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#### Details

Product Status	Active
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	CANbus, I <sup>2</sup> C, LINbus, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	39
Program Memory Size	128KB (128K x 8)
Program Memory Type	FLASH
EEPROM Size	2K x 8
RAM Size	8K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 16x12b
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	48-LQFP
Supplier Device Package	48-LQFP (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc9s08dz128mlf

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Chapter 3 Modes of Operation

### 3.6.1.2 Active BDM Enabled in Stop3 Mode

Entry into the active background mode from run mode is enabled if ENBDM in BDCSCR is set. This register is described in Chapter 17, "Development Support." If ENBDM is set when the CPU executes a STOP instruction, the system clocks to the background debug logic remain active when the MCU enters stop mode. Because of this, background debug communication remains possible. In addition, the voltage regulator does not enter its low-power standby state but maintains full internal regulation.

Most background commands are not available in stop mode. The memory-access-with-status commands do not allow memory access, but they report an error indicating that the MCU is in either stop or wait mode. The BACKGROUND command can be used to wake the MCU from stop and enter active background mode if the ENBDM bit is set. After entering background debug mode, all background commands are available.

### 3.6.2 Stop2 Mode

Stop2 mode is entered by executing a STOP instruction under the conditions as shown in Table 3-1. Most of the internal circuitry of the MCU is powered off in stop2 with the exception of the RAM. Upon entering stop2, all I/O pin control signals are latched so that the pins retain their states during stop2.

Exit from stop2 is performed by asserting  $\overline{\text{RESET}}$ .

In addition, the real-time counter (RTC) can wake the MCU from stop2, if enabled.

Upon wake-up from stop2 mode, the MCU starts up as from a power-on reset (POR):

- All module control and status registers are reset
- The LVD reset function is enabled and the MCU remains in the reset state if V<sub>DD</sub> is below the LVD trip point (low trip point selected due to POR)
- The CPU takes the reset vector

In addition to the above, upon waking up from stop2, the PPDF bit in SPMSC2 is set. This flag is used to direct user code to go to a stop2 recovery routine. PPDF remains set and the I/O pin states remain latched until a 1 is written to PPDACK in SPMSC2.

To maintain I/O states for pins that were configured as general-purpose I/O before entering stop2, the user must restore the contents of the I/O port registers, which have been saved in RAM, to the port registers before writing to the PPDACK bit. If the port registers are not restored from RAM before writing to PPDACK, then the pins will switch to their reset states when PPDACK is written.

For pins that were configured as peripheral I/O, the user must reconfigure the peripheral module that interfaces to the pin before writing to the PPDACK bit. If the peripheral module is not enabled before writing to PPDACK, the pins will be controlled by their associated port control registers when the I/O latches are opened.

### 3.6.3 On-Chip Peripheral Modules in Stop Modes

When the MCU enters any stop mode, system clocks to the internal peripheral modules are stopped. Even in the exception case (ENBDM = 1), where clocks to the background debug logic continue to operate,



#### Chapter 4 Memory

Table 4-12 shows program and erase times. The bus clock frequency and FCDIV determine the frequency of FCLK ( $f_{FCLK}$ ). The time for one cycle of FCLK is  $t_{FCLK} = 1/f_{FCLK}$ . The times are shown as a number of cycles of FCLK and as an absolute time for the case where  $t_{FCLK} = 5 \ \mu$ s. Program and erase times shown include overhead for the command state machine and enabling and disabling of program and erase voltages.

Parameter	Cycles of FCLK	Time if FCLK = 200 kHz
Byte program	9	45 μs
Burst program	4	20 μs <sup>1</sup>
Sector erase	4000	20 ms
Mass erase	20,000	100 ms
Sector erase abort	4	20 μs <sup>1</sup>

Table 4-12. Program and Erase Times

<sup>1</sup> Excluding start/end overhead

### 4.6.3 **Program and Erase Command Execution**

The FCDIV register must be initialized after any reset and any error flag is cleared before beginning command execution. The command execution steps are:

1. Write a data value to an address in the FLASH or EEPROM array. The address and data information from this write is latched into the FLASH and EEPROM interface. This write is a required first step in any command sequence. For erase and blank check commands, the value of the data is not important. For sector erase commands, the address can be any address in the sector of FLASH or EEPROM to be erased. For mass erase and blank check commands, the address can be any address in the FLASH or EEPROM memory. FLASH and EEPROM erase independently of each other.

### NOTE

Before programming a particular byte in the FLASH or EEPROM, the sector in which that particular byte resides must be erased by a mass or sector erase operation. Reprogramming bits in an already programmed byte without first performing an erase operation may disturb data stored in the FLASH or EEPROM memory.

- 2. Write the command code for the desired command to FCMD. The six valid commands are blank check (0x05), byte program (0x20), burst program (0x25), sector erase (0x40), mass erase (0x41), and sector erase abort (0x47). The command code is latched into the command buffer.
- 3. Write a 1 to the FCBEF bit in FSTAT to clear FCBEF and launch the command (including its address and data information).

