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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

Details	
Product Status	Active
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	CANbus, I <sup>2</sup> C, LINbus, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	39
Program Memory Size	128KB (128K x 8)
Program Memory Type	FLASH
EEPROM Size	2K x 8
RAM Size	8K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 16x12b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-LQFP
Supplier Device Package	48-LQFP (7x7)
Purchase URL	<a href="https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=mc9s08dz128clf">https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=mc9s08dz128clf</a>

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**Debug Module (S08DBGV3) (128K)**

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## 2.2.1 Power

$V_{DD}$  and  $V_{SS}$  are the primary power supply pins for the MCU. This voltage source supplies power to all I/O buffer circuitry and to an internal voltage regulator. The internal voltage regulator provides regulated lower-voltage source to the CPU and other internal circuitry of the MCU.

Typically, application systems have two separate capacitors across the power pins. In this case, there should be a bulk electrolytic capacitor, such as a 10- $\mu$ F tantalum capacitor, to provide bulk charge storage for the overall system and a 0.1- $\mu$ F ceramic bypass capacitor located as near to the MCU power pins as practical to suppress high-frequency noise. The MC9S08DZ128 Series has up to three  $V_{DD}$  pins. Each pin must have a bypass capacitor for best noise suppression.

$V_{DDA}$  and  $V_{SSA}$  are the analog power supply pins for the MCU. This voltage source supplies power to the ADC module. A 0.1- $\mu$ F ceramic bypass capacitor should be located as near to the MCU power pins as practical to suppress high-frequency noise.

## 2.2.2 Oscillator

Immediately after reset, the MCU uses an internally generated clock provided by the multi-purpose clock generator (MCG) module. For more information on the MCG, see [Chapter 8, “Multi-Purpose Clock Generator \(S08MCGV2\).”](#)

The oscillator (XOSC) in this MCU is a Pierce oscillator that can accommodate a crystal or ceramic resonator. Rather than a crystal or ceramic resonator, an external oscillator can be connected to the EXTAL input pin.

Refer to [Figure 2-4](#) for the following discussion.  $R_S$  (when used) and  $R_F$  should be low-inductance resistors such as carbon composition resistors. Wire-wound resistors and some metal film resistors have too much inductance. C1 and C2 normally should be high-quality ceramic capacitors that are specifically designed for high-frequency applications.

$R_F$  is used to provide a bias path to keep the EXTAL input in its linear range during crystal startup; its value is not generally critical. Typical systems use 1 M $\Omega$  to 10 M $\Omega$ . Higher values are sensitive to humidity, and lower values reduce gain and (in extreme cases) could prevent startup.

C1 and C2 are typically in the 5-pF to 25-pF range and are chosen to match the requirements of a specific crystal or resonator. Be sure to take into account printed circuit board (PCB) capacitance and MCU pin capacitance when selecting C1 and C2. The crystal manufacturer typically specifies a load capacitance which is the series combination of C1 and C2 (which are usually the same size). As a first-order approximation, use 10 pF as an estimate of combined pin and PCB capacitance for each oscillator pin (EXTAL and XTAL).

## 2.2.3 $\overline{\text{RESET}}$

$\overline{\text{RESET}}$  is a dedicated pin with a pull-up device built in. It has input hysteresis, a high current output driver, and no output slew rate control. Internal power-on reset and low-voltage reset circuitry typically make external reset circuitry unnecessary. This pin is normally connected to the standard 6-pin background debug connector so a development system can directly reset the MCU system. If desired, a manual external reset can be added by supplying a simple switch to ground (pull reset pin low to force a reset).

## 4.3 Register Addresses and Bit Assignments

The registers in the MC9S08DZ128 Series are divided into these groups:

- Direct-page registers are located in the first 128 locations in the memory map; these are accessible with efficient direct addressing mode instructions.
- High-page registers are used much less often, so they are located above 0x1800 in the memory map. This leaves more room in the direct page for more frequently used registers and RAM.
- The nonvolatile register area consists of a block of 16 locations in FLASH memory at 0xFFB0–0xFFBF. Nonvolatile register locations include:
  - NVPROT and NVOPT are loaded into working registers at reset
  - An 8-byte backdoor comparison key that optionally allows a user to gain controlled access to secure memory

Because the nonvolatile register locations are FLASH memory, they must be erased and programmed like other FLASH memory locations.

Direct-page registers can be accessed with efficient direct addressing mode instructions. Bit manipulation instructions can be used to access any bit in any direct-page register. [Table 4-2](#) is a summary of all user-accessible direct-page registers and control bits.

The direct page registers in [Table 4-2](#) can use the more efficient direct addressing mode, which requires only the lower byte of the address. Because of this, the lower byte of the address in column one is shown in bold text. In [Table 4-3](#) and [Table 4-5](#), the whole address in column one is shown in bold. In [Table 4-2](#), [Table 4-3](#), and [Table 4-5](#), the register names in column two are shown in bold to set them apart from the bit names to the right. Cells that are not associated with named bits are shaded. A shaded cell with a 0 indicates this unused bit always reads as a 0. Shaded cells with dashes indicate unused or reserved bit locations that could read as 1s or 0s.

debug interface) and verifying that FLASH is blank. To avoid returning to secure mode after the next reset, program the security bits (SEC) to the unsecured state (1:0).

