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"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	e200z0h
Core Size	32-Bit Single-Core
Speed	64MHz
Connectivity	CANbus, I²C, LINbus, SCI, SPI
Peripherals	DMA, POR, PWM, WDT
Number of I/O	149
Program Memory Size	1MB (1M x 8)
Program Memory Type	FLASH
EEPROM Size	64K x 8
RAM Size	80K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 29x10b, 5x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	176-LQFP
Supplier Device Package	176-LQFP (24x24)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/ppc5606bclu64

Figure 2 shows the MPC5607B in the 176 LQFP package.

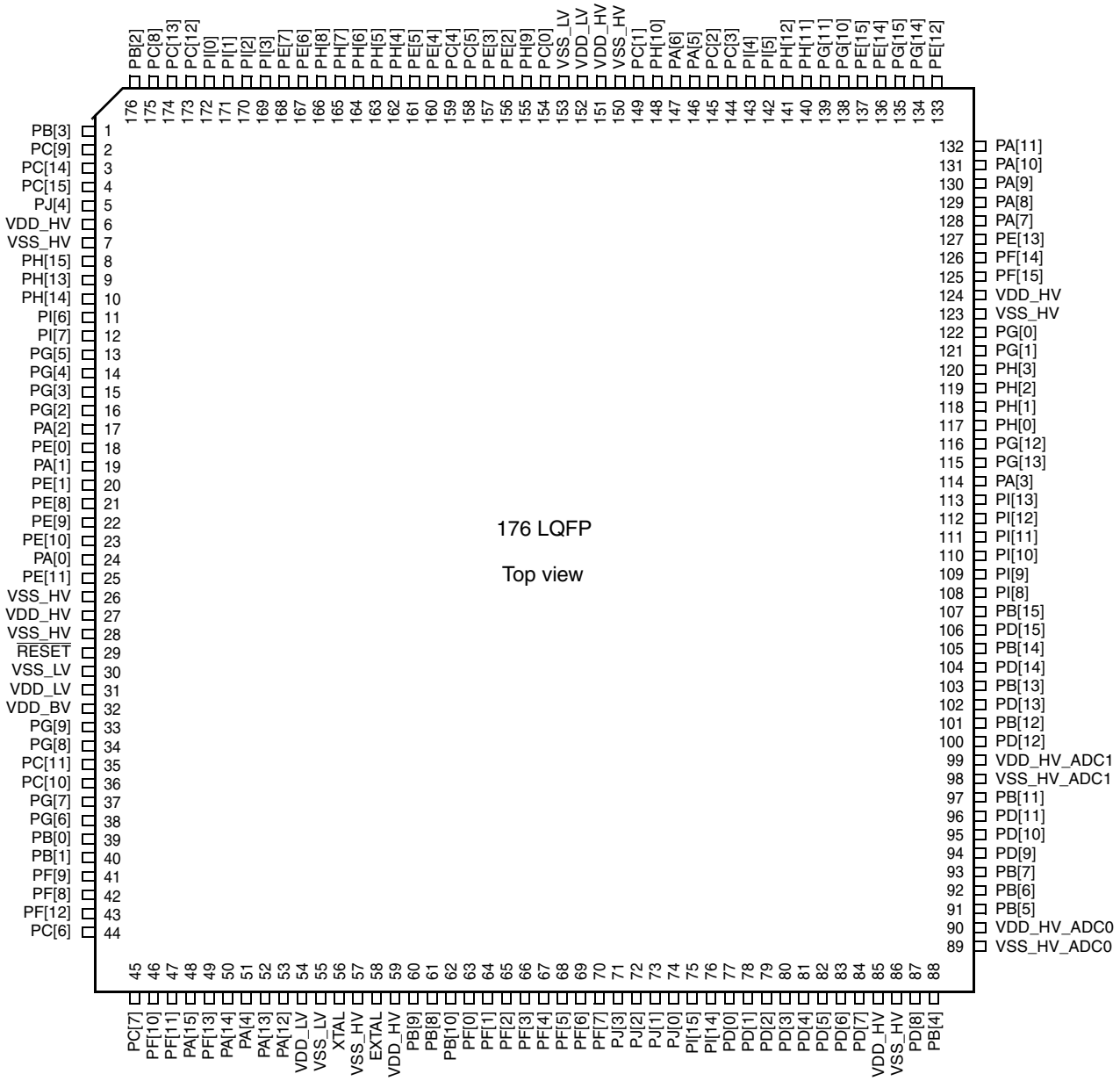


Figure 2. 176 LQFP pin configuration

Figure 5 shows the MPC5607B in the 208 MAPBGA package.

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	
A	PC[8]	PC[13]	PH[15]	PJ[4]	PH[8]	PH[4]	PC[5]	PC[0]	PI[0]	PI[1]	PC[2]	PI[4]	PE[15]	PH[11]	NC	NC	A
B	PC[9]	PB[2]	PH[13]	PC[12]	PE[6]	PH[5]	PC[4]	PH[9]	PH[10]	PI[2]	PC[3]	PG[11]	PG[15]	PG[14]	PA[11]	PA[10]	B
C	PC[14]	VDD_HV	PB[3]	PE[7]	PH[7]	PE[5]	PE[3]	VSS_LV	PC[1]	PI[3]	PA[5]	PI[5]	PE[14]	PE[12]	PA[9]	PA[8]	C
D	PH[14]	PI[6]	PC[15]	PI[7]	PH[6]	PE[4]	PE[2]	VDD_LV	VDD_HV	NC	PA[6]	PH[12]	PG[10]	PF[14]	PE[13]	PA[7]	D
E	PG[4]	PG[5]	PG[3]	PG[2]									PG[1]	PG[0]	PF[15]	VDD_HV	E
F	PE[0]	PA[2]	PA[1]	PE[1]									PH[0]	PH[1]	PH[3]	PH[2]	F
G	PE[9]	PE[8]	PE[10]	PA[0]			VSS_HV	VSS_HV	VSS_HV	VSS_HV			VDD_HV	PI[12]	PI[13]	MSEO	G
H	VSS_HV	PE[11]	VDD_HV	NC			VSS_HV	VSS_HV	VSS_HV	VSS_HV			MDO3	MDO2	MDO0	MDO1	H
J	RESET	VSS_LV	NC	NC			VSS_HV	VSS_HV	VSS_HV	VSS_HV			PI[8]	PI[9]	PI[10]	PI[11]	J
K	EVTI	NC	VDD_BV	VDD_LV			VSS_HV	VSS_HV	VSS_HV	VSS_HV			VDD_HV_ADC1	PG[12]	PA[3]	PG[13]	K
L	PG[9]	PG[8]	NC	EVTO									PB[15]	PD[15]	PD[14]	PB[14]	L
M	PG[7]	PG[6]	PC[10]	PC[11]									PB[13]	PD[13]	PD[12]	PB[12]	M
N	PB[1]	PF[9]	PB[0]	VDD_HV	PJ[0]	PA[4]	VSS_LV	EXTAL	VDD_HV	PF[0]	PF[4]	VSS_HV_ADC1	PB[11]	PD[10]	PD[9]	PD[11]	N
P	PF[8]	PJ[3]	PC[7]	PJ[2]	PJ[1]	PA[14]	VDD_LV	XTAL	PB[10]	PF[1]	PF[5]	PD[0]	PD[3]	VDD_HV_ADC0	PB[6]	PB[7]	P
R	PF[12]	PC[6]	PF[10]	PF[11]	VDD_HV	PA[15]	PA[13]	PI[14]	XTAL32	PF[3]	PF[7]	PD[2]	PD[4]	PD[7]	VSS_HV_ADC0	PB[5]	R
T	NC	NC	NC	MCKO	NC	PF[13]	PA[12]	PI[15]	EXTAL32	PF[2]	PF[6]	PD[1]	PD[5]	PD[6]	PD[8]	PB[4]	T
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	

