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#### Details

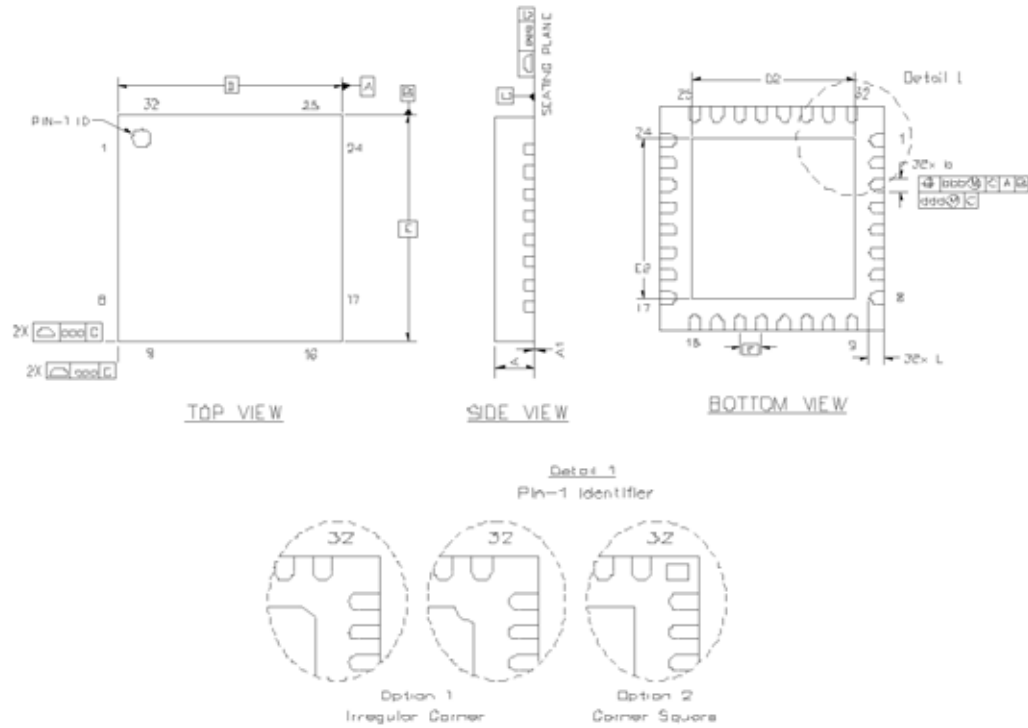
Product Status	Discontinued at Digi-Key
Core Processor	8051
Core Size	8-Bit
Speed	48 MIPS
Connectivity	I <sup>2</sup> C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, Temp Sensor, WDT
Number of I/O	25
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	2.25K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.25V
Data Converters	A/D 21x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	32-VFQFN Exposed Pad
Supplier Device Package	-
Purchase URL	<a href="https://www.e-xfl.com/product-detail/silicon-labs/c8051f38b-gmr">https://www.e-xfl.com/product-detail/silicon-labs/c8051f38b-gmr</a>

# C8051F388/9/A/B

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23.1. Baud Rate Generator .....	202
23.2. Data Format.....	203
23.3. Configuration and Operation .....	204
23.3.1. Data Transmission .....	204
23.3.2. Data Reception .....	204
23.3.3. Multiprocessor Communications .....	205
<b>24. Enhanced Serial Peripheral Interface (SPI0) .....</b>	<b>211</b>
24.1. Signal Descriptions.....	212
24.1.1. Master Out, Slave In (MOSI).....	212
24.1.2. Master In, Slave Out (MISO).....	212
24.1.3. Serial Clock (SCK) .....	212
24.1.4. Slave Select (NSS) .....	212
24.2. SPI0 Master Mode Operation .....	212
24.3. SPI0 Slave Mode Operation .....	214
24.4. SPI0 Interrupt Sources .....	215
24.5. Serial Clock Phase and Polarity .....	215
24.6. SPI Special Function Registers .....	217
<b>25. Timers .....</b>	<b>224</b>
25.1. Timer 0 and Timer 1 .....	227
25.1.1. Mode 0: 13-bit Counter/Timer .....	227
25.1.2. Mode 1: 16-bit Counter/Timer .....	228
25.1.3. Mode 2: 8-bit Counter/Timer with Auto-Reload.....	228
25.1.4. Mode 3: Two 8-bit Counter/Timers (Timer 0 Only).....	229
25.2. Timer 2 .....	235
25.2.1. 16-bit Timer with Auto-Reload.....	235
25.2.2. 8-bit Timers with Auto-Reload.....	236
25.2.3. Timer 2 Capture Modes: LFO Falling Edge .....	236
25.3. Timer 3 .....	242
25.3.1. 16-bit Timer with Auto-Reload.....	242
25.3.2. 8-bit Timers with Auto-Reload.....	243
25.3.3. Timer 3 Capture Modes: LFO Falling Edge .....	243
25.4. Timer 4 .....	249
25.4.1. 16-bit Timer with Auto-Reload.....	249
25.4.2. 8-bit Timers with Auto-Reload.....	250
25.5. Timer 5 .....	254
25.5.1. 16-bit Timer with Auto-Reload.....	254
25.5.2. 8-bit Timers with Auto-Reload.....	255
<b>26. Programmable Counter Array.....</b>	<b>259</b>
26.1. PCA Counter/Timer .....	260
26.2. PCA0 Interrupt Sources.....	261
26.3. Capture/Compare Modules .....	262
26.3.1. Edge-triggered Capture Mode.....	263
26.3.2. Software Timer (Compare) Mode.....	264
26.3.3. High-Speed Output Mode .....	265
26.3.4. Frequency Output Mode .....	266

