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Understanding [Embedded - DSP \(Digital Signal Processors\)](#)

[Embedded - DSP \(Digital Signal Processors\)](#) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of [Embedded - DSP \(Digital Signal Processors\)](#)

Details

Product Status	Active
Type	Floating Point
Interface	DAI, SPI
Clock Rate	333MHz
Non-Volatile Memory	ROM (512kB)
On-Chip RAM	384kB
Voltage - I/O	3.30V
Voltage - Core	1.20V
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	136-LFBGA, CSPBGA
Supplier Device Package	136-CSPBGA (12x12)
Purchase URL	https://www.e-xfl.com/product-detail/analog-devices/adsp-21363bbc-1aa

Data Register File

Each processing element contains a general-purpose data register file. The register files transfer data between the computation units and the data buses, and store intermediate results. These 10-port, 32-register (16 primary, 16 secondary) files, combined with the ADSP-2136x enhanced Harvard architecture, allow unconstrained data flow between computation units and internal memory. The registers in PEX are referred to as R0–R15 and in PEY as S0–S15.

Context Switch

Many of the processor's registers have secondary registers that can be activated during interrupt servicing for a fast context switch. The data registers in the register file, the DAG registers, and the multiplier result register all have secondary registers. The primary registers are active at reset, while the secondary registers are activated by control bits in a mode control register.

Universal Registers

The universal registers are general purpose registers. The USTAT (4) registers allow easy bit manipulations (Set, Clear, Toggle, Test, XOR) for all system registers (control/status) of the core.

The data bus exchange register (PX) permits data to be passed between the 64-bit PM data bus and the 64-bit DM data bus, or between the 40-bit register file and the PM/DM data bus. These registers contain hardware to handle the data width difference.

Timer

A core timer that can generate periodic software interrupts. The core timer can be configured to use FLAG3 as a timer expired signal.

Single-Cycle Fetch of Instruction and Four Operands

The processor features an enhanced Harvard architecture in which the data memory (DM) bus transfers data and the program memory (PM) bus transfers both instructions and data (see Figure 2). With its separate program and data memory buses and on-chip instruction cache, the processor can simultaneously fetch four operands (two over each data bus) and one instruction (from the cache), all in a single cycle.

Instruction Cache

The processor includes an on-chip instruction cache that enables three-bus operation for fetching an instruction and four data values. The cache is selective—only the instructions whose fetches conflict with PM bus data accesses are cached. This cache allows full-speed execution of core looped operations such as digital filter multiply-accumulates, and FFT butterfly processing.

Data Address Generators with Zero-Overhead Hardware Circular Buffer Support

The processor's two data address generators (DAGs) are used for indirect addressing and implementing circular data buffers in hardware. Circular buffers allow efficient programming of delay lines and other data structures required in digital signal

processing, and are commonly used in digital filters and Fourier transforms. The two DAGs contain sufficient registers to allow the creation of up to 32 circular buffers (16 primary register sets, 16 secondary). The DAGs automatically handle address pointer wraparound, reduce overhead, increase performance, and simplify implementation. Circular buffers can start and end at any memory location.

Flexible Instruction Set

The 48-bit instruction word accommodates a variety of parallel operations for concise programming. For example, the processor can conditionally execute a multiply, an add, and a subtract in both processing elements while branching and fetching up to four 32-bit values from memory—all in a single instruction.

On-Chip Memory

The processor contains 3M bits of internal SRAM and 4M bits of internal ROM. Each block can be configured for different combinations of code and data storage (see Table 3). Each memory block supports single-cycle, independent accesses by the core processor and I/O processor. The processor's memory architecture, in combination with its separate on-chip buses, allows two data transfers from the core and one from the I/O processor, in a single cycle.

The SRAM can be configured as a maximum of 96K words of 32-bit data, 192K words of 16-bit data, 64K words of 48-bit instructions (or 40-bit data), or combinations of different word sizes up to 3M bits. All of the memory can be accessed as 16-bit, 32-bit, 48-bit, or 64-bit words. A 16-bit floating-point storage format is supported that effectively doubles the amount of data that can be stored on-chip. Conversion between the 32-bit floating-point and 16-bit floating-point formats is performed in a single instruction. While each memory block can store combinations of code and data, accesses are most efficient when one block stores data using the DM bus for transfers, and the other block stores instructions and data using the PM bus for transfers.

Using the DM bus and PM buses, with one bus dedicated to each memory block, assures single-cycle execution with two data transfers. In this case, the instruction must be available in the cache.

On-Chip Memory Bandwidth

The internal memory architecture allows three accesses at the same time to any of the four blocks, assuming no block conflicts. The total bandwidth is gained with DMD and PMD buses (2 × 64-bits, core CLK) and the IOD bus (32-bit, PCLK).

ROM-Based Security

The processor has a ROM security feature that provides hardware support for securing user software code by preventing unauthorized reading from the internal code. When using this feature, the processor does not boot-load any external code, executing exclusively from internal ROM. Additionally, the processor is not freely accessible via the JTAG port. Instead, a unique 64-bit key, which must be scanned in through the JTAG

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Table 3. ADSP-2136x Internal Memory Space

IOP Registers 0x0000 0000–0003 FFFF			
Long Word (64 Bits)	Extended Precision Normal or Instruction Word (48 Bits)	Normal Word (32 Bits)	Short Word (16 Bits)
Block 0 ROM 0x0004 0000–0x0004 7FFF	Block 0 ROM 0x0008 0000–0x0008 AAA9	Block 0 ROM 0x0008 0000–0x0008 FFFF	Block 0 ROM 0x0010 0000–0x0011 FFFF
Reserved 0x0004 8000–0x0004 BFFF		Reserved 0x0009 0000–0x0009 7FFF	Reserved 0x0012 0000–0x0012 FFFF
Block 0 SRAM 0x0004 C000–0x0004 FFFF	Block 0 SRAM 0x0009 0000–0x0009 5554	Block 0 SRAM 0x0009 8000–0x0009 FFFF	Block 0 SRAM 0x0013 0000–0x0013 FFFF
Block 1 ROM 0x0005 0000–0x0005 7FFF	Block 1 ROM 0x000A 0000–0x000A AAA9	Block 1 ROM 0x000A 0000–0x000A FFFF	Block 1 ROM 0x0014 0000–0x0015 FFFF
Reserved 0x0005 8000–0x0005 BFFF		Reserved 0x000B 0000–0x000B 7FFF	Reserved 0x0016 0000–0x0016 FFFF
Block 1 SRAM 0x0005 C000–0x0005 FFFF	Block 1 SRAM 0x000B 0000–0x000B 5554	Block 1 SRAM 0x000B 8000–0x000B FFFF	Block 1 SRAM 0x0017 0000–0x0017 FFFF
Block 2 SRAM 0x0006 0000–0x0006 1FFF	Block 2 SRAM 0x000C 0000–0x000C 2AA9	Block 2 SRAM 0x000C 0000–0x000C 3FFF	Block 2 SRAM 0x0018 0000–0x0018 7FFF
Reserved 0x0006 2000–0x0006 FFFF		Reserved 0x000C 4000–0x000D FFFF	Reserved 0x0018 8000–0x001B FFFF
Block 3 SRAM 0x0007 0000–0x0007 1FFF	Block 3 SRAM 0x000E 0000–0x000E 2AA9	Block 3 SRAM 0x000E 0000–0x000E 3FFF	Block 3 SRAM 0x001C 0000–0x001C 7FFF
Reserved 0x0007 2000–0x0007 FFFF		Reserved 0x000E 4000–0x000F FFFF	Reserved 0x001C 8000–0x001F FFFF
			Reserved 0x0020 0000–0xFFFF FFFF

or test access port, is assigned to each customer. The device ignores a wrong key. Emulation features and external boot modes are only available after the correct key is scanned.

