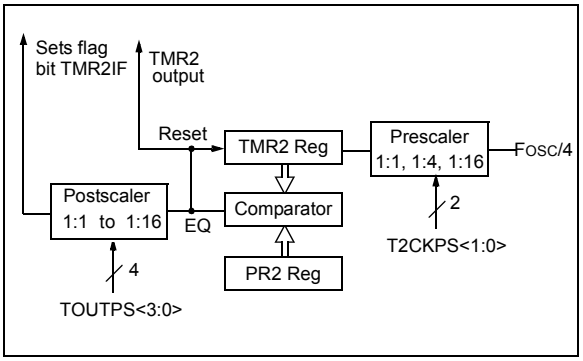
- a write to the TMR2 register
- a write to the T2CON register
- any device Reset (Power-on Reset, $\overline{MCLR}$ Reset, Watchdog Timer Reset or Brown-out Reset)

The TMR2 register is not cleared when T2CON is written.

8.2 TMR2 Output

The TMR2 output (before the postscaler) is fed to the Synchronous Serial Port module which optionally uses it to generate shift clock.

FIGURE 8-1: TIMER2 BLOCK DIAGRAM



The code example in Example 10-1 depicts the steps required to configure the Comparator module. RA3 and RA4 are configured as digital output. RA0 and RA1 are configured as the V- inputs and RA2 as the V+ input to both comparators.

EXAMPLE 10-1: INITIALIZING COMPARATOR MODULE

```
FLAG_REG     EQU              0X20
CLRF     FLAG_REG    ;Init flag register
CLRF     PORTA       ;Init PORTA
MOVF     CMCON, W    ;Load comparator bits
ANDLW    0xC0        ;Mask comparator bits
IORWF    FLAG_REG,F  ;Store bits in flag register
MOVLW    0x03        ;Init comparator mode
MOVWF    CMCON       ;CM<2:0> = 011
BSF      STATUS,RP0  ;Select Bank1
MOVLW    0x07        ;Initialize data direction
MOVWF    TRISA       ;Set RA<2:0> as inputs
                     ;RA<4:3> as outputs
                     ;TRISA<7:5> always read '0'
BCF      STATUS,RP0  ;Select Bank 0
CALL     DELAY10     ;10Ms delay
MOVF     CMCON,F     ;Read CMCONto end change
                     ;condition
BCF      PIR1,CMIF   ;Clear pending interrupts
BSF      STATUS,RP0  ;Select Bank 1
BSF      PIE1,CMIE   ;Enable comparator interrupts
BCF      STATUS,RP0  ;Select Bank 0
BSF      INTCON,PEIE ;Enable peripheral interrupts
BSF      INTCON,GIE  ;Global interrupt enable
```

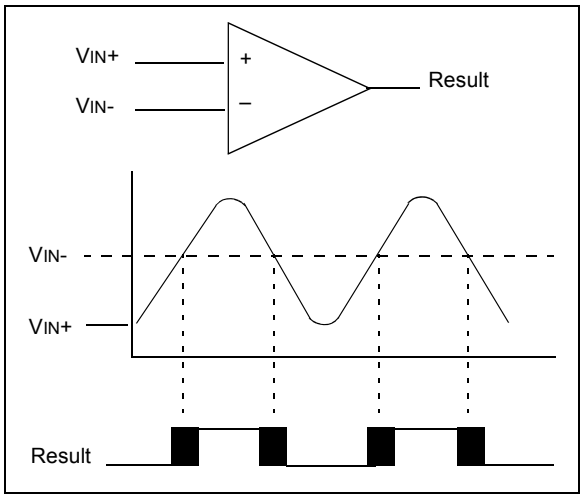
10.2 Comparator Operation

A single comparator is shown in Figure 10-2 along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input VIN-, the output of the comparator is a digital low level. When the analog input at VIN+ is greater than the analog input VIN-, the output of the comparator is a digital high level. The shaded areas of the output of the comparator in Figure 10-2 represent the uncertainty due to input offsets and response time. See Table 17-2 for Common Mode voltage.

10.3 Comparator Reference

An external or internal reference signal may be used depending on the comparator Operating mode. The analog signal that is present at VIN- is compared to the signal at VIN+, and the digital output of the comparator is adjusted accordingly (Figure 10-2).

FIGURE 10-2: SINGLE COMPARATOR



10.3.1 EXTERNAL REFERENCE SIGNAL

When external voltage references are used, the Comparator module can be configured to have the comparators operate from the same or different reference sources. However, threshold detector applications may require the same reference. The reference signal must be between VSS and VDD, and can be applied to either pin of the comparator(s).

10.3.2 INTERNAL REFERENCE SIGNAL

The Comparator module also allows the selection of an internally generated voltage reference for the comparators. **Section 11.0 "Voltage Reference Module"**, contains a detailed description of the Voltage Reference module that provides this signal. The internal reference signal is used when the comparators are in mode CM<2:0> = 010 (Figure 10-1). In this mode, the internal voltage reference is applied to the VIN+ pin of both comparators.

10.4 Comparator Response Time

Response time is the minimum time, after selecting a new reference voltage or input source, before the comparator output is to have a valid level. If the internal reference is changed, the maximum delay of the internal voltage reference must be considered when using the comparator outputs. Otherwise, the maximum delay of the comparators should be used (Table 17-2, page 142).