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**Figure 3.8. QFN-32 Package Drawing**

**Table 3.6. QFN-32 Package Dimensions**

Dimension	Min	Typ	Max	Dimension	Min	Typ	Max
A	0.80	0.85	0.90	E2	3.20	3.30	3.40
A1	0.00	0.02	0.05	L	0.35	0.40	0.45
b	0.18	0.25	0.30	aaa	—	—	0.10
D	5.00 BSC			bbb	—	—	0.10
D2	3.20	3.30	3.40	ddd	—	—	0.05
e	0.50 BSC			eee	—	—	0.08
E	5.00 BSC						

**Notes:**

1. All dimensions shown are in millimeters (mm) unless otherwise noted.
2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.
3. This drawing conforms to the JEDEC Solid State Outline MO-220, variation VHHD except for custom features D2, E2, and L which are toleranced per supplier designation.
4. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

# C8051F388/9/A/B

**Table 5.3. Port I/O DC Electrical Characteristics**

$V_{DD} = 2.7$  to  $3.6$  V,  $-40$  to  $+85$  °C unless otherwise specified.

Parameter	Test Condition	Min	Typ	Max	Unit
Output High Voltage	$I_{OH} = -3$ mA, Port I/O push-pull $I_{OH} = -10$ $\mu$ A, Port I/O push-pull $I_{OH} = -10$ mA, Port I/O push-pull	$V_{DD} - 0.7$ $V_{DD} - 0.1$ —	— — $V_{DD} - 0.8$	— — —	V
Output Low Voltage	$I_{OL} = 8.5$ mA $I_{OL} = 10$ $\mu$ A $I_{OL} = 25$ mA	— — —	— — 1.0	0.6 0.1 —	V
Input High Voltage		2.0	—	—	V
Input Low Voltage		—	—	0.8	V
Input Leakage Current	Weak Pullup Off Weak Pullup On, $V_{IN} = 0$ V	— —	— 15	$\pm 1$ 50	$\mu$ A
INT2 Detection Input Low Voltage		—	—	1.0	V
INT2 Detection Input High Voltage		3.0	—	—	V

**Table 5.4. Reset Electrical Characteristics**

$-40$  to  $+85$  °C unless otherwise specified.

Parameter	Test Condition	Min	Typ	Max	Unit
$\overline{RST}$ Output Low Voltage	$I_{OL} = 8.5$ mA, $V_{DD} = 2.7$ V to $3.6$ V	—	—	0.6	V
$\overline{RST}$ Input High Voltage		$0.7 \times V_{DD}$	—	—	V
$\overline{RST}$ Input Low Voltage		—	—	$0.3 \times V_{DD}$	V
$\overline{RST}$ Input Pullup Current	$\overline{RST} = 0.0$ V	—	15	40	$\mu$ A
$V_{DD}$ Monitor Threshold ( $V_{RST}$ )		2.60	2.65	2.70	V
Missing Clock Detector Time-out	Time from last system clock rising edge to reset initiation	80	580	800	$\mu$ s
Reset Time Delay	Delay between release of any reset source and code execution at location 0x0000	—	—	250	$\mu$ s
Minimum $\overline{RST}$ Low Time to Generate a System Reset		15	—	—	$\mu$ s
$V_{DD}$ Monitor Turn-on Time		—	—	100	$\mu$ s
$V_{DD}$ Monitor Supply Current		—	15	50	$\mu$ A

# C8051F388/9/A/B

**Table 5.11. Temperature Sensor Electrical Characteristics**

$V_{DD} = 3.0\text{ V}$ ,  $-40$  to  $+85\text{ }^{\circ}\text{C}$  unless otherwise specified.

Parameter	Test Condition	Min	Typ	Max	Unit
Linearity		—	$\pm 0.5$	—	$^{\circ}\text{C}$
Slope		—	2.87	—	$\text{mV}/^{\circ}\text{C}$
Slope Error*		—	$\pm 120$	—	$\mu\text{V}/^{\circ}\text{C}$
Offset	Temp = $0\text{ }^{\circ}\text{C}$	—	764	—	mV
Offset Error*	Temp = $0\text{ }^{\circ}\text{C}$	—	$\pm 15$	—	mV
<b>Note:</b> Represents one standard deviation from the mean.					

**Table 5.12. Voltage Reference Electrical Characteristics**

$V_{DD} = 3.0\text{ V}$ ;  $-40$  to  $+85\text{ }^{\circ}\text{C}$  unless otherwise specified.