## 4.4 Memory Management Unit

The memory management unit (MMU) allows the program and data space for the HCS08 Family of Microcontrollers to be extended beyond the 64K CPU addressable memory map. The extended memory when used for data can also be accessed linearly using a linear address pointer and data access registers.

### 4.4.1 Features

Key features of the MMU module are:

- Memory Management Unit extends the HCS08 memory space
  - up to 128K for program and data space
- Extended program space using paging scheme
  - PPAGE register used for page selection
  - fixed 16K byte memory window
  - architecture supports eight 16K pages
- Extended data space using linear address pointer
  - 17-bit linear address pointer
  - linear address pointer and data register provided in direct page allows access of complete FLASH memory map using direct page instructions
  - optional auto increment of pointer when data accessed
  - supports a 2s complement addition/subtraction to address pointer without using any math instructions or memory resources
  - supports word accesses to any address specified by the linear address pointer when using LDHX, STHX instructions

### 4.4.2 Memory Expansion

The HCS08 Core architecture limits the CPU addressable space available to 64K bytes. The Program Page (PPAGE) allows for integrating up to 128K of FLASH into the system by selecting one of the 16K byte blocks to be accessed through the Paging Window located at 0x8000-0xBFFF. The MMU module also provides a linear address pointer that allows extension of data access up to 128K.

#### 4.4.2.1 Program Space

The PPAGE register holds the page select value for the Paging Window. The value in PPAGE can be manipulated by using normal read and write instructions as well as the CALL and RTC instructions. The user should not change PPAGE directly when running from paged memory, only CALL and RTC should be used.

### 6.5.2.3 Port B Pull Enable Register (PTBPE)

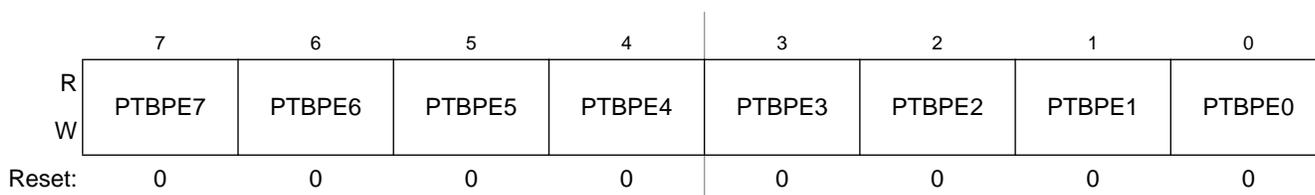


Figure 6-13. Internal Pull Enable for Port B Register (PTBPE)

Table 6-11. PTBPE Register Field Descriptions

Field	Description
7:0 PTBPE[7:0]	<p><b>Internal Pull Enable for Port B Bits</b> — Each of these control bits determines if the internal pull-up or pull-down device is enabled for the associated PTB pin. For port B pins that are configured as outputs, these bits have no effect and the internal pull devices are disabled.</p> <p>0 Internal pull-up/pull-down device disabled for port B bit n.                      1 Internal pull-up/pull-down device enabled for port B bit n.</p>

**NOTE**

Pull-down devices only apply when using pin interrupt functions, when corresponding edge select and pin select functions are configured.

### 6.5.2.4 Port B Slew Rate Enable Register (PTBSE)

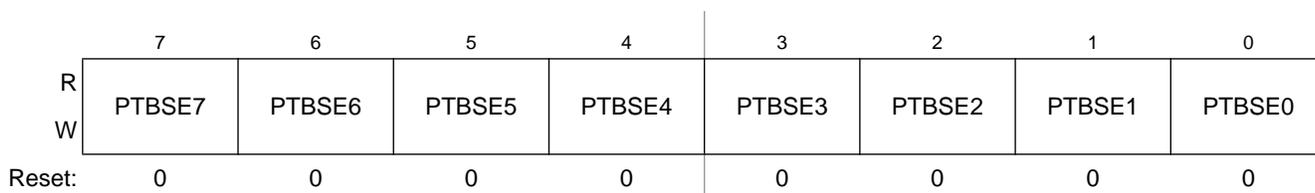


Figure 6-14. Slew Rate Enable for Port B Register (PTBSE)

Table 6-12. PTBSE Register Field Descriptions

Field	Description
7:0 PTBSE[7:0]	<p><b>Output Slew Rate Enable for Port B Bits</b> — Each of these control bits determines if the output slew rate control is enabled for the associated PTB pin. For port B pins that are configured as inputs, these bits have no effect.</p> <p>0 Output slew rate control disabled for port B bit n.                      1 Output slew rate control enabled for port B bit n.</p>

**Note:** Slew rate reset default values may differ between engineering samples and final production parts. Always initialize slew rate control to the desired value to ensure correct operation.