NOTE: The 208 MAPBGA is available only as development package for Nexus 2+.

NC = Not connected

Figure 5. 208 MAPBGA configuration

3.2 Pad configuration during reset phases

All pads have a fixed configuration under reset.

During the power-up phase, all pads are forced to tristate.

After power-up phase, all pads are tristate with the following exceptions:

- PA[9] (FAB) is pull-down. Without external strong pull-up the device starts fetching from flash.
- PA[8], PC[0] and PH[9:10] are in input weak pull-up when out of reset.
- RESET pad is driven low by the device till 40 FIRC clock cycles after phase2 completion. Minimum phase3 duration is 40 FIRC cycles.
- Nexus output pads (MDO[n], MCKO, EVTO, MSEO) are forced to output.

Table 5. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								100 LQFP	144 LQFP	176 LQFP	208 MAP BGA[4]
PA[14]	PCR[14]	AF0 AF1 AF2 AF3 —	GPIO[14] SCK_0 CS0_0 E0UC[0] EIRQ[4]	SIUL DSPI_0 DSPI_0 eMIOS_0 SIUL	I/O I/O I/O I/O I	M	Tristate	28	42	50	P6
PA[15]	PCR[15]	AF0 AF1 AF2 AF3 —	GPIO[15] CS0_0 SCK_0 E0UC[1] WKPU[10][5]	SIUL DSPI_0 DSPI_0 eMIOS_0 WKPU	I/O I/O I/O I/O I	M	Tristate	27	40	48	R6
Port B											
PB[0]	PCR[16]	AF0 AF1 AF2 AF3	GPIO[16] CAN0TX E0UC[30] LIN0TX	SIUL FlexCAN_0 eMIOS_0 LINFlex_0	I/O O I/O O	M	Tristate	23	31	39	N3
PB[1]	PCR[17]	AF0 AF1 AF2 AF3 — — —	GPIO[17] — E0UC[31] — WKPU[4][5] CAN0RX LIN0RX	SIUL — eMIOS_0 — WKPU FlexCAN_0 LINFlex_0	I/O — I/O — I I I	S	Tristate	24	32	40	N1
PB[2]	PCR[18]	AF0 AF1 AF2 AF3	GPIO[18] LIN0TX SDA E0UC[30]	SIUL LINFlex_0 I^2C_0 eMIOS_0	I/O O I/O I/O	M	Tristate	100	144	176	B2
PB[3]	PCR[19]	AF0 AF1 AF2 AF3 — —	GPIO[19] E0UC[31] SCL — WKPU[11][5] LIN0RX	SIUL eMIOS_0 I^2C_0 — WKPU LINFlex_0	I/O I/O I/O — I I	S	Tristate	1	1	1	C3
PB[4]	PCR[20]	AF0 AF1 AF2 AF3 — — —	— — — — ADC0_P[0] ADC1_P[0] GPIO[20]	— — — — ADC_0 ADC_1 SIUL	— — — — I I I	I	Tristate	50	72	88	T16

Table 5. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								100 LQFP	144 LQFP	176 LQFP	208 MAP BGA[4]
PE[9]	PCR[73]	AF0 AF1 AF2 AF3 — — —	GPIO[73] — E0UC[23] — WKPU[7][5] CAN2RX CAN3RX	SIUL — eMIOS_0 — WKPU FlexCAN_2 FlexCAN_3	I/O — I/O — I I I	S	Tristate	10	14	22	G1
PE[10]	PCR[74]	AF0 AF1 AF2 AF3 —	GPIO[74] LIN3TX CS3_1 E1UC[30] EIRQ[10]	SIUL LINFlex_3 DSPI_1 eMIOS_1 SIUL	I/O O O I/O I	S	Tristate	11	15	23	G3
PE[11]	PCR[75]	AF0 AF1 AF2 AF3 — —	GPIO[75] E0UC[24] CS4_1 — LIN3RX WKPU[14][5]	SIUL eMIOS_0 DSPI_1 — LINFlex_3 WKPU	I/O I/O O — I I	S	Tristate	13	17	25	H2
PE[12]	PCR[76]	AF0 AF1 AF2 AF3 — — —	GPIO[76] — E1UC[19][12] — EIRQ[11] SIN_2 ADC1_S[7]	SIUL — eMIOS_1 — SIUL DSPI_2 ADC_1	I/O — I/O — I I I	J	Tristate	76	109	133	C14
PE[13]	PCR[77]	AF0 AF1 AF2 AF3	GPIO[77] SOUT_2 E1UC[20] —	SIUL DSPI_2 eMIOS_1 —	I/O O I/O —	S	Tristate	—	103	127	D15
PE[14]	PCR[78]	AF0 AF1 AF2 AF3 —	GPIO[78] SCK_2 E1UC[21] — EIRQ[12]	SIUL DSPI_2 eMIOS_1 — SIUL	I/O I/O I/O — I	S	Tristate	—	112	136	C13
PE[15]	PCR[79]	AF0 AF1 AF2 AF3	GPIO[79] CS0_2 E1UC[22] —	SIUL DSPI_2 eMIOS_1 —	I/O I/O I/O —	M	Tristate	—	113	137	A13
Port F											

