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Details

Product Status	Obsolete
Core Processor	HC11
Core Size	8-Bit
Speed	3MHz
Connectivity	SCI, SPI
Peripherals	POR, WDT
Number of I/O	30
Program Memory Size	-
Program Memory Type	ROMless
EEPROM Size	512 x 8
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	4.75V ~ 5.25V
Data Converters	A/D 8x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	68-LCC (J-Lead)
Supplier Device Package	68-PLCC (24.21x24.21)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc68hc11f1cfn3

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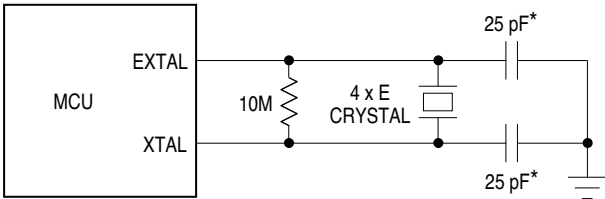
TECHNICAL DATA

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The XTAL pin is normally left unterminated when an external CMOS compatible clock is connected to the EXTAL pin. However, a 10 k Ω to 100 k Ω load resistor connected from the XTAL output to ground can be used to reduce RFI noise emission.

The XTAL output is normally used to drive a crystal. The XTAL output can be buffered with a high-impedance buffer, or it can be used to drive the EXTAL input of another M68HC11 device. Refer to **Figure 2-6**.

In all cases, use caution when designing circuitry associated with the oscillator pins. Load capacitances shown in the oscillator circuits include all stray layout capacitances. Refer to **Figure 2-4**, **Figure 2-5**, and **Figure 2-6**.



* Values include all stray capacitances.

Figure 2-4 Common Crystal Connections

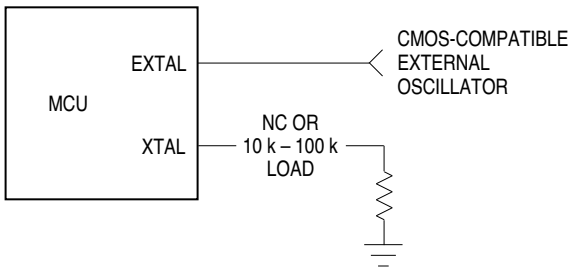
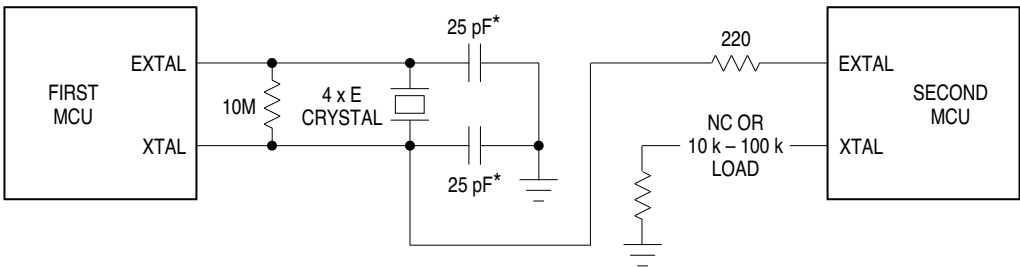


Figure 2-5 External Oscillator Connections



* Values include all stray capacitances.

Figure 2-6 One Crystal Driving Two MCUs

Table 3-2 Instruction Set (Sheet 4 of 6)