Parameter	Test Condition	Min	Typ	Max	Unit
<b>Internal Reference (REFBE = 1)</b>					
Output Voltage	$25\text{ }^{\circ}\text{C}$ ambient	2.38	2.42	2.46	V
VREF Short-Circuit Current		—	—	8	mA
VREF Temperature Coefficient		—	35	—	$\text{ppm}/^{\circ}\text{C}$
Load Regulation	Load = 0 to $200\text{ }\mu\text{A}$ to GND	—	1.5	—	$\text{ppm}/\mu\text{A}$
VREF Turn-on Time 1	$4.7\text{ }\mu\text{F}$ tantalum, $0.1\text{ }\mu\text{F}$ ceramic bypass	—	3	—	ms
VREF Turn-on Time 2	$0.1\text{ }\mu\text{F}$ ceramic bypass	—	100	—	$\mu\text{s}$
Power Supply Rejection		—	140	—	$\text{ppm}/\text{V}$
<b>External Reference (REFBE = 0)</b>					
Input Voltage Range		1	—	$V_{DD}$	V
Input Current	Sample Rate = 500 ksp/s; VREF = $3.0\text{ V}$	—	9	—	$\mu\text{A}$
<b>Power Specifications</b>					
Supply Current		—	75	—	$\mu\text{A}$

**Table 5.13. Comparator Electrical Characteristics**

$V_{DD} = 3.0\text{ V}$ ,  $-40$  to  $+85\text{ }^{\circ}\text{C}$  unless otherwise noted.

Parameter	Test Condition	Min	Typ	Max	Unit
Response Time: Mode 0, $V_{cm}^* = 1.5\text{ V}$	$CP0+ - CP0- = 100\text{ mV}$	—	100	—	ns
	$CP0+ - CP0- = -100\text{ mV}$	—	250	—	ns
Response Time: Mode 1, $V_{cm}^* = 1.5\text{ V}$	$CP0+ - CP0- = 100\text{ mV}$	—	175	—	ns
	$CP0+ - CP0- = -100\text{ mV}$	—	500	—	ns
Response Time: Mode 2, $V_{cm}^* = 1.5\text{ V}$	$CP0+ - CP0- = 100\text{ mV}$	—	320	—	ns
	$CP0+ - CP0- = -100\text{ mV}$	—	1100	—	ns
Response Time: Mode 3, $V_{cm}^* = 1.5\text{ V}$	$CP0+ - CP0- = 100\text{ mV}$	—	1050	—	ns
	$CP0+ - CP0- = -100\text{ mV}$	—	5200	—	ns
Common-Mode Rejection Ratio		—	1.5	4	mV/V
Positive Hysteresis 1	$CP0HYP1-0 = 00$	—	0	1	mV
Positive Hysteresis 2	$CP0HYP1-0 = 01$	2	5	10	mV
Positive Hysteresis 3	$CP0HYP1-0 = 10$	7	10	20	mV
Positive Hysteresis 4	$CP0HYP1-0 = 11$	15	20	30	mV
Negative Hysteresis 1	$CP0HYN1-0 = 00$	—	0	1	mV
Negative Hysteresis 2	$CP0HYN1-0 = 01$	2	5	10	mV
Negative Hysteresis 3	$CP0HYN1-0 = 10$	7	10	20	mV
Negative Hysteresis 4	$CP0HYN1-0 = 11$	15	20	30	mV
Inverting or Non-Inverting Input Voltage Range		-0.25	—	$V_{DD} + 0.25$	V
Input Capacitance		—	4	—	pF
Input Bias Current		—	0.001	—	nA
Input Offset Voltage		-10	—	+10	mV
<b>Power Supply</b>					
Power Supply Rejection		—	0.1	—	mV/V
Power-up Time		—	10	—	$\mu\text{s}$
Supply Current at DC	Mode 0	—	20	—	$\mu\text{A}$
	Mode 1	—	10	—	$\mu\text{A}$
	Mode 2	—	4	—	$\mu\text{A}$
	Mode 3	—	1	—	$\mu\text{A}$
<b>Note:</b> $V_{cm}$ is the common-mode voltage on $CP0+$ and $CP0-$ .					

## SFR Definition 6.7. ADC0LTH: ADC0 Less-Than Data High Byte

Bit	7	6	5	4	3	2	1	0
Name	ADC0LTH[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xC6; SFR Page = All Pages

Bit	Name	Function
7:0	ADC0LTH[7:0]	ADC0 Less-Than Data Word High-Order Bits.

## SFR Definition 6.8. ADC0LTL: ADC0 Less-Than Data Low Byte

Bit	7	6	5	4	3	2	1	0
Name	ADC0LTL[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xC5; SFR Page = All Pages

Bit	Name	Function
7:0	ADC0LTL[7:0]	ADC0 Less-Than Data Word Low-Order Bits.