Table 5. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								100 LQFP	144 LQFP	176 LQFP	208 MAP BGA[4]
PF[0]	PCR[80]	AF0 AF1 AF2 AF3 —	GPIO[80] E0UC[10] CS3_1 — ADC0_S[8]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	—	55	63	N10
PF[1]	PCR[81]	AF0 AF1 AF2 AF3 —	GPIO[81] E0UC[11] CS4_1 — ADC0_S[9]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	—	56	64	P10
PF[2]	PCR[82]	AF0 AF1 AF2 AF3 —	GPIO[82] E0UC[12] CS0_2 — ADC0_S[10]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O I/O — I	J	Tristate	—	57	65	T10
PF[3]	PCR[83]	AF0 AF1 AF2 AF3 —	GPIO[83] E0UC[13] CS1_2 — ADC0_S[11]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	—	58	66	R10
PF[4]	PCR[84]	AF0 AF1 AF2 AF3 —	GPIO[84] E0UC[14] CS2_2 — ADC0_S[12]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	—	59	67	N11
PF[5]	PCR[85]	AF0 AF1 AF2 AF3 —	GPIO[85] E0UC[22] CS3_2 — ADC0_S[13]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	—	60	68	P11
PF[6]	PCR[86]	AF0 AF1 AF2 AF3 —	GPIO[86] E0UC[23] CS1_1 — ADC0_S[14]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	—	61	69	T11
PF[7]	PCR[87]	AF0 AF1 AF2 AF3 —	GPIO[87] — CS2_1 — ADC0_S[15]	SIUL — DSPI_1 — ADC_0	I/O — O — I	J	Tristate	—	62	70	R11

Table 8. PAD3V5V field description[1]

Value[2]	Description
0	High voltage supply is 5.0 V
1	High voltage supply is 3.3 V

1 See the device reference manual for more information on the NVUSRO register.

2 Default manufacturing value is '1'. Value can be programmed by customer in Shadow Flash.

4.2.2 NVUSRO[OSCILLATOR_MARGIN] field description

The fast external crystal oscillator consumption is dependent on the OSCILLATOR_MARGIN bit value. Table 9 shows how NVUSRO[OSCILLATOR_MARGIN] controls the device configuration.

Table 9. OSCILLATOR_MARGIN field description[1]

Value[2]	Description
0	Low consumption configuration (4 MHz/8 MHz)
1	High margin configuration (4 MHz/16 MHz)

1 See the device reference manual for more information on the NVUSRO register.

2 Default manufacturing value is '1'. Value can be programmed by customer in Shadow Flash.

4.2.3 NVUSRO[WATCHDOG_EN] field description

The watchdog enable/disable configuration after reset is dependent on the WATCHDOG_EN bit value. Table 10 shows how NVUSRO[WATCHDOG_EN] controls the device configuration.

Table 10. WATCHDOG_EN field description

Value[1]	Description
0	Disable after reset
1	Enable after reset

1 Default manufacturing value is '1'. Value can be programmed by customer in Shadow Flash.

4.3 Absolute maximum ratings

Table 11. Absolute maximum ratings

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
V_{SS}	SR	Digital ground on VSS_HV pins	—	0	0	V
V_{DD}	SR	Voltage on VDD_HV pins with respect to ground (V_{SS})	—	–0.3	6.0	V
V_{SS_LV}	SR	Voltage on VSS_LV (low voltage digital supply) pins with respect to ground (V_{SS})	—	V_{SS} – 0.1	V_{SS} + 0.1	V
V_{DD_BV}	SR	Voltage on VDD_BV (regulator supply) pin with respect to ground (V_{SS})	—	–0.3	6.0	V
			Relative to V_{DD}	–0.3	V_{DD} + 0.3	

4.5 Thermal characteristics

4.5.1 External ballast resistor recommendations

External ballast resistor on V_{DD_BV} pin helps in reducing the overall power dissipation inside the device. This resistor is required only when maximum power consumption exceeds the limit imposed by package thermal characteristics.

As stated in Table 14 LQFP thermal characteristics, considering a thermal resistance of 144 LQFP as 48.3 °C/W, at ambient temperature T_A = 125 °C, the junction temperature T_j will cross 150 °C if the total power dissipation is greater than (150 – 125)/48.3 = 517 mW. Therefore, the total device current I_{DDMAX} at 125 °C/5.5 V must not exceed 94.1 mA (i.e., PD/VDD). Assuming an average $I_{DD}(V_{DD\_HV})$ of 15–20 mA consumption typically during device RUN mode, the LV domain consumption $I_{DD}(V_{DD\_BV})$ is thus limited to $I_{DDMAX} - I_{DD}(V_{DD\_HV})$, i.e., 80 mA.

Therefore, respecting the maximum power allowed as explained in Section 4.5.2, “Package thermal characteristics, it is recommended to use this resistor only in the 125 °C/5.5 V operating corner as per the following guidelines:

- If $I_{DD}(V_{DD\_BV}) < 80$ mA, then no resistor is required.
- If 80 mA < $I_{DD}(V_{DD\_BV}) < 90$ mA, then 4 Ω resistor can be used.
- If $I_{DD}(V_{DD\_BV}) > 90$ mA, then 8 Ω resistor can be used.

Using resistance in the range of 4–8 Ω, the gain will be around 10–20% of total consumption on V_{DD_BV}. For example, if 8 Ω resistor is used, then power consumption when $I_{DD}(V_{DD\_BV})$ is 110 mA is equivalent to power consumption when $I_{DD}(V_{DD\_BV})$ is 90 mA (approximately) when resistor not used.

In order to ensure correct power up, the minimum V_{DD_BV} to be guaranteed is 30 ms/V. If the supply ramp is slower than this value, then LVDHV3B monitoring ballast supply V_{DD_BV} pin gets triggered leading to device reset. Until the supply reaches certain threshold, this low voltage detector (LVD) generates destructive reset event in the system. This threshold depends on the maximum $I_{DD}(V_{DD\_BV})$ possible across the external resistor.

4.5.2 Package thermal characteristics

Table 14. LQFP thermal characteristics[1]

Symbol		C	Parameter	Conditions[2]	Pin count	Value			Unit
						Min	Typ	Max	
$R_{\theta JA}$	CC	D	Thermal resistance, junction-to-ambient natural convection[3]	Single-layer board — 1s	100	—	—	64	°C/W
					144	—	—	64	
					176	—	—	64	
				Four-layer board — 2s2p	100	—	—	49.7	
					144	—	—	48.3	
					176	—	—	47.3	
$R_{\theta JB}$	CC		Thermal resistance, junction-to-board[4]	Single-layer board — 1s	100	—	—	36	°C/W
					144	—	—	38	
					176	—	—	38	
				Four-layer board — 2s2p	100	—	—	33.6	
					144	—	—	33.4	
					176	—	—	33.4	

Table 22. I/O consumption (continued)

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
I_{RMSFST}	CC	D	Root mean square I/O current for FAST configuration	C_L = 25 pF, 40 MHz	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	22	mA
				C_L = 25 pF, 64 MHz		—	—	33	
				C_L = 100 pF, 40 MHz		—	—	56	
				C_L = 25 pF, 40 MHz	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	14	
				C_L = 25 pF, 64 MHz		—	—	20	
				C_L = 100 pF, 40 MHz		—	—	35	
I_{AVGSEG}	SR	D	Sum of all the static I/O current within a supply segment	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0		—	—	70	mA
				V_{DD} = 3.3 V ± 10%, PAD3V5V = 1		—	—	65	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to125 °C, unless otherwise specified

2 Stated maximum values represent peak consumption that lasts only a few ns during I/O transition.

Table 23 provides the weight of concurrent switching I/Os.