12.4 USART Synchronous Master Mode

In Synchronous Master mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit SYNC (TXSTA<4>). In addition enable bit SPEN (RCSTA<7>) is set in order to configure the RB2/TX/CK and RB1/RX/DT I/O pins to CK (clock) and DT (data) lines, respectively. The Master mode indicates that the processor transmits the master clock on the CK line. The Master mode is entered by setting bit CSRC (TXSTA<7>).

12.4.1 USART SYNCHRONOUS MASTER TRANSMISSION

The USART transmitter block diagram is shown in Figure 12-1. The heart of the transmitter is the Transmit (serial) Shift Register (TSR). The shift register obtains its data from the read/write transmit buffer register, TXREG. The TXREG register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREG (if available). Once the TXREG register transfers the data to the TSR register (occurs in one Tcycle), the TXREG is empty and interrupt bit, TXIF (PIR1<4>) is set. The interrupt can be enabled/disabled by setting/clearing enable bit TXIE (PIE1<4>). Flag bit TXIF will be set regardless of the state of enable bit TXIE and cannot be cleared in software. It will reset only when new data is loaded into the TXREG register. While flag bit TXIF indicates the status of the TXREG register, another bit TRMT (TXSTA<1>) shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR is empty. No interrupt logic is tied to this bit, so the user has to poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory so it is not available to the user.

Transmission is enabled by setting enable bit TXEN (TXSTA<5>). The actual transmission will not occur until the TXREG register has been loaded with data. The first data bit will be shifted out on the next available rising edge of the clock on the CK line. Data out is stable around the falling edge of the synchronous clock (Figure 12-8). The transmission can also be started by first loading the TXREG register and then setting bit TXEN (Figure 12-9). This is advantageous when slow baud rates are selected, since the BRG is kept in Reset when bits TXEN, CREN and SREN are clear. Setting enable bit TXEN will start the BRG, creating a shift clock immediately. Normally, when transmission is first started, the TSR register is empty, so a transfer to the TXREG register will result in an immediate transfer to TSR resulting in an empty TXREG. Back-to-back transfers are possible.

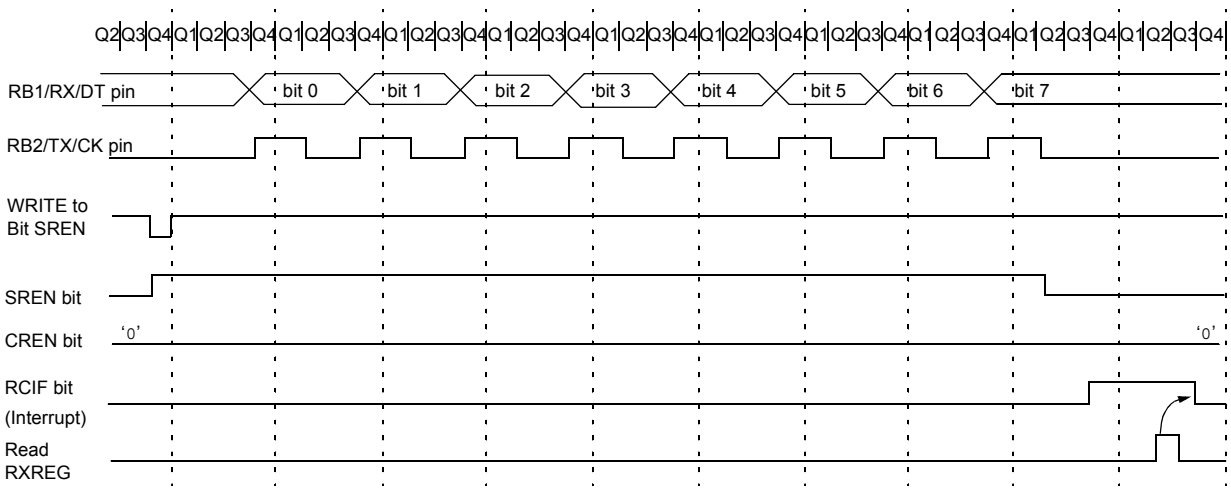
Clearing enable bit TXEN during a transmission will cause the transmission to be aborted and will reset the transmitter. The DT and CK pins will revert to high-impedance. If either bit CREN or bit SREN is set during a transmission, the transmission is aborted and the DT pin reverts to a high-impedance state (for a reception). The CK pin will remain an output if bit CSRC is set (internal clock). The transmitter logic however is not reset although it is disconnected from the pins. In order to reset the transmitter, the user has to clear bit TXEN. If bit SREN is set (to interrupt an on-going transmission and receive a single word), then after the single word is received, bit SREN will be cleared and the serial port will revert back to transmitting since bit TXEN is still set. The DT line will immediately switch from high-impedance Receive mode to transmit and start driving. To avoid this, bit TXEN should be cleared.

In order to select 9-bit transmission, the TX9 (TXSTA<6>) bit should be set and the ninth bit should be written to bit TX9D (TXSTA<0>). The ninth bit must be written before writing the 8-bit data to the TXREG register. This is because a data write to the TXREG can result in an immediate transfer of the data to the TSR register (if the TSR is empty). If the TSR was empty and the TXREG was written before writing the “new” TX9D, the “present” value of bit TX9D is loaded.