Mnemonic	Operation	Description	Addressing Mode	Instruction			Condition Codes							
				Opcode	Operand	Cycles	S	X	H	I	N	Z	V	C
INY	Increment Index Register Y	$IY + 1 \Rightarrow IY$	INH	18 08	—	4	—	—	—	—	—	Δ	—	—
JMP (opr)	Jump	See Figure 3-2	EXT IND,X IND,Y	18	7E hh ll	3	—	—	—	—	—	—	—	—
					6E ff	3	—	—	—	—	—	—	—	—
					6E ff	4	—	—	—	—	—	—	—	—
JSR (opr)	Jump to Subroutine	See Figure 3-2	DIR EXT IND,X IND,Y	18	9D dd	5	—	—	—	—	—	—	—	—
					BD hh ll	6	—	—	—	—	—	—	—	—
					AD ff	6	—	—	—	—	—	—	—	—
					AD ff	7	—	—	—	—	—	—	—	—
LDAA (opr)	Load Accumulator A	$M \Rightarrow A$	A IMM A DIR A EXT A IND,X A IND,Y	18	86 ii	2	—	—	—	—	Δ	Δ	0	—
					96 dd	3	—	—	—	—	—	—	—	—
					B6 hh ll	4	—	—	—	—	—	—	—	—
					A6 ff	4	—	—	—	—	—	—	—	—
					A6 ff	5	—	—	—	—	—	—	—	—
LDAB (opr)	Load Accumulator B	$M \Rightarrow B$	B IMM B DIR B EXT B IND,X B IND,Y	18	C6 ii	2	—	—	—	—	Δ	Δ	0	—
					D6 dd	3	—	—	—	—	—	—	—	—
					F6 hh ll	4	—	—	—	—	—	—	—	—
					E6 ff	4	—	—	—	—	—	—	—	—
					E6 ff	5	—	—	—	—	—	—	—	—
LDD (opr)	Load Double Accumulator D	$M \Rightarrow A, M + 1 \Rightarrow B$	IMM DIR EXT IND,X IND,Y	18	CC jj kk	3	—	—	—	—	Δ	Δ	0	—
					DC dd	4	—	—	—	—	—	—	—	—
					FC hh ll	5	—	—	—	—	—	—	—	—
					EC ff	5	—	—	—	—	—	—	—	—
					EC ff	6	—	—	—	—	—	—	—	—
LDS (opr)	Load Stack Pointer	$M : M + 1 \Rightarrow SP$	IMM DIR EXT IND,X IND,Y	18	8E jj kk	3	—	—	—	—	Δ	Δ	0	—
					9E dd	4	—	—	—	—	—	—	—	—
					BE hh ll	5	—	—	—	—	—	—	—	—
					AE ff	5	—	—	—	—	—	—	—	—
					AE ff	6	—	—	—	—	—	—	—	—
LDX (opr)	Load Index Register X	$M : M + 1 \Rightarrow IX$	IMM DIR EXT IND,X IND,Y	18	CE jj kk	3	—	—	—	—	Δ	Δ	0	—
					DE dd	4	—	—	—	—	—	—	—	—
					FE hh ll	5	—	—	—	—	—	—	—	—
					EE ff	5	—	—	—	—	—	—	—	—
					EE ff	6	—	—	—	—	—	—	—	—
LDY (opr)	Load Index Register Y	$M : M + 1 \Rightarrow IY$	IMM DIR EXT IND,X IND,Y	18	CE jj kk	4	—	—	—	—	Δ	Δ	0	—
					DE dd	5	—	—	—	—	—	—	—	—
					FE hh ll	6	—	—	—	—	—	—	—	—
					1A ff	6	—	—	—	—	—	—	—	—
					EE ff	6	—	—	—	—	—	—	—	—
LSL (opr)	Logical Shift Left		EXT IND,X IND,Y	18	78 hh ll	6	—	—	—	—	Δ	Δ	Δ	Δ
					68 ff	6	—	—	—	—	—	—	—	—
					68 ff	7	—	—	—	—	—	—	—	—
LSLA	Logical Shift Left A		A	INH	48	—	—	—	—	—	Δ	Δ	Δ	Δ
LSLB	Logical Shift Left B		B	INH	58	—	—	—	—	—	Δ	Δ	Δ	Δ
LSLD	Logical Shift Left Double		INH		05	—	—	—	—	—	Δ	Δ	Δ	Δ
LSR (opr)	Logical Shift Right		EXT IND,X IND,Y	18	74 hh ll	6	—	—	—	—	0	Δ	Δ	Δ
					64 ff	6	—	—	—	—	—	—	—	—
					64 ff	7	—	—	—	—	—	—	—	—
LSRA	Logical Shift Right A		A	INH	44	—	—	—	—	—	0	Δ	Δ	Δ
LSRB	Logical Shift Right B		B	INH	54	—	—	—	—	—	0	Δ	Δ	Δ
LSRD	Logical Shift Right Double		INH		04	—	—	—	—	—	0	Δ	Δ	Δ
MUL	Multiply 8 by 8	$A * B \Rightarrow D$	INH		3D	10	—	—	—	—	—	—	—	Δ
NEG (opr)	Two's Complement Memory Byte	$0 - M \Rightarrow M$	EXT IND,X IND,Y	18	70 hh ll	6	—	—	—	—	Δ	Δ	Δ	Δ
					60 ff	6	—	—	—	—	—	—	—	—
					60 ff	7	—	—	—	—	—	—	—	—
NEGA	Two's Complement A	$0 - A \Rightarrow A$	A	INH	40	—	—	—	—	—	Δ	Δ	Δ	Δ
NEGB	Two's Complement B	$0 - B \Rightarrow B$	B	INH	50	—	—	—	—	—	Δ	Δ	Δ	Δ

Table 4-7 EEPROM Erase Mode Control

BYTE	ROW	Action
0	0	Bulk Erase (All 512 Bytes)
0	1	Row Erase (16 Bytes)
1	0	Byte Erase
1	1	Byte Erase

ERASE — Erase/Normal Control for EEPROM

Can be read or written any time.

0 = Normal read or program mode

1 = Erase mode

EELAT — EEPROM Latch Control

Can be read or written any time. When EELAT equals one, writes to EEPROM cause address and data to be latched.

0 = EEPROM address and data bus configured for normal reads

1 = EEPROM address and data bus configured for programming or erasing

EEPGM — EEPROM Program Command

Can be read any time. Can only be written while EELAT = 1.

0 = Program or erase voltage switched off to EEPROM array

1 = Program or erase voltage switched on to EEPROM array

4.4.3 CONFIG Register Programming

Because the CONFIG register is implemented with EEPROM cells, use EEPROM procedures to erase and program this register. The procedure for programming is the same as for programming a byte in the EEPROM array, except that the CONFIG register address is used. CONFIG can be programmed or erased (including byte erase) while the MCU is operating in any mode, provided that PTCON in BPROT is clear. To change the value in the CONFIG register, complete the following procedure. Do not initiate a reset until the procedure is complete. The new value will not take effect until after the next reset sequence.

1. Erase the CONFIG register.
2. Program the new value to the CONFIG address.
3. Initiate reset.

CONFIG — System Configuration Register

\$103F

	Bit 7	6	5	4	3	2	1	Bit 0	
	EE3	EE2	EE1	EE0	—	NOCOP	—	EEON	
RESET:	1	1	1	1	1	P	1	1	Single Chip
	1	1	1	1	1	P(L)	1	1	Bootstrap
	P	P	P	P	1	P	1	P	Expanded
	P	P	P	P	1	P(L)	1	0	Special Test

P indicates a previously programmed bit. P(L) indicates that the bit resets to the logic level held in the EEPROM bit prior to reset, but the function of COP is controlled by DISR bit in TEST1 register.

1. POR or $\overline{\text{RESET}}$ pin
2. Clock monitor reset
3. COP watchdog reset
4. $\overline{\text{XIRQ}}$ interrupt
5. Illegal opcode interrupt
6. Software interrupt (SWI)

The maskable interrupt sources have the following priority arrangement:

1. $\overline{\text{IRQ}}$
2. Real-time interrupt
3. Timer input capture 1
4. Timer input capture 2
5. Timer input capture 3
6. Timer output compare 1
7. Timer output compare 2
8. Timer output compare 3
9. Timer output compare 4
10. Timer input capture 4/output compare 5
11. Timer overflow
12. Pulse accumulator overflow
13. Pulse accumulator input edge
14. SPI transfer complete
15. SCI system (refer to **Figure 5-5**)

Any one of these interrupts can be assigned the highest maskable interrupt priority by writing the appropriate value to the PSEL bits in the HPRIO register. Otherwise, the priority arrangement remains the same. An interrupt that is assigned highest priority is still subject to global masking by the I bit in the CCR, or by any associated local bits. Interrupt vectors are not affected by priority assignment. To avoid race conditions, HPRIO can only be written while I-bit interrupts are inhibited.