Due to the dynamic current limitations, the sum of the weight of concurrent switching I/Os on a single segment must not exceed 100% to ensure device functionality.

Table 23. I/O weight[1]

Supply segment			Pad	176 LQFP				144/100 LQFP			
				Weight 5 V		Weight 3.3 V		Weight 5 V		Weight 3.3 V	
176 LQFP	144 LQFP	100 LQFP		SRC[2] = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1
6	4	4	PB[3]	5%	—	6%	—	13%	—	15%	—
			PC[9]	4%	—	5%	—	13%	—	15%	—
			PC[14]	4%	—	4%	—	13%	—	15%	—
			PC[15]	3%	4%	4%	4%	12%	18%	15%	16%
	—	—	PJ[4]	3%	4%	3%	3%	—	—	—	—

Table 23. I/O weight[1] (continued)

Supply segment			Pad	176 LQFP				144/100 LQFP			
				Weight 5 V		Weight 3.3 V		Weight 5 V		Weight 3.3 V	
176 LQFP	144 LQFP	100 LQFP		SRC[2] = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1	SRC = 0	SRC = 1
3	2	2	PB[9]	1%	—	1%	—	1%	—	1%	—
			PB[8]	1%	—	1%	—	1%	—	1%	—
			PB[10]	5%	—	6%	—	6%	—	7%	—
		—	PF[0]	5%	—	6%	—	6%	—	8%	—
		—	PF[1]	5%	—	6%	—	7%	—	8%	—
		—	PF[2]	6%	—	7%	—	7%	—	9%	—
		—	PF[3]	6%	—	7%	—	8%	—	9%	—
		—	PF[4]	6%	—	7%	—	8%	—	10%	—
		—	PF[5]	6%	—	7%	—	9%	—	10%	—
		—	PF[6]	6%	—	7%	—	9%	—	11%	—
		—	PF[7]	6%	—	7%	—	9%	—	11%	—
	—	—	PJ[3]	6%	—	7%	—	—	—	—	—
	—	—	PJ[2]	6%	—	7%	—	—	—	—	—
	—	—	PJ[1]	6%	—	7%	—	—	—	—	—
	—	—	PJ[0]	6%	—	7%	—	—	—	—	—
	—	—	PI[15]	6%	—	7%	—	—	—	—	—
	—	—	PI[14]	6%	—	7%	—	—	—	—	—
	2	2	PD[0]	1%	—	1%	—	1%	—	1%	—
			PD[1]	1%	—	1%	—	1%	—	1%	—
			PD[2]	1%	—	1%	—	1%	—	1%	—
			PD[3]	1%	—	1%	—	1%	—	1%	—
			PD[4]	1%	—	1%	—	1%	—	1%	—
			PD[5]	1%	—	1%	—	1%	—	1%	—
			PD[6]	1%	—	1%	—	1%	—	2%	—
			PD[7]	1%	—	1%	—	1%	—	2%	—

Table 24. Reset electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
V_{IH}	SR	P	Input High Level CMOS (Schmitt Trigger)	—	$0.65V_{DD}$	—	V_{DD} + 0.4	V
V_{IL}	SR	P	Input low Level CMOS (Schmitt Trigger)	—	–0.4	—	$0.35V_{DD}$	V
V_{HYS}	CC	C	Input hysteresis CMOS (Schmitt Trigger)	—	$0.1V_{DD}$	—	—	V
V_{OL}	CC	P	Output low level	Push Pull, I_{OL} = 2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	—	—	$0.1V_{DD}$	V
				Push Pull, I_{OL} = 1 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[2]	—	—	$0.1V_{DD}$	
				Push Pull, I_{OL} = 1 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	—	—	0.5	
T_{tr}	CC	D	Output transition time output pin[3] MEDIUM configuration	C_L = 25 pF, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	10	ns
				C_L = 50 pF, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	20	
				C_L = 100 pF, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	40	
				C_L = 25 pF, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	12	
				C_L = 50 pF, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	25	
				C_L = 100 pF, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	40	
W_{FRST}	SR	P	$\overline{\text{RESET}}$ input filtered pulse	—	—	—	40	ns
W_{NFRST}	SR	P	$\overline{\text{RESET}}$ input not filtered pulse	—	1000	—	—	ns
$\lvert I_{WPU} \rvert$	CC	P	Weak pull-up current absolute value	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	10	—	150	µA
		D		V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	10	—	150	
		P		V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[4]	10	—	250	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 This is a transient configuration during power-up, up to the end of reset PHASE2 (refer to RGM module section of the device reference manual).

3 C_L includes device and package capacitance (C_{PKG} < 5 pF).

4 The configuration PAD3V5 = 1 when V_{DD} = 5 V is only transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

Table 26. Low voltage detector electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
V_{PORUP}	SR	P	Supply for functional POR module	T_A = 25 °C, after trimming	1.0	—	5.5	V
V_{PORH}	CC	P	Power-on reset threshold		1.5	—	2.6	
$V_{LVDHV3H}$	CC	T	LVDHV3 low voltage detector high threshold		—	—	2.95	
$V_{LVDHV3L}$	CC	P	LVDHV3 low voltage detector low threshold		2.7	—	2.9	
$V_{LVDHV3BH}$	CC	P	LVDHV3B low voltage detector high threshold		—	—	2.95	
$V_{LVDHV3BL}$	CC	P	LVDHV3B low voltage detector low threshold		2.7	—	2.9	
$V_{LVDHV5H}$	CC	T	LVDHV5 low voltage detector high threshold		—	—	4.5	
$V_{LVDHV5L}$	CC	P	LVDHV5 low voltage detector low threshold		3.8	—	4.4	
$V_{LVDLVCORL}$	CC	P	LVDLVCOR low voltage detector low threshold		1.08	—	1.16	
$V_{LVDLVBKPL}$	CC	P	LVDLVBKP low voltage detector low threshold		1.08	—	1.16	

[1] V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

4.9 Power consumption

Table 27 provides DC electrical characteristics for significant application modes. These values are indicative values; actual consumption depends on the application.