# C8051F388/9/A/B

## SFR Definition 8.2. CPT0MD: Comparator0 Mode Selection

Bit	7	6	5	4	3	2	1	0
Name			CP0RIE	CP0FIE			CP0MD[1:0]	
Type	R	R	R/W	R/W	R	R	R/W	
Reset	0	0	0	0	0	0	1	0

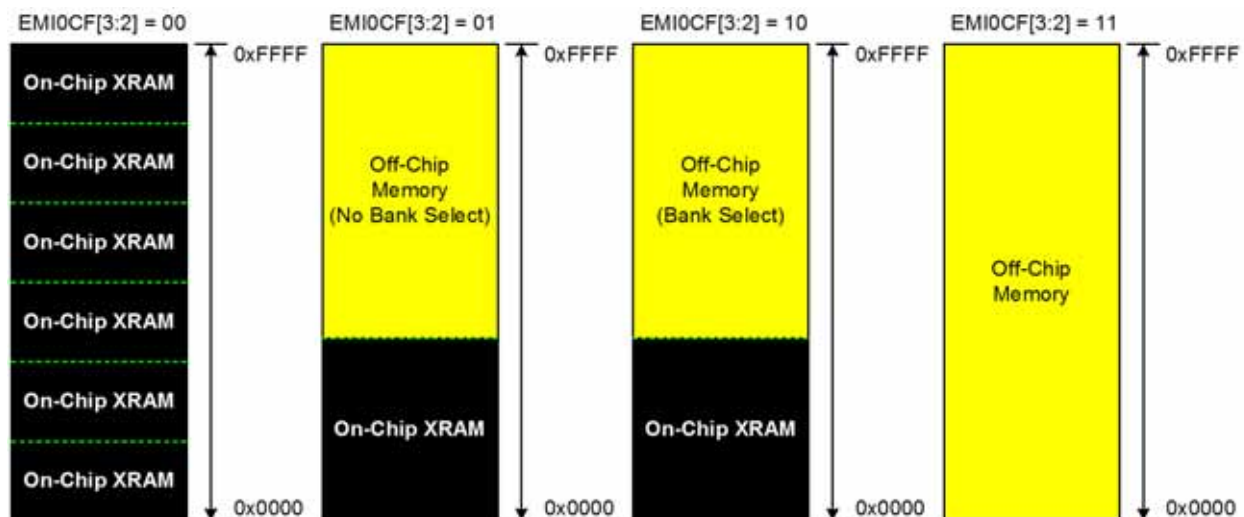
SFR Address = 0x9D; SFR Page = All Pages

Bit	Name	Function
7:6	Unused	Read = 00b, Write = don't care.
5	CP0RIE	<b>Comparator0 Rising-Edge Interrupt Enable.</b> 0: Comparator0 Rising-edge interrupt disabled. 1: Comparator0 Rising-edge interrupt enabled.
4	CP0FIE	<b>Comparator0 Falling-Edge Interrupt Enable.</b> 0: Comparator0 Falling-edge interrupt disabled. 1: Comparator0 Falling-edge interrupt enabled.
3:2	Unused	Read = 00b, Write = don't care.
1:0	CP0MD[1:0]	<b>Comparator0 Mode Select.</b> These bits affect the response time and power consumption for Comparator0. 00: Mode 0 (Fastest Response Time, Highest Power Consumption) 01: Mode 1 10: Mode 2 11: Mode 3 (Slowest Response Time, Lowest Power Consumption)



## 14.5. Memory Mode Selection

The external data memory space can be configured in one of four modes, shown in Figure 14.3, based on the EMIF Mode bits in the EMI0CF register (SFR Definition 14.4). These modes are summarized below. More information about the different modes can be found in Section “14.6. Timing” on page 96.



**Figure 14.3. EMIF Operating Modes**

### 14.5.1. Internal XRAM Only

When EMI0CF.[3:2] are set to 00, all MOVX instructions will target the internal XRAM space on the device. Memory accesses to addresses beyond the populated space will wrap on 2k or 4k boundaries (depending on the RAM available on the device). As an example, the addresses 0x1000 and 0x2000 both evaluate to address 0x0000 in on-chip XRAM space.

