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Details

Product Status	Obsolete
Core Processor	ARM926EJ-S
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	454MHz
Co-Processors/DSP	Data; DCP
RAM Controllers	DRAM
Graphics Acceleration	No
Display & Interface Controllers	LCD, Touchscreen
Ethernet	-
SATA	-
USB	USB 2.0 + PHY (1)
Voltage - I/O	2.0V, 2.5V, 2.7V, 3.0V, 3.3V
Operating Temperature	-40°C ~ 85°C (TA)
Security Features	Cryptography, Hardware ID
Package / Case	128-LQFP
Supplier Device Package	128-LQFP (14x14)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mcimx233cag4b

Table 5-189. HW ICOLL INTERRUPT84 Bit Field Descriptions

Freescale Semiconductor

```
pCurCmd = (hw_apbh_chn_cmd_t *) HW_APBH_CHN_CURCMDAR_RD(1); // read the whole register, since there
is only one field
pCurCmd = (hw_apbh_chn_cmd_t *) BF_RDn(APBH_CHN_CURCMDAR, 1, CMD_ADDR); // or, use multi-register
bitfield read macro
pCurCmd = (hw_apbh_chn_cmd_t *) HW_APBH_CHN_CURCMDAR(1).CMD_ADDR; // or, assign from bitfield of
indexed register's struct
```

The APBH DMA Channel 1 Next Command Address register contains the address of the next multiword command to be executed. Commands are threaded on the command address. Set CHAIN to 1 in the DMA command word to process command lists.

Table 10-28. HW APBH CH1 NXTCMPDAR

Table 10-29. HW APBH CH1 NXCMDAR Bit Field Descriptions

APBH DMA Channel 1 is controlled by a variable sized command structure. Software loads this register with the address of the first command structure to process and increments the Channel 1 semaphore to start processing. This register points to the next command structure to be executed when the current command is completed.

```
HW_APBH_CHn_NXTCMDAR_WR(1, (reg32_t) pCommandTwoStructure); // write the entire register, since
there is only one field
BF_WrN(APBH_CHn_NXTCMDAR, 1, (reg32_t) pCommandTwoStructure); // or, use multi-register bitfield
write macro
HW_APBH_CHn_NXTCMDAR(1).CMD_ADDR = (reg32_t) pCommandTwoStructure; // or, assign to bitfield of
indexed register's struct
```

The APBH DMA Channel 1 command register specifies the cycle to perform for the current command chain item.

```
HW APBH CH1 CMD                                0x0D0
```

APBX DMA Channel 12 is controlled by a variable sized command structure. This register points to the command structure currently being executed.

Empty example.

11.5.91 APBX DMA Channel 12 Next Command Address Register Description

The APBX DMA Channel 12 next command address register points to the next multiword command to be executed. Commands are threaded on the command address. Set CHAIN to one to process command lists.

HW APBX CH12 NXTCMDAR 0x650

Table 11-184. HW APBX CH12 NXTCMDAR

3	3	2	2	2	2	2	2	2	2	2	1	1	1	1	1	1	1	1	1	0	0	0	0	0	0	0	0	0	0		
1	0	9	8	7	6	5	4	3	2	1	0	9	8	7	6	5	4	3	2	1	0	9	8	7	6	5	4	3	2	1	0
CMD_ADDR																															

Table 11-185. HW_APBX_CH12_NXTCMDAR Bit Field Descriptions

BITS	LABEL	RW	RESET	DEFINITION
31:0	CMD_ADDR	RW	0x00000000	Pointer to next command structure for Channel 12.

APBX DMA Channel 12 is controlled by a variable sized command structure. Software loads this register with the address of the first command structure to process and increments the Channel 12 semaphore to start processing. This register points to the next command structure to be executed when the current command is completed.

Empty example.

11.5.92 APBX DMA Channel 12 Command Register Description

The APBX DMA Channel 12 command register specifies the cycle to perform for the current command chain item.

```
HW APBX CH12 CMD                                0x660
```

If a different LOWPOWER_CONTROL bit is set to 1 while in one of the low-power modes, or on clearing of the original bit to 0, the memory controller exits the current low-power mode. There will be at least a 15 cycle delay before the memory controller is fully operational or enters the new low-power mode.

NOTE: There is a deadlock possibility that exists when using the manual low-power mode entry. If a read cycle from the ARM core occurs to the DRAM when a manual low-power mode is active, the ARM cycle does not complete. There is no other device within the SOC that can deactivate the low-power mode. Thus, the system will be deadlocked. The same can occur with multiple write cycles that will fill the two-command deep write buffer of the memory controller.

12.2.6.5 Register Programming

The low-power modes of the memory controller are controlled through the LOWPOWER_CONTROL and LOWPOWER_AUTO_ENABLE bit fields in HW_DRAM_CTL16. These five-bit bit fields each contain one bit for controlling each low-power mode. The LOWPOWER_CONTROL bit field enables the associated low-power mode, and the LOWPOWER_AUTO_ENABLE bit field sets the entry method into that mode as manual or automatic. Table 12-1 shows the relationship between the five bits of the lowpower_control and lowpower_auto_enable bit fields and the various low-power modes.

Table 12-1. Low-Power Mode Bit Fields

Low-Power Mode	Enable	Entry
Memory Power-Down (Mode 1)	LOWPOWER_CONTROL [4] =1	LOWPOWER_AUTO_ENABLE [4] • 0 = Manual • 1 = Automatic
Memory Power-Down with Memory Clock Gating (Mode 2)	LOWPOWER_CONTROL [3] =1	LOWPOWER_AUTO_ENABLE [3] • 0 = Manual • 1 = Automatic
Memory Self-Refresh (Mode 3)	LOWPOWER_CONTROL [2] =1	LOWPOWER_AUTO_ENABLE [2] • 0 = Manual • 1 = Automatic
Memory Self-Refresh with Memory Clock Gating (Mode 4)	LOWPOWER_CONTROL [1] =1	LOWPOWER_AUTO_ENABLE [1] • 0 = Manual • 1 = Automatic
Memory Self-Refresh with Memory and Controller Clock Gating (Mode 5)	LOWPOWER_CONTROL [0] =1	LOWPOWER_AUTO_ENABLE [0] • 0 = Manual • 1 = Automatic

When a LOWPOWER_CONTROL bit field bit is set to 1 by the user, the memory controller checks the LOWPOWER_AUTO_ENABLE bit field.

- If the associated bit in the LOWPOWER_AUTO_ENABLE bit field is set to 1, then the memory controller watches the associated counter for expiration, and then enters that low-power mode. Table 12-2 shows the correlation between the low-power modes and the counters that control each mode's automatic entry.