Follow these steps when setting up a Synchronous Master Transmission:

1. TRISB<1> and TRISB<2> should both be set to ‘1’ to configure the RB1/RX/DT and RB2/TX/CK pins as inputs. Output drive, when required, is controlled by the peripheral circuitry.
2. Initialize the SPBRG register for the appropriate baud rate (**Section 12.1 “USART Baud Rate Generator (BRG)”**).
3. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
4. If interrupts are desired, then set enable bit TXIE.
5. If 9-bit transmission is desired, then set bit TX9.
6. Enable the transmission by setting bit TXEN.
7. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
8. Start each transmission by loading data to the TXREG register.

FIGURE 12-10: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)



Note: Timing diagram demonstrates Sync Master mode with bit SREN = 1 and bit BRG = 0.

12.5 USART Synchronous Slave Mode

Synchronous Slave mode differs from the Master mode in the fact that the shift clock is supplied externally at the RB2/TX/CK pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in Sleep mode. Slave mode is entered by clearing bit CSRC (TXSTA<7>).

12.5.1 USART SYNCHRONOUS SLAVE TRANSMIT

The operation of the Synchronous Master and Slave modes are identical except in the case of the Sleep mode.

If two words are written to the TXREG and then the SLEEP instruction is executed, the following will occur:

a) The first word will immediately transfer to the TSR register and transmit.
b) The second word will remain in TXREG register.
c) Flag bit TXIF will not be set.
d) When the first word has been shifted out of TSR, the TXREG register will transfer the second word to the TSR and flag bit TXIF will now be set.
e) If enable bit TXIE is set, the interrupt will wake the chip from Sleep and if the global interrupt is enabled, the program will branch to the interrupt vector (0004h).

Follow these steps when setting up a Synchronous Slave Transmission:

1. TRISB<1> and TRISB<2> should both be set to '1' to configure the RB1/RX/DT and RB2/TX/CK pins as inputs. Output drive, when required, is controlled by the peripheral circuitry.
2. Enable the synchronous slave serial port by setting bits SYNC and SPEN and clearing bit CSRC.
3. Clear bits CREN and SREN.
4. If interrupts are desired, then set enable bit TXIE.
5. If 9-bit transmission is desired, then set bit TX9.
6. Enable the transmission by setting enable bit TXEN.
7. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
8. Start transmission by loading data to the TXREG register.

REGISTER 14-1: CONFIG – CONFIGURATION WORD REGISTER

$\overline{CP}$	—	—	—	—	$\overline{CPD}$	LVP	BOREN	MCLRE	FOSC2	$\overline{PWRTE}$	WDTE	F0SC1	F0SC0
bit 13													bit 0

bit 13: **$\overline{CP}$:** Flash Program Memory Code Protection bit(2)
(PIC16F648A)
1 = Code protection off
0 = 0000h to 0FFFh code-protected
(PIC16F628A)
1 = Code protection off
0 = 0000h to 07FFh code-protected
(PIC16F627A)
1 = Code protection off
0 = 0000h to 03FFh code-protected

bit 12-9: **Unimplemented**: Read as '0'

bit 8: **$\overline{CPD}$:** Data Code Protection bit(3)
1 = Data memory code protection off
0 = Data memory code-protected

bit 7: **LVP**: Low-Voltage Programming Enable bit
1 = RB4/PGM pin has PGM function, low-voltage programming enabled
0 = RB4/PGM is digital I/O, HV on $\overline{MCLR}$ must be used for programming

bit 6: **BOREN**: Brown-out Reset Enable bit (1)
1 = BOR Reset enabled
0 = BOR Reset disabled

bit 5: **MCLRE**: RA5/$\overline{MCLR}$/V_{PP} Pin Function Select bit
1 = RA5/$\overline{MCLR}$/V_{PP} pin function is $\overline{MCLR}$
0 = RA5/$\overline{MCLR}$/V_{PP} pin function is digital Input, $\overline{MCLR}$ internally tied to V_{DD}

bit 3: **$\overline{PWRTE}$**: Power-up Timer Enable bit (1)
1 = PWRT disabled
0 = PWRT enabled