A partial command sequence can be aborted manually by writing a 0 to FCBEF any time after the write to the memory array and before writing the 1 that clears FCBEF and launches the complete command. Aborting a command in this way sets the FACCERR access error flag which must be cleared before starting a new command.

NP

**Chapter 4 Memory** 



Figure 4-12. Burst Program Flowchart

MC9S08DZ128 Series Data Sheet, Rev. 1



### 6.5.4.3 Port D Pull Enable Register (PTDPE)



Figure 6-26. Internal Pull Enable for Port D Register (PTDPE)

#### Table 6-24. PTDPE Register Field Descriptions

Field	Description
7:0	Internal Pull Enable for Port D Bits — Each of these control bits determines if the internal pull-up or pull-down
PTDPE[7:0]	device is enabled for the associated PTD pin. For port D pins that are configured as outputs, these bits have no
	effect and the internal pull devices are disabled.
	0 Internal pull-up/pull-down device disabled for port D bit n.
	1 Internal pull-up/pull-down device enabled for port D bit n.

#### NOTE

Pull-down devices only apply when using pin interrupt functions, when corresponding edge select and pin select functions are configured.

### 6.5.4.4 Port D Slew Rate Enable Register (PTDSE)



Figure 6-27. Slew Rate Enable for Port D Register (PTDSE)

#### Table 6-25. PTDSE Register Field Descriptions

Field	Description
7:0 PTDSE[7:0]	<ul> <li>Output Slew Rate Enable for Port D Bits — Each of these control bits determines if the output slew rate control is enabled for the associated PTD pin. For port D pins that are configured as inputs, these bits have no effect.</li> <li>Output slew rate control disabled for port D bit n.</li> <li>Output slew rate control enabled for port D bit n.</li> </ul>

Note: Slew rate reset default values may differ between engineering samples and final production parts. Always initialize slew rate control to the desired value to ensure correct operation.



**Chapter 6 Parallel Input/Output Control** 

### 6.5.7 Port G Registers

Port G is controlled by the registers listed below.

### 6.5.7.1 Port G Data Register (PTGD)



#### Figure 6-42. Port G Data Register (PTGD)

#### Table 6-40. PTGD Register Field Descriptions

Field	Description
7:0 PTGD[7:0]	<b>Port G Data Register Bits</b> — For port G pins that are inputs, reads return the logic level on the pin. For port G pins that are configured as outputs, reads return the last value written to this register. Writes are latched into all bits of this register. For port G pins that are configured as outputs, the logic level is driven out the corresponding MCU pin. Reset forces PTGD to all 0s, but these 0s are not driven out the corresponding pins because reset also configures all port pins as high-impedance inputs with pull-ups disabled.

### 6.5.7.2 Port G Data Direction Register (PTGDD)

	7	6	5	4	3	2	1	0
R	PTGDD7	PTGDD6	PTGDD5	PTGDD4	PTGDD3	PTGDD2	PTGDD1	PTGDD0
W								
Reset:	0	0	0	0	0	0	0	0

#### Figure 6-43. Port G Data Direction Register (PTGDD)

#### Table 6-41. PTGDD Register Field Descriptions

Field	Description
7:0 PTGDD[7:0]	<b>Data Direction for Port G Bits</b> — These read/write bits control the direction of port G pins and what is read for PTGD reads.
	<ol> <li>Input (output driver disabled) and reads return the pin value.</li> <li>Output driver enabled for port G bit n and PTGD reads return the contents of PTGDn.</li> </ol>



### 7.3.6.5 Indexed, 16-Bit Offset (IX2)

This variation of indexed addressing uses the 16-bit value in the H:X index register pair plus a 16-bit offset included in the instruction as the address of the operand needed to complete the instruction.

### 7.3.6.6 SP-Relative, 8-Bit Offset (SP1)

This variation of indexed addressing uses the 16-bit value in the stack pointer (SP) plus an unsigned 8-bit offset included in the instruction as the address of the operand needed to complete the instruction.

### 7.3.6.7 SP-Relative, 16-Bit Offset (SP2)

This variation of indexed addressing uses the 16-bit value in the stack pointer (SP) plus a 16-bit offset included in the instruction as the address of the operand needed to complete the instruction.

### 7.4 Special Operations

The CPU performs a few special operations that are similar to instructions but do not have opcodes like other CPU instructions. In addition, a few instructions such as STOP and WAIT directly affect other MCU circuitry. This section provides additional information about these operations.

### 7.4.1 Reset Sequence

Reset can be caused by a power-on-reset (POR) event, internal conditions such as the COP (computer operating properly) watchdog, or by assertion of an external active-low reset pin. When a reset event occurs, the CPU immediately stops whatever it is doing (the MCU does not wait for an instruction boundary before responding to a reset event). For a more detailed discussion about how the MCU recognizes resets and determines the source, refer to the Resets, Interrupts, and System Configuration chapter.