Table 14-6. HW_ECC8_STATUS1 Bit Field Descriptions

BITS	LABEL	RW	RESET	DEFINITION
31:28	STATUS_PAYLOAD7	RO	0xc	Count of symbols in error during processing of payload area 7. 0xF indicates uncorrectable. NO_ERRORS = 0x0 No errors occurred. ONE_CORRECTABLE = 0x1 One correctable error occurred. TWO_CORRECTABLE = 0x2 Two correctable errors occurred. THREE_CORRECTABLE = 0x3 Three correctable errors occurred. FOUR_CORRECTABLE = 0x4 Four correctable errors occurred. FIVE_CORRECTABLE = 0x5 Five correctable errors occurred. SIX_CORRECTABLE = 0x6 Six correctable errors occurred. SEVEN_CORRECTABLE = 0x7 Seven correctable errors occurred. EIGHT_CORRECTABLE = 0x8 Eight correctable errors occurred. NOT_CHECKED = 0xC This block was not examined by the ECC8. UNCORRECTABLE = 0xE Errors occurred that were uncorrectable. ALL_ONES = 0xF All bits are 1.
27:24	STATUS_PAYLOAD6	RO	0xc	Count of symbols in error during processing of payload area 6. 0xF indicates uncorrectable. NO_ERRORS = 0x0 No errors occurred. ONE_CORRECTABLE = 0x1 One correctable error occurred. TWO_CORRECTABLE = 0x2 Two correctable errors occurred. THREE_CORRECTABLE = 0x3 Three correctable errors occurred. FOUR_CORRECTABLE = 0x4 Four correctable errors occurred. FIVE_CORRECTABLE = 0x5 Five correctable errors occurred. SIX_CORRECTABLE = 0x6 Six correctable errors occurred. SEVEN_CORRECTABLE = 0x7 Seven correctable errors occurred. EIGHT_CORRECTABLE = 0x8 Eight correctable errors occurred. NOT_CHECKED = 0xC This block was not examined by the ECC8. UNCORRECTABLE = 0xE Errors occurred that were uncorrectable. ALL_ONES = 0xF All bits are 1.
23:20	STATUS_PAYLOAD5	RO	0xc	Count of symbols in error during processing of payload area 5. 0xF indicates uncorrectable. NO_ERRORS = 0x0 No errors occurred. ONE_CORRECTABLE = 0x1 One correctable error occurred. TWO_CORRECTABLE = 0x2 Two correctable errors occurred. THREE_CORRECTABLE = 0x3 Three correctable errors occurred. FOUR_CORRECTABLE = 0x4 Four correctable errors occurred. FIVE_CORRECTABLE = 0x5 Five correctable errors occurred. SIX_CORRECTABLE = 0x6 Six correctable errors occurred. SEVEN_CORRECTABLE = 0x7 Seven correctable errors occurred. EIGHT_CORRECTABLE = 0x8 Eight correctable errors occurred. NOT_CHECKED = 0xC This block was not examined by the ECC8. UNCORRECTABLE = 0xE Errors occurred that were uncorrectable. ALL_ONES = 0xF All bits are 1.
19:16	STATUS_PAYLOAD4	RO	0xc	Count of symbols in error during processing of payload area 4. 0xF indicates uncorrectable. NO_ERRORS = 0x0 No errors occurred. ONE_CORRECTABLE = 0x1 One correctable error occurred. TWO_CORRECTABLE = 0x2 Two correctable errors occurred. THREE_CORRECTABLE = 0x3 Three correctable errors occurred. FOUR_CORRECTABLE = 0x4 Four correctable errors occurred. FIVE_CORRECTABLE = 0x5 Five correctable errors occurred. SIX_CORRECTABLE = 0x6 Six correctable errors occurred. SEVEN_CORRECTABLE = 0x7 Seven correctable errors occurred. EIGHT_CORRECTABLE = 0x8 Eight correctable errors occurred. NOT_CHECKED = 0xC This block was not examined by the ECC8. UNCORRECTABLE = 0xE Errors occurred that were uncorrectable. ALL_ONES = 0xF All bits are 1.
15:12	STATUS_PAYLOAD3	RO	0xc	Count of symbols in error during processing of payload area 3. 0xF indicates uncorrectable. NO_ERRORS = 0x0 No errors occurred. ONE_CORRECTABLE = 0x1 One correctable error occurred. TWO_CORRECTABLE = 0x2 Two correctable errors occurred. THREE_CORRECTABLE = 0x3 Three correctable errors occurred. FOUR_CORRECTABLE = 0x4 Four correctable errors occurred. FIVE_CORRECTABLE = 0x5 Five correctable errors occurred. SIX_CORRECTABLE = 0x6 Six correctable errors occurred. SEVEN_CORRECTABLE = 0x7 Seven correctable errors occurred. EIGHT_CORRECTABLE = 0x8 Eight correctable errors occurred. NOT_CHECKED = 0xC This block was not examined by the ECC8. UNCORRECTABLE = 0xE Errors occurred that were uncorrectable. ALL_ONES = 0xF All bits are 1.

detected within a block, the BCH hardware activates the error correction logic to determine where bit errors have occurred and ultimately correct them in the data buffer in system memory. After an entire flash page has been read and corrected, the BCH will signal an interrupt to the CPU.

Figure 15-2 indicates how data read from the GPMI is operated on within the BCH hardware. As the BCH receives data from the GPMI (top row) it is written to memory by the BCH's Bus Interface Unit (BIU) (second row). For blocks requiring correction, the KES logic will be activated after the entire block has been received. Once the error locator polynomial has been computed, the corrections are determined by the Chien Search and fed back to the BIU, which performs a read/modify/write operation on the buffer in memory to correct the data.

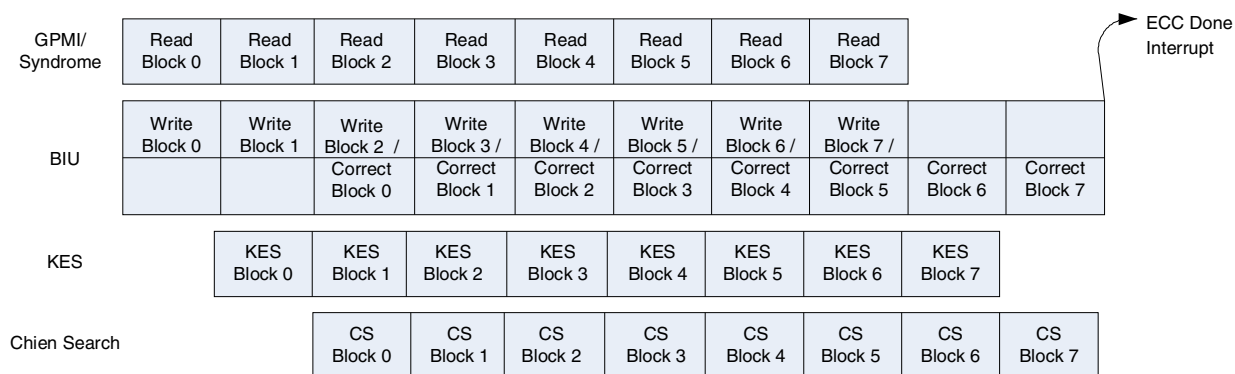


Figure 15-2. Block Pipeline while Reading Flash

15.2.1 BCH Limitations and Assumptions

- The BCH is programmable to support 2, 4, 6, 8, 10, 12, 14, 16, 18, and 20 bit error correction. ECC0 is supported as a passthrough, non-correcting mode.
- Data block sizes must be a multiple of 4 bytes and must be 4-byte aligned in system memory.
- The BCH supports a programmable number of metadata/auxiliary data bytes, from 0 to 255.
- Metadata will be written at the beginning of the flash page to facilitate fast access for filesystem operations.
- Metadata may be treated as an independent block for ECC purposes or combined with the first data block to conserve bits in the flash.
- The BCH does not support a partial page write.
- Flash read operations can read the entire page or the first block on the page.
- The BCH also supports a memory-to-memory mode of operation that does not require the use of DMA or the GPMI.