### 6.5.8.3 Port H Pull Enable Register (PTHPE)

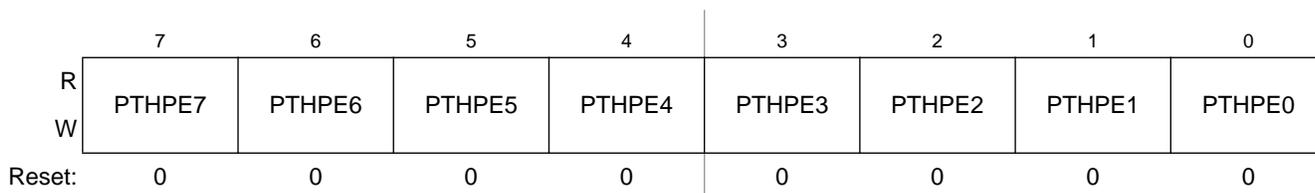


Figure 6-49. Internal Pull Enable for Port H Register (PTHPE)

Table 6-47. PTHPE Register Field Descriptions

Field	Description
7:0 PTHPE[7:0]	<p><b>Internal Pull Enable for Port H Bits</b> — Each of these control bits determines if the internal pull-up device is enabled for the associated PTH pin. For port H pins that are configured as outputs, these bits have no effect and the internal pull devices are disabled.</p> <p>0 Internal pull-up device disabled for port H bit n. 1 Internal pull-up device enabled for port H bit n.</p>

**NOTE**

Pull-down devices only apply when using pin interrupt functions, when corresponding edge select and pin select functions are configured.

### 6.5.8.4 Port H Slew Rate Enable Register (PTHSE)

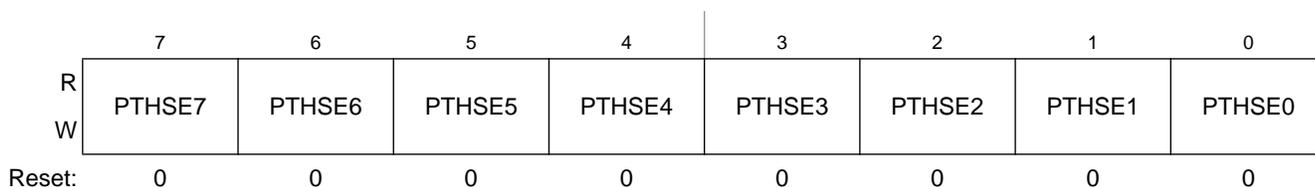


Figure 6-50. Slew Rate Enable for Port H Register (PTHSE)

Table 6-48. PTHSE Register Field Descriptions

Field	Description
7:0 PTHSE[7:0]	<p><b>Output Slew Rate Enable for Port H Bits</b> — Each of these control bits determines if the output slew rate control is enabled for the associated PTH pin. For port H pins that are configured as inputs, these bits have no effect.</p> <p>0 Output slew rate control disabled for port H bit n. 1 Output slew rate control enabled for port H bit n.</p>

**Note:** Slew rate reset default values may differ between engineering samples and final production parts. Always initialize slew rate control to the desired value to ensure correct operation.

### 7.3.6.5 Indexed, 16-Bit Offset (IX2)

This variation of indexed addressing uses the 16-bit value in the H:X index register pair plus a 16-bit offset included in the instruction as the address of the operand needed to complete the instruction.

### 7.3.6.6 SP-Relative, 8-Bit Offset (SP1)

This variation of indexed addressing uses the 16-bit value in the stack pointer (SP) plus an unsigned 8-bit offset included in the instruction as the address of the operand needed to complete the instruction.

### 7.3.6.7 SP-Relative, 16-Bit Offset (SP2)

This variation of indexed addressing uses the 16-bit value in the stack pointer (SP) plus a 16-bit offset included in the instruction as the address of the operand needed to complete the instruction.

## 7.4 Special Operations

The CPU performs a few special operations that are similar to instructions but do not have opcodes like other CPU instructions. In addition, a few instructions such as STOP and WAIT directly affect other MCU circuitry. This section provides additional information about these operations.

### 7.4.1 Reset Sequence

Reset can be caused by a power-on-reset (POR) event, internal conditions such as the COP (computer operating properly) watchdog, or by assertion of an external active-low reset pin. When a reset event occurs, the CPU immediately stops whatever it is doing (the MCU does not wait for an instruction boundary before responding to a reset event). For a more detailed discussion about how the MCU recognizes resets and determines the source, refer to the [Resets, Interrupts, and System Configuration](#) chapter.

The reset event is considered concluded when the sequence to determine whether the reset came from an internal source is done and when the reset pin is no longer asserted. At the conclusion of a reset event, the CPU performs a 6-cycle sequence to fetch the reset vector from 0xFFFFE and 0xFFFF and to fill the instruction queue in preparation for execution of the first program instruction.

### 7.4.2 Interrupt Sequence

When an interrupt is requested, the CPU completes the current instruction before responding to the interrupt. At this point, the program counter is pointing at the start of the next instruction, which is where the CPU should return after servicing the interrupt. The CPU responds to an interrupt by performing the same sequence of operations as for a software interrupt (SWI) instruction, except the address used for the vector fetch is determined by the highest priority interrupt that is pending when the interrupt sequence started.

The CPU sequence for an interrupt is:

1. Store the contents of PCL, PCH, X, A, and CCR on the stack, in that order.
2. Set the I bit in the CCR.

while the CPU is in stop mode, CPU clocks will resume and the CPU will enter active background mode where other serial background commands can be processed. This ensures that a host development system can still gain access to a target MCU even if it is in stop mode.

Recovery from stop mode depends on the particular HCS08 and whether the oscillator was stopped in stop mode. Refer to the [Modes of Operation](#) chapter for more details.