The reset event is considered concluded when the sequence to determine whether the reset came from an internal source is done and when the reset pin is no longer asserted. At the conclusion of a reset event, the CPU performs a 6-cycle sequence to fetch the reset vector from 0xFFFE and 0xFFFF and to fill the instruction queue in preparation for execution of the first program instruction.

### 7.4.2 Interrupt Sequence

When an interrupt is requested, the CPU completes the current instruction before responding to the interrupt. At this point, the program counter is pointing at the start of the next instruction, which is where the CPU should return after servicing the interrupt. The CPU responds to an interrupt by performing the same sequence of operations as for a software interrupt (SWI) instruction, except the address used for the vector fetch is determined by the highest priority interrupt that is pending when the interrupt sequence started.

The CPU sequence for an interrupt is:

- 1. Store the contents of PCL, PCH, X, A, and CCR on the stack, in that order.
- 2. Set the I bit in the CCR.



Field	Description
7 ADPC23	ADC Pin Control 23. ADPC23 controls the pin associated with channel AD23. 0 AD23 pin I/O control enabled 1 AD23 pin I/O control disabled
6 ADPC22	<ul> <li>ADC Pin Control 22. ADPC22 controls the pin associated with channel AD22.</li> <li>AD22 pin I/O control enabled</li> <li>AD22 pin I/O control disabled</li> </ul>
5 ADPC21	ADC Pin Control 21. ADPC21 controls the pin associated with channel AD21. 0 AD21 pin I/O control enabled 1 AD21 pin I/O control disabled
4 ADPC20	ADC Pin Control 20. ADPC20 controls the pin associated with channel AD20. 0 AD20 pin I/O control enabled 1 AD20 pin I/O control disabled
3 ADPC19	ADC Pin Control 19. ADPC19 controls the pin associated with channel AD19. 0 AD19 pin I/O control enabled 1 AD19 pin I/O control disabled
2 ADPC18	ADC Pin Control 18. ADPC18 controls the pin associated with channel AD18. 0 AD18 pin I/O control enabled 1 AD18 pin I/O control disabled
1 ADPC17	ADC Pin Control 17. ADPC17 controls the pin associated with channel AD17. 0 AD17 pin I/O control enabled 1 AD17 pin I/O control disabled
0 ADPC16	ADC Pin Control 16. ADPC16 controls the pin associated with channel AD16. 0 AD16 pin I/O control enabled 1 AD16 pin I/O control disabled

#### Table 10-12. APCTL3 Register Field Descriptions

### **10.4** Functional Description

The ADC module is disabled during reset or when the ADCH bits are all high. The module is idle when a conversion has completed and another conversion has not been initiated. When idle, the module is in its lowest power state.

The ADC can perform an analog-to-digital conversion on any of the software selectable channels. In 12-bit and 10-bit mode, the selected channel voltage is converted by a successive approximation algorithm into a 12-bit digital result. In 8-bit mode, the selected channel voltage is converted by a successive approximation algorithm into a 9-bit digital result.

When the conversion is completed, the result is placed in the data registers (ADCRH and ADCRL). In 10-bit mode, the result is rounded to 10 bits and placed in the data registers (ADCRH and ADCRL). In 8-bit mode, the result is rounded to 8 bits and placed in ADCRL. The conversion complete flag (COCO) is then set and an interrupt is generated if the conversion complete interrupt has been enabled (AIEN = 1).

The ADC module has the capability of automatically comparing the result of a conversion with the contents of its compare registers. The compare function is enabled by setting the ACFE bit and operates with any of the conversion modes and configurations.

Field	Description				
7–6 MULT	<b>IIC Multiplier Factor</b> . The MULT bits define the multiplier factor, mul. This factor, along with the SC generates the IIC baud rate. The multiplier factor mul as defined by the MULT bits is provided below 00 mul = 01 01 mul = 02 10 mul = 04 11 Reserved	L divider, v.			
5–0 ICR	5–0 ICR IIC Clock Rate. The ICR bits are used to prescale the bus clock for bit rate selection. These bits and bits determine the IIC baud rate, the SDA hold time, the SCL Start hold time, and the SCL Stop hold Table 11-5 provides the SCL divider and hold values for corresponding values of the ICR.				
	The SCL divider multiplied by multiplier factor mul generates IIC baud rate.				
	IIC baud rate = $\frac{\text{bus speed (Hz)}}{\text{mul} \times \text{SCLdivider}}$	Eqn. 11-1			
	SDA hold time is the delay from the falling edge of SCL (IIC clock) to the changing of SDA (IIC data).				
	SDA hold time = bus period (s) $\times$ mul $\times$ SDA hold value	Eqn. 11-2			
	SCL start hold time is the delay from the falling edge of SDA (IIC data) while SCL is high (Start condition) to the falling edge of SCL (IIC clock).				
	SCL Start hold time = bus period (s) $\times$ mul $\times$ SCL Start hold value	Eqn. 11-3			
	SCL stop hold time is the delay from the rising edge of SCL (IIC clock) to the rising edge of SDA SDA (IIC data) while SCL is high (Stop condition).				
	SCL Stop hold time = bus period (s) $\times$ mul $\times$ SCL Stop hold value	Eqn. 11-4			

#### Table 11-3. IICxF Field Descriptions

For example, if the bus speed is 8 MHz, the table below shows the possible hold time values with different ICR and MULT selections to achieve an IIC baud rate of 100kbps.