FAMILY PERIPHERAL ARCHITECTURE

The ADSP-2136x family contains a rich set of peripherals that support a wide variety of applications, including high quality audio, medical imaging, communications, military, test equipment, 3D graphics, speech recognition, monitor control, imaging, and other applications.

Parallel Port

The parallel port provides interfaces to SRAM and peripheral devices. The multiplexed address and data pins (AD15–0) can access 8-bit devices with up to 24 bits of address, or 16-bit devices with up to 16 bits of address. In either mode, 8-bit or 16-bit, the maximum data transfer rate is $f_{PCLK}/4$.

DMA transfers are used to move data to and from internal memory. Access to the core is also facilitated through the parallel port register read/write functions. The \overline{RD} , \overline{WR} , and ALE (address latch enable) pins are the control pins for the parallel port.

Serial Peripheral (Compatible) Interface

The processors contain two serial peripheral interface ports (SPIs). The SPI is an industry-standard synchronous serial link, enabling the processor's SPI-compatible port to communicate with other SPI-compatible devices. The SPI consists of two data pins, one device select pin, and one clock pin. It is a full-duplex synchronous serial interface, supporting both master and slave modes and can operate at a maximum baud rate of $f_{PCLK}/4$.

The SPI port can operate in a multimaster environment by interfacing with up to four other SPI-compatible devices, either acting as a master or slave device. The ADSP-2136x SPI-compatible peripheral implementation also features programmable baud rate, clock phase, and polarities. The SPI-compatible port uses open drain drivers to support a multimaster configuration and to avoid data contention.

Pulse-Width Modulation

The entire PWM module has four groups of four PWM outputs each. Therefore, this module generates 16 PWM outputs in total. Each PWM group produces two pairs of PWM signals on the four PWM outputs.

The PWM module is a flexible, programmable, PWM waveform generator that can be programmed to generate the required switching patterns for various applications related to motor and engine control or audio power control. The PWM generator can

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audio channels in I2S, left-justified sample pair, or right-justified mode. One frame sync cycle indicates one 64-bit left/right pair, but data is sent to the FIFO as 32-bit words (that is, one-half of a frame at a time). The processor supports 24- and 32-bit I²S, 24- and 32-bit left-justified, and 24-, 20-, 18- and 16-bit right-justified formats.

Precision Clock Generator (PCG)

The precision clock generators (PCG) consist of two units, each of which generates a pair of signals (clock and frame sync) derived from a clock input signal. The units, A and B, are identical in functionality and operate independently of each other. The two signals generated by each unit are normally used as a serial bit clock/frame sync pair.

Peripheral Timers

The following three general-purpose timers can generate periodic interrupts and be independently set to operate in one of three modes:

- Pulse waveform generation mode
- Pulse width count/capture mode
- External event watchdog mode

Each general-purpose timer has one bidirectional pin and four registers that implement its mode of operation: a 6-bit configuration register, a 32-bit count register, a 32-bit period register, and a 32-bit pulse width register. A single control and status register enables or disables all three general-purpose timers independently.

I/O PROCESSOR FEATURES

The processor's I/O provides many channels of DMA and controls the extensive set of peripherals described in the previous sections.

DMA Controller

The processor's on-chip DMA controllers allow data transfers without processor intervention. The DMA controller operates independently and invisibly to the processor core, allowing DMA operations to occur while the core is simultaneously executing its program instructions. DMA transfers can occur between the processor's internal memory and its serial ports, the SPI-compatible (serial peripheral interface) ports, the IDP (input data port), the parallel data acquisition port (PDAP), or the parallel port (PP). See [Table 4](#).

Table 4. DMA Channels

Peripheral	ADSP-2136x
SPORTs	12
IDP/PDAP	8
SPI	2
MTM/DTCP	2
Parallel Port	1
Total DMA Channels	25

SYSTEM DESIGN

The following sections provide an introduction to system design options and power supply issues.

Program Booting

The internal memory of the processor boots at system power-up from an 8-bit EPROM via the parallel port, an SPI master, an SPI slave, or an internal boot. Booting is determined by the boot configuration (BOOT_CFG1–0) pins in [Table 5](#). Selection of the boot source is controlled via the SPI as either a master or slave device, or it can immediately begin executing from ROM.

Table 5. Boot Mode Selection

BOOT_CFG1–0	Booting Mode
00	SPI Slave Boot
01	SPI Master Boot
10	Parallel Port Boot via EPROM
11	No booting occurs. Processor executes from internal ROM after reset.

Phase-Locked Loop

The processors use an on-chip phase-locked loop (PLL) to generate the internal clock for the core. On power-up, the CLK_CFG1–0 pins are used to select ratios of 32:1, 16:1, and 6:1. After booting, numerous other ratios can be selected via software control.

The ratios are made up of software configurable numerator values from 1 to 64 and software configurable divisor values of 1, 2, 4, and 8.

Power Supplies

The processor has a separate power supply connection for the internal (V_{DDINT}), external (V_{DDEXT}), and analog (A_{VDD}/A_{VSS}) power supplies. The internal and analog supplies must meet the 1.2 V requirement for K, B, and Y grade models, and the 1.0 V requirement for Y models. (For information on the temperature ranges offered for this product, see [Operating Conditions on Page 14](#), [Package Information on Page 16](#), and [Ordering Guide on Page 56](#).) The external supply must meet the 3.3 V requirement. All external supply pins must be connected to the same power supply.

Note that the analog supply pin (A_{VDD}) powers the processor's internal clock generator PLL. To produce a stable clock, it is recommended that PCB designs use an external filter circuit for the A_{VDD} pin. Place the filter components as close as possible to the A_{VDD}/A_{VSS} pins. For an example circuit, see [Figure 3](#). (A recommended ferrite chip is the muRata BLM18AG102SN1D.) To reduce noise coupling, the PCB should use a parallel pair of power and ground planes for V_{DDINT} and GND. Use wide traces to connect the bypass capacitors to the analog power (A_{VDD}) and ground (A_{VSS}) pins. Note that the A_{VDD} and A_{VSS} pins specified in [Figure 3](#) are inputs to the processor and not the analog ground plane on the board—the A_{VSS} pin should connect directly to digital ground (GND) at the chip.

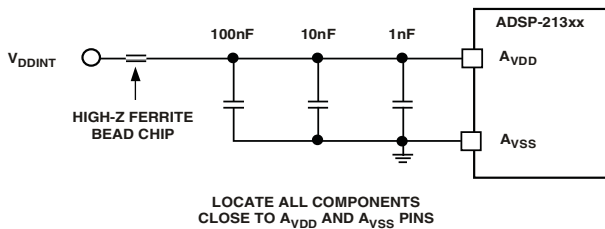


Figure 3. Analog Power (A_{VDD}) Filter Circuit

Target Board JTAG Emulator Connector

Analog Devices' DSP Tools product line of JTAG emulators uses the IEEE 1149.1 JTAG test access port of the processor to monitor and control the target board processor during emulation. Analog Devices' DSP Tools product line of JTAG emulators provides emulation at full processor speed, allowing inspection and modification of memory, registers, and processor stacks. The processor's JTAG interface ensures that the emulator does not affect target system loading or timing.

For complete information on Analog Devices' SHARC DSP Tools product line of JTAG emulator operation, refer to the appropriate emulator user's guide.

DEVELOPMENT TOOLS

Analog Devices supports its processors with a complete line of software and hardware development tools, including integrated development environments (which include CrossCore[®] Embedded Studio and/or VisualDSP++[®]), evaluation products, emulators, and a wide variety of software add-ins.

Integrated Development Environments (IDEs)

For C/C++ software writing and editing, code generation, and debug support, Analog Devices offers two IDEs.

The newest IDE, CrossCore Embedded Studio, is based on the Eclipse™ framework. Supporting most Analog Devices processor families, it is the IDE of choice for future processors, including multicore devices. CrossCore Embedded Studio seamlessly integrates available software add-ins to support real time operating systems, file systems, TCP/IP stacks, USB stacks, algorithmic software modules, and evaluation hardware board support packages. For more information visit www.analog.com/cces.