Table 27. Power consumption on VDD_BV and VDD_HV

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
I_{DDMAX}[2]	CC	D	RUN mode maximum average current	—		—	115	140[3]	mA
I_{DDRUN}[4]	CC	T	RUN mode typical average current[5]	f_{CPU} = 8 MHz		—	12	—	mA
		T		f_{CPU} = 16 MHz		—	27	—	
		T		f_{CPU} = 32 MHz		—	43	—	
		P		f_{CPU} = 48 MHz		—	56	100	
		P		f_{CPU} = 64 MHz		—	70	125	
I_{DDHALT}	CC	C	HALT mode current[6]	Slow internal RC oscillator (128 kHz) running	T_A = 25 °C	—	10	18	mA
		P			T_A = 125 °C	—	17	28	
I_{DDSTOP}	CC	P	STOP mode current[7]	Slow internal RC oscillator (128 kHz) running	T_A = 25 °C	—	350	900[8]	µA
		D			T_A = 55 °C	—	750	—	
		D			T_A = 85 °C	—	2	7	mA
		D			T_A = 105 °C	—	4	10	
		P			T_A = 125 °C	—	7	14	

Table 34. ESD absolute maximum ratings[1,2]

Symbol	Ratings	Conditions	Class	Max value[3]	Unit
$V_{ESD(HBM)}$	Electrostatic discharge voltage (Human Body Model)	T_A = 25 °C conforming to AEC-Q100-002	H1C	2000	V
$V_{ESD(MM)}$	Electrostatic discharge voltage (Machine Model)	T_A = 25 °C conforming to AEC-Q100-003	M2	200	
$V_{ESD(CDM)}$	Electrostatic discharge voltage (Charged Device Model)	T_A = 25 °C conforming to AEC-Q100-011	C3A	500	
				750 (corners)	

1 All ESD testing is in conformity with CDF-AEC-Q100 Stress Test Qualification for Automotive Grade Integrated Circuits.

2 A device will be defined as a failure if after exposure to ESD pulses the device no longer meets the device specification requirements. Complete DC parametric and functional testing shall be performed per applicable device specification at room temperature followed by hot temperature, unless specified otherwise in the device specification.

3 Data based on characterization results, not tested in production

4.11.3.2 Static latch-up (LU)

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin.
- A current injection is applied to each input, output and configurable I/O pin.

These tests are compliant with the EIA/JESD 78 IC latch-up standard.

Table 35. Latch-up results

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T_A = 125 °C conforming to JESD 78	II level A

4.12 Fast external crystal oscillator (4 to 16 MHz) electrical characteristics

The device provides an oscillator/resonator driver. Figure 11 describes a simple model of the internal oscillator driver and provides an example of a connection for an oscillator or a resonator.

Table 36 provides the parameter description of 4 MHz to 16 MHz crystals used for the design simulations.

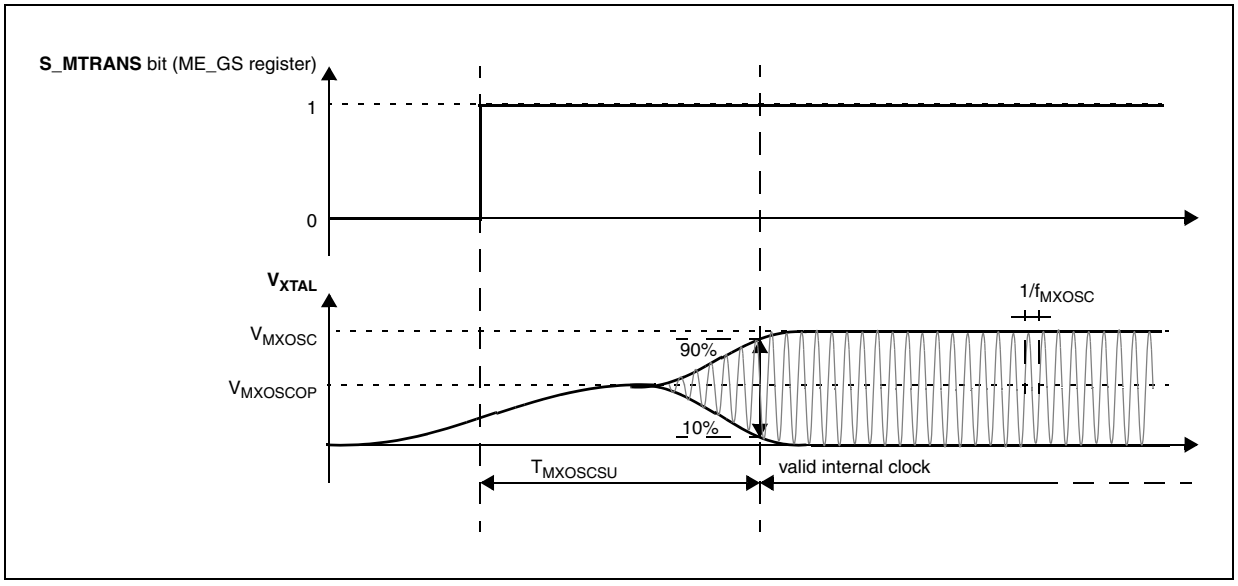


Figure 12. Fast external crystal oscillator (4 to 16 MHz) timing diagram

Table 37. Fast external crystal oscillator (4 to 16 MHz) electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
f_{FXOSC}	SR	—	Fast external crystal oscillator frequency	—	4.0	—	16.0	MHz
g_{mFXOSC}	CC	C	Fast external crystal oscillator transconductance	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 OSCILLATOR_MARGIN = 0	2.2	—	8.2	mA/V
	CC	P		V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 OSCILLATOR_MARGIN = 0	2.0	—	7.4	
	CC	C		V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 OSCILLATOR_MARGIN = 1	2.7	—	9.7	
	CC	C		V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 OSCILLATOR_MARGIN = 1	2.5	—	9.2	
V_{FXOSC}	CC	T	Oscillation amplitude at EXTAL	f_{OSC} = 4 MHz, OSCILLATOR_MARGIN = 0	1.3	—	—	V
				f_{OSC} = 16 MHz, OSCILLATOR_MARGIN = 1	1.3	—	—	
$V_{FXOSCOP}$	CC	C	Oscillation operating point	—	—	0.95	—	V
I_{FXOSC}[2]	CC	T	Fast external crystal oscillator consumption	—	—	2	3	mA