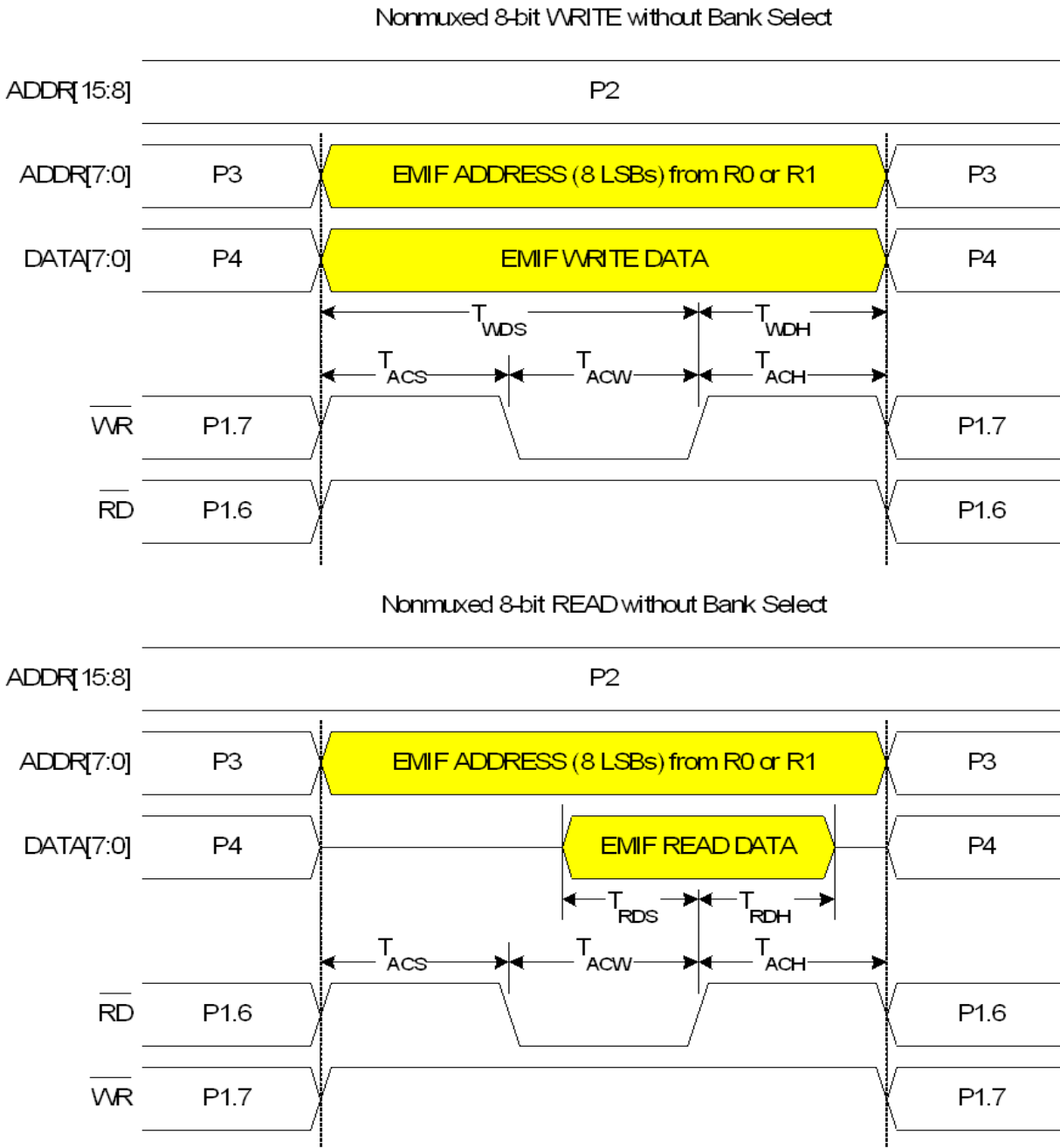
- 8-bit MOVX operations use the contents of EMI0CN to determine the high-byte of the effective address and R0 or R1 to determine the low-byte of the effective address.
- 16-bit MOVX operations use the contents of the 16-bit DPTR to determine the effective address.

### 14.5.2. Split Mode without Bank Select

When EMI0CF.[3:2] are set to 01, the XRAM memory map is split into two areas, on-chip space and off-chip space.

- Effective addresses below the internal XRAM size boundary will access on-chip XRAM space.
- Effective addresses above the internal XRAM size boundary will access off-chip space.
- 8-bit MOVX operations use the contents of EMI0CN to determine whether the memory access is on-chip or off-chip. However, in the “No Bank Select” mode, an 8-bit MOVX operation will not drive the upper 8-bits A[15:8] of the Address Bus during an off-chip access. This allows the user to manipulate the upper address bits at will by setting the Port state directly via the port latches. This behavior is in contrast with “Split Mode with Bank Select” described below. The lower 8-bits of the Address Bus A[7:0] are driven, determined by R0 or R1.
- 16-bit MOVX operations use the contents of DPTR to determine whether the memory access is on-chip or off-chip, and unlike 8-bit MOVX operations, the full 16-bits of the Address Bus A[15:0] are driven during the off-chip transaction.

## 14.6.1.2. 8-bit MOVX without Bank Select: EMI0CF[4:2] = 101 or 111



**Figure 14.5. Non-Multiplexed 8-bit MOVX without Bank Select Timing**

**Table 15.2. Special Function Registers (Continued)**

SFRs are listed in alphabetical order. All undefined SFR locations are reserved

Register	Address	Page	Description	Page
<b>TMR5RLL</b>	0xCA	F	Timer/Counter 5 Reload Low	257
<b>VDM0CN</b>	0xFF	All Pages	V <sub>DD</sub> Monitor Control	126
<b>XBR0</b>	0xE1	All Pages	Port I/O Crossbar Control 0	153
<b>XBR1</b>	0xE2	All Pages	Port I/O Crossbar Control 1	154
<b>XBR2</b>	0xE3	All Pages	Port I/O Crossbar Control 2	155