The other Analog Devices IDE, VisualDSP++, supports processor families introduced prior to the release of CrossCore Embedded Studio. This IDE includes the Analog Devices VDK real time operating system and an open source TCP/IP stack. For more information visit www.analog.com/visualdsp. Note that VisualDSP++ will not support future Analog Devices processors.

EZ-KIT Lite Evaluation Board

For processor evaluation, Analog Devices provides wide range of EZ-KIT Lite[®] evaluation boards. Including the processor and key peripherals, the evaluation board also supports on-chip emulation capabilities and other evaluation and development

features. Also available are various EZ-Extenders[®], which are daughter cards delivering additional specialized functionality, including audio and video processing. For more information visit www.analog.com and search on "ezkit" or "ezextender".

EZ-KIT Lite Evaluation Kits

For a cost-effective way to learn more about developing with Analog Devices processors, Analog Devices offer a range of EZ-KIT Lite evaluation kits. Each evaluation kit includes an EZ-KIT Lite evaluation board, directions for downloading an evaluation version of the available IDE(s), a USB cable, and a power supply. The USB controller on the EZ-KIT Lite board connects to the USB port of the user's PC, enabling the chosen IDE evaluation suite to emulate the on-board processor in-circuit. This permits the customer to download, execute, and debug programs for the EZ-KIT Lite system. It also supports in-circuit programming of the on-board Flash device to store user-specific boot code, enabling standalone operation. With the full version of CrossCore Embedded Studio or VisualDSP++ installed (sold separately), engineers can develop software for supported EZ-KITs or any custom system utilizing supported Analog Devices processors.

Software Add-Ins for CrossCore Embedded Studio

Analog Devices offers software add-ins which seamlessly integrate with CrossCore Embedded Studio to extend its capabilities and reduce development time. Add-ins include board support packages for evaluation hardware, various middleware packages, and algorithmic modules. Documentation, help, configuration dialogs, and coding examples present in these add-ins are viewable through the CrossCore Embedded Studio IDE once the add-in is installed.

Board Support Packages for Evaluation Hardware

Software support for the EZ-KIT Lite evaluation boards and EZ-Extender daughter cards is provided by software add-ins called Board Support Packages (BSPs). The BSPs contain the required drivers, pertinent release notes, and select example code for the given evaluation hardware. A download link for a specific BSP is located on the web page for the associated EZ-KIT or EZ-Extender product. The link is found in the *Product Download* area of the product web page.

Middleware Packages

Analog Devices separately offers middleware add-ins such as real time operating systems, file systems, USB stacks, and TCP/IP stacks. For more information see the following web pages:

- www.analog.com/ucos3
- www.analog.com/ucfs
- www.analog.com/ucusb
- www.analog.com/lwip

Algorithmic Modules

To speed development, Analog Devices offers add-ins that perform popular audio and video processing algorithms. These are available for use with both CrossCore Embedded Studio and

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ELECTRICAL CHARACTERISTICS

Parameter	Description	Test Conditions	Min	Max	Unit
V_{OH}^1	High Level Output Voltage	@ $V_{DDEXT} = \text{Min}$, $I_{OH} = -1.0 \text{ mA}^2$	2.4		V
V_{OL}^1	Low Level Output Voltage	@ $V_{DDEXT} = \text{Min}$, $I_{OL} = 1.0 \text{ mA}^2$		0.4	V
$I_{IH}^{3,4}$	High Level Input Current	@ $V_{DDEXT} = \text{Max}$, $V_{IN} = V_{DDEXT} \text{ Max}$		10	μA
I_{IL}^3	Low Level Input Current	@ $V_{DDEXT} = \text{Max}$, $V_{IN} = 0 \text{ V}$		10	μA
I_{ILPU}^4	Low Level Input Current Pull-Up	@ $V_{DDEXT} = \text{Max}$, $V_{IN} = 0 \text{ V}$		200	μA
$I_{OZH}^{5,6}$	Three-State Leakage Current	@ $V_{DDEXT} = \text{Max}$, $V_{IN} = V_{DDEXT} \text{ Max}$		10	μA
I_{OZL}^5	Three-State Leakage Current	@ $V_{DDEXT} = \text{Max}$, $V_{IN} = 0 \text{ V}$		10	μA
I_{OZLPU}^6	Three-State Leakage Current Pull-Up	@ $V_{DDEXT} = \text{Max}$, $V_{IN} = 0 \text{ V}$		200	μA
$I_{DD-INTYP}^{7,8}$	Supply Current (Internal)	$t_{CLK} = \text{Min}$, $V_{DDINT} = \text{Nom}$		800	mA
I_{AVDD}^9	Supply Current (Analog)	$A_{VDD} = \text{Max}$		10	mA
$C_{IN}^{10,11}$	Input Capacitance	$f_{IN} = 1 \text{ MHz}$, $T_{CASE} = 25^\circ\text{C}$, $V_{IN} = 1.2 \text{ V}$		4.7	pF

¹ Applies to output and bidirectional pins: AD15-0, \overline{RD} , \overline{WR} , ALE, FLAG3-0, DAI_Px, SPICLK, MOSI, MISO, \overline{EMU} , TDO, and XTAL.

² See [Output Drive Currents on Page 46](#) for typical drive current capabilities.

³ Applies to input pins: \overline{SPIDS} , BOOT_CFGx, CLK_CFGx, TCK, RESET, and CLKIN.

⁴ Applies to input pins with 22.5 k Ω internal pull-ups: \overline{TRST} , TMS, TDI.

⁵ Applies to three-stateable pins: FLAG3-0.

⁶ Applies to three-stateable pins with 22.5 k Ω pull-ups: AD15-0, DAI_Px, SPICLK, \overline{EMU} , MISO, and MOSI.

⁷ Typical internal current data reflects nominal operating conditions.

⁸ See the Engineer-to-Engineer Note "Estimating Power for the ADSP-21362 SHARC Processors" (EE-277) for further information.

⁹ Characterized, but not tested.

¹⁰ Applies to all signal pins.

¹¹ Guaranteed, but not tested.

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PACKAGE INFORMATION

The information presented in [Figure 4](#) provides details about the package branding for the ADSP-2136x processor. For a complete listing of product availability, see [Ordering Guide on Page 56](#).



Figure 4. Typical Package Brand

Table 7. Package Brand Information

Brand Key	Field Description
t	Temperature Range
pp	Package Type
Z	RoHS Compliant Designation
cc	See Ordering Guide
vvvvv.x	Assembly Lot Code
n.n	Silicon Revision
#	RoHS Compliant Designation
yyww	Date Code

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

MAXIMUM POWER DISSIPATION

See the Engineer-to-Engineer Note “*Estimating Power for the ADSP-21362 SHARC Processors*” (EE-277) for detailed thermal and power information regarding maximum power dissipation. For information on package thermal specifications, see [Thermal Characteristics on Page 47](#).

ABSOLUTE MAXIMUM RATINGS

Stresses greater than those listed in [Table 8](#) may cause permanent damage to the device. These are stress ratings only; functional operation of the device at these or any other conditions greater than those indicated in the operational sections of

this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Table 8. Absolute Maximum Ratings

Parameter	Rating
Internal (Core) Supply Voltage (V_{DDINT})	-0.3 V to +1.5 V
Analog (PLL) Supply Voltage (A_{VDD})	-0.3 V to +1.5 V
External (I/O) Supply Voltage (V_{DDEXT})	-0.3 V to +4.6 V
Input Voltage	-0.5 V to +3.8 V
Output Voltage Swing	-0.5 V to $V_{DDEXT} + 0.5$ V
Load Capacitance	200 pF
Storage Temperature Range	-65°C to +150°C
Junction Temperature While Biased	125°C

TIMING SPECIFICATIONS

Use the exact timing information given. Do not attempt to derive parameters from the addition or subtraction of others. While addition or subtraction would yield meaningful results for an individual device, the values given in this data sheet reflect statistical variations and worst cases. Consequently, it is not meaningful to add parameters to derive longer times. For voltage reference levels, see [Figure 39 on Page 46](#) under [Test Conditions](#).