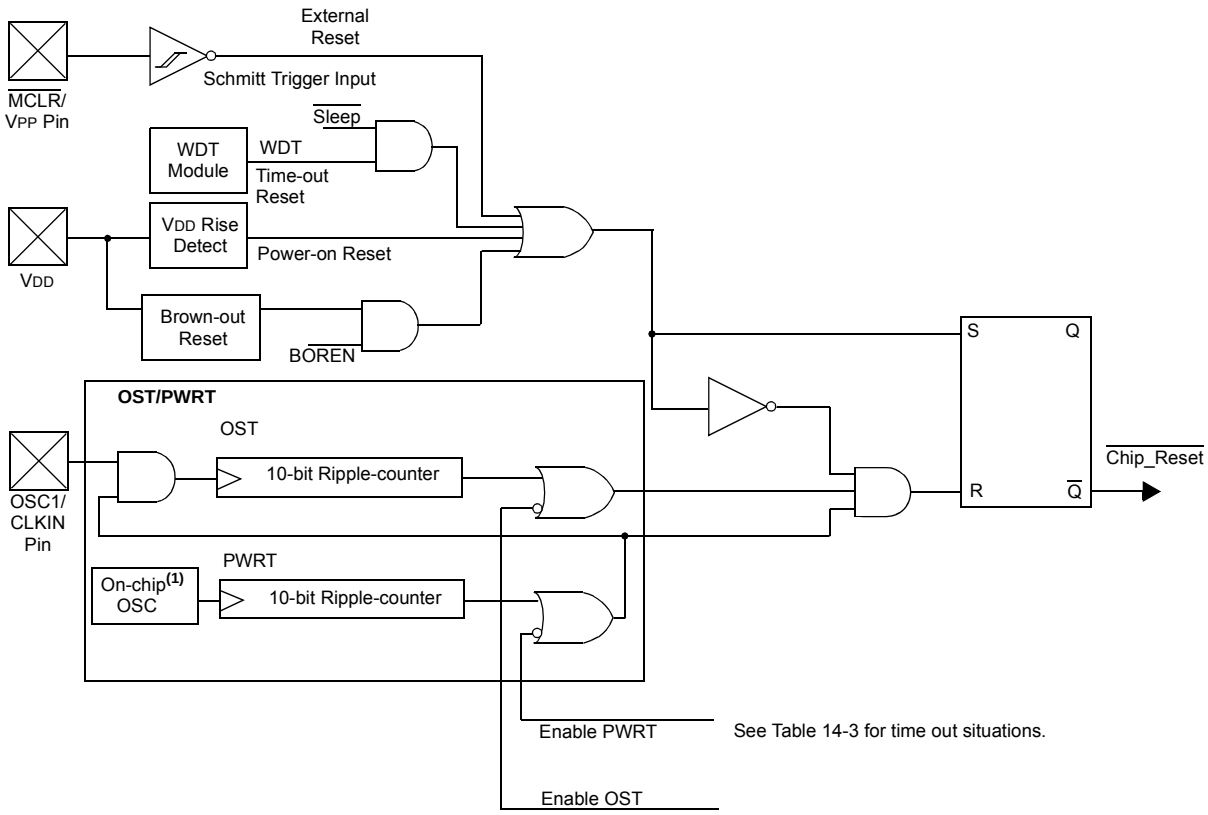
bit 2: **WDTE**: Watchdog Timer Enable bit
1 = WDT enabled
0 = WDT disabled

bit 4, 1-0: **FOSC<2:0>**: Oscillator Selection bits(4)
111 = RC oscillator: CLKOUT function on RA6/OSC2/CLKOUT pin, Resistor and Capacitor on RA7/OSC1/CLKIN
110 = RC oscillator: I/O function on RA6/OSC2/CLKOUT pin, Resistor and Capacitor on RA7/OSC1/CLKIN
101 = INTOSC oscillator: CLKOUT function on RA6/OSC2/CLKOUT pin, I/O function on RA7/OSC1/CLKIN
100 = INTOSC oscillator: I/O function on RA6/OSC2/CLKOUT pin, I/O function on RA7/OSC1/CLKIN
011 = EC: I/O function on RA6/OSC2/CLKOUT pin, CLKIN on RA7/OSC1/CLKIN
010 = HS oscillator: High-speed crystal/resonator on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN
001 = XT oscillator: Crystal/resonator on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN
000 = LP oscillator: Low-power crystal on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN

Note 1: Enabling Brown-out Reset does not automatically enable the Power-up Timer (PWRT) the way it does on the PIC16F627/628 devices.
2: The code protection scheme has changed from the code protection scheme used on the PIC16F627/628 devices. The entire Flash program memory needs to be bulk erased to set the $\overline{CP}$ bit, turning the code protection off. See "*PIC16F627A/628A/648A EEPROM Memory Programming Specification*" (DS41196) for details.
3: The entire data EEPROM needs to be bulk erased to set the $\overline{CPD}$ bit, turning the code protection off. See "*PIC16F627A/628A/648A EEPROM Memory Programming Specification*" (DS41196) for details.
4: When $\overline{MCLR}$ is asserted in INTOSC mode, the internal clock oscillator is disabled.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = bit is set	'0' = bit is cleared	x = bit is unknown

FIGURE 14-6: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



Note 1: This is a separate oscillator from the INTOSC/RC oscillator.

XORLW **Exclusive OR Literal with W**

Syntax: [*label*] XORLW k

Operands: $0 \le k \le 255$

Operation: (W) .XOR. k → (W)

Status Affected: Z

Encoding:

11	1010	kkkk	kkkk

Description: The contents of the W register are XOR'ed with the eight-bit literal 'k'. The result is placed in the W register.