мшт	ICP	Hold Times (μs)			
MOLI		SDA	SCL Start	SCL Stop	
0x2	0x00	3.500	3.000	5.500	
0x1	0x07	2.500	4.000	5.250	
0x1	0x0B	2.250	4.000	5.250	
0x0	0x14	2.125	4.250	5.125	
0x0	0x18	1.125	4.750	5.125	

Table 11-4. Hold Time Values for 8 MHz Bus Speed



Field	Description
1 SLPRQ <sup>5</sup>	Sleep Mode Request — This bit requests the MSCAN to enter sleep mode, which is an internal power saving mode (see Section 12.5.5.4, "MSCAN Sleep Mode"). The sleep mode request is serviced when the CAN bus is idle, i.e., the module is not receiving a message and all transmit buffers are empty. The module indicates entry to sleep mode by setting SLPAK = 1 (see Section 12.3.2, "MSCAN Control Register 1 (CANCTL1)"). SLPRQ cannot be set while the WUPIF flag is set (see Section 12.3.4.1, "MSCAN Receiver Flag Register (CANRFLG)"). Sleep mode will be active until SLPRQ is cleared by the CPU or, depending on the setting of WUPE, the MSCAN detects activity on the CAN bus and clears SLPRQ itself. 0 Running — The MSCAN functions normally 1 Sleep mode request — The MSCAN enters sleep mode when CAN bus idle
0 INITRQ <sup>6,7</sup>	<b>Initialization Mode Request</b> — When this bit is set by the CPU, the MSCAN skips to initialization mode (see Section 12.5.5.5, "MSCAN Initialization Mode"). Any ongoing transmission or reception is aborted and synchronization to the CAN bus is lost. The module indicates entry to initialization mode by setting INITAK = 1 (Section 12.3.2, "MSCAN Control Register 1 (CANCTL1)"). The following registers enter their hard reset state and restore their default values: CANCTL0 <sup>8</sup> , CANRFLG <sup>9</sup> , CANRIER <sup>10</sup> , CANTFLG, CANTIER, CANTARQ, CANTAAK, and CANTBSEL. The registers CANCTL1, CANBTR0, CANBTR1, CANIDAC, CANIDAR0-7, and CANIDMR0-7 can only be written by the CPU when the MSCAN is in initialization mode. (INITRQ = 1 and INITAK = 1). The values of the error counters are not affected by initialization mode. When this bit is cleared by the CPU, the MSCAN restarts and then tries to synchronize to the CAN bus. If the MSCAN is not in bus-off state, it synchronizes after 11 consecutive recessive bits on the CAN bus; if the MSCAN is in bus-off state, it continues to wait for 128 occurrences of 11 consecutive recessive bits. Writing to other bits in CANCTL0, CANRFLG, CANRIER, CANTFLG, or CANTIER must be done only after initialization mode is exited, which is INITRQ = 0 and INITAK = 0. 0 Normal operation 1 MSCAN in initialization mode

#### Table 12-1. CANCTL0 Register Field Descriptions (continued)

<sup>1</sup> The MSCAN must be in normal mode for this bit to become set.

<sup>2</sup> See the Bosch CAN 2.0A/B specification for a detailed definition of transmitter and receiver states.

- <sup>3</sup> In order to protect from accidentally violating the CAN protocol, the TXCAN pin is immediately forced to a recessive state when the CPU enters wait (CSWAI = 1) or stop mode (see Section 12.5.5.2, "Operation in Wait Mode" and Section 12.5.5.3, "Operation in Stop Mode").
- <sup>4</sup> The CPU has to make sure that the WUPE bit and the WUPIE wake-up interrupt enable bit (see Section 12.3.5, "MSCAN Receiver Interrupt Enable Register (CANRIER)) is enabled, if the recovery mechanism from stop or wait is required.
- <sup>5</sup> The CPU cannot clear SLPRQ before the MSCAN has entered sleep mode (SLPRQ = 1 and SLPAK = 1).
- <sup>6</sup> The CPU cannot clear INITRQ before the MSCAN has entered initialization mode (INITRQ = 1 and INITAK = 1).
- <sup>7</sup> In order to protect from accidentally violating the CAN protocol, the TXCAN pin is immediately forced to a recessive state when the initialization mode is requested by the CPU. Thus, the recommended procedure is to bring the MSCAN into sleep mode (SLPRQ = 1 and SLPAK = 1) before requesting initialization mode.
- <sup>8</sup> Not including WUPE, INITRQ, and SLPRQ.
- <sup>9</sup> TSTAT1 and TSTAT0 are not affected by initialization mode.