5.3.1 Highest Priority Interrupt and Miscellaneous Register

HPRIO — Highest Priority I-Bit Interrupt and Miscellaneous

\$103C

	Bit 7	6	5	4	3	2	1	Bit 0	
	RBOOT*	SMOD*	MDA*	IRV	PSEL3	PSEL2	PSEL1	PSEL0	
RESET:	0	0	0	0	0	1	0	1	Single Chip
	0	0	1	1	0	1	0	1	Expanded
	1	1	0	0	0	1	0	1	Bootstrap
	0	1	1	1	0	1	0	1	Special Test

*The values of the RBOOT, SMOD, MDA, and IRV reset bits depend on the operating mode selected during power-up. Refer to Table 4–3.

RBOOT — Read Bootstrap ROM

Set to one out of reset in bootstrap mode. Valid while in special modes only. Can be read any time. Can only be written in special modes. Refer to **SECTION 4 OPERATING MODES AND ON-CHIP MEMORY** for more information.

SECTION 6 PARALLEL INPUT/OUTPUT

The MC68HC11F1 MCU has up to 54 input/output lines, depending on the operating mode. The data bus of this microcontroller is nonmultiplexed. I/O lines are organized into seven parallel ports. Ports with bidirectional pins have an associated data direction control register. This register (DDRx) contains a data direction control bit for each bidirectional port line. The following table is a summary of the configuration and features of each port.

Table 6-1 I/O Port Configuration

Port	Input Pins	Output Pins	Bidirectional Pins	Shared Functions
Port A	—	—	8	Timer
Port B	—	8	—	High-Order Address
Port C	—	—	8	Data Bus
Port D	—	—	6	SCI and SPI
Port E	8	—	—	A/D Converter
Port F	—	8	—	Low-Order Address
Port G	—	—	8	Chip Select Outputs

Port pin function is mode dependent. Do not confuse pin function with the electrical state of the pin at reset. Port pins are either driven to a specified logic level or are configured as high impedance inputs. I/O pins configured as high-impedance inputs have port data that is indeterminate. The contents of the corresponding latches are dependent upon the electrical state of the pins during reset. In port descriptions, an “I” indicates this condition. Port pins that are driven to a known logic level during reset are shown with a value of either one or zero. Some control bits are unaffected by reset. Reset states for these bits are indicated with a “U”.

6.1 Port A

Port A has eight bidirectional I/O pins and shares functions with the timer system.

PORTA — Port A Data

\$1000

	Bit 7	6	5	4	3	2	1	Bit 0
	PA7	PA6	PA5	PA4	PA3	PA2	PA1	PA0
RESET:	I	I	I	I	I	I	I	I
Alt. Pin Func.:	PAI	OC2	OC3	OC4	IC4/OC5	IC1	IC2	IC3
And/or:	OC1	OC1	OC1	OC1	OC1	—	—	—