### 7.4.5 BGND Instruction

The BGND instruction is new to the HCS08 compared to the M68HC08. BGND would not be used in normal user programs because it forces the CPU to stop processing user instructions and enter the active background mode. The only way to resume execution of the user program is through reset or by a host debug system issuing a GO, TRACE1, or TAGGO serial command through the background debug interface.

Software-based breakpoints can be set by replacing an opcode at the desired breakpoint address with the BGND opcode. When the program reaches this breakpoint address, the CPU is forced to active background mode rather than continuing the user program.

## 7.5 CALL and RTC Instructions

The CALL is similar to a jump-to-subroutine (JSR) instruction, but the subroutine that is called can be located anywhere in the normal 64-Kbyte address space or on any page of program expansion memory. When CALL is executed, a return address is calculated, then it and the current program page register value are stacked, and a new instruction-supplied value is written to PPAGE. The PPAGE value controls which of the possible 16-Kbyte pages is visible through the window in the 64-Kbyte memory map. Execution continues at the address of the called subroutine.

The actual sequence of operations that occur during execution of CALL is:

1. CPU calculates the address of the next instruction after the CALL instruction (the return address) and pushes this 16-bit value onto the stack, low byte first.
2. CPU reads the old PPAGE value and pushes it onto the stack.
3. CPU writes the new instruction-supplied page select value to PPAGE. This switches the destination page into the program overlay window in the CPU address range 0x8000 0xBFFF.
4. Instruction queue is refilled starting from the destination address, and execution begins at the new address.

This sequence of operations is an uninterruptable CPU instruction. There is no need to inhibit interrupts during CALL execution. In addition, a CALL can be performed from any address in memory to any other address. This is a big improvement over other bank-switching schemes, where the page switch operation can be performed only by a program outside the overlay window.

For all practical purposes, the PPAGE value supplied by the instruction can be considered to be part of the effective address. The new page value is provided by an immediate operand in the instruction.



## 8.1.1 Features

Key features of the MCG module are:

- Frequency-locked loop (FLL)
  - Internal or external reference can be used to control the FLL
- Phase-locked loop (PLL)
  - Voltage-controlled oscillator (VCO)
  - Modulo VCO frequency divider
  - Phase/Frequency detector
  - Integrated loop filter
  - Lock detector with interrupt capability
- Internal reference clock
  - Nine trim bits for accuracy
  - Can be selected as the clock source for the MCU
- External reference clock
  - Control for external oscillator
  - Clock monitor with reset capability
  - Can be selected as the clock source for the MCU
- Reference divider is provided
- Clock source selected can be divided down by 1, 2, 4, or 8
- BDC clock (MCGLCLK) is provided as a constant divide by 2 of the DCO output whether in an FLL or PLL mode.
- Two selectable digitally controlled oscillators (DCOs) optimized for different frequency ranges.
- Option to maximize DCO output frequency for a 32,768 Hz external reference clock source.

reference can achieve a high-range maximum DCO output of 39.85 MHz with a multiplier of 1216. When the DRS bit is clear, the 32.768 kHz reference can achieve a mid-range maximum DCO output of 19.92 MHz with a multiplier of 608.

In FBI and FEI modes, setting the DMX32 bit is not recommended. If the internal reference is trimmed to a frequency above 32.768 kHz, the greater FLL multiplication factor could potentially push the microcontroller system clock out of specification and damage the part.

### 8.5.3 MCG Mode Switching

When switching between operational modes of the MCG, certain configuration bits must be changed in order to properly move from one mode to another. Each time any of these bits are changed (PLLS, IREFS, CLKS, or EREFS), the corresponding bits in the MCGSC register (PLLST, IREFST, CLKST, or OSCINIT) must be checked before moving on in the application software.

Additionally, care must be taken to ensure that the reference clock divider (RDIV) is set properly for the mode being switched to. For instance, in PEE mode, if using a 4 MHz crystal, RDIV must be set to %001 (divide-by-2) or %010 (divide -by-4) in order to divide the external reference down to the required frequency between 1 and 2 MHz.

If switching to FBE or FEE mode, first setting the DIV32 bit will ensure a proper reference frequency is sent to the FLL clock at all times.

In FBE, FEE, FBI, and FEI modes, at any time, the application can switch the FLL multiplication factor between 1024 and 512 with the DRS bit in MCGT. Writes to DRS will be ignored if LP=1 or PLLS=1.

The RDIV and IREFS bits should always be set properly before changing the PLLS bit so that the FLL or PLL clock has an appropriate reference clock frequency to switch to. The table below shows MCGOUT frequency calculations using RDIV, BDIV, and VDIV settings for each clock mode. The bus frequency is equal to MCGOUT divided by 2.