## 19.5.2. External RC Example

If an RC network is used as an external oscillator source for the MCU, the circuit should be configured as shown in Figure 19.1, “RC Mode”. The capacitor should be no greater than 100 pF; however, for very small capacitors, the total capacitance may be dominated by parasitic capacitance in the PCB layout. To determine the required External Oscillator Frequency Control value (XFCN) in the OSCXCN Register, first select the RC network value to produce the desired frequency of oscillation, according to Equation 19.1, where  $f$  = the frequency of oscillation in MHz,  $C$  = the capacitor value in pF, and  $R$  = the pull-up resistor value in k $\Omega$ .

$$f = 1.23 \times 10^3 \sqrt{R \times C}$$

### Equation 19.1. RC Mode Oscillator Frequency

For example: If the frequency desired is 100 kHz, let  $R = 246 \text{ k}\Omega$  and  $C = 50 \text{ pF}$ :

$$f = 1.23(10^3) / \sqrt{RC} = 1.23(10^3) / \sqrt{246 \times 50} = 0.1 \text{ MHz} = 100 \text{ kHz}$$

Referring to the table in SFR Definition 19.6, the required XFCN setting is 010b.

## 19.5.3. External Capacitor Example

If a capacitor is used as an external oscillator for the MCU, the circuit should be configured as shown in Figure 19.1, “C Mode”. The capacitor should be no greater than 100 pF; however, for very small capacitors, the total capacitance may be dominated by parasitic capacitance in the PCB layout. To determine the required External Oscillator Frequency Control value (XFCN) in the OSCXCN Register, select the capacitor to be used and find the frequency of oscillation according to Equation 19.2, where  $f$  = the frequency of oscillation in MHz,  $C$  = the capacitor value in pF, and  $V_{DD}$  = the MCU power supply in Volts.

$$f = (KF) / (C \times V_{DD})$$

### Equation 19.2. C Mode Oscillator Frequency

For example: Assume  $V_{DD} = 3.0 \text{ V}$  and  $f = 150 \text{ kHz}$ :

$$f = KF / (C \times V_{DD})$$

$$0.150 \text{ MHz} = KF / (C \times 3.0)$$

Since the frequency of roughly 150 kHz is desired, select the K Factor from the table in SFR Definition 19.6 (OSCXCN) as  $KF = 22$ :

$$0.150 \text{ MHz} = 22 / (C \times 3.0)$$

$$C \times 3.0 = 22 / 0.150 \text{ MHz}$$

$$C = 146.6 / 3.0 \text{ pF} = 48.8 \text{ pF}$$

Therefore, the XFCN value to use in this example is 011b and  $C = 50 \text{ pF}$ .

**SFR Definition 20.18. P3MDOUT: Port 3 Output Mode**

Bit	7	6	5	4	3	2	1	0
Name	P3MDOUT[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xA7; SFR Page = All Pages

Bit	Name	Function
7:0	P3MDOUT[7:0]	<b>Output Configuration Bits for P3.7–P3.0 (respectively).</b> These bits are ignored if the corresponding bit in register P3MDIN is logic 0. 0: Corresponding P3.n Output is open-drain. 1: Corresponding P3.n Output is push-pull.

**SFR Definition 20.19. P3SKIP: Port 3 Skip**

Bit	7	6	5	4	3	2	1	0
Name	P3SKIP[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xDF; SFR Page = All Pages

Bit	Name	Function
7:0	P3SKIP[3:0]	<b>Port 3 Crossbar Skip Enable Bits.</b> These bits select Port 3 pins to be skipped by the Crossbar Decoder. Port pins used for analog, special functions or GPIO should be skipped by the Crossbar. 0: Corresponding P3.n pin is not skipped by the Crossbar. 1: Corresponding P3.n pin is skipped by the Crossbar.

## SFR Definition 21.7. SMB0ADM: SMBus0 Slave Address Mask

Bit	7	6	5	4	3	2	1	0
Name	SLVM0[6:0]							EHACK0
Type	R/W							R/W
Reset	1	1	1	1	1	1	1	0

SFR Address = 0xCE; SFR Page = 0

Bit	Name	Function
7:1	SLVM0[6:0]	<b>SMBus0 Slave Address Mask.</b> Defines which bits of register SMB0ADR are compared with an incoming address byte, and which bits are ignored. Any bit set to 1 in SLVM0[6:0] enables comparisons with the corresponding bit in SLV0[6:0]. Bits set to 0 are ignored (can be either 0 or 1 in the incoming address).
0	EHACK0	<b>Hardware Acknowledge Enable.</b> Enables hardware acknowledgement of slave address and received data bytes. 0: Firmware must manually acknowledge all incoming address and data bytes. 1: Automatic Slave Address Recognition and Hardware Acknowledge is Enabled.