Switching Characteristics specify how the processor changes its signals. Circuitry external to the processor must be designed for compatibility with these signal characteristics. Switching characteristics describe what the processor will do in a given circumstance. Use switching characteristics to ensure that any timing requirement of a device connected to the processor (such as memory) is satisfied.

Timing Requirements apply to signals that are controlled by circuitry external to the processor, such as the data input for a read operation. Timing requirements guarantee that the processor operates correctly with other devices.

Core Clock Requirements

The processor’s internal clock (a multiple of CLKIN) provides the clock signal for timing internal memory, processor core, and serial ports. During reset, program the ratio between the processor’s internal clock frequency and external (CLKIN) clock frequency with the CLK_CFG1–0 pins.

The processor’s internal clock switches at higher frequencies than the system input clock (CLKIN). To generate the internal clock, the processor uses an internal phase-locked loop (PLL, see [Figure 5](#)). This PLL-based clocking minimizes the skew between the system clock (CLKIN) signal and the processor’s internal clock.

Voltage Controlled Oscillator

In application designs, the PLL multiplier value should be selected in such a way that the VCO frequency never exceeds f_{VCO} specified in [Table 11](#).

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Power-Up Sequencing

The timing requirements for processor startup are given in Table 10. Note that during power-up, when the V_{DDINT} power supply comes up after V_{DDEXT} , a leakage current of the order of

three-state leakage current pull-up, pull-down, may be observed on any pin, even if that is an input only (for example the \overline{RESET} pin) until the V_{DDINT} rail has powered up.

Table 10. Power-Up Sequencing Timing Requirements (Processor Startup)

Parameter		Min	Max	Unit
<i>Timing Requirements</i>				
t_{RSTVDD}	\overline{RESET} Low Before V_{DDINT}/V_{DDEXT} On	0		ns
$t_{IVDDEVDD}$	V_{DDINT} On Before V_{DDEXT}	-50	+200	ms
t_{CLKVDD}^1	CLKIN Valid After V_{DDINT}/V_{DDEXT} Valid	0	200	ms
t_{CLKRST}	CLKIN Valid Before \overline{RESET} Deasserted	10^2		μ s
t_{PLLST}	PLL Control Setup Before \overline{RESET} Deasserted	20		μ s
<i>Switching Characteristic</i>				
$t_{CORERST}$	Core Reset Deasserted After \overline{RESET} Deasserted	$4096t_{CK} + 2 t_{CCLK}^{3,4}$		

¹ Valid V_{DDINT}/V_{DDEXT} assumes that the supplies are fully ramped to their 1.2 V rails and 3.3 V rails. Voltage ramp rates can vary from microseconds to hundreds of milliseconds, depending on the design of the power supply subsystem.

² Assumes a stable CLKIN signal, after meeting worst-case start-up timing of crystal oscillators. Refer to your crystal oscillator manufacturer's data sheet for start-up time. Assume a 25 ms maximum oscillator start-up time if using the XTAL pin and internal oscillator circuit in conjunction with an external crystal.

³ Applies after the power-up sequence is complete. Subsequent resets require a minimum of 4 CLKIN cycles for \overline{RESET} to be held low to properly initialize and propagate default states at all I/O pins.

⁴ The 4096 cycle count depends on t_{SRST} specification in Table 12. If setup time is not met, 1 additional CLKIN cycle can be added to the core reset time, resulting in 4097 cycles maximum.

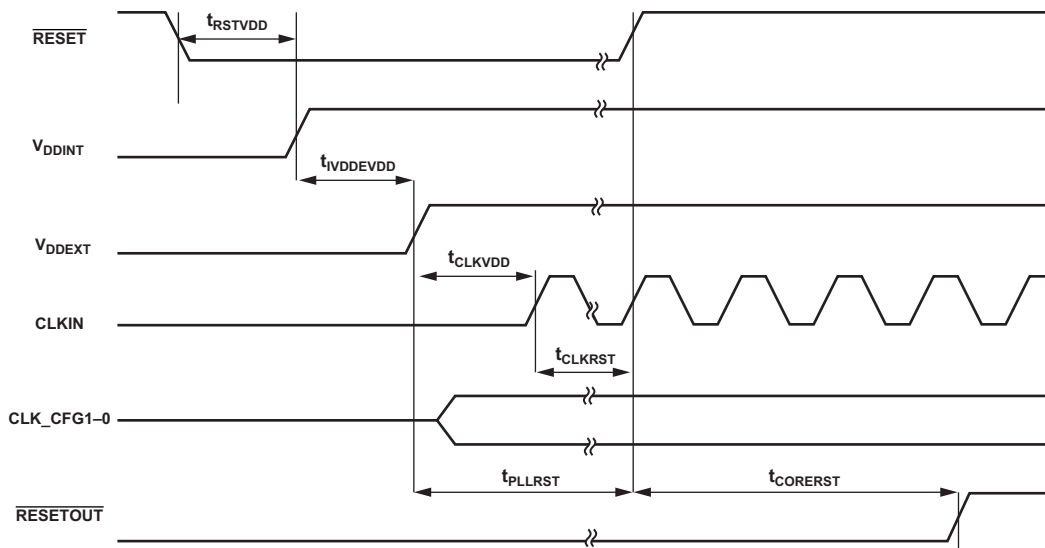


Figure 6. Power-Up Sequencing

Core Timer

The following timing specification applies to FLAG3 when it is configured as the core timer (TMREXP pin).

Table 14. Core Timer

Parameter	Min	Unit
<i>Switching Characteristic</i>		
t_{WCTIM} TMREXP Pulse Width	$2 \times t_{PCLK} - 1$	ns

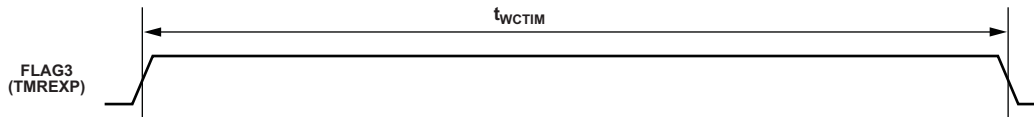


Figure 11. Core Timer

Timer PWM_OUT Cycle Timing

The following timing specification applies to Timer0, Timer1, and Timer2 in PWM_OUT (pulse-width modulation) mode. Timer signals are routed to the DAI_P20-1 pins through the SRU. Therefore, the timing specifications provided below are valid at the DAI_P20-1 pins.

Table 15. Timer PWM_OUT Timing

Parameter	Min	Max	Unit
<i>Switching Characteristic</i>			
t_{PWMO} Timer Pulse Width Output	$2 \times t_{PCLK} - 1$	$2 \times (2^{31} - 1) \times t_{PCLK}$	ns