Words: 1

Cycles: 1

Example:

```
XORLW   0xAF
```

Before Instruction

W = 0xB5

After Instruction

W = 0x1A

XORWF **Exclusive OR W with f**

Syntax: [*label*] XORWF f,d

Operands: $0 \le f \le 127$
$d \in [0,1]$

Operation: (W) .XOR. (f) → (dest)

Status Affected: Z

Encoding:

00	0110	dfff	ffff

Description: Exclusive OR the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

Words: 1

Cycles: 1

Example

```
XORWF    REG1, 1
```

Before Instruction

REG1 = 0xAF
W = 0xB5

After Instruction

REG1 = 0x1A
W = 0xB5

16.11 PICkit 2 Development Programmer/Debugger and PICkit 2 Debug Express

The PICkit™ 2 Development Programmer/Debugger is a low-cost development tool with an easy to use interface for programming and debugging Microchip's Flash families of microcontrollers. The full featured Windows® programming interface supports baseline (PIC10F, PIC12F5xx, PIC16F5xx), midrange (PIC12F6xx, PIC16F), PIC18F, PIC24, dsPIC30, dsPIC33, and PIC32 families of 8-bit, 16-bit, and 32-bit microcontrollers, and many Microchip Serial EEPROM products. With Microchip's powerful MPLAB Integrated Development Environment (IDE) the PICkit™ 2 enables in-circuit debugging on most PIC® microcontrollers. In-Circuit-Debugging runs, halts and single steps the program while the PIC microcontroller is embedded in the application. When halted at a breakpoint, the file registers can be examined and modified.

The PICkit 2 Debug Express include the PICkit 2, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

16.12 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at V_{DDMIN} and V_{DDMAX} for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages and a modular, detachable socket assembly to support various package types. The ICSP™ cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices and incorporates an MMC card for file storage and data applications.

16.13 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM™ and dsPICDEM™ demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ® security ICs, CAN, IrDA®, PowerSmart battery management, SEEVAL® evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

17.3 DC Characteristics: PIC16F627A/628A/648A (Extended)

DC CHARACTERISTICS		Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ Ta ≤ +125°C for extended					
Param No.	Device Characteristics	Min†	Typ	Max	Units	Conditions	
						VDD	Note
Supply Voltage (VDD)							
D001	—	3.0	—	5.5	V	—	
Power-down Base Current (IPD)							
D020E	—	—	0.01	4	µA	3.0	WDT, BOR, Comparators, VREF and T1OSC: disabled
		—	0.02	8	µA	5.0	
Peripheral Module Current (ΔIMOD)(1)							
D021E	—	—	2	9	µA	3.0	WDT Current
		—	9	20	µA	5.0	
D022E	—	—	29	52	µA	4.5	BOR Current
		—	30	55	µA	5.0	
D023E	—	—	22	37	µA	3.0	Comparator Current (Both comparators enabled)
		—	44	68	µA	5.0	
D024E	—	—	50	75	µA	3.0	VREF Current
		—	83	110	µA	5.0	
D025E	—	—	1.3	4	µA	3.0	T1OSC Current
		—	1.8	6	µA	5.0	
Supply Current (IDD)							
D010E	—	—	15	28	µA	3.0	FOSC = 32 kHz LP Oscillator Mode
		—	28	54	µA	5.0	
D011E	—	—	175	340	µA	3.0	FOSC = 1 MHz XT Oscillator Mode
		—	320	520	µA	5.0	
D012E	—	—	450	650	µA	3.0	FOSC = 4 MHz XT Oscillator Mode
		—	0.710	1.1	mA	5.0	
D012AE	—	—	565	785	µA	3.0	FOSC = 4 MHz INTOSC
		—	0.895	1.3	mA	5.0	
D013E	—	—	2.5	2.9	mA	4.5	FOSC = 20 MHz HS Oscillator Mode
		—	2.75	3.5	mA	5.0	

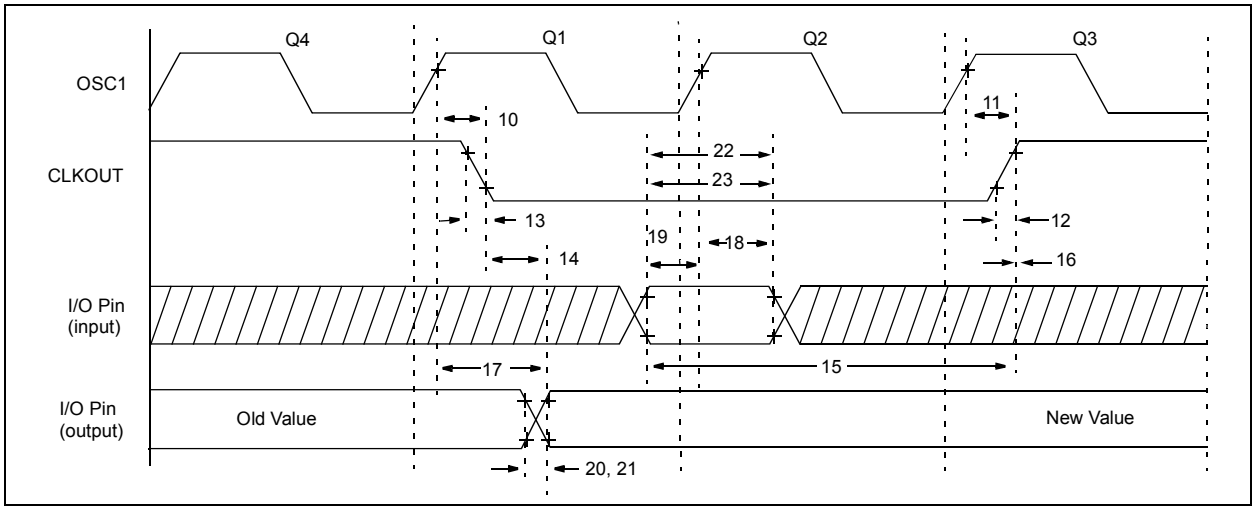
Note 1: The "Δ" current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement. Max values should be used when calculating total current consumption.

