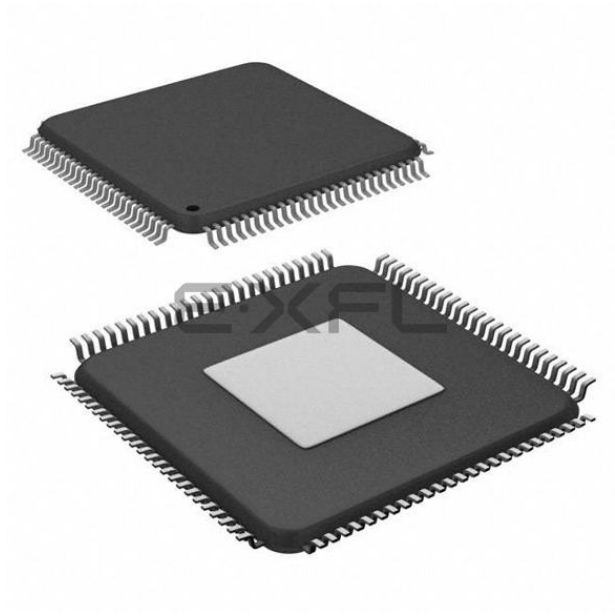
Infineon Technologies - XMC4300F100F256AAXQMA1 Datasheet



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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	ARM® Cortex®-M4
Core Size	32-Bit Single-Core
Speed	144MHz
Connectivity	CANbus, Ethernet, I²C, LINbus, MMC/SD, SPI, UART/USART, USB OTG, USIC
Peripherals	DMA, I²S, LED, POR, Touch-Sense, WDT
Number of I/O	73
Program Memory Size	256KB (256K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	128K x 8
Voltage - Supply (Vcc/Vdd)	3.13V ~ 3.63V
Data Converters	A/D 16x12b; D/A 2x12b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP Exposed Pad
Supplier Device Package	PG-LQFP-100-25
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/xmc4300f100f256aaxqma1

Table 2 Features of XMC4300 Device Types

Derivative[1)]	LED TS Intf.	SD MMC Intf.	ETH Intf.	ECAT Slave Intf.	USB Intf.	USIC Chan.	MultiCAN Nodes, MO
XMC4300-F100x256	1	1	RMII	2 x MII	1	2 x 2	N0, N1 MO[0..63]

1) x is a placeholder for the supported temperature range.

Table 3 Features of XMC4300 Device Types

Derivative[1)]	ADC Chan.	DAC Chan.	CCU4 Slice	CCU8 Slice
XMC4300-F100x256	16	2	2 x 4	1 x 4

1) x is a placeholder for the supported temperature range.

1.4 Definition of Feature Variants

The XMC4300 types are offered with several memory sizes and number of available VADC channels. **Table 4** describes the location of the available Flash memory, **Table 5** describes the location of the available SRAMs, **Table 6** the available VADC channels.

Table 4 Flash Memory Ranges

Total Flash Size	Cached Range	Uncached Range
256 Kbytes	$0800\ 0000_H$ – $0803\ FFFF_H$	$0C00\ 0000_H$ – $0C03\ FFFF_H$

Table 5 SRAM Memory Ranges

Total SRAM Size	Program SRAM	System Data SRAM
128 Kbytes	$1FFF\ 0000_H$ – $1FFF\ FFFF_H$	$2000\ 0000_H$ – $2000\ FFFF_H$

Table 6 ADC Channels[1)]

Package	VADC G0	VADC G1
PG-LQFP-100	CH0..CH7	CH0..CH7

1) Some pins in a package may be connected to more than one channel. For the detailed mapping see the Port I/O Function table.

2.2 Pin Configuration and Definition

The following figures summarize all pins, showing their locations on the four sides of the different packages.