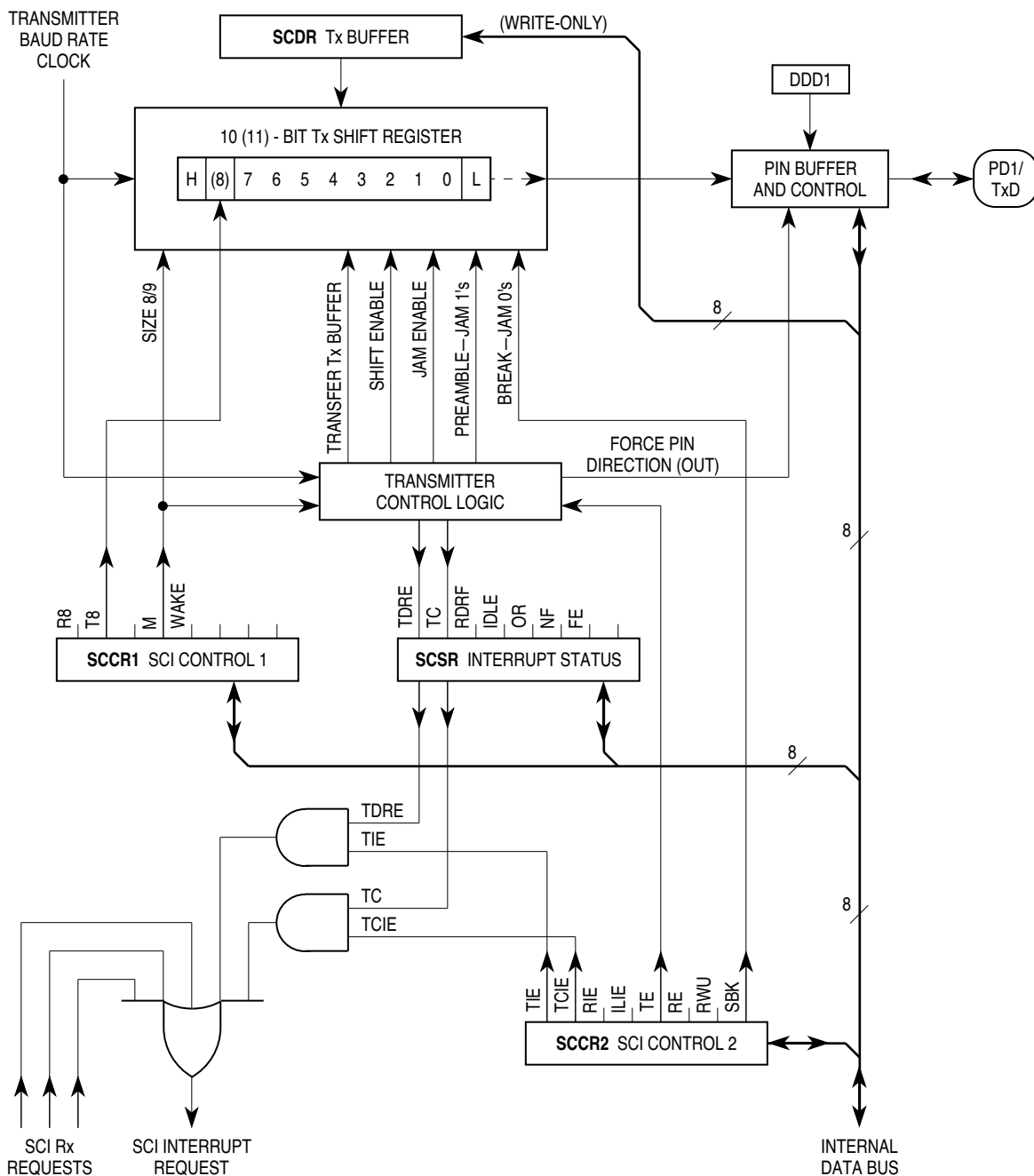


Figure 7-1 SCI Transmitter Block Diagram

7.3 Receive Operation

During receive operations, the transmit sequence is reversed. The serial shift register receives data and transfers it to a parallel receive data register (SCDR) as a complete word. This double buffered operation allows a character to be shifted in serially while another character is already in the SCDR. An advanced data recovery scheme distinguishes valid data from noise in the serial data stream. The data input is selectively sampled to detect receive data, and a majority voting circuit determines the value and integrity of each bit.

R8 — Receive Data Bit 8

If M bit is set, R8 stores the ninth bit in the receive data character.

T8 — Transmit Data Bit 8

If M bit is set, T8 stores the ninth bit in the transmit data character.

M — Mode (Select Character Format)

0 = Start bit, 8 data bits, 1 stop bit

1 = Start bit, 9 data bits, 1 stop bit

WAKE — Wakeup by Address Mark/Idle

0 = Wakeup by IDLE line recognition

1 = Wakeup by address mark (most significant data bit set)

7.6.3 Serial Communications Control Register 2

The SCCR2 register provides the control bits that enable or disable individual SCI functions.

SCCR2 — SCI Control Register 2

\$102D

	Bit 7	6	5	4	3	2	1	Bit 0
	TIE	TCIE	RIE	ILIE	TE	RE	RWU	SBK
RESET:	0	0	0	0	0	0	0	0

TIE — Transmit Interrupt Enable

0 = TDRE interrupts disabled

1 = SCI interrupt requested when TDRE status flag is set

TCIE — Transmit Complete Interrupt Enable

0 = TC interrupts disabled

1 = SCI interrupt requested when TC status flag is set

RIE — Receiver Interrupt Enable

0 = RDRF and OR interrupts disabled

1 = SCI interrupt requested when RDRF flag or the OR status flag is set

ILIE — Idle-Line Interrupt Enable

0 = IDLE interrupts disabled

1 = SCI interrupt requested when IDLE status flag is set

TE — Transmitter Enable

When TE goes from zero to one, one unit of idle character time (logic one) is queued as a preamble.

0 = Transmitter disabled

1 = Transmitter enabled

RE — Receiver Enable

0 = Receiver disabled

1 = Receiver enabled

8.3.1 Master In Slave Out

MISO is one of two unidirectional serial data signals. It is an input to a master device and an output from a slave device. The MISO line of a slave device is placed in the high-impedance state if the slave device is not selected.

8.3.2 Master Out Slave In

The MOSI line is the second of the two unidirectional serial data signals. It is an output from a master device and an input to a slave device. The master device places data on the MOSI line a half-cycle before the clock edge that the slave device uses to latch the data.

8.3.3 Serial Clock

SCK, an input to a slave device, is generated by the master device and synchronizes data movement in and out of the device through the MOSI and MISO lines. Master and slave devices are capable of exchanging a byte of information during a sequence of eight clock cycles.

There are four possible timing relationships that can be chosen by using control bits CPOL and CPHA in the serial peripheral control register (SPCR). Both master and slave devices must operate with the same timing. The SPI clock rate select bits, SPR[1:0], in the SPCR of the master device, select the clock rate. In a slave device, SPR[1:0] have no effect on the operation of the SPI.

8.3.4 Slave Select

The slave select ($\overline{SS}$) input of a slave device must be externally asserted before a master device can exchange data with the slave device. $\overline{SS}$ must be low before data transactions and must stay low for the duration of the transaction.

The $\overline{SS}$ line of the master must be held high. If it goes low, a mode fault error flag (MODF) is set in the serial peripheral status register (SPSR). To disable the mode fault circuit, write a one in bit 5 of the port D data direction register. This sets the $\overline{SS}$ pin to act as a general-purpose output rather than the dedicated input to the slave select circuit, thus inhibiting the mode fault flag. The other three lines are dedicated to the SPI whenever the serial peripheral interface is on.

The state of the master and slave CPHA bits affects the operation of $\overline{SS}$. CPHA settings should be identical for master and slave. When CPHA = 0, the shift clock is the OR of $\overline{SS}$ with SCK. In this clock phase mode, $\overline{SS}$ must go high between successive characters in an SPI message. When CPHA = 1, $\overline{SS}$ can be left low between successive SPI characters. In cases where there is only one SPI slave MCU, its $\overline{SS}$ line can be tied to V_{SS} as long as only CPHA = 1 clock mode is used.

8.4 SPI System Errors

Two system errors can be detected by the SPI system. The first type of error arises in a multiple-master system when more than one SPI device simultaneously tries to be a master. This error is called a mode fault. The second type of error, write collision, indicates that an attempt was made to write data to the SPDR while a transfer was in progress.