**Table 8-10. MCGOUT Frequency Calculation Options**

Clock Mode	$f_{MCGOUT}^1$	Note
FEI (FLL engaged internal)	$(f_{int} * F) / B$	Typical $f_{MCGOUT} = 16$ MHz immediately after reset.
FEE (FLL engaged external)	$(f_{ext} / R * F) / B$	$f_{ext} / R$ must be in the range of 31.25 kHz to 39.0625 kHz
FBE (FLL bypassed external)	$f_{ext} / B$	$f_{ext} / R$ must be in the range of 31.25 kHz to 39.0625 kHz
FBI (FLL bypassed internal)	$f_{int} / B$	Typical $f_{int} = 32$ kHz
PEE (PLL engaged external)	$[(f_{ext} / R) * M] / B$	$f_{ext} / R$ must be in the range of 1 MHz to 2 MHz
PBE (PLL bypassed external)	$f_{ext} / B$	$f_{ext} / R$ must be in the range of 1 MHz to 2 MHz
BLPI (Bypassed low power internal)	$f_{int} / B$	
BLPE (Bypassed low power external)	$f_{ext} / B$	

## 9.1.2 Features

The ACMP has the following features:

- Full rail to rail supply operation.
- Selectable interrupt on rising edge, falling edge, or either rising or falling edges of comparator output.
- Option to compare to fixed internal bandgap reference voltage.
- Option to allow comparator output to be visible on a pin, ACMPxO.
- Can operate in stop3 mode

## 9.1.3 Modes of Operation

This section defines the ACMP operation in wait, stop and background debug modes.

### 9.1.3.1 ACMP in Wait Mode

The ACMP continues to run in wait mode if enabled before executing the WAIT instruction. Therefore, the ACMP can be used to bring the MCU out of wait mode if the ACMP interrupt, ACIE is enabled. For lowest possible current consumption, the ACMP should be disabled by software if not required as an interrupt source during wait mode.

### 9.1.3.2 ACMP in Stop Modes

#### 9.1.3.2.1 Stop3 Mode Operation

The ACMP continues to operate in Stop3 mode if enabled and compare operation remains active. If ACOPE is enabled, comparator output operates as in the normal operating mode and comparator output is placed onto the external pin. The MCU is brought out of stop when a compare event occurs and ACIE is enabled; ACF flag sets accordingly.

If stop is exited with a reset, the ACMP will be put into its reset state.

#### 9.1.3.2.2 Stop2 and Stop1 Mode Operation

During either Stop2 and Stop1 mode, the ACMP module will be fully powered down. Upon wake-up from Stop2 or Stop1 mode, the ACMP module will be in the reset state.

### 9.1.3.3 ACMP in Active Background Mode

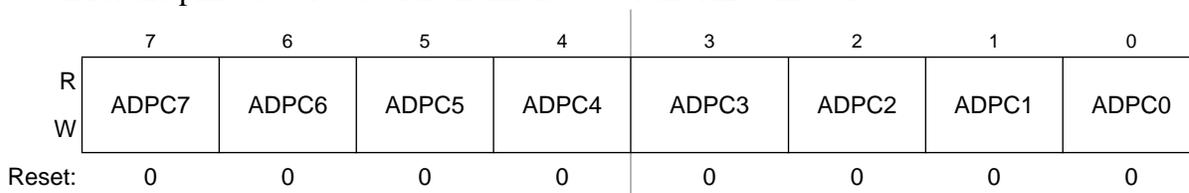
When the microcontroller is in active background mode, the ACMP will continue to operate normally.

## 9.1.4 Block Diagram

The block diagram for the Analog Comparator module is shown [Figure 9-2](#).



used to control the pins associated with channels 0–7 of the ADC module.



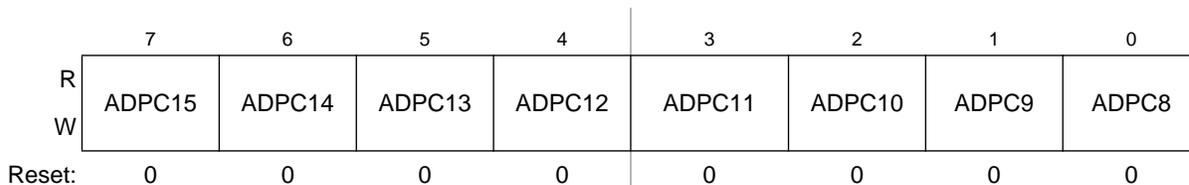
**Figure 10-10. Pin Control 1 Register (APCTL1)**

**Table 10-10. APCTL1 Register Field Descriptions**

Field	Description
7 ADPC7	ADC Pin Control 7. ADPC7 controls the pin associated with channel AD7. 0 AD7 pin I/O control enabled 1 AD7 pin I/O control disabled
6 ADPC6	ADC Pin Control 6. ADPC6 controls the pin associated with channel AD6. 0 AD6 pin I/O control enabled 1 AD6 pin I/O control disabled
5 ADPC5	ADC Pin Control 5. ADPC5 controls the pin associated with channel AD5. 0 AD5 pin I/O control enabled 1 AD5 pin I/O control disabled
4 ADPC4	ADC Pin Control 4. ADPC4 controls the pin associated with channel AD4. 0 AD4 pin I/O control enabled 1 AD4 pin I/O control disabled
3 ADPC3	ADC Pin Control 3. ADPC3 controls the pin associated with channel AD3. 0 AD3 pin I/O control enabled 1 AD3 pin I/O control disabled
2 ADPC2	ADC Pin Control 2. ADPC2 controls the pin associated with channel AD2. 0 AD2 pin I/O control enabled 1 AD2 pin I/O control disabled
1 ADPC1	ADC Pin Control 1. ADPC1 controls the pin associated with channel AD1. 0 AD1 pin I/O control enabled 1 AD1 pin I/O control disabled
0 ADPC0	ADC Pin Control 0. ADPC0 controls the pin associated with channel AD0. 0 AD0 pin I/O control enabled 1 AD0 pin I/O control disabled

### 10.3.9 Pin Control 2 Register (APCTL2)

APCTL2 controls channels 8–15 of the ADC module.