15.2.2 Flash Page Layout

The BCH supports a fully programmable flash page layout, versus the hardwired modes supported in the former ECC8 engine. The BCH maintains 4 independent layout registers that can describe four completely different NAND devices or layouts. When the BCH initiates an operation, it selects one of the lay-

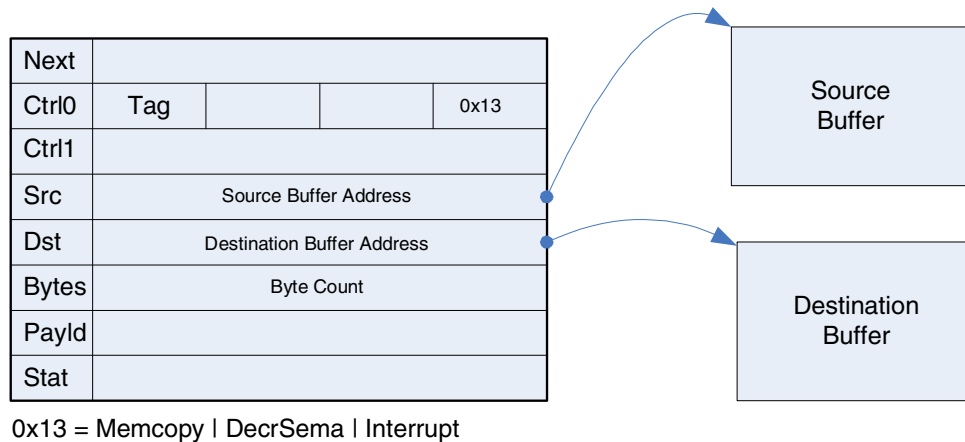


Figure 16-6. Basic Memory Copy Operation

<code>

```
typedef struct _dcp_descriptor
{
    u32          *next;
    hw_dcp_packet1_t  ctrl0;
    hw_dcp_packet2_t  ctrl1;
    u32          *src,
                *dst,
                buf_size,
                *payload,
                stat;
} DCP_DESCRIPTOR;

DCP_DESCRIPTOR dcp1;
u32 *srcbuffer, *dstbuffer;

// set up control packet
dcp1.next = 0; // single packet in chain
dcp1.ctrl0.U = 0; // clear ctrl0 field
dcp1.ctrl0.B.ENABLE_MEMCOPY = 1; // enable memcopy
dcp1.ctrl0.B.DECR_SEMAPHORE = 1; // decrement semaphore
dcp1.ctrl0.B.INTERRUPT = 1; // interrupt
dcp1.ctrl1.U = 0; // clear ctrl1
dcp1.src = srcbuffer; // source buffer
dcp1.dst = dstbuffer; // destination buffer
dcp1.buf_size = 512; // 512 bytes
dcp1.payload = NULL; // not required
dcp1.status = 0; // clear status

// Enable channel 0
HW_DCP_CHnCMDPTR_WR(0, dcp1); // write packet address to pointer register
HW_DCP_CHnSEMA_WR(0, 1); // increment semaphore by 1

// now wait for interrupt or poll
// polling code
while ( (HW_DCP_STAT_RD() & 0x01) == 0x00 );

// now check/clear channel status
if ( (HW_DCP_CHnSTAT_RD(0) & 0xFF) != 0 ) {
    // an error occurred
    HW_DCP_CHnSTAT_CLR(0, 0xff);
}
// clear interrupt register
HW_DCP_STAT_CLR(1);
```

</code>

16.2.6.2 Basic Hash Operation Programming Example

To perform a basic hash operation, only a single descriptor is required. The DCP simply reads data from the source buffer and computes the hash value on the contents. This process is illustrated in [Figure 16-7](#).

Table 16-36. HW DCP PACKET3 Bit Field Descriptions

DESCRIPTION:

EXAMPLE:

Empty Example.

16.3.13 DCP Work Packet 4 Status Register Description

This register displays the values for the current work packet offset 0x10 (Destination Address) field.

HW DCP PACKET4

0x0C0

Table 16-37. HW DCP PACKET4

Table 16-38. HW DCP PACKET4 Bit Field Descriptions

DESCRIPTION:

This register shows the contents of the Destination Address register from the packet being processed.

EXAMPLE:

Empty Example.

16.3.14 DCP Work Packet 5 Status Register Description

This register displays the values for the current work packet offset 0x14 (Buffer Size) field.