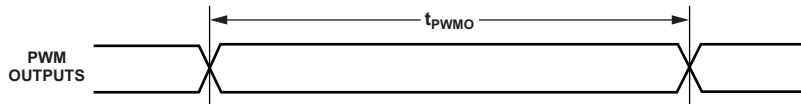


Figure 12. Timer PWM_OUT Timing

ADSP-21362/ADSP-21363/ADSP-21364/ADSP-21365/ADSP-21366

Table 21. 16-Bit Memory Read Cycle

Parameter	K and B Grade		Y Grade		Unit
	Min	Max	Min	Max	
<i>Timing Requirements</i>					
t_{DRS}	AD15-0 Data Setup Before \overline{RD} High		3.3	4.5	ns
t_{DRH}	AD15-0 Data Hold After \overline{RD} High		0	0	ns
<i>Switching Characteristics</i>					
t_{ALEW}	ALE Pulse Width		$2 \times t_{PCLK} - 2.0$	$2 \times t_{PCLK} - 2.0$	ns
t_{ADAS}^1	AD15-0 Address Setup Before ALE Deasserted		$t_{PCLK} - 2.5$	$t_{PCLK} - 2.5$	ns
t_{ALERW}	ALE Deasserted to Read Asserted		$2 \times t_{PCLK} - 3.8$	$2 \times t_{PCLK} - 3.8$	ns
t_{RRH}^2	Delay Between \overline{RD} Rising Edge to Next Falling Edge		$H + t_{PCLK} - 1.4$	$H + t_{PCLK} - 1.4$	ns
t_{RWALE}	Read Deasserted to ALE Asserted		$F + H + 0.5$	$F + H + 0.5$	ns
t_{RDDR}	ALE Address Drive After Read High		$F + H + t_{PCLK} - 2.3$	$F + H + t_{PCLK} - 2.3$	ns
t_{ADAH}^1	AD15-0 Address Hold After ALE Deasserted		$t_{PCLK} - 2.3$	$t_{PCLK} - 2.3$	ns
t_{ALEHZ1}	ALE Deasserted to Address/Data15-0 in High-Z		t_{PCLK}	$t_{PCLK} + 3.0$	ns
t_{RW}	\overline{RD} Pulse Width		$D - 2.0$	$D - 2.0$	ns

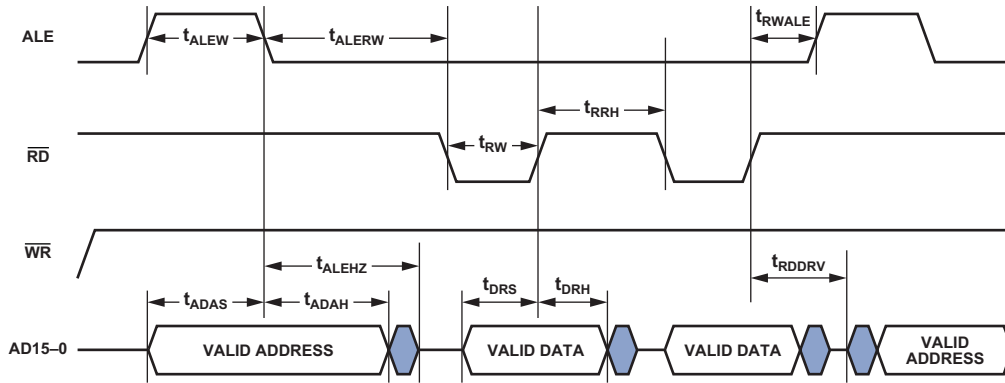
$D = (\text{The value set by the PPDUR Bits (5-1) in the PPCTL register}) \times t_{PCLK}$

$H = t_{PCLK}$ (if a hold cycle is specified, else $H = 0$)

$F = 7 \times t_{PCLK}$ (if FLASH_MODE is set, else $F = 0$)

¹ On reset, ALE is an active high cycle. However, it can be configured by software to be active low.

² This parameter is only available when in EMPP = 0 mode.



NOTE: FOR 16-BIT MEMORY READS, WHEN EMPP ≠ 0, ONLY ONE \overline{RD} PULSE OCCURS BETWEEN ALE CYCLES. WHEN EMPP = 0, MULTIPLE \overline{RD} PULSES OCCUR BETWEEN ALE CYCLES. FOR COMPLETE INFORMATION, SEE THE ADSP-2136x SHARC PROCESSOR HARDWARE REFERENCE.

Figure 18. Read Cycle for 16-Bit Memory Timing

Memory Write—Parallel Port

Use these specifications for asynchronous interfacing to memories (and memory-mapped peripherals) when the processor is accessing external memory space.

Table 22. 8-Bit Memory Write Cycle

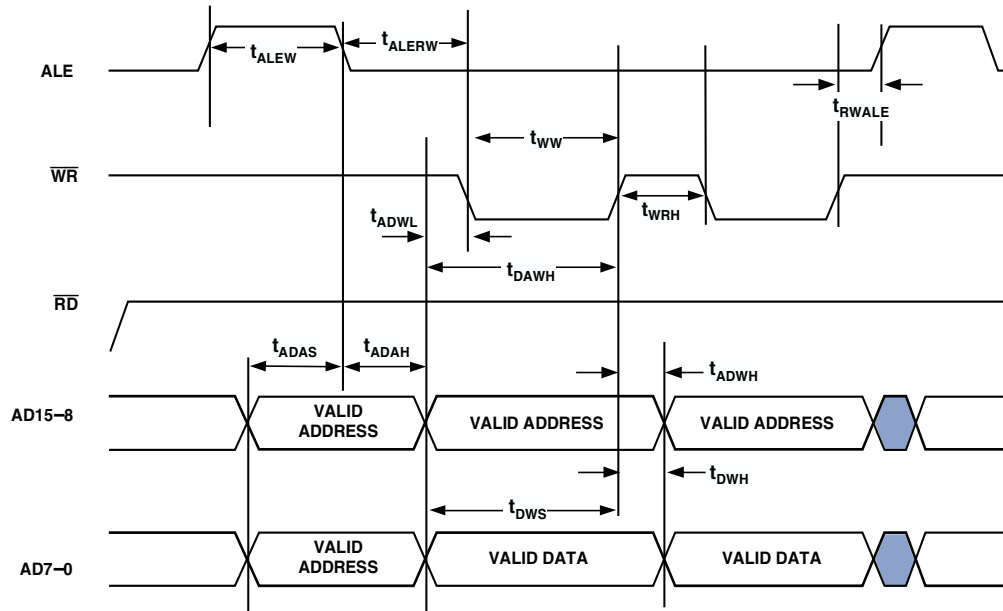
Parameter		K and B Grade	Y Grade	Unit
		Min	Min	
<i>Switching Characteristics</i>				
t_{ALEW}	ALE Pulse Width	$2 \times t_{PCLK} - 2.0$	$2 \times t_{PCLK} - 2.0$	ns
t_{ADAS}^1	AD15–0 Address Setup Before ALE Deasserted	$t_{PCLK} - 2.8$	$t_{PCLK} - 2.8$	ns
t_{ALERW}	ALE Deasserted to Write Asserted	$2 \times t_{PCLK} - 3.8$	$2 \times t_{PCLK} - 3.8$	ns
t_{RWALE}	Write Deasserted to ALE Asserted	$H + 0.5$	$H + 0.5$	ns
t_{WRH}	Delay Between \overline{WR} Rising Edge to Next \overline{WR} Falling Edge	$F + H + t_{PCLK} - 2.3$	$F + H + t_{PCLK} - 2.3$	ns
t_{ADAH}^1	AD15–0 Address Hold After ALE Deasserted	$t_{PCLK} - 0.5$	$t_{PCLK} - 0.5$	ns
t_{WW}	\overline{WR} Pulse Width	$D - F - 2.0$	$D - F - 2.0$	ns
t_{ADWL}	AD15–8 Address to \overline{WR} Low	$t_{PCLK} - 2.8$	$t_{PCLK} - 3.5$	ns
t_{ADWH}	AD15–8 Address Hold After \overline{WR} High	H	H	ns
t_{DWS}	AD7–0 Data Setup Before \overline{WR} High	$D - F + t_{PCLK} - 4.0$	$D - F + t_{PCLK} - 4.0$	ns
t_{DWH}	AD7–0 Data Hold After \overline{WR} High	H	H	ns
t_{DAWH}	AD15–8 Address to \overline{WR} High	$D - F + t_{PCLK} - 4.0$	$D - F + t_{PCLK} - 4.0$	ns

$D = (\text{The value set by the PPDUR Bits (5–1) in the PPCTL register}) \times t_{PCLK}$.

$H = t_{PCLK}$ (if a hold cycle is specified, else $H = 0$)

$F = 7 \times t_{PCLK}$ (if FLASH_MODE is set, else $F = 0$). If FLASH_MODE is set, D must be $\geq 9 \times t_{PCLK}$.