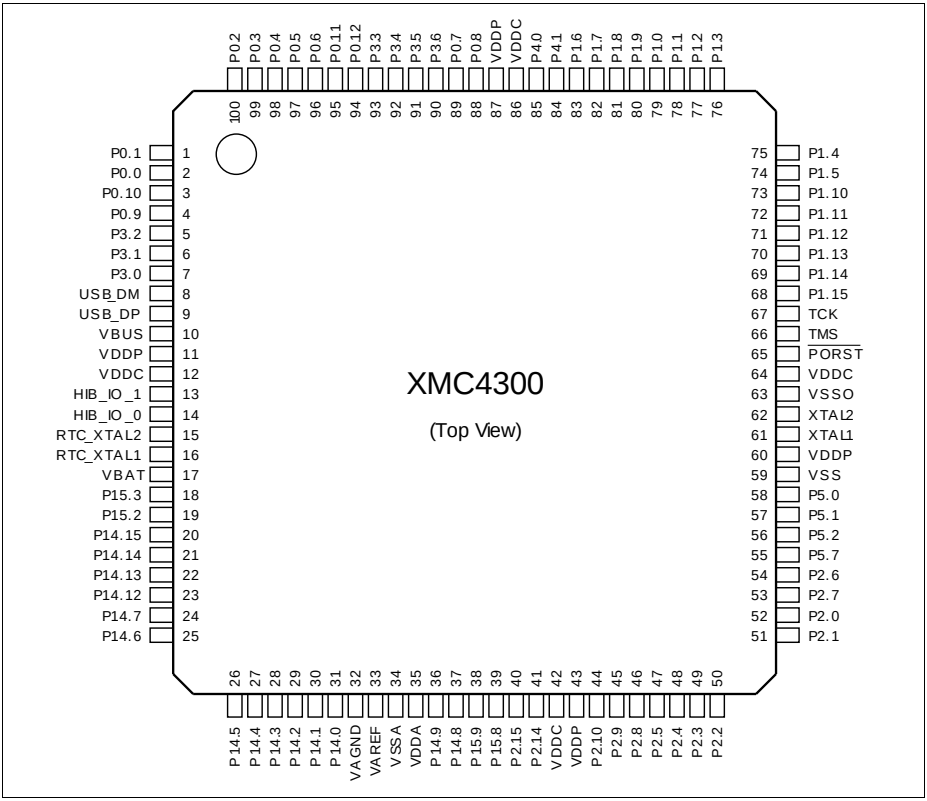

Figure 3 XMC4300 PG-LQFP-100 Pin Configuration (top view)

Table 9 Package Pin Mapping (cont'd)

Function	LQFP-100	Pad Type	Notes
P3.2	5	A2	
P3.3	93	A1+	
P3.4	92	A1+	
P3.5	91	A2	
P3.6	90	A2	
P4.0	85	A2	
P4.1	84	A2	
P5.0	58	A1+	
P5.1	57	A1+	
P5.2	56	A1+	
P5.7	55	A1+	
P14.0	31	AN/DIG_IN	
P14.1	30	AN/DIG_IN	
P14.2	29	AN/DIG_IN	
P14.3	28	AN/DIG_IN	
P14.4	27	AN/DIG_IN	
P14.5	26	AN/DIG_IN	
P14.6	25	AN/DIG_IN	
P14.7	24	AN/DIG_IN	
P14.8	37	AN/DAC/DIG_IN	
P14.9	36	AN/DAC/DIG_IN	
P14.12	23	AN/DIG_IN	
P14.13	22	AN/DIG_IN	
P14.14	21	AN/DIG_IN	
P14.15	20	AN/DIG_IN	
P15.2	19	AN/DIG_IN	
P15.3	18	AN/DIG_IN	
P15.8	39	AN/DIG_IN	
P15.9	38	AN/DIG_IN	

Table 9 Package Pin Mapping (cont'd)

Function	LQFP-100	Pad Type	Notes
HIB_IO_0	14	A1 special	At the first power-up and with every reset of the hibernate domain this pin is configured as open-drain output and drives "0". As output the medium driver mode is active.
HIB_IO_1	13	A1 special	At the first power-up and with every reset of the hibernate domain this pin is configured as input with no pull device active. As output the medium driver mode is active.
USB_DP	9	special	
USB_DM	8	special	
TCK	67	A1	Weak pull-down active.
TMS	66	A1+	Weak pull-up active. As output the strong-soft driver mode is active.
$\overline{\text{PORST}}$	65	special	Weak pull-up permanently active, strong pull-down controlled by EVR.
XTAL1	61	clock_IN	
XTAL2	62	clock_O	
RTC_XTAL1	16	clock_IN	
RTC_XTAL2	15	clock_O	
VBAT	17	Power	When VDDP is supplied VBAT has to be supplied as well.
VBUS	10	special	
VAREF	33	AN_Ref	
VAGND	32	AN_Ref	
VDDA	35	AN_Power	
VSSA	34	AN_Power	
VDDC	12	Power	
VDDC	42	Power	
VDDC	64	Power	
VDDC	86	Power	
VDDP	11	Power	

2.3 Power Connection Scheme

Figure 5. shows a reference power connection scheme for the XMC4300.