TABLE 17-5: PRECISION INTERNAL OSCILLATOR PARAMETERS

Parameter No.	Sym	Characteristic	Min	Typ	Max	Units	Conditions
F10	FIOSC	Oscillator Center frequency	—	4	—	MHz	
F13	ΔIOSC	Oscillator Accuracy	3.96	4	4.04	MHz	VDD = 3.5 V, 25°C
			3.92	4	4.08	MHz	2.0V ≤ VDD ≤ 5.5V 0°C ≤ TA ≤ +85°C
			3.80	4	4.20	MHz	2.0V ≤ VDD ≤ 5.5V -40°C ≤ TA ≤ +85°C (IND) -40°C ≤ TA ≤ +125°C (EXT)
F14*	TIOSCST	Oscillator Wake-up from Sleep start-up time	—	6	8	μs	VDD = 2.0V, -40°C to +85°C
			—	4	6	μs	VDD = 3.0V, -40°C to +85°C
			—	3	5	μs	VDD = 5.0V, -40°C to +85°C

Legend: TBD = To Be Determined.
* Characterized but not tested.

FIGURE 17-5: CLKOUT AND I/O TIMING



18.0 DC AND AC CHARACTERISTICS GRAPHS AND TABLES

The graphs and tables provided in this section are for **design guidance** and are **not tested**.

In some graphs or tables, the data presented are **outside specified operating range** (i.e., outside specified VDD range). This is for **information only** and devices are ensured to operate properly only within the specified range.

The data presented in this section is a **statistical summary** of data collected on units from different lots over a period of time and matrix samples. 'Typical' represents the mean of the distribution at 25°C. 'Max' or 'Min' represents (mean + 3σ) or (mean - 3σ) respectively, where σ is standard deviation, over the whole temperature range.

FIGURE 18-1: TYPICAL BASELINE IPD VS. VDD (-40°C TO 25°C)