## SFR Definition 21.8. SMB1ADR: SMBus1 Slave Address

Bit	7	6	5	4	3	2	1	0
Name	SLV1[6:0]							GC1
Type	R/W							R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xCF; SFR Page = F

Bit	Name	Function
7:1	SLV1[6:0]	<b>SMBus1 Hardware Slave Address.</b> Defines the SMBus1 Slave Address(es) for automatic hardware acknowledgment. Only address bits which have a 1 in the corresponding bit position in SLVM1[6:0] are checked against the incoming address. This allows multiple addresses to be recognized.
0	GC1	<b>General Call Address Enable.</b> When hardware address recognition is enabled (EHACK1 = 1), this bit will determine whether the General Call Address (0x00) is also recognized by hardware. 0: General Call Address is ignored. 1: General Call Address is recognized.

**Table 21.5. SMBus Status Decoding: Hardware ACK Disabled (EHACK = 0) (Continued)**

Mode	Values Read				Current SMBus State	Typical Response Options	Values to Write			Next Status Vector Expected
	Status Vector	ACKRQ	ARBLOST	ACK			STA	STO	ACK	
Slave Transmitter	0100	0	0	0	A slave byte was transmitted; NACK received.	No action required (expecting STOP condition).	0	0	X	0001
		0	0	1	A slave byte was transmitted; ACK received.	Load SMB0DAT with next data byte to transmit.	0	0	X	0100
		0	1	X	A Slave byte was transmitted; error detected.	No action required (expecting Master to end transfer).	0	0	X	0001
	0101	0	X	X	An illegal STOP or bus error was detected while a Slave Transmission was in progress.	Clear STO.	0	0	X	—
Slave Receiver	0010	1	0	X	A slave address + R/W was received; ACK requested.	If Write, Acknowledge received address	0	0	1	0000
						If Read, Load SMB0DAT with data byte; ACK received address	0	0	1	0100
						NACK received address.	0	0	0	—
		1	1	X	Lost arbitration as master; slave address + R/W received; ACK requested.	If Write, Acknowledge received address	0	0	1	0000
						If Read, Load SMB0DAT with data byte; ACK received address	0	0	1	0100
						NACK received address.	0	0	0	—
	0001	0	0	X	A STOP was detected while addressed as a Slave Transmitter or Slave Receiver.	Reschedule failed transfer; NACK received address.	1	0	0	1110
						Clear STO.	0	0	X	—
						No action required (transfer complete/aborted).	0	0	0	—
	0000	1	0	X	A slave byte was received; ACK requested.	Acknowledge received byte; Read SMB0DAT.	0	0	1	0000
						NACK received byte.	0	0	0	—

# C8051F388/9/A/B

**Table 22.1. Timer Settings for Standard Baud Rates Using Internal Oscillator**

Target Baud Rate (bps)	Actual Baud Rate (bps)	Baud Rate Error	Oscillator Divide Factor	Timer Clock Source	SCA1-SCA0 (pre-scale select*)	T1M	Timer 1 Reload Value (hex)
<b>Note:</b> SCA1-SCA0 and T1M define the Timer Clock Source. X = Don't care							



1 at the end of the transfer. If interrupts are enabled, an interrupt request is generated when the SPIF flag is set. While the SPI0 master transfers data to a slave on the MOSI line, the addressed SPI slave device simultaneously transfers the contents of its shift register to the SPI master on the MISO line in a full-duplex operation. Therefore, the SPIF flag serves as both a transmit-complete and receive-data-ready flag. The data byte received from the slave is transferred MSB-first into the master's shift register. When a byte is fully shifted into the register, it is moved to the receive buffer where it can be read by the processor by reading SPI0DAT.

When configured as a master, SPI0 can operate in one of three different modes: multi-master mode, 3-wire single-master mode, and 4-wire single-master mode. The default, multi-master mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 1. In this mode, NSS is an input to the device, and is used to disable the master SPI0 when another master is accessing the bus. When NSS is pulled low in this mode, MSTEN (SPI0CN.6) and SPIEN (SPI0CN.0) are set to 0 to disable the SPI master device, and a Mode Fault is generated (MODF, SPI0CN.5 = 1). Mode Fault will generate an interrupt if enabled. SPI0 must be manually re-enabled in software under these circumstances. In multi-master systems, devices will typically default to being slave devices while they are not acting as the system master device. In multi-master mode, slave devices can be addressed individually (if needed) using general-purpose I/O pins. Figure 24.2 shows a connection diagram between two master devices in multiple-master mode.

3-wire single-master mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 0. In this mode, NSS is not used, and is not mapped to an external port pin through the crossbar. Any slave devices that must be addressed in this mode should be selected using general-purpose I/O pins. Figure 24.3 shows a connection diagram between a master device in 3-wire master mode and a slave device.

4-wire single-master mode is active when NSSMD1 (SPI0CN.3) = 1. In this mode, NSS is configured as an output pin, and can be used as a slave-select signal for a single SPI device. In this mode, the output value of NSS is controlled (in software) with the bit NSSMD0 (SPI0CN.2). Additional slave devices can be addressed using general-purpose I/O pins. Figure 24.4 shows a connection diagram for a master device in 4-wire master mode and two slave devices.