¹On reset, ALE is an active high cycle. However, it can be configured by software to be active low.



NOTE: MEMORY WRITES ALWAYS OCCUR IN GROUPS OF FOUR BETWEEN ALE CYCLES. THIS FIGURE SHOWS ONLY TWO MEMORY WRITES TO PROVIDE THE NECESSARY TIMING INFORMATION.

Figure 19. Write Cycle for 8-Bit Memory Timing

ADSP-21362/ADSP-21363/ADSP-21364/ADSP-21365/ADSP-21366

Serial Ports

To determine whether communication is possible between two devices at clock speed n , the following specifications must be confirmed: 1) frame sync (FS) delay and frame sync setup and hold, 2) data delay and data setup and hold, and 3) serial clock (SCLK) width.

Serial port signals are routed to the DAI_P20–1 pins using the SRU. Therefore, the timing specifications provided below are valid at the DAI_P20–1 pins.

Table 24. Serial Ports—External Clock

Parameter	K and B Grade		Y Grade	Unit
	Min	Max	Max	
<i>Timing Requirements</i>				
t_{SFSE}^1 Frame Sync Setup Before SCLK (Externally Generated Frame Sync in Either Transmit or Receive Mode)	2.5			ns
t_{HFSE}^1 Frame Sync Hold After SCLK (Externally Generated Frame Sync in Either Transmit or Receive Mode)	2.5			ns
t_{SDRE}^1 Receive Data Setup Before Receive SCLK	2.5			ns
t_{HDRE}^1 Receive Data Hold After SCLK	2.5			ns
t_{SCLKW} SCLK Width	$(t_{PCLK} \times 4) \div 2 - 2$			ns
t_{SCLK} SCLK Period	$t_{PCLK} \times 4$			ns
<i>Switching Characteristics</i>				
t_{DFSE}^2 Frame Sync Delay After SCLK (Internally Generated Frame Sync in Either Transmit or Receive Mode)		9.5	11	ns
t_{HOFSE}^2 Frame Sync Hold After SCLK (Internally Generated Frame Sync in Either Transmit or Receive Mode)	2			ns
t_{DDTE}^2 Transmit Data Delay After Transmit SCLK		9.5	11	ns
t_{HDTE}^2 Transmit Data Hold After Transmit SCLK	2			ns

¹Referenced to sample edge.

²Referenced to drive edge.

Table 25. Serial Ports—Internal Clock

Parameter	K and B Grade		Y Grade	Unit
	Min	Max	Max	
<i>Timing Requirements</i>				
t_{SFSI}^1 Frame Sync Setup Before SCLK (Externally Generated Frame Sync in Either Transmit or Receive Mode)	7			ns
t_{HFSI}^1 Frame Sync Hold After SCLK (Externally Generated Frame Sync in Either Transmit or Receive Mode)	2.5			ns
t_{SDRI}^1 Receive Data Setup Before SCLK	7			ns
t_{HDRI}^1 Receive Data Hold After SCLK	2.5			ns
<i>Switching Characteristics</i>				
t_{DFSI}^2 Frame Sync Delay After SCLK (Internally Generated Frame Sync in Transmit Mode)		3	3.5	ns
t_{HOFSI}^2 Frame Sync Hold After SCLK (Internally Generated Frame Sync in Transmit Mode)	-1.0			ns
t_{DFSIR}^2 Frame Sync Delay After SCLK (Internally Generated Frame Sync in Receive Mode)		8	9.5	ns
$t_{HOF SIR}^2$ Frame Sync Hold After SCLK (Internally Generated Frame Sync in Receive Mode)	-1.0			ns
t_{DDTI}^2 Transmit Data Delay After SCLK		3	4.0	ns
t_{HDTI}^2 Transmit Data Hold After SCLK	-1.0			ns
t_{SCLKIW} Transmit or Receive SCLK Width	$2 \times t_{PCLK} - 2$	$2 \times t_{PCLK} + 2$	$2 \times t_{PCLK} + 2$	ns

¹Referenced to the sample edge.

²Referenced to drive edge.

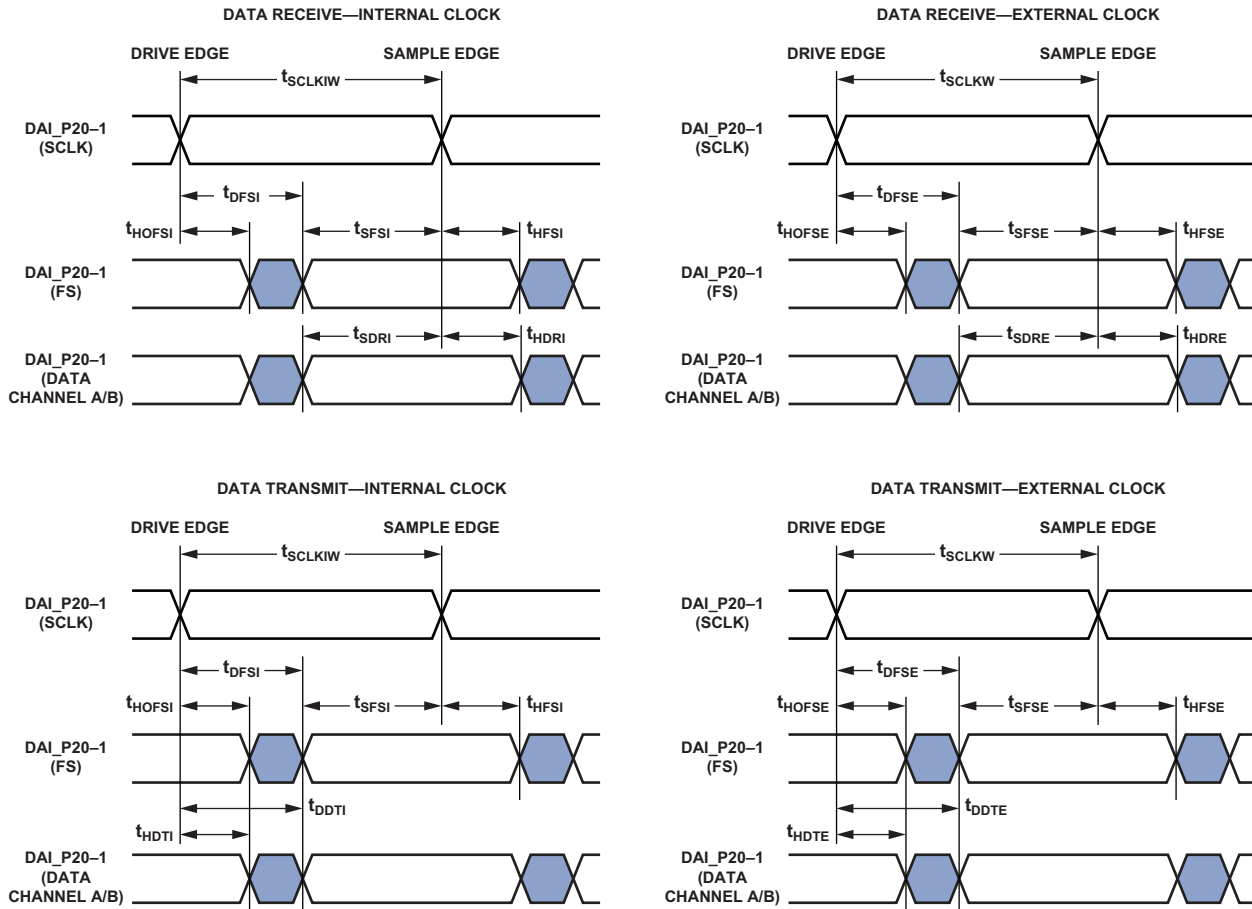


Figure 21. Serial Ports

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Parallel Data Acquisition Port (PDAP)

The timing requirements for the PDAP are provided in Table 29. PDAP is the parallel mode operation of Channel 0 of the IDP. For details on the operation of the IDP, refer to the *ADSP-2136x SHARC Processor Hardware Reference*, “Input Data Port” chapter.