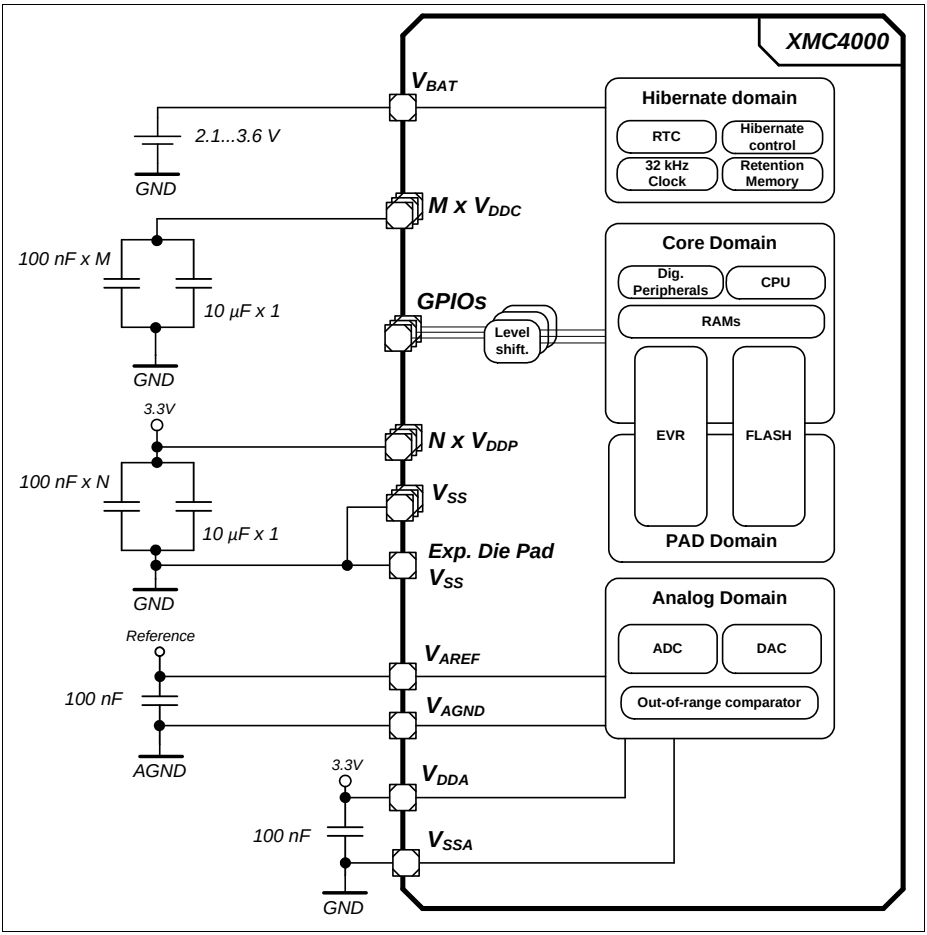

Figure 5 Power Connection Scheme

Every power supply pin needs to be connected. Different pins of the same supply need also to be externally connected. As example, all V_{DDP} pins must be connected externally to one V_{DDP} net. In this reference scheme one 100 nF capacitor is connected at each supply pin against V_{SS}. An additional 10 µF capacitor is connected to the V_{DDP} nets and an additional 10 uF capacitor to the V_{DDC} nets.

3.2 DC Parameters

3.2.1 Input/Output Pins

The digital input stage of the shared analog/digital input pins is identical to the input stage of the standard digital input/output pins.
The Pull-up on the $\overline{\text{PORST}}$ pin is identical to the Pull-up on the standard digital input/output pins.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Table 19 Standard Pad Parameters

Parameter	Symbol	Values		Unit	Note / Test Condition
		Min.	Max.		
Pin capacitance (digital inputs/outputs)	C_{IO} CC	–	10	pF	
Pull-down current	$\lvert I_{PDL}\rvert$ SR	150	–	µA	1) $V_{IN} \geq 0.6 \times V_{DDP}$
		–	10	µA	2) $V_{IN} \leq 0.36 \times V_{DDP}$
Pull-Up current	$\lvert I_{PUH}\rvert$ SR	–	10	µA	2) $V_{IN} \geq 0.6 \times V_{DDP}$
		100	–	µA	1) $V_{IN} \leq 0.36 \times V_{DDP}$
Input Hysteresis for pads of all A classes[3]	$HYSA$ CC	$0.1 \times V_{DDP}$	–	V	
$\overline{\text{PORST}}$ spike filter always blocked pulse duration	t_{SF1} CC	–	10	ns	
$\overline{\text{PORST}}$ spike filter pass-through pulse duration	t_{SF2} CC	100	–	ns	
$\overline{\text{PORST}}$ pull-down current	$\lvert I_{PPD}\rvert$ CC	13	–	mA	V_{IN} = 1.0 V

1) Current required to override the pull device with the opposite logic level (“force current”). With active pull device, at load currents between force and keep current the input state is undefined.