HW DCP PACKET5

0x0D0

Table 16-39. HW DCP PACKET5

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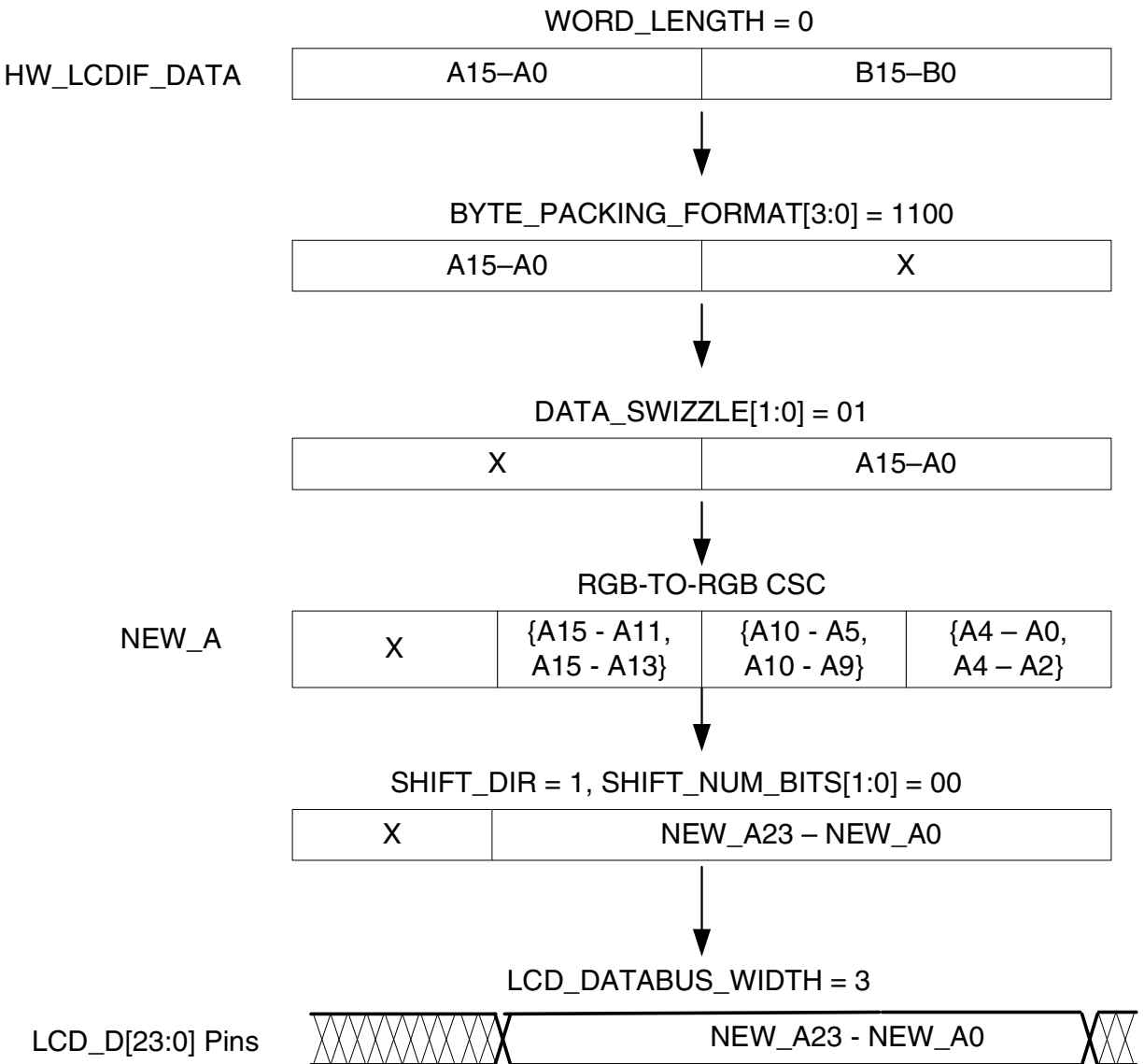


Figure 18-6. 16-Bit LCDIF Register Programming—Example B

18.2.3 LCDIF Interrupts

The LCDIF supports a number of interrupts to aid controlling and status reporting of the block. All the interrupts have individual mask bits for enabling or disabling each of them. They all get funneled through a single interrupt line connected to the interrupt collector (ICOLL). The following list describes the different interrupts supported by the LCDIF:

- Underflow interrupt is asserted when the clock domain crossing FIFO (TXFIFO) becomes empty but the block is in active display portion at that time. Software should take corrective action to make sure that this does not happen. This interrupt is of value only in DOTCLK and DVI modes.
- Overflow interrupt will be asserted in PIO mode if software writes to LFIFO while it is full.

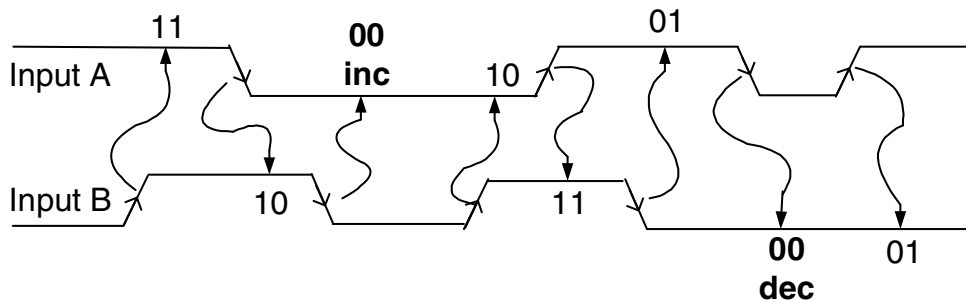


Figure 22-7. Rotary Decoding Mode—Input Transitions

Figure 22-7 shows that each detected edge causes a transition in the decoder state machine. Not all transitions are legal (see Table 22-2). For example, there is no legal way to transition directly from state 11 to 00 using normal inputs. In the cases where this occurs, the state machine goes to an alternate set of states and follows the input sequence until a valid sequence leading to state 00 is detected. No increment or decrement action is taken from the alternate state sequence.

Table 22-2. Rotary Decoder State Machine Transitions

CURRENT STATE	“INPUT” BA=00	“INPUT” BA=01	“INPUT” BA=10	“INPUT” BA=11
00	00	01	10	error
01	00, dec	01	error	11
10	00, inc	error	10	11
11	error	01	10	11

22.3.1 Testing the Rotary Decoder

To test the rotary decoder, select PWM1 and PWM2 as inputs to ROTARYA and ROTARYB. Since PWM1 and PWM2 can be started with known phase offsets and duty cycles, a continuous increment or decrement stream can be generated. Since PWM1 and PWM2 can be used as GPIO devices, the final part of the test is to generate and test a sequence of clockwise and counter-clockwise rotations to cover the entire state machine transitions, including the error conditions.

22.3.2 Behavior During Reset

A soft reset (SFTRST) can take multiple clock periods to complete, so do NOT set CLKGATE when setting SFTRST. The reset process gates the clocks automatically. See [Section 39.3.10, “Correct Way to Soft Reset a Block,”](#) for additional information on using the SFTRST and CLKGATE bit fields.

22.4 Programmable Registers

The following registers describe the programming interface for the timers and the rotary decoder.

Table 22-20. HW_TIMROT_TIMCTRL3 Bit Field Descriptions

BITS	LABEL	RW	RESET	DEFINITION
31:20	RSRVD2	RO	0x0	Always write zeroes to this bit field.
19:16	TEST_SIGNAL	RW	0x0	Selects the source of the signal to be measured in duty cycle mode. NEVER_TICK = 0x0 Never tick. Freeze the count. PWM0 = 0x1 Input from PWM0. PWM1 = 0x2 Input from PWM1. PWM2 = 0x3 Input from PWM2. PWM3 = 0x4 Input from PWM3. PWM4 = 0x5 Input from PWM4. ROTARYA = 0x6 Input from Rotary A. ROTARYB = 0x7 Input from Rotary B. 32KHZ_XTAL = 0x8 Input from 32-kHz crystal. 8KHZ_XTAL = 0x9 Input from 8 kHz (divided from 32-kHz crystal). 4KHZ_XTAL = 0xA Input from 4 kHz (divided from 32-kHz crystal). 1KHZ_XTAL = 0xB Input from 1 kHz (divided from 32-kHz crystal). TICK_ALWAYS = 0xC Always tick.
15	IRQ	RW	0x0	This bit is set to one when Timer 3 decrements to zero. Write a zero to clear it or use Clear SCT mode.
14	IRQ_EN	RW	0x0	Set this bit to one to enable the generation of a CPU interrupt when the count reaches zero in normal counter mode.
13:11	RSRVD1	RO	0x0	Always write zeroes to this bit field.
10	DUTY_VALID	RO	0x0	This bit is set and cleared by the hardware. It is set only when in duty cycle measuring mode and the HW_TIMROT_TIMCOUNT3 has valid duty cycle data to be read. This register will be cleared if not in duty cycle mode or on writes to this register. In the case that it is written while in duty cycle mode, this bit will clear but will again be set at the appropriate time for reading the count register.
9	DUTY_CYCLE	RW	0x0	Set this bit to one to cause the timer to operate in duty cycle measuring mode.
8	POLARITY	RW	0x0	Set this bit to one to invert the input to the edge detector. 0: Positive edge detection. 1: Invert to negative edge detection.
7	UPDATE	RW	0x0	Set this bit to one to cause the running count to be written from the CPU at the same time a new fixed count register value is written.
6	RELOAD	RW	0x0	Set this bit to one to cause the timer to reload its current count from its fixed count value whenever the current count decrements to zero. When set to zero, the timer enters a mode that freezes at a count of zero. When the fixed count is zero, setting this bit to one causes a continuous reload of the fixed count register so that writing a non-zero value will start the timer.