<sup>10</sup> RSTAT1 and RSTAT0 are not affected by initialization mode.



Field	Description
1 SLPAK	<ul> <li>Sleep Mode Acknowledge — This flag indicates whether the MSCAN module has entered sleep mode (see Section 12.5.5.4, "MSCAN Sleep Mode"). It is used as a handshake flag for the SLPRQ sleep mode request. Sleep mode is active when SLPRQ = 1 and SLPAK = 1. Depending on the setting of WUPE, the MSCAN will clear the flag if it detects activity on the CAN bus while in sleep mode.CPU clearing the SLPRQ bit will also reset the SLPAK bit.</li> <li>0 Running — The MSCAN operates normally</li> <li>1 Sleep mode active — The MSCAN has entered sleep mode</li> </ul>
0 INITAK	Initialization Mode Acknowledge — This flag indicates whether the MSCAN module is in initialization mode (see Section 12.5.5.5, "MSCAN Initialization Mode"). It is used as a handshake flag for the INITRQ initialization mode request. Initialization mode is active when INITRQ = 1 and INITAK = 1. The registers CANCTL1, CANBTR0, CANBTR1, CANIDAC, CANIDAR0–CANIDAR7, and CANIDMR0–CANIDMR7 can be written only by the CPU when the MSCAN is in initialization mode. 0 Running — The MSCAN operates normally 1 Initialization mode active — The MSCAN is in initialization mode

#### Table 12-2. CANCTL1 Register Field Descriptions (continued)

### 12.3.3 MSCAN Bus Timing Register 0 (CANBTR0)

The CANBTR0 register configures various CAN bus timing parameters of the MSCAN module.

	7	6	5	4	3	2	1	0
R W	SJW1	SJW0	BRP5	BRP4	BRP3	BRP2	BRP1	BRP0
Reset:	0	0	0	0	0	0	0	0

Figure 12-6. MSCAN Bus Timing Register 0 (CANBTR0)

#### Read: Anytime Write: Anytime in initialization mode (INITRQ = 1 and INITAK = 1)

#### Table 12-3. CANBTR0 Register Field Descriptions

Field	Description
7:6 SJW[1:0]	<b>Synchronization Jump Width</b> — The synchronization jump width defines the maximum number of time quanta (Tq) clock cycles a bit can be shortened or lengthened to achieve resynchronization to data transitions on the CAN bus (see Table 12-4).
5:0 BRP[5:0]	<b>Baud Rate Prescaler</b> — These bits determine the time quanta (Tq) clock which is used to build up the bit timing (see Table 12-5).

#### Table 12-4. Synchronization Jump Width

SJW1	SJW0	Synchronization Jump Width
0	0	1 Tq clock cycle
0	1	2 Tq clock cycles
1	0	3 Tq clock cycles
1	1	4 Tq clock cycles



Table 12-30. IDR1	<b>Register Field</b>	Descriptions
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Field	Description
7:5 ID[2:0]	<b>Standard Format Identifier</b> — The identifiers consist of 11 bits (ID[10:0]) for the standard format. ID10 is the most significant bit and is transmitted first on the CAN bus during the arbitration procedure. The priority of an identifier is defined to be highest for the smallest binary number. See also ID bits in Table 12-29.
4 RTR	<ul> <li>Remote Transmission Request — This flag reflects the status of the Remote Transmission Request bit in the CAN frame. In the case of a receive buffer, it indicates the status of the received frame and supports the transmission of an answering frame in software. In the case of a transmit buffer, this flag defines the setting of the RTR bit to be sent.</li> <li>0 Data frame</li> <li>1 Remote frame</li> </ul>
3 IDE	<ul> <li>ID Extended — This flag indicates whether the extended or standard identifier format is applied in this buffer. In the case of a receive buffer, the flag is set as received and indicates to the CPU how to process the buffer identifier registers. In the case of a transmit buffer, the flag indicates to the MSCAN what type of identifier to send.</li> <li>0 Standard format (11 bit)</li> <li>1 Extended format (29 bit)</li> </ul>



Figure 12-32. Identifier Register 3 — Standard Mapping

### 12.4.3 Data Segment Registers (DSR0-7)

The eight data segment registers, each with bits DB[7:0], contain the data to be transmitted or received. The number of bytes to be transmitted or received is determined by the data length code in the corresponding DLR register.



### 12.6.2 Bus-Off Recovery

The bus-off recovery is user configurable. The bus-off state can either be exited automatically or on user request.

For reasons of backwards compatibility, the MSCAN defaults to automatic recovery after reset. In this case, the MSCAN will become error active again after counting 128 occurrences of 11 consecutive recessive bits on the CAN bus (See the Bosch CAN specification for details).

If the MSCAN is configured for user request (BORM set in Section 12.3.2, "MSCAN Control Register 1 (CANCTL1)"), the recovery from bus-off starts after both independent events have become true:

- 128 occurrences of 11 consecutive recessive bits on the CAN bus have been monitored
- BOHOLD in Section 12.3.12, "MSCAN Miscellaneous Register (CANMISC) has been cleared by the user

These two events may occur in any order.