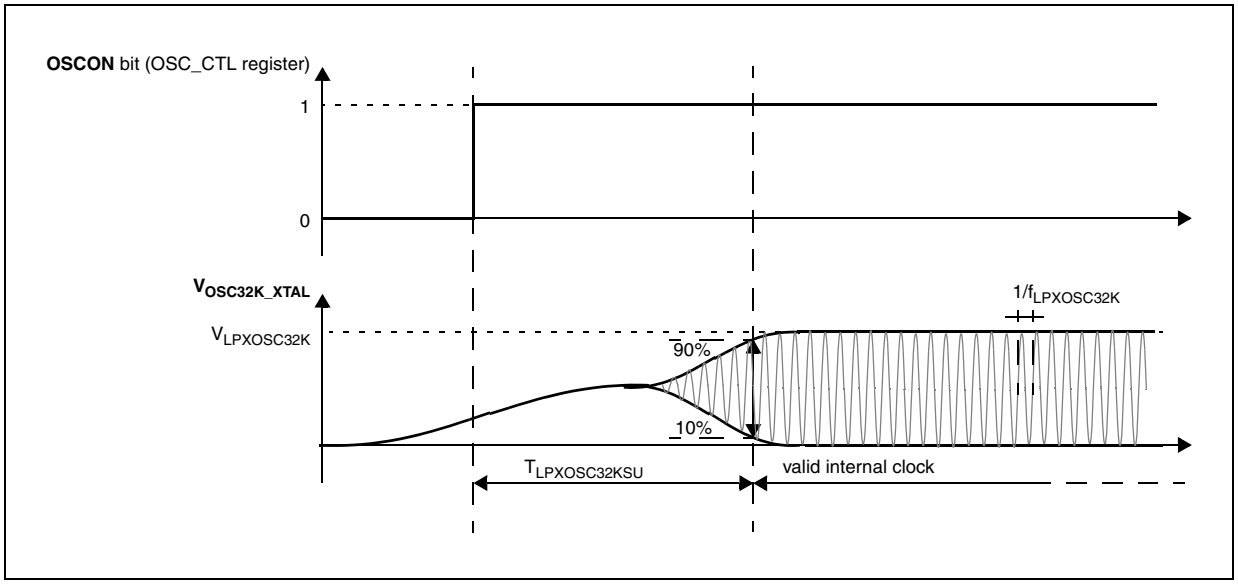


Figure 15. Slow external crystal oscillator (32 kHz) timing diagram

Table 39. Slow external crystal oscillator (32 kHz) electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
f_{SXOSC}	SR	—	Slow external crystal oscillator frequency	—	32	32.768	40	kHz
V_{SXOSC}	CC	T	Oscillation amplitude	—	—	2.1	—	V
$I_{SXOSCBIAS}$	CC	T	Oscillation bias current	—		2.5		µA
I_{SXOSC}	CC	T	Slow external crystal oscillator consumption	—	—	—	8	µA
$t_{SXOSCSU}$	CC	T	Slow external crystal oscillator start-up time	—	—	—	2[2]	s

[1] V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified. Values are specified for no neighbor GPIO pin activity. If oscillator is enabled (OSC32K_XTAL and OSC32K_EXTAL pins), neighboring pins should not toggle.

[2] Start-up time has been measured with EPSON TOYOCOM MC306 crystal. Variation may be seen with other crystal.

4.14 FMPLL electrical characteristics

The device provides a frequency modulated phase locked loop (FMPLL) module to generate a fast system clock from the main oscillator driver.

Table 40. FMPLL electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
f_{PLLIN}	SR	—	FMPLL reference clock[2]	—	4	—	64	MHz

Table 42. Slow internal RC oscillator (128 kHz) electrical characteristics (continued)

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
$\Delta_{SIRCVAR}$	CC	C	Slow internal RC oscillator variation in temperature and supply with respect to f_{SIRC} at T_A = 55 °C in high frequency configuration	High frequency configuration	–10	—	10	%

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 This does not include consumption linked to clock tree toggling and peripherals consumption when RC oscillator is ON.

4.17 ADC electrical characteristics

4.17.1 Introduction

The device provides two Successive Approximation Register (SAR) analog-to-digital converters (10-bit and 12-bit).

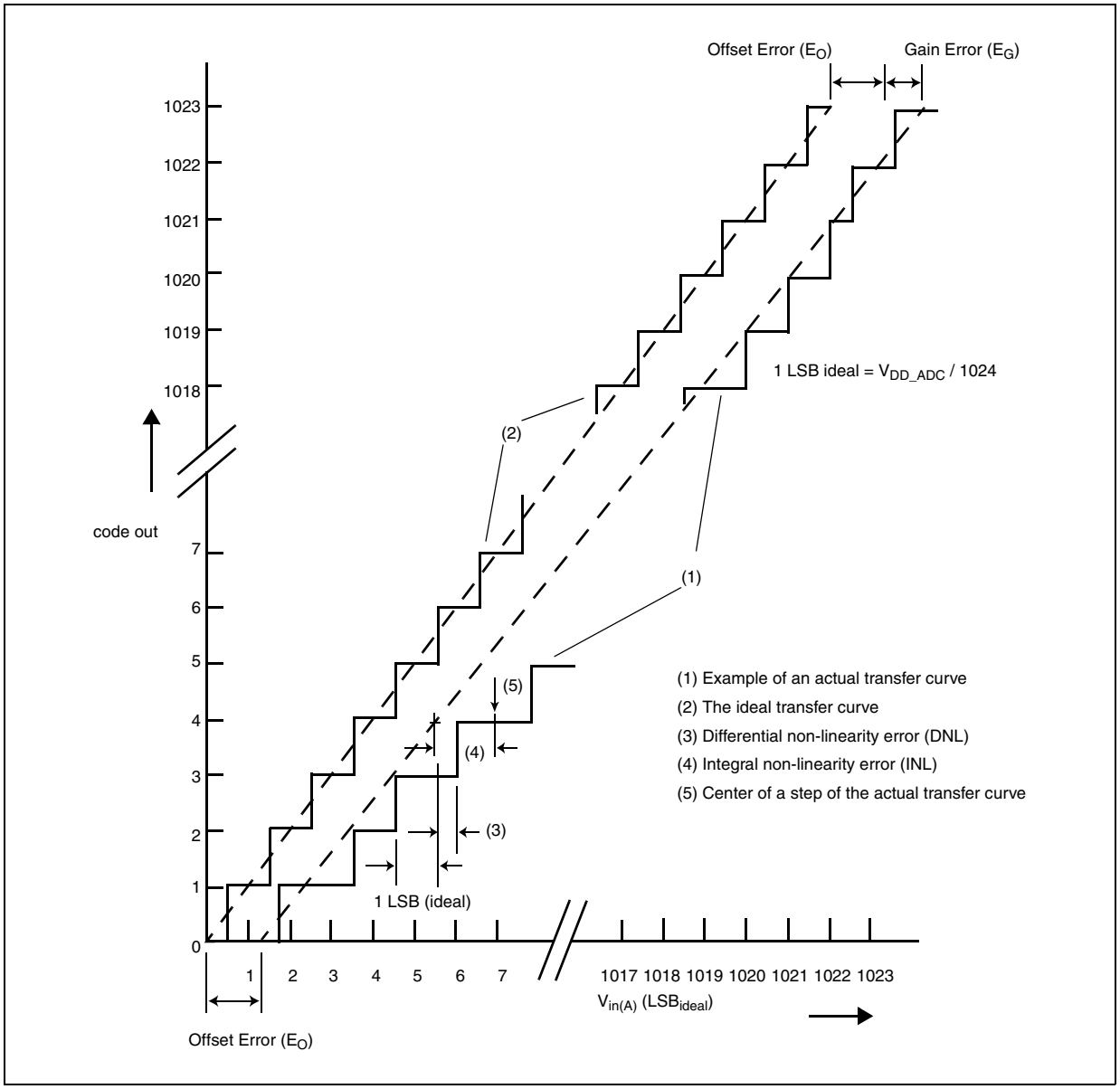


Figure 16. ADC_0 characteristic and error definitions

4.17.2 Input impedance and ADC accuracy

In the following analysis, the input circuit corresponding to the precise channels is considered.