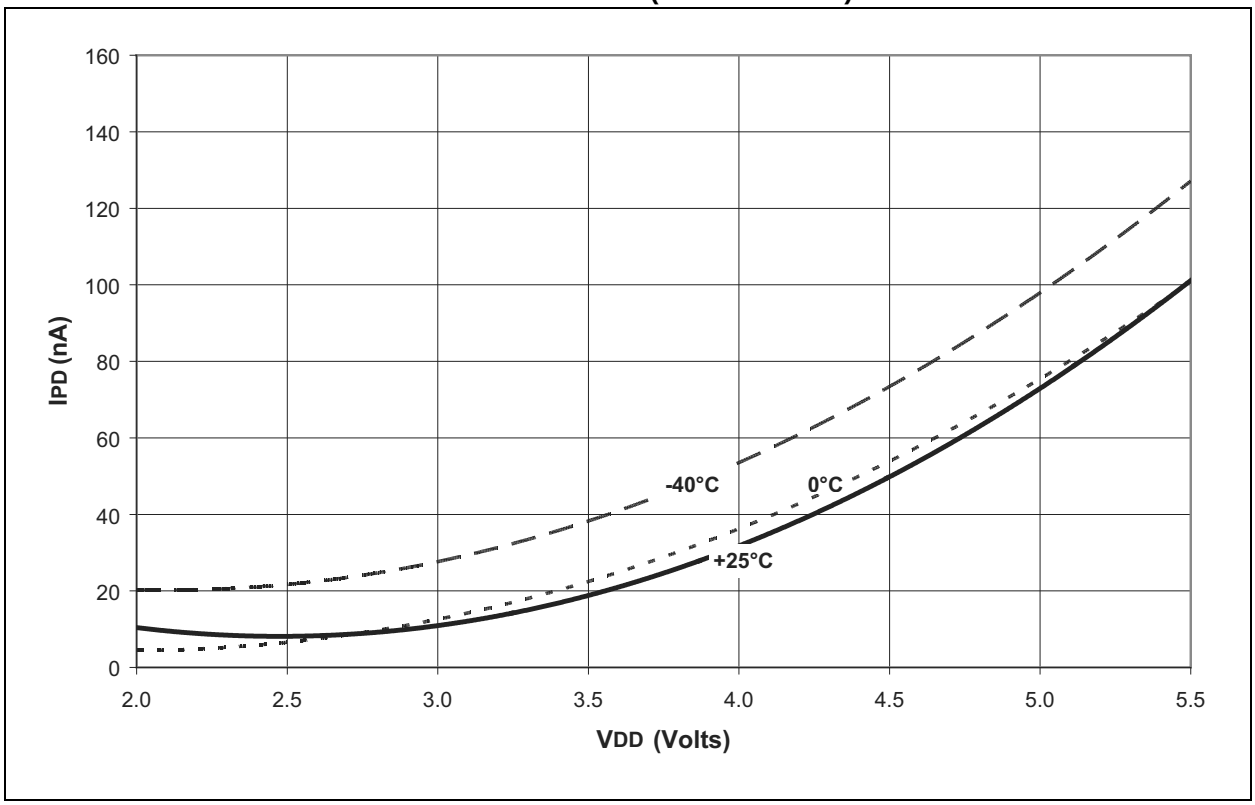


FIGURE 18-16: INTERNAL OSCILLATOR IDD vs. VDD – SLOW MODE

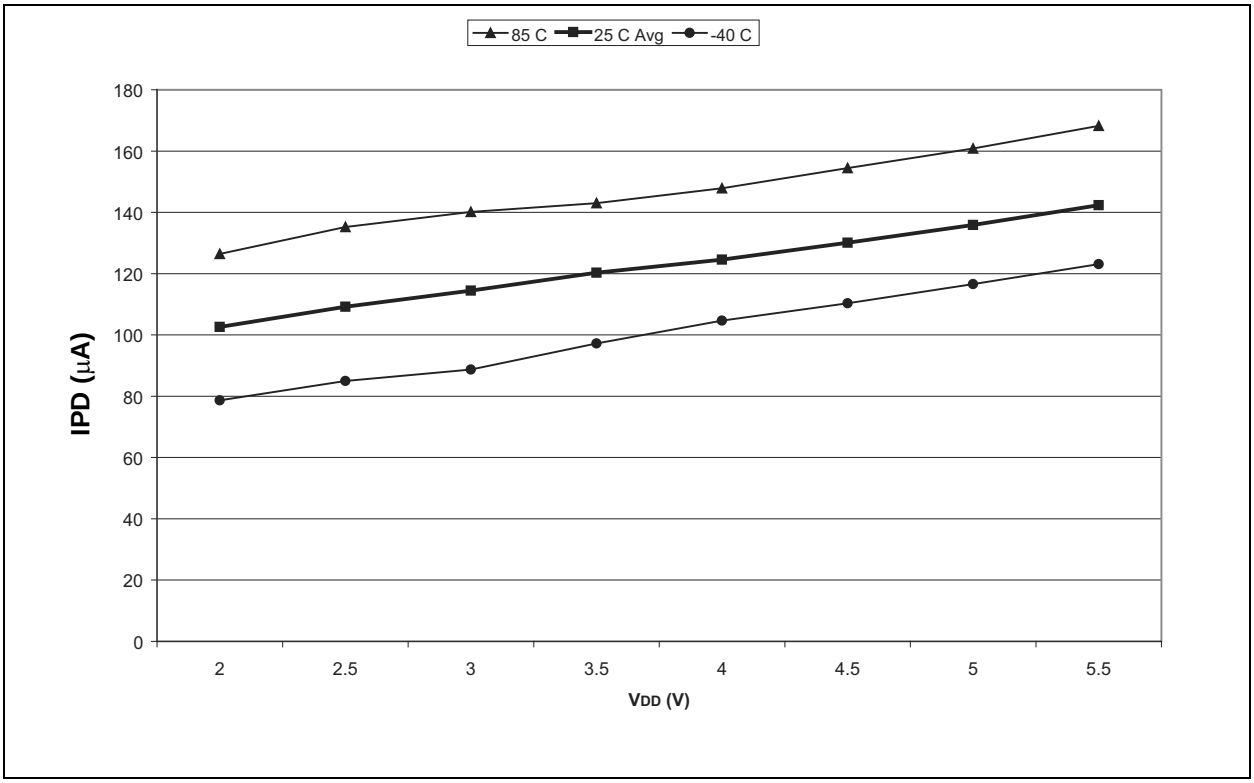
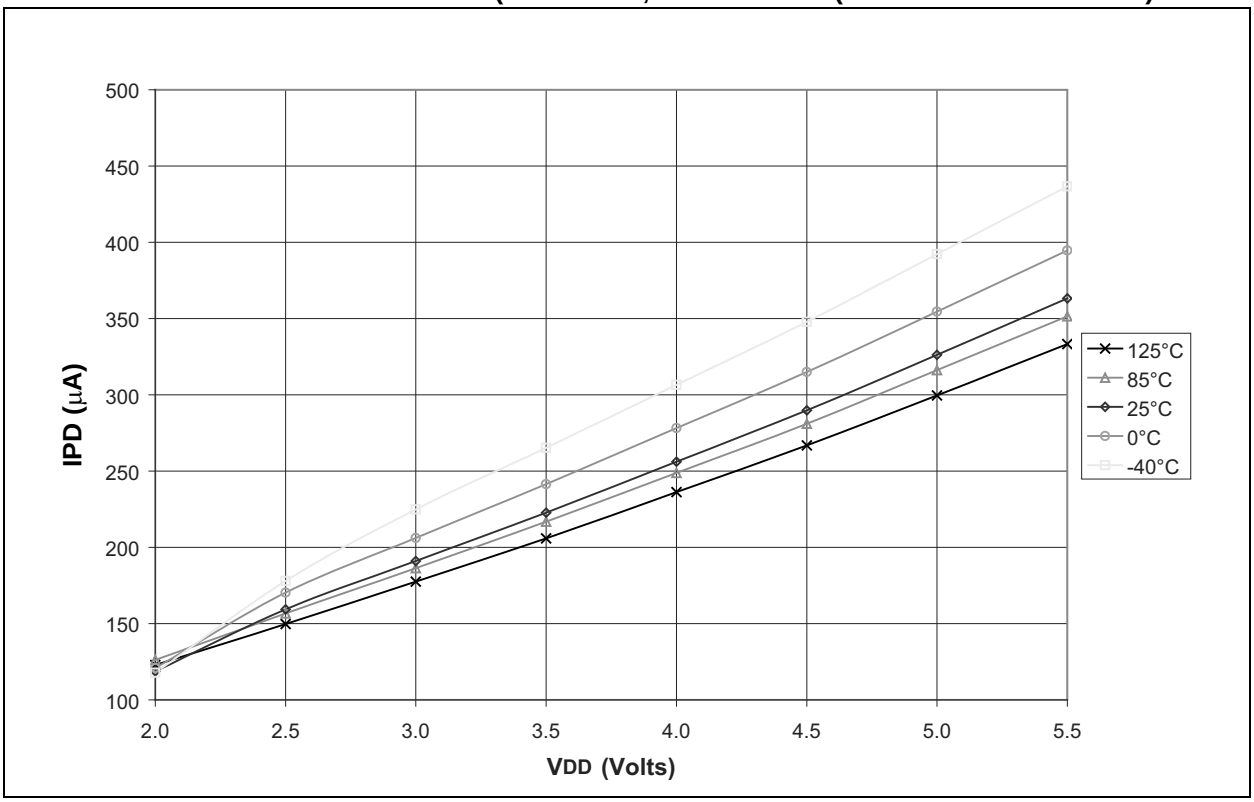
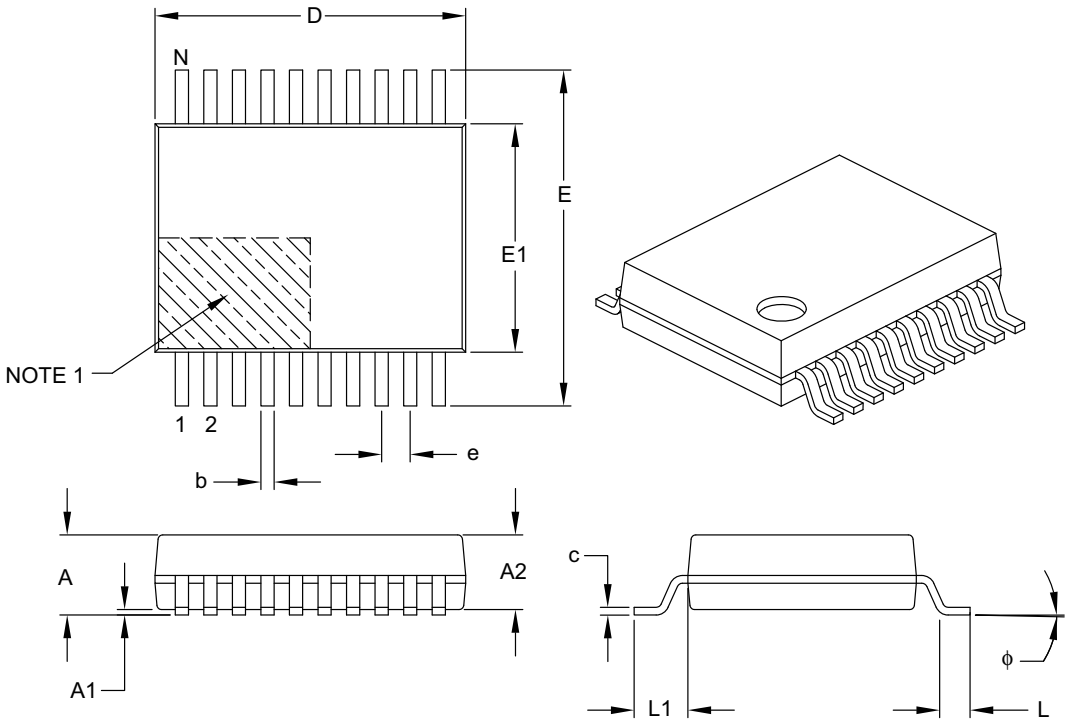


FIGURE 18-17: SUPPLY CURRENT (IDD vs. VDD, FOSC = 1 MHz (XT OSCILLATOR MODE)



20-Lead Plastic Shrink Small Outline (SS) – 5.30 mm Body [SSOP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		20	
Pitch	e		0.65 BSC	
Overall Height	A	–	–	2.00
Molded Package Thickness	A2	1.65	1.75	1.85
Standoff	A1	0.05	–	–
Overall Width	E	7.40	7.80	8.20
Molded Package Width	E1	5.00	5.30	5.60
Overall Length	D	6.90	7.20	7.50
Foot Length	L	0.55	0.75	0.95
Footprint	L1		1.25 REF	
Lead Thickness	c	0.09	–	0.25
Foot Angle	ϕ	0°	4°	8°
Lead Width	b	0.22	–	0.38

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.20 mm per side.
3. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.
REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-072B