### 14.1.2 Features

Features of SCI module include:

- Full-duplex, standard non-return-to-zero (NRZ) format
- Double-buffered transmitter and receiver with separate enables
- Programmable baud rates (13-bit modulo divider)
- Interrupt-driven or polled operation:
  - Transmit data register empty and transmission complete
  - Receive data register full
  - Receive overrun, parity error, framing error, and noise error
  - Idle receiver detect
  - Active edge on receive pin
  - Break detect supporting LIN
- Hardware parity generation and checking
- Programmable 8-bit or 9-bit character length
- Receiver wakeup by idle-line or address-mark
- Optional 13-bit break character generation / 11-bit break character detection
- Selectable transmitter output polarity

### 14.1.3 Modes of Operation

See Section 14.3, "Functional Description," For details concerning SCI operation in these modes:

- 8- and 9-bit data modes
- Stop mode operation
- Loop mode
- Single-wire mode



Chapter 14 Serial Communications Interface (S08SCIV4)

### 14.2 Register Definition

The SCI has eight 8-bit registers to control baud rate, select SCI options, report SCI status, and for transmit/receive data.

Refer to the direct-page register summary in the Memory chapter of this data sheet for the absolute address assignments for all SCI registers. This section refers to registers and control bits only by their names. A Freescale-provided equate or header file is used to translate these names into the appropriate absolute addresses.

### 14.2.1 SCI Baud Rate Registers (SCIxBDH, SCIxBDL)

This pair of registers controls the prescale divisor for SCI baud rate generation. To update the 13-bit baud rate setting [SBR12:SBR0], first write to SCIxBDH to buffer the high half of the new value and then write to SCIxBDL. The working value in SCIxBDH does not change until SCIxBDL is written.

SCIxBDL is reset to a non-zero value, so after reset the baud rate generator remains disabled until the first time the receiver or transmitter is enabled (RE or TE bits in SCIxC2 are written to 1).



### Figure 14-4. SCI Baud Rate Register (SCIxBDH)

Field	Description
7 LBKDIE	<ul> <li>LIN Break Detect Interrupt Enable (for LBKDIF)</li> <li>0 Hardware interrupts from LBKDIF disabled (use polling).</li> <li>1 Hardware interrupt requested when LBKDIF flag is 1.</li> </ul>
6 RXEDGIE	<b>RxD Input Active Edge Interrupt Enable (for RXEDGIF)</b> 0Hardware interrupts from RXEDGIF disabled (use polling).1Hardware interrupt requested when RXEDGIF flag is 1.
4:0 SBR[12:8]	<b>Baud Rate Modulo Divisor</b> — The 13 bits in SBR[12:0] are referred to collectively as BR, and they set the modulo divide rate for the SCI baud rate generator. When BR = 0, the SCI baud rate generator is disabled to reduce supply current. When BR = 1 to 8191, the SCI baud rate = BUSCLK/( $16 \times BR$ ). See also BR bits in Table 14-3.



Field	Description
4 TXINV <sup>1</sup>	<ul> <li>Transmit Data Inversion — Setting this bit reverses the polarity of the transmitted data output.</li> <li>0 Transmit data not inverted</li> <li>1 Transmit data inverted</li> </ul>
3 ORIE	<ul> <li>Overrun Interrupt Enable — This bit enables the overrun flag (OR) to generate hardware interrupt requests.</li> <li>0 OR interrupts disabled (use polling).</li> <li>1 Hardware interrupt requested when OR = 1.</li> </ul>
2 NEIE	<ul> <li>Noise Error Interrupt Enable — This bit enables the noise flag (NF) to generate hardware interrupt requests.</li> <li>0 NF interrupts disabled (use polling).</li> <li>1 Hardware interrupt requested when NF = 1.</li> </ul>
1 FEIE	<ul> <li>Framing Error Interrupt Enable — This bit enables the framing error flag (FE) to generate hardware interrupt requests.</li> <li>0 FE interrupts disabled (use polling).</li> <li>1 Hardware interrupt requested when FE = 1.</li> </ul>
0 PEIE	<ul> <li>Parity Error Interrupt Enable — This bit enables the parity error flag (PF) to generate hardware interrupt requests.</li> <li>0 PF interrupts disabled (use polling).</li> <li>1 Hardware interrupt requested when PF = 1.</li> </ul>

#### Table 14-8. SCIxC3 Field Descriptions (continued)

<sup>1</sup> Setting TXINV inverts the TxD output for all cases: data bits, start and stop bits, break, and idle.

### 14.2.7 SCI Data Register (SCIxD)

This register is actually two separate registers. Reads return the contents of the read-only receive data buffer and writes go to the write-only transmit data buffer. Reads and writes of this register are also involved in the automatic flag clearing mechanisms for the SCI status flags.