Table 24-22. HW_PWM_PERIOD4 Bit Field Descriptions

BITS	LABEL	RW	RESET	DEFINITION
22:20	CDIV	RW	0x0	Clock divider ratio to apply to the crystal clock frequency (24.0 MHz) that times the PWM output signal. DIV_1 = 0x0 Divide by 1. DIV_2 = 0x1 Divide by 2. DIV_4 = 0x2 Divide by 4. DIV_8 = 0x3 Divide by 8. DIV_16 = 0x4 Divide by 16. DIV_64 = 0x5 Divide by 64. DIV_256 = 0x6 Divide by 256. DIV_1024 = 0x7 Divide by 1024.
19:18	INACTIVE_STATE	RW	0x0	The logical inactive state that is mapped to the PWM output signal. Note that the undefined state of 0x1 is mapped to high-impedance. HI_Z = 0x0 Inactive state sets PWM output to high-impedance. 0 = 0x2 Inactive state sets PWM output to 0 (low). 1 = 0x3 Inactive state sets PWM output to 1 (high).
17:16	ACTIVE_STATE	RW	0x0	The logical active state is mapped to the PWM output signal. Note that the undefined state of 0x1 is mapped to high-impedance. HI_Z = 0x0 Active state sets PWM output to high-impedance. 0 = 0x2 Active state sets PWM output to 0 (low). 1 = 0x3 Active state sets PWM output to 1 (high).
15:0	PERIOD	RW	0x0	Number of divided XTAL clock cycles in the entire period of the PWM waveform, minus 1. For example, to obtain 6 clock cycles in the actual period, set this field to 5.

DESCRIPTION:

This register contains the programming parameters for multi-chip attachment mode, clock divider value, active high, low values and period for channel 4.

EXAMPLE:

```
HW_PWM_PERIODn_WR(4, 0x00000b1f); // Set up period and active/inactive output states
```

24.4.12 PWM Version Register Description

This register indicates the version of the block for debug purposes.

HW_PWM_VERSION

0x0b0

Table 24-23. HW_PWM_VERSION

3	3	2	2	2	2	2	2	2	2	2	2	1	1	1	1	1	1	1	1	0	0	0	0	0	0	0	0	0			
1	0	9	8	7	6	5	4	3	2	1	0	9	8	7	6	5	4	3	2	1	0	9	8	7	6	5	4	3	2	1	0
MAJOR								MINOR								STEP															

dropped (decimated) to find the next sample at which to produce a filtered multibit sigma delta A/D value to send to the DMA. Whenever the whole number part (bits 23:16) is zero, then a sample is produced.

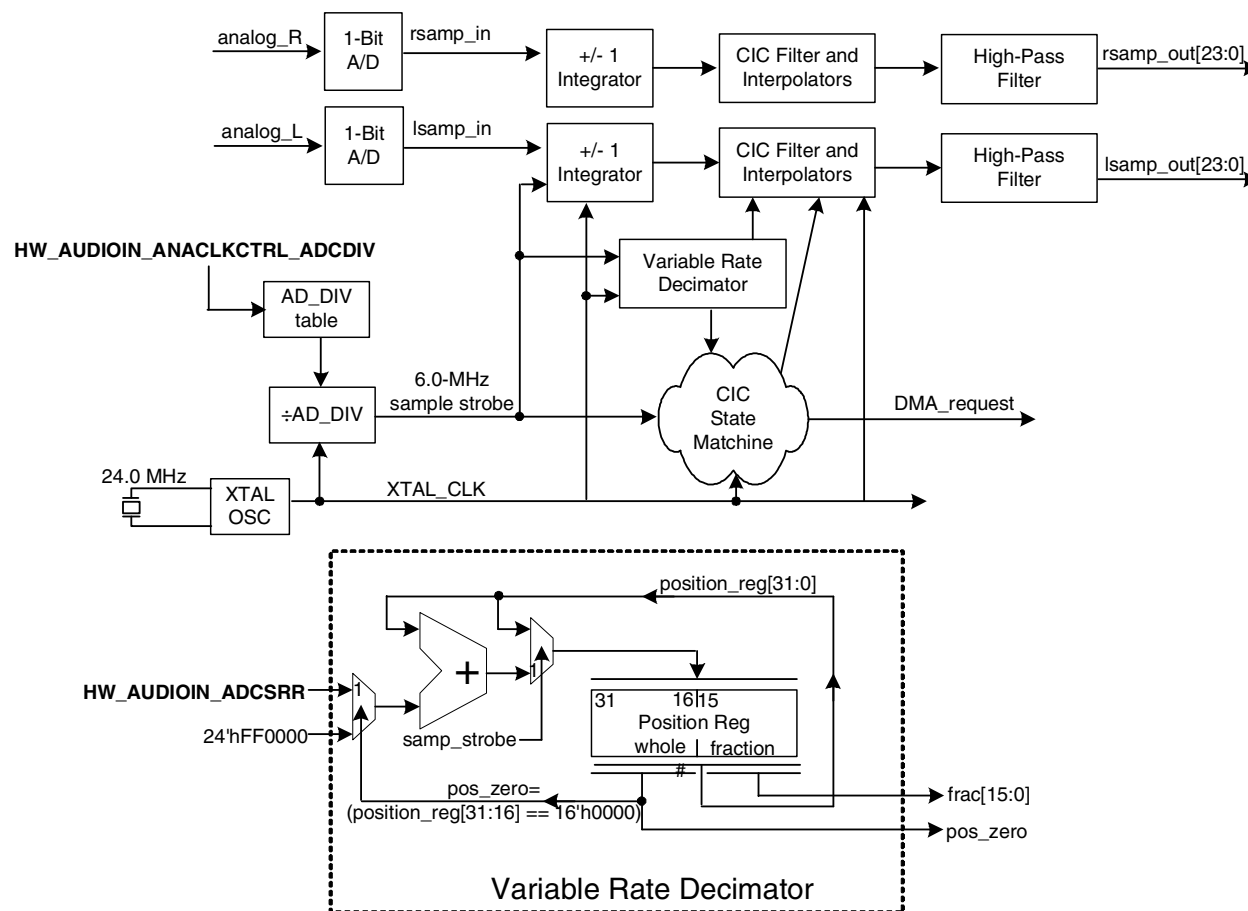


Figure 28-3. Variable-Rate A/D Converter

The range of values of the samples stored into the on-chip RAM is proportional to the square of the over-sample rate (OSR) used in the capture process. The larger the OSR, the longer period the integrators run in the ADC. As a result, the range of values seen for the same signal wave form captured at the same sample rate but with two different OSR will be different.