TOC1–TOC4 — Timer Output Compare
\$1016–\$101D

\$1016	Bit 15	14	13	12	11	10	9	Bit 8	TOC1 (High)
\$1017	Bit 7	6	5	4	3	2	1	Bit 0	TOC1 (Low)
\$1018	Bit 15	14	13	12	11	10	9	Bit 8	TOC2 (High)
\$1019	Bit 7	6	5	4	3	2	1	Bit 0	TOC2 (Low)
\$101A	Bit 15	14	13	12	11	10	9	Bit 8	TOC3 (High)
\$101B	Bit 7	6	5	4	3	2	1	Bit 0	TOC3 (Low)
\$101C	Bit 15	14	13	12	11	10	9	Bit 8	TOC4 (High)
\$101D	Bit 7	6	5	4	3	2	1	Bit 0	TOC4 (Low)

All TOCx register pairs reset to ones (\$FFFF).

9.3.2 Timer Compare Force Register

The CFORC register allows forced early compares. FOC[1:5] correspond to the five output compares. These bits are set for each output compare that is to be forced. The action taken as a result of a forced compare is the same as if there were a match between the OCx register and the free-running counter, except that the corresponding interrupt status flag bits are not set. The forced channels trigger their programmed pin actions to occur at the next timer count transition after the write to CFORC.

The CFORC bits should not be used on an output compare function that is programmed to toggle its output on a successful compare because a normal compare that occurs immediately before or after the force can result in an undesirable operation.

CFORC — Timer Compare Force
\$100B

	Bit 7	6	5	4	3	2	1	Bit 0
	FOC1	FOC2	FOC3	FOC4	FOC5	—	—	—
RESET:	0	0	0	0	0	0	0	0

FOC[1:5] — Force Output Comparison

When the FOC bit associated with an output compare circuit is set, the output compare circuit immediately performs the action it is programmed to do when an output match occurs.

0 = Not affected

1 = Output x action occurs

Bits [2:0] — Not implemented

Always read zero

9.3.3 Output Compare Mask Registers

Use OC1M with OC1 to specify the bits of port A that are affected by a successful OC1 compare. The bits of the OC1M register correspond to PA[7:3].

TCTL1 — Timer Control 1

\$1020

	Bit 7	6	5	4	3	2	1	Bit 0
	OM2	OL2	OM3	OL3	OM4	OL4	OM5	OL5
RESET:	0	0	0	0	0	0	0	0

OM[2:5] — Output Mode

OL[2:5] — Output Level

These control bit pairs are encoded to specify the action taken after a successful OCx compare. OC5 functions only if the I4/O5 bit in the PACTL register is clear. Refer to **Table 9-3** for the coding.

Table 9-2 Timer Output Compare Configuration

OMx	OLx	Action Taken on Successful Compare
0	0	Timer disconnected from output pin logic
0	1	Toggle OCx output line
1	0	Clear OCx output line to zero
1	1	Set OCx output line to one

9.3.7 Timer Interrupt Mask Register 1

Use this 8-bit register to enable or inhibit the timer input capture and output compare interrupts.

TMSK1 — Timer Interrupt Mask 1

\$1022

	Bit 7	6	5	4	3	2	1	Bit 0
	OC1I	OC2I	OC3I	OC4I	I4/O5I	IC1I	IC2I	IC3I
RESET:	0	0	0	0	0	0	0	0

OC1I–OC4I — Output Compare x Interrupt Enable

If the OCxI enable bit is set when the OCxF flag bit is set, a hardware interrupt sequence is requested.

I4/O5I — Input Capture 4/Output Compare 5 Interrupt Enable

When I4/O5 in PACTL is one, I4/O5I is the input capture 4 interrupt enable bit. When I4/O5 in PACTL is zero, I4/O5I is the output compare 5 interrupt enable bit.

IC1I–IC3I — Input Capture x Interrupt Enable

If the ICxI enable bit is set when the ICxF flag bit is set, a hardware interrupt sequence is requested.

NOTE

Bits in TMSK1 correspond bit for bit with flag bits in TFLG1. Ones in TMSK1 enable the corresponding interrupt sources.

9.3.8 Timer Interrupt Flag Register 1

Bits in this register indicate when timer system events have occurred. Coupled with the bits of TMSK1, the bits of TFLG1 allow the timer subsystem to operate in either a polled or interrupt driven system. Each bit of TFLG1 corresponds to a bit in TMSK1 in the same position.

TFLG1 — Timer Interrupt Flag 1

\$1023

	Bit 7	6	5	4	3	2	1	Bit 0
	OC1F	OC2F	OC3F	OC4F	I4/O5F	IC1F	IC2F	IC3F
RESET:	0	0	0	0	0	0	0	0

Clear flags by writing a one to the corresponding bit position(s).

OC1F–OC4F — Output Compare x Flag

Set each time the counter matches output compare x value

I4/O5F — Input Capture 4/Output Compare 5 Flag

Set by IC4 or OC5, depending on the function enabled by I4/O5 bit in PACTL

IC1F–IC3F — Input Capture x Flag

Set each time a selected active edge is detected on the ICx input line

9.3.9 Timer Interrupt Mask Register 2

Use this 8-bit register to enable or inhibit timer overflow and real-time interrupts. The timer prescaler control bits are included in this register.

TMSK2 — Timer Interrupt Mask 2

\$1024

	Bit 7	6	5	4	3	2	1	Bit 0
	TOI	RTII	PAOVI	PAII	—	—	PR1	PR0
RESET:	0	0	0	0	0	0	0	0

TOI — Timer Overflow Interrupt Enable

0 = TOF interrupts disabled

1 = Interrupt requested when TOF is set to one

RTII — Real-Time Interrupt Enable

Refer to **9.4 Real-Time Interrupt**.

PAOVI — Pulse Accumulator Overflow Interrupt Enable

Refer to **9.6.3 Pulse Accumulator Status and Interrupt Bits**.