	7	6	5	4	3	2	1	0
R	R7	R6	R5	R4	R3	R2	R1	R0
W	T7	Т6	T5	T4	Т3	T2	T1	T0
Reset	0	0	0	0	0	0	0	0

Figure 14-11. SCI Data Register (SCIxD)

### 14.3 Functional Description

The SCI allows full-duplex, asynchronous, NRZ serial communication among the MCU and remote devices, including other MCUs. The SCI comprises a baud rate generator, transmitter, and receiver block. The transmitter and receiver operate independently, although they use the same baud rate generator. During normal operation, the MCU monitors the status of the SCI, writes the data to be transmitted, and processes received data. The following describes each of the blocks of the SCI.

### 14.3.1 Baud Rate Generation

As shown in Figure 14-12, the clock source for the SCI baud rate generator is the bus-rate clock.



### 16.1.1 Features

The TPM includes these distinctive features:

- One to eight channels:
  - Each channel may be input capture, output compare, or edge-aligned PWM
  - Rising-Edge, falling-edge, or any-edge input capture trigger
  - Set, clear, or toggle output compare action
  - Selectable polarity on PWM outputs
- Module may be configured for buffered, center-aligned pulse-width-modulation (CPWM) on all channels
- Timer clock source selectable as prescaled bus clock, fixed system clock, or an external clock pin
  - Prescale taps for divide-by 1, 2, 4, 8, 16, 32, 64, or 128
  - Fixed system clock source are synchronized to the bus clock by an on-chip synchronization circuit
  - External clock pin may be shared with any timer channel pin or a separated input pin
- 16-bit free-running or modulo up/down count operation
- Timer system enable
- One interrupt per channel plus terminal count interrupt

### 16.1.2 Modes of Operation

In general, TPM channels may be independently configured to operate in input capture, output compare, or edge-aligned PWM modes. A control bit allows the whole TPM (all channels) to switch to center-aligned PWM mode. When center-aligned PWM mode is selected, input capture, output compare, and edge-aligned PWM functions are not available on any channels of this TPM module.

When the microcontroller is in active BDM background or BDM foreground mode, the TPM temporarily suspends all counting until the microcontroller returns to normal user operating mode. During stop mode, all system clocks, including the main oscillator, are stopped; therefore, the TPM is effectively disabled until clocks resume. During wait mode, the TPM continues to operate normally. Provided the TPM does not need to produce a real time reference or provide the interrupt source(s) needed to wake the MCU from wait mode, the user can save power by disabling TPM functions before entering wait mode.

• Input capture mode

When a selected edge event occurs on the associated MCU pin, the current value of the 16-bit timer counter is captured into the channel value register and an interrupt flag bit is set. Rising edges, falling edges, any edge, or no edge (disable channel) may be selected as the active edge which triggers the input capture.

• Output compare mode

When the value in the timer counter register matches the channel value register, an interrupt flag bit is set, and a selected output action is forced on the associated MCU pin. The output compare action may be selected to force the pin to zero, force the pin to one, toggle the pin, or ignore the pin (used for software timing functions).



## Chapter 17 Development Support

### 17.1 Introduction

Development support systems in the HCS08 include the background debug controller (BDC) and the on-chip debug module (DBG). The BDC provides a single-wire debug interface to the target MCU that provides a convenient interface for programming the on-chip FLASH and other nonvolatile memories. The BDC is also the primary debug interface for development and allows non-intrusive access to memory data and traditional debug features such as CPU register modify, breakpoints, and single instruction trace commands.

In the HCS08 Family, address and data bus signals are not available on external pins (not even in test modes). Debug is done through commands fed into the target MCU via the single-wire background debug interface. The debug module provides a means to selectively trigger and capture bus information so an external development system can reconstruct what happened inside the MCU on a cycle-by-cycle basis without having external access to the address and data signals.

### 17.1.1 Forcing Active Background

The method for forcing active background mode depends on the specific HCS08 derivative. For the MC9S08DZ128 Series, you can force active background after a power-on reset by holding the BKGD pin low as the device exits the reset condition. You can also force active background by driving BKGD low immediately after a serial background command that writes a one to the BDFR bit in the SBDFR register. Other causes of reset including an external pin reset or an internally generated error reset ignore the state of the BKGD pin and reset into normal user mode. If no debug pod is connected to the BKGD pin, the MCU will always reset into normal operating mode.



<sup>1</sup> When BRKEN = 0, TAG is do not care (x in the table).

 $^{2}$  In end trace configurations (BEGIN = 0) where a CPU breakpoint is enabled (BRKEN = 1), TRGSEL should agree with TAG. In this case, where TRGSEL = 0 to select no opcode tracking qualification and TAG = 1 to specify a tag-type CPU breakpoint, the CPU breakpoint would not take effect until sometime after the FIFO stopped storing values. Depending on program loops or interrupts, the delay could be very long.

 $^{3}$  In end trace configurations (BEGIN = 0) where a CPU breakpoint is enabled (BRKEN = 1), TRGSEL should agree with TAG. In this case, where TRGSEL = 1 to select opcode tracking qualification and TAG = 0 to specify a force-type CPU breakpoint, the CPU breakpoint would erroneously take effect before the FIFO stopped storing values and the debug run would not complete normally.