For example:

- An 8-kHz microphone captured at $F_{ADC} = 6.0$ MHz will be 36 times smaller than the values resulting from capturing the same source signal at $F_{ADC} = 1.0$ MHz.
- The peak range of values seen in a capture of a signal at 44.1 kHz with $F_{analogADC} = 6.0$ MHz is ± 3200 .
- The oversample ratio in this case is $OSR = 136.054$.
- Calculate a magnitude constant, K_{filter} for ADC's filter from this as $K_{filter} = OSR^2 / \text{Peak Value} = (136.054)^2 / 3200 = 5.7846$.
- For any OSR in any sample rate, the peak value can be approximated by $\text{Value}_{peak} = OSR^2 / K_{filter}$

shots them and starts a new conversion operation for all scheduled channels. Thus, one can set the schedule bits for four channels on the same clock edge. The channel with the largest channel number is converted last and has its interrupt status bit set last. If that channel is the only one of the four with an interrupt enable bit set, then it interrupts the ARM after all four channels have been converted, effectively ganging four channels together.

33.2.4 Delay Channels

To minimize the interrupt load on the ARM processor, four delay channels are provided. Each has an 11-bit counter that increments at 2 kHz. A delay channel can be kicked off either by an ARM store instruction or at the completion of a delay channel time-out. At time-out, each channel has the option of kicking off any combination of LRADC conversions, as well as any combination of delay channels.

NOTE: The DELAY fields in HW_LRADC_DELAY0, HW_LRADC_DELAY1, HW_LRADC_DELAY2, and HW_LRADC_DELAY3 must be non-zero; otherwise, the LRADC will not trigger the delay group. The ACCUMULATE bit in the appropriate channel register HW_LRADC_CHn must be set to 1 if NUM_SAMPLES is greater than 0; otherwise, the IRQs will not fire.

Consider the case of a touch-screen that requires 4x oversampling of its coordinate values. Further, suppose you wish to receive an oversampled X or Y coordinate approximately every 5 ms and that the oversampling should be spaced at 1-ms intervals.

- In the touch-screen, first select either X or Y drive, then set up the appropriate LRADC.
- In setting up the LRADC, clear the accumulator associated with it by setting the ACCUMULATE bit and set the NUM_SAMPLES field to 3 (4 samples before interrupt request).
- Next, set up two delay channels.
 - Delay Channel 1 is set to delay 1 ms (DELAY = 1, two ticks) and then kick the schedule bit for LRADC 4. Its LOOP_COUNT bit field is also set to 3, so that four kicks of LRADC 4 occur, each spaced by 1 ms.
 - Delay Channel 0 is set to delay 1 ms with LOOP_COUNT = 0, i.e., one time. Its TRIGGER_DELAYS field is set to trigger Delay Channel 1 when it times out. The ISR routine kicks off Delay Channel 0 immediately before it does its return from interrupt. Another interrupt (LRADC4_IRQ) is asserted once the entire 4x oversample data capture is complete. A sample timeline for such a sequence is shown in [Figure 33-3](#).

NOTE: If a delay group schedules channels to be sampled and a manual write to the schedule field in CTRL0 occurs while the block is discarding samples, the LRADC will switch to the new schedule and will not sample the channels that were previously scheduled. The time window for this to happen is very small and lasts only while the LRADC is discarding samples.

WARNING: The pad ESD protection limits the voltage on the LRADC0-LRADC6 inputs to VDDIO. The BATT and 5V inputs to the LRADC have built-in dividers to handle the higher voltages.

Table 33-28. HW_LRADC_DELAY0 Bit Field Descriptions

BITS	LABEL	RW	RESET	DEFINITION
19:16	TRIGGER_DELAYS	RW	0x0	Setting a bit in this bit field to one causes the delay controller to trigger the corresponding delay channel. This trigger occurs when the delay count of this delay channel reaches zero. Note that all four delay channels can be triggered at the same time, including the one that issues the trigger. This can have the effect of automatically retriggering a delay channel.
15:11	LOOP_COUNT	RW	0x00	This bit field specifies the number of times this delay counter will count down and then trigger its designated targets. This is particularly useful for scheduling multiple samples of an LRADC channel set. If this field is set to 0x0, then exactly one delay loop will be generated with exactly one event triggering the target LRADC and/or delay channels. Note setting the loop count to 0x01 will yield two conversions.
10:0	DELAY	RW	0x000	This 11-bit field counts down to zero. At zero it triggers either a set of LRADC channel conversions or another delay channel, or both. It can trigger up to all eight LRADCs and all four delay channels in a single even. This counter operates on a 2KHz clock derived from crystal clock.

DESCRIPTION:

The LRADC Delay Channel provides control by which LRADC channels and delay channels (including itself) may be triggered. The triggering of the selected delay and LRADC channel(s) is delayed by the DELAY field value which counts down on a 2 kHz clock. It is possible to use delay channels chained together to configure dependent timing of channel conversions as in the example provided in introduction to this block. A delay channel may also be configured to trigger itself. In this case, it could be used to simultaneously trigger an LRADC channel, providing continuous acquisitions of the conversions executed, delayed by the value specified in the DELAY field. The delay channel is started by setting the KICK bit to one.

EXAMPLE:

```
HW_LRADC_DELAYn_WR(0, (BF_LRADC_DELAYn_TRIGGER_LRADCs(0x05) | // LRADC channel 0 and 2
BF_LRADC_DELAYn_KICK(1) | // Start the Delay channel
BF_LRADC_DELAYn_TRIGGER_DELAYS(0x1) | // restart delay channel 0 each time
BF_LRADC_DELAYn_DELAY(0x0E45) ) ); // delay 3653 periods of 2 kHz clock
// ... do other things until the triggered LRADC channels report an interrupt.
```

33.4.15 LRADC Scheduling Delay 1 Description

The LRADC scheduling delay 1 register controls one delay operation. At the end of this delay, this channel can trigger one or more LRADC channels or one or more Scheduling delay channels .