PAII — Pulse Accumulator Input Edge Interrupt Enable

Refer to **9.6.3 Pulse Accumulator Status and Interrupt Bits**.

PR[1:0] — Timer Prescaler Select

These bits are used to select the prescaler divide-by ratio. In normal modes, PR[1:0] can only be written once, and the write must be within 64 cycles after reset. Refer to **Table 9-1** and **Table 9-4** for specific timing values.

Table 9-3 Timer Prescaler Selection

PR[1:0]	Prescaler
0 0	1
0 1	4
1 0	8
1 1	16

NOTE

Bits in TMSK2 correspond bit for bit with flag bits in TFLG2. Ones in TMSK2 enable the corresponding interrupt sources.

9.3.10 Timer Interrupt Flag Register 2

Bits in this register indicate when certain timer system events have occurred. Coupled with the four high-order bits of TMSK2, the bits of TFLG2 allow the timer subsystem to operate in either a polled or interrupt driven system. Each bit of TFLG2 corresponds to a bit in TMSK2 in the same position.

TFLG2 — Timer Interrupt Flag 2 **\$1025**

	Bit 7	6	5	4	3	2	1	Bit 0
	TOF	RTIF	PAOVF	PAIF	—	—	—	—
RESET:	0	0	0	0	0	0	0	0

Clear flags by writing a one to the corresponding bit position(s).

TOF — Timer Overflow Interrupt Flag
Set when TCNT changes from \$FFFF to \$0000

RTIF — Real-Time (Periodic) Interrupt Flag
Refer to **9.4 Real-Time Interrupt**.

PAOVF — Pulse Accumulator Overflow Interrupt Flag
Refer to **9.6 Pulse Accumulator**.

PAIF — Pulse Accumulator Input Edge Interrupt Flag
Refer to **9.6 Pulse Accumulator**.

Bits [3:0] — Not implemented
Always read zero

9.4 Real-Time Interrupt

The real-time interrupt (RTI) feature, used to generate hardware interrupts at a fixed periodic rate, is controlled and configured by two bits (RTR1 and RTR0) in the pulse accumulator control (PACTL) register. The RTII bit in the TMSK2 register enables the interrupt capability. The four different rates available are a product of the MCU oscillator frequency and the value of bits RTR[1:0]. Refer to **Table 9-4**, which shows the periodic real-time interrupt rates.

10.4 Channel Assignments

The multiplexer allows the A/D converter to select one of sixteen analog signals. Eight of these channels correspond to port E input lines, four of the channels are internal reference points or test functions, and four channels are reserved. Refer to **Table 10-1**.

Table 10-1 A/D Converter Channel Assignments

Channel Number	Channel Signal	Result in ADRx if MULT = 1
1	AN0	ADR1
2	AN1	ADR2
3	AN2	ADR3
4	AN3	ADR4
5	AN4	ADR1
6	AN5	ADR2
7	AN6	ADR3
8	AN7	ADR4
9	Reserved	—
10	Reserved	—
11	Reserved	—
12	Reserved	—
13	V _{RH} *	ADR1
14	V _{RL} *	ADR2
15	(V _{RH})/2*	ADR3
16	Reserved*	ADR4

*Used for factory testing

10.5 Single-Channel Operation

There are two types of single-channel operation. When SCAN = 0, the first type, the single selected channel is converted four consecutive times. The first result is stored in A/D result register 1 (ADR1), and the fourth result is stored in ADR4. After the fourth conversion is complete, all conversion activity is halted until a new conversion command is written to the ADCTL register. In the second type of single-channel operation, SCAN = 1, conversions continue to be performed on the selected channel with the fifth conversion being stored in register ADR1 (overwriting the first conversion result), the sixth conversion overwriting ADR2, and so on.

10.6 Multiple-Channel Operation

There are two types of multiple-channel operation. When SCAN = 0, the first type, a selected group of four channels is converted one time each. The first result is stored in A/D result register 1 (ADR1), and the fourth result is stored in ADR4. After the fourth conversion is complete, all conversion activity is halted until a new conversion command is written to the ADCTL register. In the second type of multiple-channel operation, SCAN = 1, conversions continue to be performed on the selected group of channels with the fifth conversion being stored in register ADR1 (replacing the earlier conversion result for the first channel in the group), the sixth conversion overwriting ADR2, and so on.

NOTE

When the multiple-channel continuous scan mode is used, extra care is needed in the design of circuitry driving the A/D inputs. The charge on the capacitive DAC array before the sample time is related to the voltage on the previously converted channel. A charge share situation exists between the internal DAC capacitance and the external circuit capacitance. Although the amount of charge involved is small, the rate at which it is repeated is every 64 μ s for an E clock of 2 MHz. The RC charging rate of the external circuit must be balanced against this charge sharing effect to avoid errors in accuracy. Refer to *M68HC11 Reference Manual* (M68HC11RM/AD) for further information.

CD–CA — Channel Selects D–A

Refer to **Table 10-2**. When a multiple channel mode is selected (MULT = 1), the two least significant channel select bits (CB and CA) have no meaning and the CD and CC bits specify which group of four channels is to be converted.