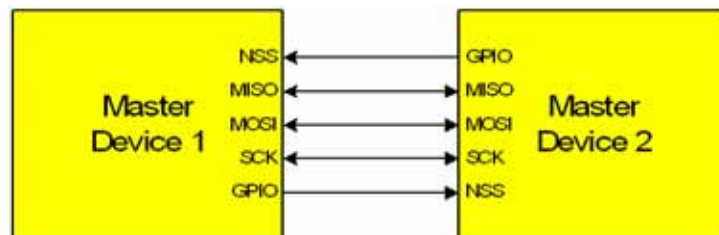


Figure 24.2. Multiple-Master Mode Connection Diagram



Figure 24.3. 3-Wire Single Master and 3-Wire Single Slave Mode Connection Diagram

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## SFR Definition 25.13. TMR2H Timer 2 High Byte

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Bit	7	6	5	4	3	2	1	0
Name	TMR2H[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xCD; SFR Page = 0

Bit	Name	Function
7:0	TMR2H[7:0]	<b>Timer 2 Low Byte.</b> In 16-bit mode, the TMR2H register contains the high byte of the 16-bit Timer 2. In 8-bit mode, TMR2H contains the 8-bit high byte timer value.



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**SFR Definition 25.23. TMR4H Timer 4 High Byte**

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Bit	7	6	5	4	3	2	1	0
Name	TMR4H[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x95; SFR Page = F

Bit	Name	Function
7:0	TMR4H[7:0]	<b>Timer 4 High Byte.</b> In 16-bit mode, the TMR4H register contains the high byte of the 16-bit Timer 4. In 8-bit mode, TMR4H contains the 8-bit high byte timer value.

## 26.3.6. 16-Bit Pulse Width Modulator Mode

A PCA module may also be operated in 16-Bit PWM mode. In this mode, the 16-bit capture/compare module defines the number of PCA clocks for the low time of the PWM signal. When the PCA counter matches the module contents, the output on CEXn is asserted high; when the 16-bit counter overflows, CEXn is asserted low. To output a varying duty cycle, new value writes should be synchronized with PCA CCFn match interrupts. 16-Bit PWM Mode is enabled by setting the ECOMn, PWMn, and PWM16n bits in the PCA0CPMn register. For a varying duty cycle, match interrupts should be enabled (ECCFn = 1 AND MATn = 1) to help synchronize the capture/compare register writes. If the MATn bit is set to 1, the CCFn flag for the module will be set each time a 16-bit comparator match (rising edge) occurs. The CF flag in PCA0CN can be used to detect the overflow (falling edge). The duty cycle for 16-Bit PWM Mode is given by Equation 26.3.

**Important Note About Capture/Compare Registers:** When writing a 16-bit value to the PCA0 Capture/Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to 0; writing to PCA0CPHn sets ECOMn to 1.

$$\text{Duty Cycle} = \frac{(65536 - \text{PCA0CPn})}{65536}$$

### Equation 26.3. 16-Bit PWM Duty Cycle

Using Equation 26.3, the largest duty cycle is 100% (PCA0CPn = 0), and the smallest duty cycle is 0.0015% (PCA0CPn = 0xFFFF). A 0% duty cycle may be generated by clearing the ECOMn bit to 0.

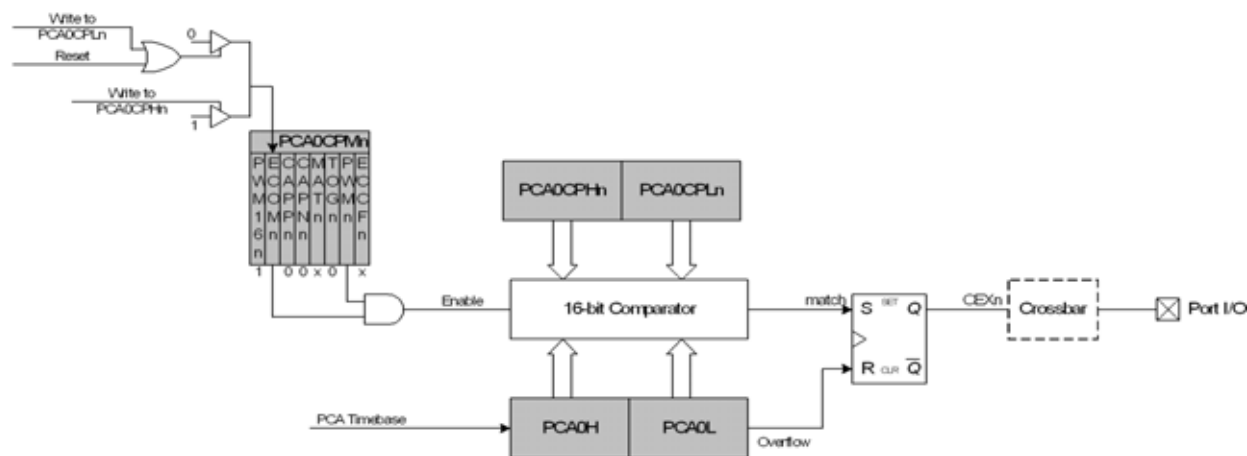


Figure 26.9. PCA 16-Bit PWM Mode