**Figure 10-11. Pin Control 2 Register (APCTL2)**

ALTCLK for this MCU. Consult the module introduction for information on ALTCLK specific to this MCU.

A conversion complete event sets the COCO and generates an ADC interrupt to wake the MCU from wait mode if the ADC interrupt is enabled (AIEN = 1).

## 10.4.7 MCU Stop3 Mode Operation

Stop mode is a low power-consumption standby mode during which most or all clock sources on the MCU are disabled.

### 10.4.7.1 Stop3 Mode With ADACK Disabled

If the asynchronous clock, ADACK, is not selected as the conversion clock, executing a stop instruction aborts the current conversion and places the ADC in its idle state. The contents of ADCRH and ADCRL are unaffected by stop3 mode. After exiting from stop3 mode, a software or hardware trigger is required to resume conversions.

### 10.4.7.2 Stop3 Mode With ADACK Enabled

If ADACK is selected as the conversion clock, the ADC continues operation during stop3 mode. For guaranteed ADC operation, the MCU's voltage regulator must remain active during stop3 mode. Consult the module introduction for configuration information for this MCU.

If a conversion is in progress when the MCU enters stop3 mode, it continues until completion. Conversions can be initiated while the MCU is in stop3 mode by means of the hardware trigger or if continuous conversions are enabled.

A conversion complete event sets the COCO and generates an ADC interrupt to wake the MCU from stop3 mode if the ADC interrupt is enabled (AIEN = 1).

#### NOTE

The ADC module can wake the system from low-power stop and cause the MCU to begin consuming run-level currents without generating a system level interrupt. To prevent this scenario, software should ensure the data transfer blocking mechanism (discussed in [Section 10.4.4.2, “Completing Conversions”](#)) is cleared when entering stop3 and continuing ADC conversions.

## 10.4.8 MCU Stop2 Mode Operation

The ADC module is automatically disabled when the MCU enters stop2 mode. All module registers contain their reset values following exit from stop2. Therefore, the module must be re-enabled and re-configured following exit from stop2.

**Table 12-1. CANCTL0 Register Field Descriptions**

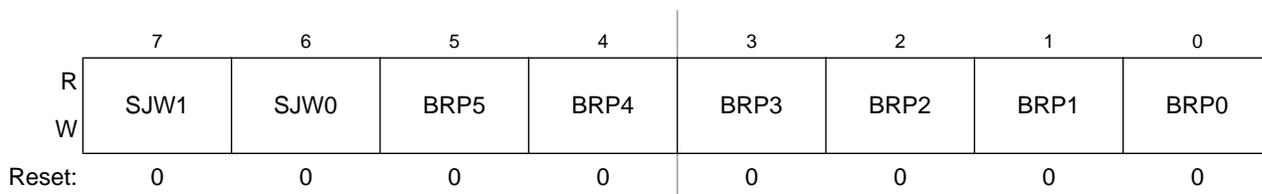
Field	Description
7 RXFRM <sup>1</sup>	<b>Received Frame Flag</b> — This bit is read and clear only. It is set when a receiver has received a valid message correctly, independently of the filter configuration. After it is set, it remains set until cleared by software or reset. Clearing is done by writing a 1. Writing a 0 is ignored. This bit is not valid in loopback mode. 0 No valid message was received since last clearing this flag 1 A valid message was received since last clearing of this flag
6 RXACT	<b>Receiver Active Status</b> — This read-only flag indicates the MSCAN is receiving a message. The flag is controlled by the receiver front end. This bit is not valid in loopback mode. 0 MSCAN is transmitting or idle <sup>2</sup> 1 MSCAN is receiving a message (including when arbitration is lost) <sup>2</sup>
5 CSWAI <sup>3</sup>	<b>CAN Stops in Wait Mode</b> — Enabling this bit allows for lower power consumption in wait mode by disabling all the clocks at the CPU bus interface to the MSCAN module. 0 The module is not affected during wait mode 1 The module ceases to be clocked during wait mode
4 SYNCH	<b>Synchronized Status</b> — This read-only flag indicates whether the MSCAN is synchronized to the CAN bus and able to participate in the communication process. It is set and cleared by the MSCAN. 0 MSCAN is not synchronized to the CAN bus 1 MSCAN is synchronized to the CAN bus
3 TIME	<b>Timer Enable</b> — This bit activates an internal 16-bit wide free running timer which is clocked by the bit clock rate. If the timer is enabled, a 16-bit time stamp will be assigned to each transmitted/received message within the active TX/RX buffer. As soon as a message is acknowledged on the CAN bus, the time stamp will be written to the highest bytes (0x000E, 0x000F) in the appropriate buffer (see <a href="#">Section 12.4, “Programmer’s Model of Message Storage”</a> ). The internal timer is reset (all bits set to 0) when disabled. This bit is held low in initialization mode. 0 Disable internal MSCAN timer 1 Enable internal MSCAN timer
2 WUPE <sup>4</sup>	<b>Wake-Up Enable</b> — This configuration bit allows the MSCAN to restart from sleep mode when traffic on CAN is detected (see <a href="#">Section 12.5.5.4, “MSCAN Sleep Mode”</a> ). This bit must be configured before sleep mode entry for the selected function to take effect. 0 Wake-up disabled — The MSCAN ignores traffic on CAN 1 Wake-up enabled — The MSCAN is able to restart