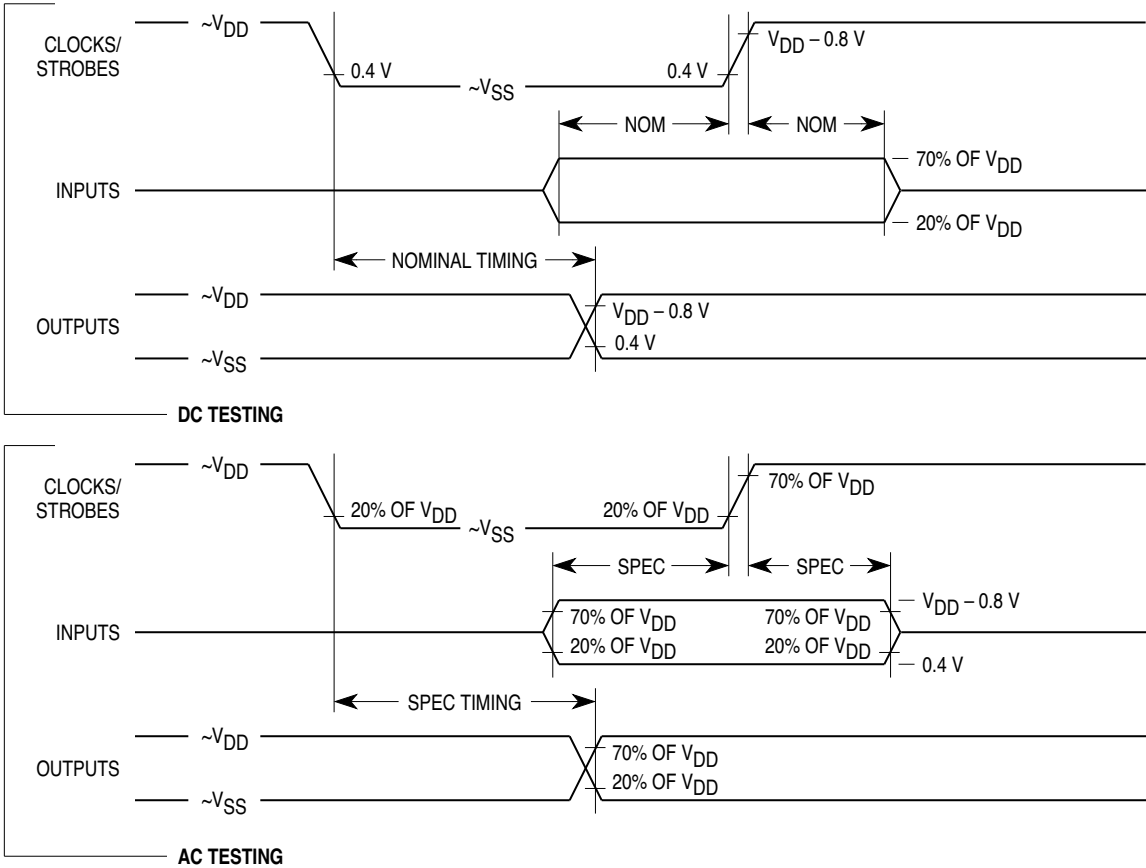
Table 10-2 A/D Converter Channel Selection

Channel Select Control Bits	Channel Signal	Result in ADRx if MULT = 1
CD:CC:CB:CA		
0000	AN0	ADR1
0001	AN1	ADR2
0010	AN2	ADR3
0011	AN3	ADR4
0100	AN4	ADR1
0101	AN5	ADR2
0110	AN6	ADR3
0111	AN7	ADR4
1000	Reserved	—
1001	Reserved	—
1010	Reserved	—
1011	Reserved	—
1100	V _{RH} *	ADR1
1101	V _{RL} *	ADR2
1110	(V _{RH})/2*	ADR3
1111	Reserved*	ADR4

*Used for factory testing

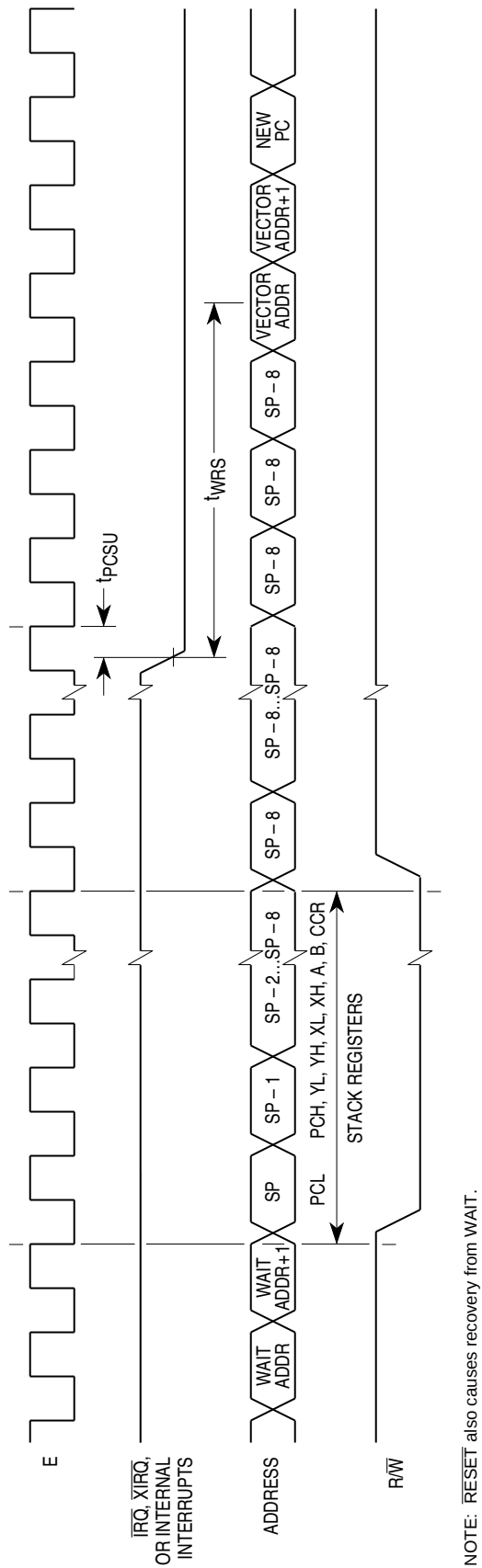
10.9 A/D Converter Result Registers

These read-only registers hold an 8-bit conversion result. Writes to these registers have no effect. Data in the A/D converter result registers is valid when the CCF flag in the ADCTL register is set, indicating a conversion sequence is complete. If conversion results are needed sooner, refer to **Figure 10-3**, which shows the A/D conversion sequence diagram.