HW_LRADC_DELAY1	0x0E0
HW_LRADC_DELAY1_SET	0x0E4
HW_LRADC_DELAY1_CLR	0x0E8
HW_LRADC_DELAY1_TOG	0x0EC

- When the synchronized edge is recognized by the on-chip SJTAG controller, it pulls the DEBUG line back down clock to Q to pad after the rising edge of its 24-MHz clock. This is the critical timing mark that is detected in the FPGA/CPLD and used to time data in next phase.
- The timing mark phase ends when the on-chip SJTAG stops driving the serial JTAG wire low for one cycle.

34.2.3 Debugger Send TDI, Mode Phase

- During the first 24-MHz clock period of this phase, the FPGA/CPLD sends a one clock-wide signal that either tells the on-chip SJTAG that it is present and a JTAG clock begins, or it tells the on-chip SJTAG to do a JTAG reset operation to the ARM JTAG TAP controller.
- If a noise glitch falsely triggered the ASYNC Start Phase, then the on-chip SJTAG will treat it as a TAP controller reset in most cases.
- If the debugger is performing a JTAG clock cycle operation then, it next sends the state of the debugger TDI and MODE pins sequentially on the wire, i.e., one in each of the following two 24-MHz clocks. Notice that for this phase, the FPGA/CPLD knows the correct timing to change these three data elements on the wire because of the timing information it learned from the Timing Mark Phase.
- This phase ends after the FPGA/CPLD drives the serial wire low on the fourth 24-MHz clock of this phase.

To review, the first data element sent is the signal that distinguishes clock cycles from TAP reset cycles. The next two bits sent are the JTAG MODE and TDI bits from the debugger. Finally, the line is driven low and pulled down for half a 24-MHz clock and the driver is turned off and the pulldown left on. This phase ends when the half-clock pulldown is complete. The rising edge of the JTAG clock is sent to the ARM TAP controller during this phase.

34.2.4 i.MX23 Wait For Return Clock Phase

During this phase:

- The on-chip SJTAG controller waits for the ARM TAP controller to send back the return clock. This is an asynchronous event for both the on-chip TAP controller and the FPGA/CPLD controller.
- The on-chip controller drives the serial wire high for one 24-MHz clock period to tell the FPGA/CPLD that the variable length wait for return clock period is complete.
- This phase ends when the on-chip SJTAG detects the return clock going high and drives the serial high for one 24-MHz clock.

34.2.5 i.MX23 Sends TDO and Return Clock Timing Phase

During this phase:

- The on-chip SJTAG controller sends the value of the ARM TAP controller's TDO signal on the wire for one full 24-MHz period that begins on the rising edge of the on-chip 24-MHz clock.
- This phase ends when the TDO value has been sent.

35.8.3.1 NAND Config Block

The primary NAND Config Block (NCB) resides on the NAND attached to GPMI_CE0. The NCB is the first sector in the first good block. In the single NAND configuration, a copy of the NCB also resides on the NAND—its location is immediately after NCB1 - 2nd search area. In the case of multiple NAND devices, copies of the NCB, LDLB, and DBBT are located on the first two NANDs. This case is shown in more detail in [Section 35.8.3.4, “Firmware Layout on the NAND.”](#)

The layout of first good page containing NCB is described in [Figure 35-6](#).

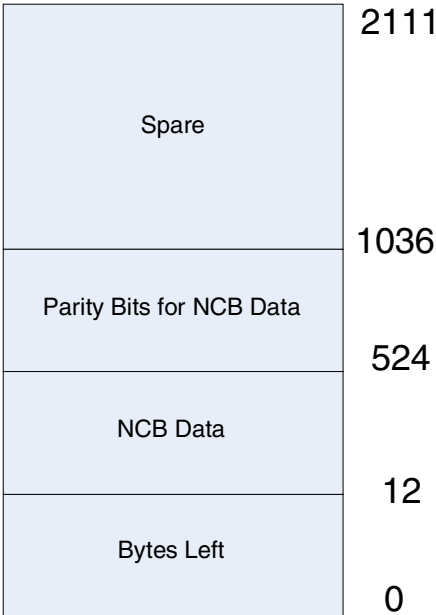


Figure 35-6. Layout of Boot Page Containing NCB

The NCB is located on the first good page of the NAND; minimum size of a page is 2112 bytes. The first 12 bytes of NCB page are reserved and left blank (all 0s), next 512 bytes are reserved for NCB data. The remaining 512 bytes are available for software ECC and the rest are all 0s. NCB is protected using SEC-DED Hamming codes.

35.8.3.2 Single Error Correct and Double Error Detect (SEC-DED) Hamming

For each 8 bits of data in 512 byte NCB, 5-bit parity is used. Each byte of parity area contains 5 parity bits (LSB) and 3 unused bits (MSB).

For each 8 bits of NCB data, parity is calculated and compared with corresponding parity bits read from the parity area of NCB page to detect errors and correct single error.

If errors are more than one then flag shall be raised against the block.

To determine a good NCB, all three fingerprints must match and the ECC must not fail.

Table 36-5. 169-Pin BGA Pin Definitions by Pin Name (continued)

Number	Pin Name	Group	Type	Description 1	Description 2	Description 3
A11	BATT	DCDC	A	Battery Input		
B11	DCDC_BATTERY	DCDC	A	DCDC Battery		
A13	DCDC_GND	DCDC	A	DCDC Ground		
B13	DCDC_LN1	DCDC	A	DCDC Inductor N 1		
A12	DCDC_LP	DCDC	A	DCDC Inductor P		
B12	DCDC_VDDA	DCDC	A	DCDC Analog Power		
D13	DCDC_VDDD	DCDC	A	DCDC Digital Core Power		
C13	DCDC_VDDIO	DCDC	A	DCDC I/O Power		
E12	DEBUG	SYSTEM	DIO	1-Wire Debug Port		
K13	EMI_A00	EMI	DIO	EMI Address 0		
K12	EMI_A01	EMI	DIO	EMI Address 1		
J13	EMI_A02	EMI	DIO	EMI Address 2		
K11	EMI_A03	EMI	DIO	EMI Address 3		
G10	EMI_A04	EMI	DIO	EMI Address 4		
G12	EMI_A05	EMI	DIO	EMI Address 5		
F12	EMI_A06	EMI	DIO	EMI Address 6		
G11	EMI_A07	EMI	DIO	EMI Address 7		
G13	EMI_A08	EMI	DIO	EMI Address 8		
F13	EMI_A09	EMI	DIO	EMI Address 9		
J12	EMI_A10	EMI	DIO	EMI Address 10		
H10	EMI_A11	EMI	DIO	EMI Address 11		
H13	EMI_A12	EMI	DIO	EMI Address 12		
L13	EMI_BA0	EMI	DIO	DRAM Bank Address 0		
L12	EMI_BA1	EMI	DIO	DRAM Bank Address 1		
M13	EMI_CASN	EMI	DIO	EMI Column Address Strobe#		
J10	EMI_CE0N	EMI	DIO	EMI Chip Enable 0#		
H12	EMI_CE1N	EMI	DIO	EMI Chip Enable 1#		
F11	EMI_CKE	EMI	DIO	EMI Clock Enable		
M6	EMI_CLK	EMI	DIO	EMI Clock		
N6	EMI_CLKN	EMI	DIO	EMI Clock#		
K10	EMI_D00	EMI	DIO	EMI Data 0		
M10	EMI_D01	EMI	DIO	EMI Data 1		
N11	EMI_D02	EMI	DIO	EMI Data 2		
N10	EMI_D03	EMI	DIO	EMI Data 3		
M11	EMI_D04	EMI	DIO	EMI Data 4		
K9	EMI_D05	EMI	DIO	EMI Data 5		
L11	EMI_D06	EMI	DIO	EMI Data 6		
L9	EMI_D07	EMI	DIO	EMI Data 7		
N9	EMI_D08	EMI	DIO	EMI Data 8		
N8	EMI_D09	EMI	DIO	EMI Data 9		
M8	EMI_D10	EMI	DIO	EMI Data 10		
K7	EMI_D11	EMI	DIO	EMI Data 11		
L7	EMI_D12	EMI	DIO	EMI Data 12		
K6	EMI_D13	EMI	DIO	EMI Data 13		
N7	EMI_D14	EMI	DIO	EMI Data 14		
M7	EMI_D15	EMI	DIO	EMI Data 15		