4 In begin trace configurations (BEGIN = 1) where a CPU breakpoint is enabled (BRKEN = 1), TAG should not be set to 1. In begin trace debug runs, the CPU breakpoint corresponds to the FIFO full condition which does not correspond to a taggable instruction fetch.

### 18.4.5 FIFO

The FIFO is an eight word deep FIFO. In all trigger modes except for event only, the data stored in the FIFO will be change of flow addresses. In the event only trigger modes only the data bus value corresponding to the event is stored. In event only trigger modes, the high byte of the valid data from the FIFO will always read a 0x00 and the extended information byte in DBGFX will always read 0x00.

### 18.4.5.1 Storing Data in FIFO

In all trigger modes except for the event only modes, the address stored in the FIFO will be determined by the change of flow indicators from the core. The signal core\_cof[1] indicates the current core address is the destination address of an indirect JSR or JMP instruction, or a RTS, RTC, or RTI instruction or interrupt vector and the destination address should be stored. The signal core\_cof[0] indicates that a conditional branch was taken and that the source address of the conditional branch should be stored.

### 18.4.5.2 Storing with Begin-Trigger

Storing with Begin-Trigger can be used in all trigger modes. Once the DBG module is enabled and armed in the begin-trigger mode, data is not stored in the FIFO until the trigger condition is met. Once the trigger condition is met the DBG module will remain armed until 8 words are stored in the FIFO. If the core\_cof[1] signal becomes asserted, the current address is stored in the FIFO. If the core\_cof[0] signal becomes asserted, the address registered during the previous last cycle is decremented by two and stored in the FIFO.

### 18.4.5.3 Storing with End-Trigger

Storing with End-Trigger cannot be used in event-only trigger modes. Once the DBG module is enabled and armed in the end-trigger mode, data is stored in the FIFO until the trigger condition is met. If the core\_cof[1] signal becomes asserted, the current address is stored in the FIFO. If the core\_cof[0] signal becomes asserted, the address registered during the previous last cycle is decremented by two and stored in the FIFO. When the trigger condition is met, the ARM and ARMF will be cleared and no more data will be stored. In non-event only end-trigger modes, if the trigger is at a change of flow address the trigger event will be stored in the FIFO.

### 18.4.5.4 Reading Data from FIFO

The data stored in the FIFO can be read using BDM commands provided the DBG module is enabled and not armed (DBGEN=1 and ARM=0). The FIFO data is read out first-in-first-out. By reading the CNT bits



#### **Appendix A Electrical Characteristics**

maximum-rated voltages to this high-impedance circuit. Reliability of operation is enhanced if unused inputs are tied to an appropriate logic voltage level (for instance, either  $V_{SS}$  or  $V_{DD}$ ).

Num	Rating	Symbol	Value	Unit
1	Supply voltage	V <sub>DD</sub>	-0.3 to + 5.8	V
2	Input voltage	V <sub>In</sub>	– 0.3 to V <sub>DD</sub> + 0.3	V
3	Instantaneous maximum current Single pin limit (applies to all port pins) <sup>1, 2, 3</sup>	Ι <sub>D</sub>	± 25	mA
4	Maximum current into V <sub>DD</sub>	I <sub>DD</sub>	120	mA
5	Storage temperature	T <sub>stg</sub>	-55 to +150	°C

 Table A-2. Absolute Maximum Ratings

<sup>1</sup> Input must be current limited to the value specified. To determine the value of the required current-limiting resistor, calculate resistance values for positive (V<sub>DD</sub>) and negative (V<sub>SS</sub>) clamp voltages, then use the larger of the two resistance values.

 $^2\,$  All functional non-supply pins are internally clamped to V\_{SS} and V\_{DD}

<sup>3</sup> Power supply must maintain regulation within operating V<sub>DD</sub> range during instantaneous and operating maximum current conditions. If positive injection current (V<sub>In</sub> > V<sub>DD</sub>) is greater than I<sub>DD</sub>, the injection current may flow out of V<sub>DD</sub> and could result in external power supply going out of regulation. Ensure external V<sub>DD</sub> load will shunt current greater than maximum injection current. This will be the greatest risk when the MCU is not consuming power. Examples are: if no system clock is present, or if the clock rate is very low which would reduce overall power consumption.

### A.4 Thermal Characteristics

This section provides information about operating temperature range, power dissipation, and package thermal resistance. Power dissipation on I/O pins is usually small compared to the power dissipation in on-chip logic and it is user-determined rather than being controlled by the MCU design. In order to take  $P_{I/O}$  into account in power calculations, determine the difference between actual pin voltage and  $V_{SS}$  or  $V_{DD}$  and multiply by the pin current for each I/O pin. Except in cases of unusually high pin current (heavy loads), the difference between pin voltage and  $V_{SS}$  or  $V_{DD}$  will be very small.