2) Load current at which the pull device still maintains the valid logic level (“keep current”). With active pull device, at load currents between force and keep current the input state is undefined.

3) Hysteresis is implemented to avoid metastable states and switching due to internal ground bounce. It can not be guaranteed that it suppresses switching due to external system noise.

Table 24 VADC Parameters (Operating Conditions apply)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Total Unadjusted Error	TUE CC	-4	–	4	LSB	12-bit resolution; V_{DDA} = 3.3 V; $V_{AREF} = V_{DDA}$ 7)
Differential Non-Linearity Error 8)	EA_{DNL} CC	-3	–	3	LSB	
Gain Error 8)	EA_{GAIN} CC	-4	–	4	LSB	
Integral Non-Linearity 8)	EA_{INL} CC	-3	–	3	LSB	
Offset Error 8)	EA_{OFF} CC	-4	–	4	LSB	
Worst case ADC V_{DDA} power supply current per active converter	I_{DDAA} CC	–	1.5	2	mA	during conversion V_{DDP} = 3.6 V, T_J = 150 °C
Charge consumption on V_{AREF} per conversion 5)	Q_{CONV} CC	–	30	–	pC	0 V ≤ V_{AREF} ≤ V_{DDA} 9)
ON resistance of the analog input path	R_{AIN} CC	–	600	1 200	Ohm	
ON resistance for the ADC test (pull down for AIN7)	R_{AIN7T} CC	180	550	900	Ohm	
Resistance of the reference voltage input path	R_{AREF} CC	–	700	1 700	Ohm	

1) A running conversion may become imprecise in case the normal conditions are violated (voltage overshoot).

2) If the analog reference voltage is below V_{DDA}, then the ADC converter errors increase. If the reference voltage is reduced by the factor k (k<1), TUE, DNL, INL, Gain, and Offset errors increase also by the factor 1/k.

3) The leakage current definition is a continuous function, as shown in figure ADCx Analog Inputs Leakage. The numerical values defined determine the characteristic points of the given continuous linear approximation - they do not define step function (see **Figure 12**).

4) The sampling capacity of the conversion C-network is pre-charged to V_{AREF}/2 before the sampling moment. Because of the parasitic elements, the voltage measured at AINx can deviate from V_{AREF}/2.

5) Applies to AINx, when used as alternate reference input.

6) This represents an equivalent switched capacitance. This capacitance is not switched to the reference voltage at once. Instead, smaller capacitances are successively switched to the reference voltage.

7) For 10-bit conversions, the errors are reduced to 1/4; for 8-bit conversions, the errors are reduced to 1/16. Never less than ±1 LSB.

8) The sum of DNL/INL/GAIN/OFF errors does not exceed the related total unadjusted error TUE.

9) The resulting current for a conversion can be calculated with $I_{AREF} = Q_{CONV} / t_c$.
The fastest 12-bit post-calibrated conversion of t_c = 459 ns results in a typical average current of I_{AREF} = 65.4 µA.

3.2.4 Out-of-Range Comparator (ORC)

The Out-of-Range Comparator (ORC) triggers on analog input voltages (V_{AIN}) above the analog reference[1] (V_{AREF}) on selected input pins (GxORCy) and generates a service request trigger (GxORCOUTy).

Note: These parameters are not subject to production test, but verified by design and/or characterization.

The parameters in **Table 27** apply for the maximum reference voltage $V_{AREF} = V_{DDA} + 50$ mV.

Table 27 ORC Parameters (Operating Conditions apply)

Parameter	Symbol		Values			Unit	Note / Test Condition
			Min.	Typ.	Max.		
DC Switching Level	V_{ODC}	CC	100	125	210	mV	$V_{AIN} \geq V_{AREF} + V_{ODC}$
Hysteresis	V_{OHYS}	CC	50	–	V_{ODC}	mV	
Detection Delay of a persistent Overvoltage	t_{ODD}	CC	50	–	450	ns	$V_{AIN} \geq V_{AREF} + 210$ mV
			45	–	105	ns	$V_{AIN} \geq V_{AREF} + 400$ mV
Always detected Overvoltage Pulse	t_{OPDD}	CC	440	–	–	ns	$V_{AIN} \geq V_{AREF} + 210$ mV
			90	–	–	ns	$V_{AIN} \geq V_{AREF} + 400$ mV
Never detected Overvoltage Pulse	t_{OPDN}	CC	–	–	45	ns	$V_{AIN} \geq V_{AREF} + 210$ mV
			–	–	30	ns	$V_{AIN} \geq V_{AREF} + 400$ mV
Release Delay	t_{ORD}	CC	65	–	105	ns	$V_{AIN} \leq V_{AREF}$
Enable Delay	t_{OED}	CC	–	100	200	ns	

1) Always the standard VADC reference, alternate references do not apply to the ORC.