Table 36-7. 169-Pin BGA Ball Map

	1	2	3	4	5	6	7	8	9	10	11	12	13	
A	USB_DP	VSSA2	VAG	XTALJ	XTALO	LINE1_INR	HP_VGN0	LRADC6	VDD4P2	LRADC4	BATT	DCDC_LP	DCDC_GND	A
B	VDD6	USB_DM	PSWITCH	RTC_XTALJ	LINE1_INL	MIC	HPL	HPR	LRADC1	LRADC5	DCDC_BATTERY	DCDC_VDDA	DCDC_LN1	B
C	SPEAKERP	VDDXTAL	PWM0	RTC_XTALO	I2C_SCL	I2C_SDA	VDDA1	VSSA1	LRADC2	VDDM	VDAC1	PWM4	DCDC_VDDIO	C
D	SPEAKERN	PWM1	LCD_D00	VSSA4	VDD01/3	VSS01/3	ROTARYA	VDD01/2	LRADC3	SSP1_DETECT	PWM3	SSP1_SCK	DCDC_VDD0	D
E	LCD_D17	LCD_D01	LCD_D16	LCD_D02	LCD_D15	GPML_RDY3	GPML_WRN	ROTARYB	VSS01/3	SSP1_DATA3	VDDIO03_1/2/3	DEBUG	VDDSV	E
F	LCD_D03	LCD_D14	LCD_D04	LCD_D13	LCD_D05	GPML_RDY2	GPML_WPN	GPML_CE1N	SSP1_DATA1	SSP1_DATA2	EMI_CKE	EMI_A06	EMI_A09	F
G	LCD_D12	LCD_D06	VDDIO03_1/2/3	LCD_D11	LCD_D07	GPML_RDY0	AUART1_CTS	GPML_CE0N	SSP1_DATA0	EMI_A04	EMI_A07	EMI_A05	EMI_A08	G
H	LCD_D10	LCD_D08	LCD_D09	LCD_CS	LCD_ENABLE	GPML_RDN	AUART1_RTS	AUART1_RX	SSP1_CMD	EMI_A11	PWM2	EMI_CE1N	EMI_A12	H
J	LCD_VSYNC	LCD_HSYNC	LCD_RS	LCD_WR	LCD_RESET	GPML_RDY1	VSSIO_EMI1/2	AUART1_TX	VSSIO_EMI1/2	EMI_CE0N	EMI_WEN	EMI_A10	EMI_A02	J
K	LCD_DOTCK	GPML_D00	GPML_CLE	GPML_ALE	GPML_CE2N	EMI_D13	EMI_D11	EMI_DQS1	EMI_D05	EMI_D00	EMI_A03	EMI_A01	EMI_A00	K
L	GPML_D01	GPML_D02	GPML_D03	GPML_D05	GPML_D04	VDDIO_EMI0	EMI_D12	VDDIO_EMI1/2	EMI_D07	VDDIO_EMI1/2	EMI_D06	EMI_BA1	EMI_BA0	L
M	GPML_D07	GPML_D06	GPML_D08	GPML_D09	GPML_D10	EMI_CLK	EMI_D15	EMI_D10	EMI_DQM1	EMI_D01	EMI_D04	EMI_DQM0	EMI_CASN	M
N	GPML_D11	GPML_D12	GPML_D13	GPML_D14	GPML_D15	EMI_CLKN	EMI_D14	EMI_D09	EMI_D08	EMI_D03	EMI_D02	EMI_DQS0	EMI_RASN	N
	1	2	3	4	5	6	7	8	9	10	11	12	13	

Table 37-10. HW_PINCTRL_MUXSEL1 Bit Field Descriptions

BITS	LABEL	RW	RESET	DEFINITION
27:26	BANK0_PIN29	RW	0x3	Pin 69, AUART1_TX pin function selection: 00= auart1_tx; 01= reserved; 10= ssp1_d7; 11= GPIO.
25:24	BANK0_PIN28	RW	0x3	Pin 68, AUART1_RX pin function selection: 00= auart1_rx; 01= reserved; 10= ssp1_d6; 11= GPIO.
23:22	BANK0_PIN27	RW	0x3	Pin 67, AUART1_RTS pin function selection: 00= auart1_rts; 01= reserved; 10= ssp1_d5; 11= GPIO.
21:20	BANK0_PIN26	RW	0x3	Pin 66, AUART1_CTS pin function selection: 00= auart1_cts; 01= reserved; 10= ssp1_d4; 11= GPIO.
19:18	BANK0_PIN25	RW	0x3	Pin 60, GPML_RDN pin function selection: 00= gpml_rdn; 01= reserved; 10= reserved; 11= GPIO.
17:16	BANK0_PIN24	RW	0x3	Pin 65, GPML_WRN pin function selection: 00= gpml_wrn; 01= reserved; 10= ssp2_sck; 11= GPIO.
15:14	BANK0_PIN23	RW	0x3	Pin 64, GPML_WPN pin function selection: 00= gpml_wpn; 01= reserved; 10= reserved; 11= GPIO.
13:12	BANK0_PIN22	RW	0x3	Pin 63, GPML_RDY3 pin function selection: 00= gpml_ready3; 01= reserved; 10= reserved; 11= GPIO.
11:10	BANK0_PIN21	RW	0x3	Pin 62, GPML_RDY2 pin function selection: 00= gpml_ready2; 01= reserved; 10= reserved; 11= GPIO.
9:8	BANK0_PIN20	RW	0x3	Pin 43, GPML_RDY1 pin function selection: 00= gpml_ready1; 01= reserved; 10= ssp2_cmd; 11= GPIO.