Note that the most significant 16 bits of external 20-bit PDAP data can be provided through either the parallel port AD15–0 pins or the DAI_P20–5 pins. The remaining 4 bits can only be sourced through DAI_P4–1. The timing below is valid at the DAI_P20–1 pins or at the AD15–0 pins.

Table 29. Parallel Data Acquisition Port (PDAP)

Parameter		Min	Unit
<i>Timing Requirements</i>			
$t_{SPCLKEN}^1$	PDAP_CLKEN Setup Before PDAP_CLK Sample Edge	2.5	ns
$t_{HPCLKEN}^1$	PDAP_CLKEN Hold After PDAP_CLK Sample Edge	2.5	ns
t_{PDSD}^1	PDAP_DAT Setup Before SCLK PDAP_CLK Sample Edge	3.0	ns
t_{PDHD}^1	PDAP_DAT Hold After SCLK PDAP_CLK Sample Edge	2.5	ns
t_{PDCLKW}	Clock Width	$(t_{PCLK} \times 4) \div 2 - 3$	ns
t_{PDCLK}	Clock Period	$t_{PCLK} \times 4$	ns
<i>Switching Characteristics</i>			
t_{PDHLDD}	Delay of PDAP Strobe After Last PDAP_CLK Capture Edge for a Word	$2 \times t_{PCLK} - 1$	ns
t_{PDSTRB}	PDAP Strobe Pulse Width	$2 \times t_{PCLK} - 1.5$	ns

¹Data source pins are AD15–0 and DAI_P4–1, or DAI pins. Source pins for serial clock and frame sync are DAI pins.

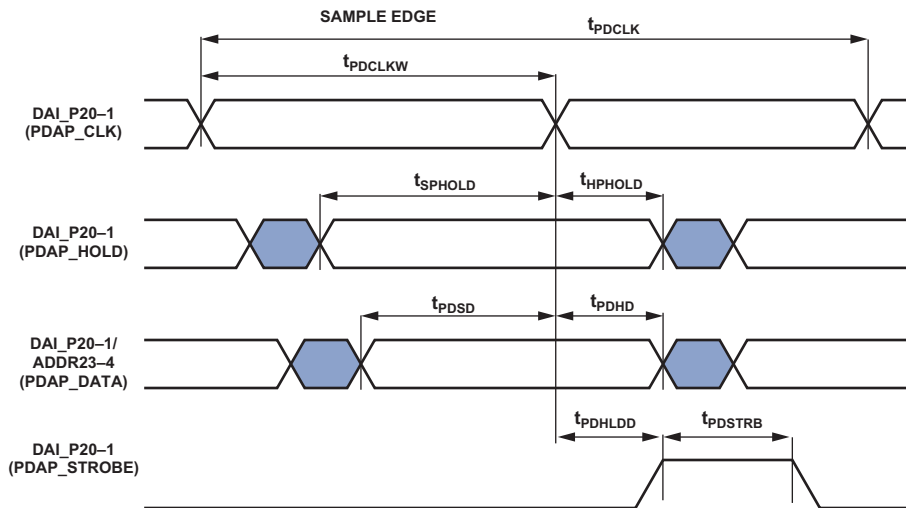


Figure 25. PDAP Timing

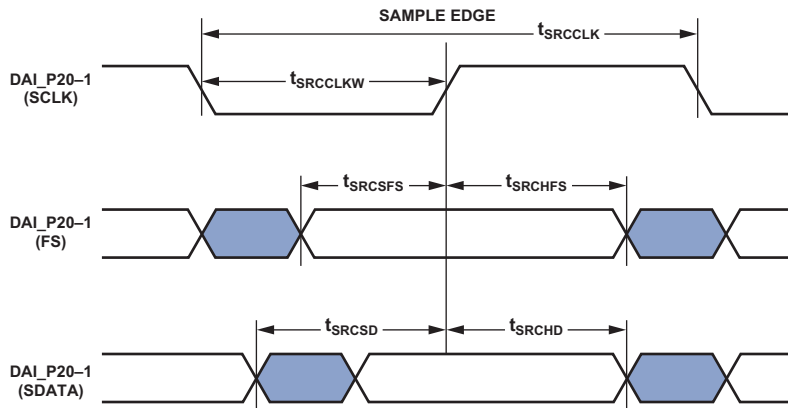


Figure 27. SRC Serial Input Port Timing

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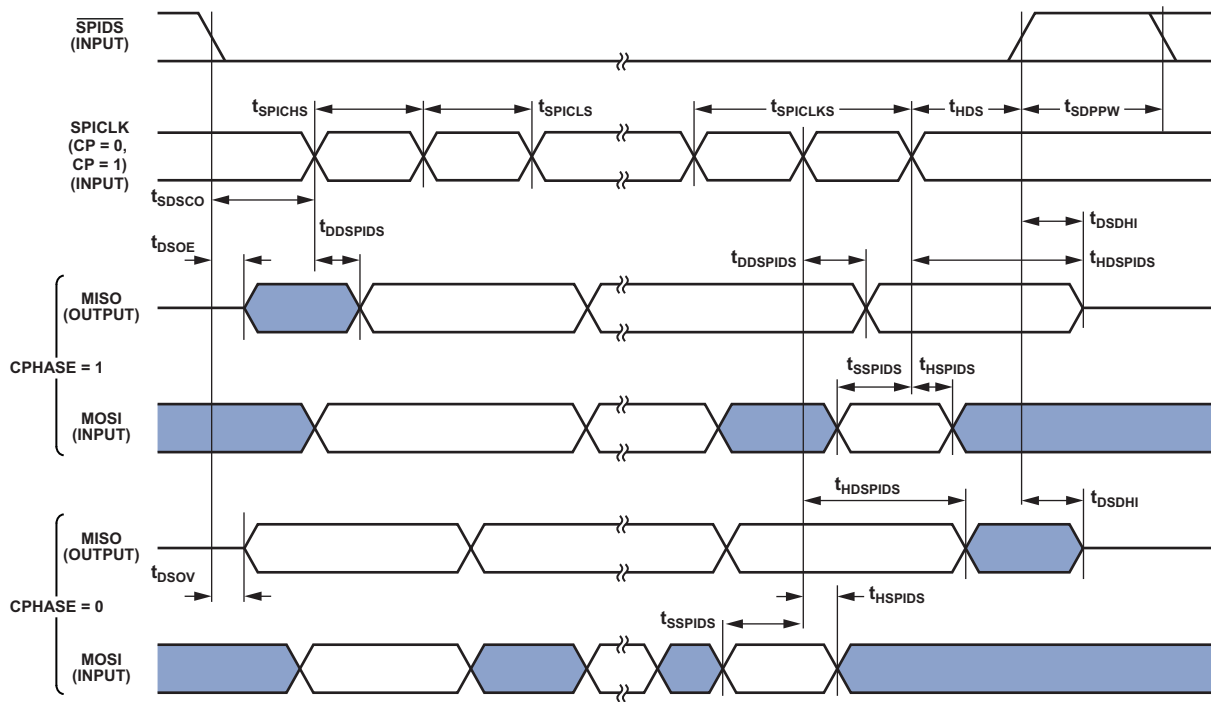


Figure 35. SPI Slave Timing

JTAG Test Access Port and Emulation

Table 41. JTAG Test Access Port and Emulation

Parameter	Min	Max	Unit
<i>Timing Requirements</i>			
t_{TCK} TCK Period	t_{CK}		ns
t_{STAP} TDI, TMS Setup Before TCK High	5		ns
t_{HTAP} TDI, TMS Hold After TCK High	6		ns
t_{SSYS}^1 System Inputs Setup Before TCK High	7		ns
t_{HSYS}^1 System Inputs Hold After TCK High	18		ns
t_{TRSTW} \overline{TRST} Pulse Width	$4 \times t_{CK}$		ns
<i>Switching Characteristics</i>			
t_{DTDO} TDO Delay from TCK Low		7	ns
t_{DSYS}^2 System Outputs Delay After TCK Low		$t_{CK} \div 2 + 7$	ns

¹System Inputs = ADDR15-0, \overline{SPIDS} , CLK_CFG1-0, \overline{RESET} , BOOT_CFG1-0, MISO, MOSI, SPICLK, DAI_Px, and FLAG3-0.