To preserve the accuracy of the A/D converter, it is necessary that analog input pins have low AC impedance. Placing a capacitor with good high frequency characteristics at the input pin of the device can be effective: the capacitor should be as large as possible, ideally infinite. This capacitor contributes to attenuating the noise present on the input pin; furthermore, it sources charge during the sampling phase, when the analog signal source is a high-impedance source.

A real filter can typically be obtained by using a series resistance with a capacitor on the input pin (simple RC filter). The RC filtering may be limited according to the value of source impedance of the transducer or circuit supplying the analog signal to be measured. The filter at the input pins must be designed taking into account the dynamic characteristics of the input signal (bandwidth) and the equivalent input impedance of the ADC itself.

Eqn. 5

$$\tau_1 = (R_{SW} + R_{AD}) \bullet \frac{C_P \bullet C_S}{C_P + C_S}$$

Equation 5 can again be simplified considering only C_S as an additional worst condition. In reality, the transient is faster, but the A/D converter circuitry has been designed to be robust also in the very worst case: the sampling time t_S is always much longer than the internal time constant:

Eqn. 6

$$\tau_1 < (R_{SW} + R_{AD}) \bullet C_S \ll t_s$$

The charge of C_{P1} and C_{P2} is redistributed also on C_S, determining a new value of the voltage V_{A1} on the capacitance according to Equation 7:

Eqn. 7

$$V_{A1} \bullet (C_S + C_{P1} + C_{P2}) = V_A \bullet (C_{P1} + C_{P2})$$

2. A second charge transfer involves also C_F (that is typically bigger than the on-chip capacitance) through the resistance R_L: again considering the worst case in which C_{P2} and C_S were in parallel to C_{P1} (since the time constant in reality would be faster), the time constant is:

Eqn. 8

$$\tau_2 < R_L \bullet (C_S + C_{P1} + C_{P2})$$

In this case, the time constant depends on the external circuit: in particular imposing that the transient is completed well before the end of sampling time t_s, a constraints on R_L sizing is obtained:

Eqn. 9

$$10 \bullet \tau_2 = 10 \bullet R_L \bullet (C_S + C_{P1} + C_{P2}) < t_s$$

Of course, R_L shall be sized also according to the current limitation constraints, in combination with R_S (source impedance) and R_F (filter resistance). Being C_F definitively bigger than C_{P1}, C_{P2} and C_S, then the final voltage V_{A2} (at the end of the charge transfer transient) will be much higher than V_{A1}. Equation 10 must be respected (charge balance assuming now C_S already charged at V_{A1}):

Eqn. 10

$$V_{A2} \bullet (C_S + C_{P1} + C_{P2} + C_F) = V_A \bullet C_F + V_{A1} \bullet (C_{P1} + C_{P2} + C_S)$$

The two transients above are not influenced by the voltage source that, due to the presence of the $R_F C_F$ filter, is not able to provide the extra charge to compensate the voltage drop on C_S with respect to the ideal source V_A; the time constant $R_F C_F$ of the filter is very high with respect to the sampling time (t_s). The filter is typically designed to act as antialiasing.

Table 44. ADC_0 conversion characteristics (10-bit ADC_0) (continued)

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
R_{SW1}	CC	D	Internal resistance of analog source	—		—	—	3	kΩ
R_{SW2}	CC	D	Internal resistance of analog source	—		—	—	2	kΩ
R_{AD}	CC	D	Internal resistance of analog source	—		—	—	2	kΩ
I_{INJ}	SR	—	Input current Injection	Current injection on one ADC_0 input, different from the converted one	V_{DD} = 3.3 V ± 10%	–5	—	5	mA
					V_{DD} = 5.0 V ± 10%	–5	—	5	
| INL |	CC	T	Absolute integral nonlinearity	No overload		—	0.5	1.5	LSB
| DNL |	CC	T	Absolute differential nonlinearity	No overload		—	0.5	1.0	LSB
| E_O |	CC	T	Absolute offset error	—		—	0.5	—	LSB
| E_G |	CC	T	Absolute gain error	—		—	0.6	—	LSB
TUEP	CC	P	Total unadjusted error[7] for precise channels, input only pins	Without current injection		–2	0.6	2	LSB
		T		With current injection		–3	—	3	
TUEX	CC	T	Total unadjusted error[7] for extended channel	Without current injection		–3	1	3	LSB
		T		With current injection		–4		4	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified.

2 Analog and digital V_{SS} **must** be common (to be tied together externally).

3 V_{AINx} may exceed V_{SS_ADC0} and V_{DD_ADC0} limits, remaining on absolute maximum ratings, but the results of the conversion will be clamped respectively to 0x000 or 0x3FF.

4 Duty cycle is ensured by using system clock without prescaling. When ADCLKSEL = 0, the duty cycle is ensured by internal divider by 2.