- NOTES:
1. Full test loads are applied during all DC electrical tests and AC timing measurements.
 2. During AC timing measurements, inputs are driven to 0.4 volts and $V_{DD} - 0.8$ volts while timing measurements are taken at the 20% and 70% of V_{DD} points.

Figure A-1 Test Methods



NOTE: $\overline{RESET}$ also causes recovery from WAIT.

Figure A-5 WAIT Recovery from Interrupt Timing Diagram

Table A-5 Peripheral Port Timing

$V_{DD} = 5.0 \text{ Vdc} \pm 5\%$, $V_{SS} = 0 \text{ Vdc}$, $T_A = T_L \text{ to } T_H$

Characteristic	Symbol	2.0 MHz		3.0 MHz		4.0 MHz		Unit
		Min	Max	Min	Max	Min	Max	
Frequency of Operation (E-Clock Frequency)	f_o	dc	2.0	dc	3.0	dc	4.0	MHz
E-Clock Period	t_{cyc}	500	—	333	—	250	—	ns
Peripheral Data Setup Time (MCU Read of Ports A, C, D, E, G)	t_{PDSU}	100	—	100	—	100	—	ns
Peripheral Data Hold Time (MCU Read of Ports A, C, D, E, G)	t_{PDH}	50	—	50	—	50	—	ns
Delay Time, Peripheral Data Write (MCU Write to Port A) (MCU Write to Ports B, C, D, F, and G $t_{PWD} = 1/4 t_{cyc} + 100 \text{ ns}$)	t_{PWD}	—	200	—	200	—	200	ns
		—	225	—	183	—	162	

NOTES:

- Ports C, D, and G timing is valid for active drive (CWOM, DWOM, and GWOM bits **cleared**).
- All timing is shown with respect to 20% V_{DD} and 70% V_{DD} , unless otherwise noted.

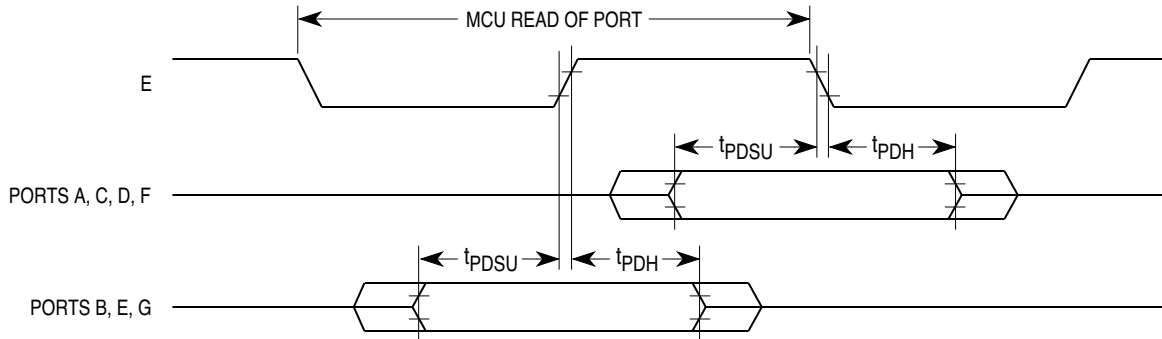


Figure A-7 Port Read Timing Diagram

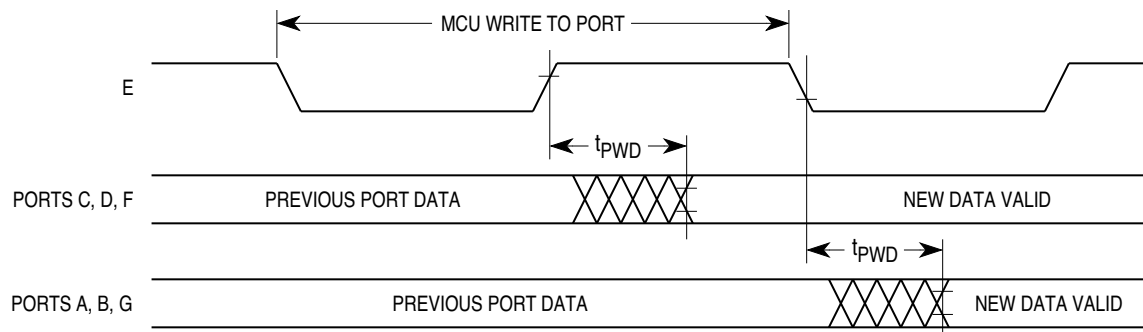


Figure A-8 Port Write Timing Diagram

B.3 Ordering Information

Use the information in **Table B-1** to specify the appropriate device when placing an order.

Table B-1 Device Ordering Information

Description	Package	Temperature	Frequency	MC Order Number
NO ROM, 512 Bytes EEPROM, 1024 Bytes RAM	80-Pin LQFP (14 mm X 14 mm, 1.4 mm thick)	– 40° to + 85° C	2 MHz	MC68HC11F1CPU2
			3 MHz	MC68HC11F1CPU3
			4 MHz	MC68HC11F1CPU4
		– 40° to + 105° C	2 MHz	MC68HC11F1VPU2
			3 MHz	MC68HC11F1VPU3
			4 MHz	MC68HC11F1VPU4
		– 40° to + 125° C	2 MHz	MC68HC11F1MPU2
			3 MHz	MC68HC11F1MPU3
	68-Pin PLCC	– 40° to + 85° C	2 MHz	MC68HC11F1CFN2
			3 MHz	MC68HC11F1CFN3
			4 MHz	MC68HC11F1CFN4
		– 40° to + 105° C	2 MHz	MC68HC11F1VFN2
			3 MHz	MC68HC11F1VFN3
			4 MHz	MC68HC11F1VFN4
		– 40° to + 125° C	2 MHz	MC68HC11F1MFN2
			3 MHz	MC68HC11F1MFN3