²System Outputs = MISO, MOSI, SPICLK, DAI_Px, ADDR15-0, \overline{RD} , \overline{WR} , FLAG3-0, \overline{EMU} , and ALE.

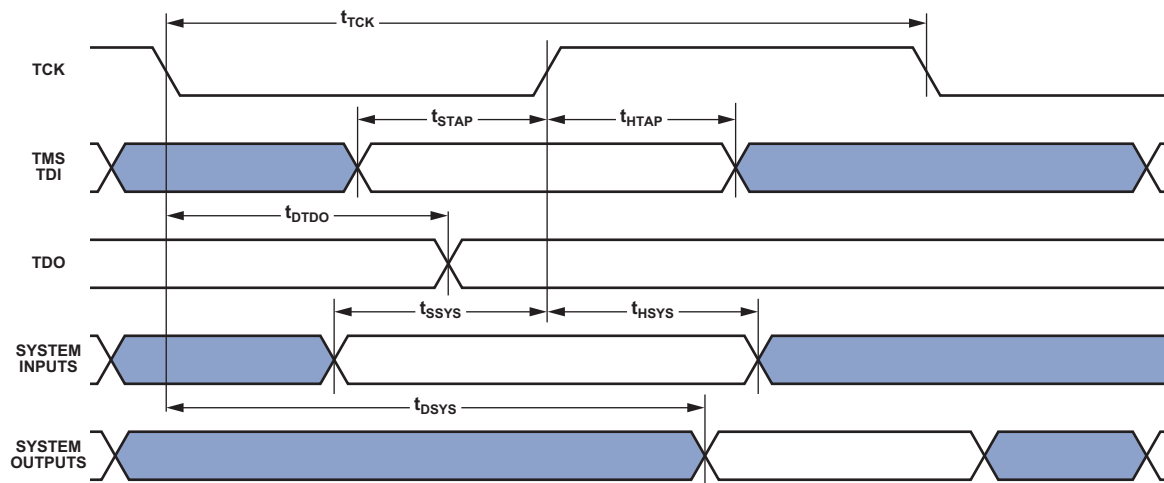
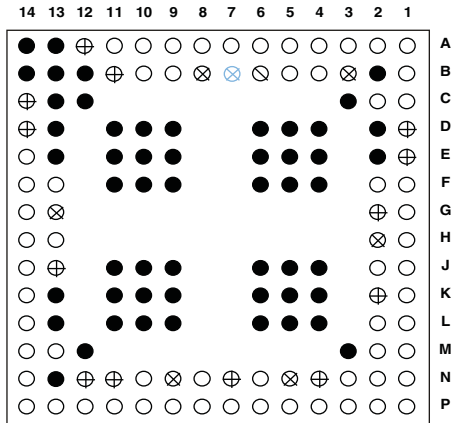


Figure 36. IEEE 1149.1 JTAG Test Access Port

ADSP-21362/ADSP-21363/ADSP-21364/ADSP-21365/ADSP-21366



KEY

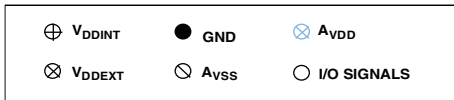
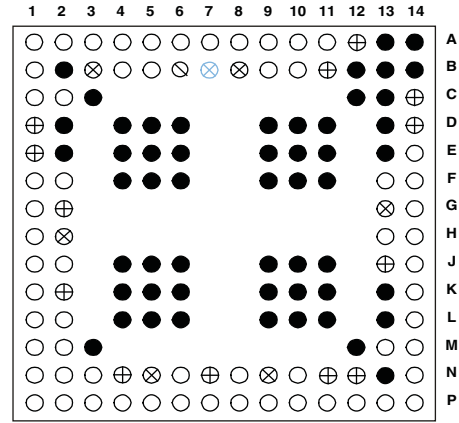


Figure 45. BGA Pin Assignments (Bottom View, Summary)



KEY

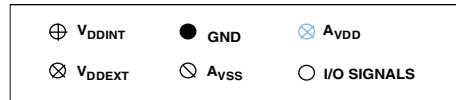


Figure 46. BGA Pin Assignments (Top View, Summary)

ADSP-21362/ADSP-21363/ADSP-21364/ADSP-21365/ADSP-21366

ORDERING GUIDE

Model ¹	Notes	Temperature Range ²	Instruction Rate	On-Chip SRAM	ROM	Package Description	Package Option
ADSP-21363KBC-1AA		0°C to +70°C	333 MHz	3M Bit	4M Bit	136-Ball CSP_BGA	BC-136-1
ADSP-21363KBCZ-1AA		0°C to +70°C	333 MHz	3M Bit	4M Bit	136-Ball CSP_BGA	BC-136-1
ADSP-21363KSWZ-1AA		0°C to +70°C	333 MHz	3M Bit	4M Bit	144-Lead LQFP_EP	SW-144-1
ADSP-21363BBC-1AA		-40°C to +85°C	333 MHz	3M Bit	4M Bit	136-Ball CSP_BGA	BC-136-1
ADSP-21363BBCZ-1AA		-40°C to +85°C	333 MHz	3M Bit	4M Bit	136-Ball CSP_BGA	BC-136-1
ADSP-21363BSWZ-1AA		-40°C to +85°C	333 MHz	3M Bit	4M Bit	144-Lead LQFP_EP	SW-144-1
ADSP-21363YSWZ-2AA	3	-40°C to +105°C	200 MHz	3M Bit	4M Bit	144-Lead LQFP_EP	SW-144-1
ADSP-21364KBCZ-1AA		0°C to +70°C	333 MHz	3M Bit	4M Bit	136-Ball CSP_BGA	BC-136-1
ADSP-21364KSWZ-1AA		0°C to +70°C	333 MHz	3M Bit	4M Bit	144-Lead LQFP_EP	SW-144-1
ADSP-21364BBCZ-1AA		-40°C to +85°C	333 MHz	3M Bit	4M Bit	136-Ball CSP_BGA	BC-136-1
ADSP-21364BSWZ-1AA		-40°C to +85°C	333 MHz	3M Bit	4M Bit	144-Lead LQFP_EP	SW-144-1
ADSP-21364YSWZ-2AA		-40°C to +105°C	200 MHz	3M Bit	4M Bit	144-Lead LQFP_EP	SW-144-1
ADSP-21366KBCZ-1AR	3, 4, 5	0°C to +70°C	333 MHz	3M Bit	4M Bit	136-Ball CSP_BGA	BC-136-1
ADSP-21366KBCZ-1AA	3, 4	0°C to +70°C	333 MHz	3M Bit	4M Bit	136-Ball CSP_BGA	BC-136-1
ADSP-21366KSWZ-1AA	3, 4	0°C to +70°C	333 MHz	3M Bit	4M Bit	144-Lead LQFP_EP	SW-144-1

¹Z = RoHS compliant part.

²Referenced temperature is ambient temperature. The ambient temperature is not a specification. Please see [Operating Conditions on Page 14](#) for junction temperature (T_j) specification which is the only temperature specification.

³License from Dolby Laboratories, Inc., and Digital Theater Systems (DTS) required for these products.

⁴Available with a wide variety of audio algorithm combinations sold as part of a chipset and bundled with necessary software. For a complete list, visit our website at www.analog.com/sharc.

⁵R = Tape and reel.

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