**Table 12-2. CANCTL1 Register Field Descriptions (continued)**

Field	Description
1 SLPAK	<p><b>Sleep Mode Acknowledge</b> — This flag indicates whether the MSCAN module has entered sleep mode (see Section 12.5.5.4, “MSCAN Sleep Mode”). It is used as a handshake flag for the SLPRQ sleep mode request. Sleep mode is active when SLPRQ = 1 and SLPAK = 1. Depending on the setting of WUPE, the MSCAN will clear the flag if it detects activity on the CAN bus while in sleep mode. CPU clearing the SLPRQ bit will also reset the SLPAK bit.</p> <p>0 Running — The MSCAN operates normally            1 Sleep mode active — The MSCAN has entered sleep mode</p>
0 INITAK	<p><b>Initialization Mode Acknowledge</b> — This flag indicates whether the MSCAN module is in initialization mode (see Section 12.5.5.5, “MSCAN Initialization Mode”). It is used as a handshake flag for the INITRQ initialization mode request. Initialization mode is active when INITRQ = 1 and INITAK = 1. The registers CANCTL1, CANBTR0, CANBTR1, CANIDAC, CANIDAR0–CANIDAR7, and CANIDMR0–CANIDMR7 can be written only by the CPU when the MSCAN is in initialization mode.</p> <p>0 Running — The MSCAN operates normally            1 Initialization mode active — The MSCAN is in initialization mode</p>

### 12.3.3 MSCAN Bus Timing Register 0 (CANBTR0)

The CANBTR0 register configures various CAN bus timing parameters of the MSCAN module.



**Figure 12-6. MSCAN Bus Timing Register 0 (CANBTR0)**

Read: Anytime

Write: Anytime in initialization mode (INITRQ = 1 and INITAK = 1)

**Table 12-3. CANBTR0 Register Field Descriptions**

Field	Description
7:6 SJW[1:0]	<b>Synchronization Jump Width</b> — The synchronization jump width defines the maximum number of time quanta (Tq) clock cycles a bit can be shortened or lengthened to achieve resynchronization to data transitions on the CAN bus (see Table 12-4).
5:0 BRP[5:0]	<b>Baud Rate Prescaler</b> — These bits determine the time quanta (Tq) clock which is used to build up the bit timing (see Table 12-5).

**Table 12-4. Synchronization Jump Width**

SJW1	SJW0	Synchronization Jump Width
0	0	1 Tq clock cycle
0	1	2 Tq clock cycles
1	0	3 Tq clock cycles
1	1	4 Tq clock cycles

The MSCAN is able to leave sleep mode (wake up) only when:

- CAN bus activity occurs and  $WUPE = 1$   
or
- the CPU clears the SLPRQ bit

**NOTE**

The CPU cannot clear the SLPRQ bit before sleep mode ( $SLPRQ = 1$  and  $SLPAK = 1$ ) is active.

After wake-up, the MSCAN waits for 11 consecutive recessive bits to synchronize to the CAN bus. As a consequence, if the MSCAN is woken-up by a CAN frame, this frame is not received.

The receive message buffers (RxFG and RxBG) contain messages if they were received before sleep mode was entered. All pending actions will be executed upon wake-up; copying of RxBG into RxFG, message aborts and message transmissions. If the MSCAN remains in bus-off state after sleep mode was exited, it continues counting the 128 occurrences of 11 consecutive recessive bits.