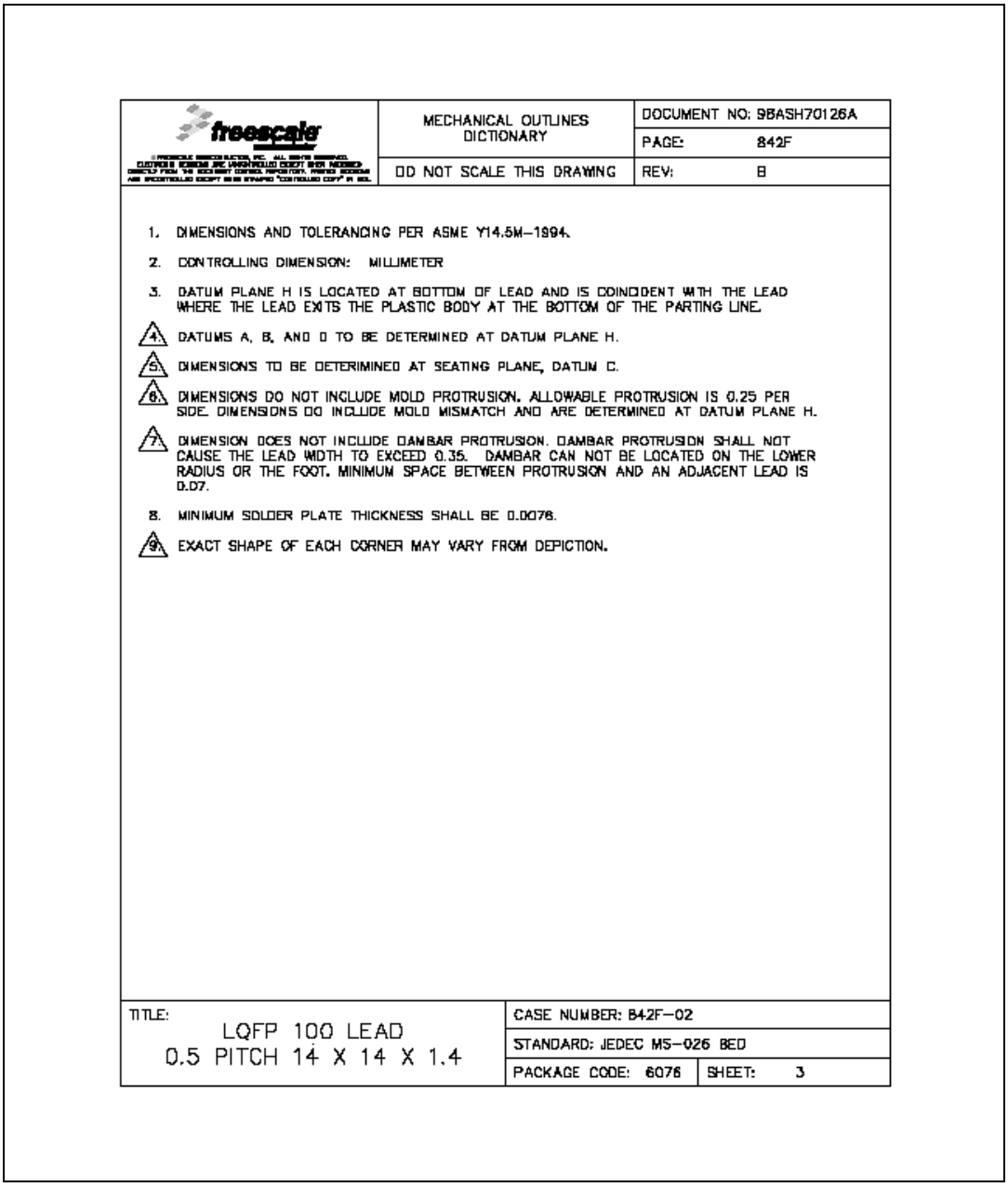
5 During the sampling time the input capacitance C_S can be charged/discharged by the external source. The internal resistance of the analog source must allow the capacitance to reach its final voltage level within t_{ADC0_S}. After the end of the sampling time t_{ADC0_S}, changes of the analog input voltage have no effect on the conversion result. Values for the sampling clock t_{ADC0_S} depend on programming.

6 This parameter does not include the sampling time t_{ADC0_S}, but only the time for determining the digital result and the time to load the result's register with the conversion result.

7 Total Unadjusted Error: The maximum error that occurs without adjusting Offset and Gain errors. This error is a combination of Offset, Gain and Integral Linearity errors.

Table 45. ADC_1 conversion characteristics (12-bit ADC_1) (continued)

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
V_{AINx}	SR	—	Analog input voltage[3]	—	V_{SS_ADC1} – 0.1	—	V_{DD_ADC1} + 0.1	V
$I_{ADC1pwd}$	SR	—	ADC_1 consumption in power down mode	—	—	—	50	µA
$I_{ADC1run}$	SR	—	ADC_1 consumption in running mode	—	—	—	6	mA
f_{ADC1}	SR	—	ADC_1 analog frequency	V_{DD} = 3.3 V	3.33	—	20 + 4%	MHz
				V_{DD} = 5 V	3.33	—	32 + 4%	
t_{ADC1_PU}	SR	—	ADC_1 power up delay	—	—	—	1.5	µs
t_{ADC1_S}	CC	T	Sampling time[4] V_{DD} = 3.3 V	f_{ADC1} = 20 MHz, INPSAMP = 12	600	—	—	ns
			Samplingtime[4] V_{DD} = 5.0 V	f_{ADC1} = 32 MHz, INPSAMP = 17	500	—	—	
			Sampling time[4] V_{DD} = 3.3 V	f_{ADC1} = 3.33 MHz, INPSAMP = 255	—	—	76.2	µs
			Sampling time[4] V_{DD} = 5.0 V	f_{ADC1} = 3.33 MHz, INPSAMP = 255	—	—	76.2	
t_{ADC1_C}	CC	P	Conversion time[5] V_{DD} = 3 .3 V	f_{ADC1} = 20 MHz, INPCMP = 0	2.4	—	—	µs
			Conversion time[5] V_{DD} = 5.0 V	$f_{ADC\ 1}$ = 32 MHz, INPCMP = 0	1.5	—	—	µs
			Conversion time[5] V_{DD} = 3.3 V	$f_{ADC\ 1}$ = 13.33 MHz, INPCMP = 0	—	—	3.6	µs
			Conversion time[5] V_{DD} = 5.0 V	f_{ADC1} = 13.33 MHz, INPCMP = 0	—	—	3.6	µs
Δ_{ADC1_SYS}	SR	—	ADC_1 digital clock duty cycle	ADCLKSEL = 1[6]	45	—	55	%
C_S	CC	D	ADC_1 input sampling capacitance	—	—	—	5	pF
C_{P1}	CC	D	ADC_1 input pin capacitance 1	—	—	—	3	pF
C_{P2}	CC	D	ADC_1 input pin capacitance 2	—	—	—	1	pF
C_{P3}	CC	D	ADC_1 input pin capacitance 3	—	—	—	1.5	pF
R_{SW1}	CC	D	Internal resistance of analog source	—	—	—	1	kΩ
R_{SW2}	CC	D	Internal resistance of analog source	—	—	—	2	kΩ
R_{AD}	CC	D	Internal resistance of analog source	—	—	—	0.3	kΩ



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1. DIMENSIONS AND TOLERANCING PER ASME Y14.5M–1994.
2. CONTROLLING DIMENSION: MILLIMETER
3. DATUM PLANE H IS LOCATED AT BOTTOM OF LEAD AND IS COINCIDENT WITH THE LEAD WHERE THE LEAD EXITS THE PLASTIC BODY AT THE BOTTOM OF THE PARTING LINE.
4. DATUMS A, B, AND D TO BE DETERMINED AT DATUM PLANE H.
5. DIMENSIONS TO BE DETERIMINED AT SEATING PLANE, DATUM C.
6. DIMENSIONS DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS 0.25 PER SIDE. DIMENSIONS DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT DATUM PLANE H.
7. DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION. DAMBAR PROTRUSION SHALL NOT CAUSE THE LEAD WIDTH TO EXCEED 0.35. DAMBAR CAN NOT BE LOCATED ON THE LOWER RADIUS OR THE FOOT. MINIMUM SPACE BETWEEN PROTRUSION AND AN ADJACENT LEAD IS 0.07.
8. MINIMUM SOLDER PLATE THICKNESS SHALL BE 0.0076.
9. EXACT SHAPE OF EACH CORNER MAY VARY FROM DEPICTION.

TITLE: LQFP 100 LEAD 0.5 PITCH 14 X 14 X 1.4	CASE NUMBER: 842F–02	
	STANDARD: JEDEC MS–026 BED	
	PACKAGE CODE: 6076	SHEET: 3

Figure 40. 100 LQFP package mechanical drawing (Part 3 of 3)