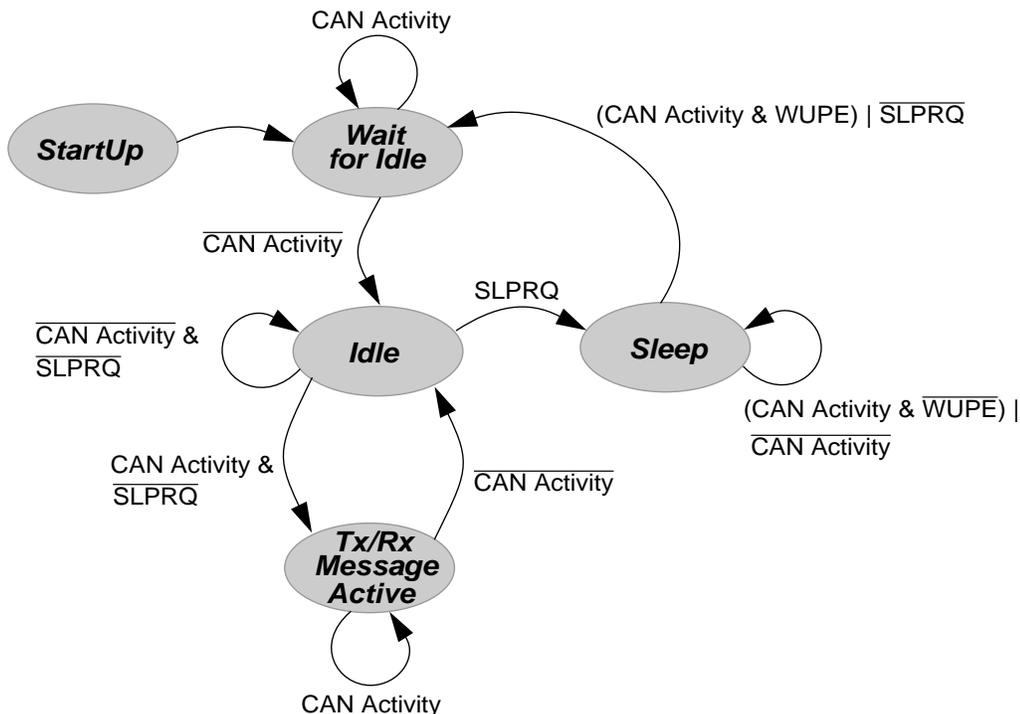


Figure 12-45. Simplified State Transitions for Entering/Leaving Sleep Mode

## 12.6.2 Bus-Off Recovery

The bus-off recovery is user configurable. The bus-off state can either be exited automatically or on user request.

For reasons of backwards compatibility, the MSCAN defaults to automatic recovery after reset. In this case, the MSCAN will become error active again after counting 128 occurrences of 11 consecutive recessive bits on the CAN bus (See the Bosch CAN specification for details).

If the MSCAN is configured for user request (BORM set in [Section 12.3.2, “MSCAN Control Register 1 \(CANCTL1\)”](#)), the recovery from bus-off starts after both independent events have become true:

- 128 occurrences of 11 consecutive recessive bits on the CAN bus have been monitored
- BOHOLD in [Section 12.3.12, “MSCAN Miscellaneous Register \(CANMISC\)”](#) has been cleared by the user

These two events may occur in any order.

the TPM counter is a free-running counter then the update is made when the TPM counter changes from 0xFFFFE to 0xFFFF.

#### 16.4.2.4 Center-Aligned PWM Mode

This type of PWM output uses the up/down counting mode of the timer counter (CPWMS=1). The output compare value in TPMxCnVH:TPMxCnVL determines the pulse width (duty cycle) of the PWM signal while the period is determined by the value in TPMxMODH:TPMxMODL. TPMxMODH:TPMxMODL should be kept in the range of 0x0001 to 0x7FFF because values outside this range can produce ambiguous results. ELSnA will determine the polarity of the CPWM output.

$$\text{pulse width} = 2 \times (\text{TPMxCnVH:TPMxCnVL})$$

$$\text{period} = 2 \times (\text{TPMxMODH:TPMxMODL}); \text{TPMxMODH:TPMxMODL} = 0x0001 - 0x7FFF$$

If the channel-value register TPMxCnVH:TPMxCnVL is zero or negative (bit 15 set), the duty cycle will be 0%. If TPMxCnVH:TPMxCnVL is a positive value (bit 15 clear) and is greater than the (non-zero) modulus setting, the duty cycle will be 100% because the duty cycle compare will never occur. This implies the usable range of periods set by the modulus register is 0x0001 through 0x7FFE (0x7FFF if you do not need to generate 100% duty cycle). This is not a significant limitation. The resulting period would be much longer than required for normal applications.

TPMxMODH:TPMxMODL=0x0000 is a special case that should not be used with center-aligned PWM mode. When CPWMS=0, this case corresponds to the counter running free from 0x0000 through 0xFFFF, but when CPWMS=1 the counter needs a valid match to the modulus register somewhere other than at 0x0000 in order to change directions from up-counting to down-counting.

The output compare value in the TPM channel registers (times 2) determines the pulse width (duty cycle) of the CPWM signal (Figure 16-16). If ELSnA=0, a compare occurred while counting up forces the CPWM output signal low and a compare occurred while counting down forces the output high. The counter counts up until it reaches the modulo setting in TPMxMODH:TPMxMODL, then counts down until it reaches zero. This sets the period equal to two times TPMxMODH:TPMxMODL.

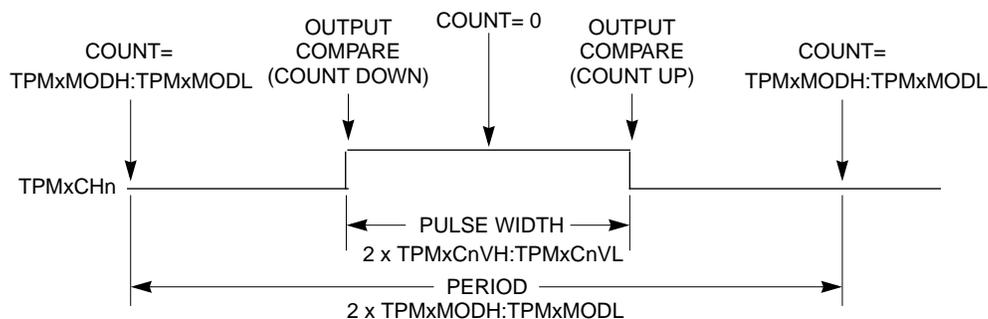


Figure 16-16. CPWM Period and Pulse Width (ELSnA=0)

Center-aligned PWM outputs typically produce less noise than edge-aligned PWMs because fewer I/O pin transitions are lined up at the same system clock edge. This type of PWM is also required for some types of motor drives.