3.2.9 Flash Memory Parameters

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Table 35 Flash Memory Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Erase Time per 256 Kbyte Sector	t_{ERP} CC	–	5	5.5	s	
Erase Time per 64 Kbyte Sector	t_{ERP} CC	–	1.2	1.4	s	
Erase Time per 16 Kbyte Logical Sector	t_{ERP} CC	–	0.3	0.4	s	
Program time per page[1]	t_{PRP} CC	–	5.5	11	ms	
Erase suspend delay	t_{FL_ErSusp} CC	–	–	15	ms	
Wait time after margin change	$t_{FL\_MarginDel}$ CC	10	–	–	µs	
Wake-up time	t_{WU} CC	–	–	270	µs	
Read access time	t_a CC	22	–	–	ns	For operation with 1 / f_{CPU} < t_a wait states must be configured[2]
Data Retention Time, Physical Sector[3][4]	t_{RET} CC	20	–	–	years	Max. 1000 erase/program cycles
Data Retention Time, Logical Sector[3][4]	t_{RETL} CC	20	–	–	years	Max. 100 erase/program cycles
Data Retention Time, User Configuration Block (UCB)[3][4]	t_{RTU} CC	20	–	–	years	Max. 4 erase/program cycles per UCB
Endurance on 64 Kbyte Physical Sector PS4	N_{EPS4} CC	10000	–	–	cycles	Cycling distributed over life time[5]

1) In case the Program Verify feature detects weak bits, these bits will be programmed once more. The reprogramming takes an additional time of 5.5 ms.
2) The following formula applies to the wait state configuration: FCON.WSPFLASH × (1 / f_{CPU}) ≥ t_a.
3) Storage and inactive time included.
4) Values given are valid for an average weighted junction temperature of T_J = 110°C.
5) Only valid with robust EEPROM emulation algorithm, equally cycling the logical sectors. For more details see the Reference Manual.

3.3.2 Power-Up and Supply Monitoring

$\overline{\text{PORST}}$ is always asserted when V_{DDP} and/or V_{DDC} violate the respective thresholds.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

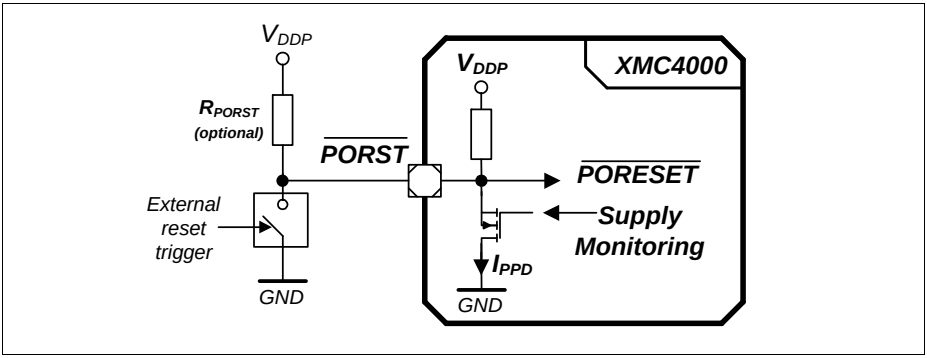

Figure 21 $\overline{\text{PORST}}$ Circuit

Table 36 Supply Monitoring Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Digital supply voltage reset threshold	V_{POR} CC	2.79[1]	–	3.05[2]	V	[3]
Core supply voltage reset threshold	V_{PV} CC	–	–	1.17	V	
V_{DDP} voltage to ensure defined pad states	V_{DDPPA} CC	–	1.0	–	V	
$\overline{\text{PORST}}$ rise time	t_{PR} SR	–	–	2	µs	[4]
Startup time from power-on reset with code execution from Flash	t_{SSW} CC	–	2.5	3.5	ms	Time to the first user code instruction
V_{DDC} ramp up time	t_{VCR} CC	–	550	–	µs	Ramp up after power-on or after a reset triggered by a violation of V_{POR} or V_{PV}

1) Minimum threshold for reset assertion.

2) Maximum threshold for reset deassertion.
3) The V_{DDP} monitoring has a typical hysteresis of V_{PORHYS} = 180 mV.
4) If t_{PR} is not met, low spikes on $\overline{\text{PORST}}$ may be seen during start up (e.g. reset pulses generated by the supply monitoring due to a slow ramping V_{DDP}).

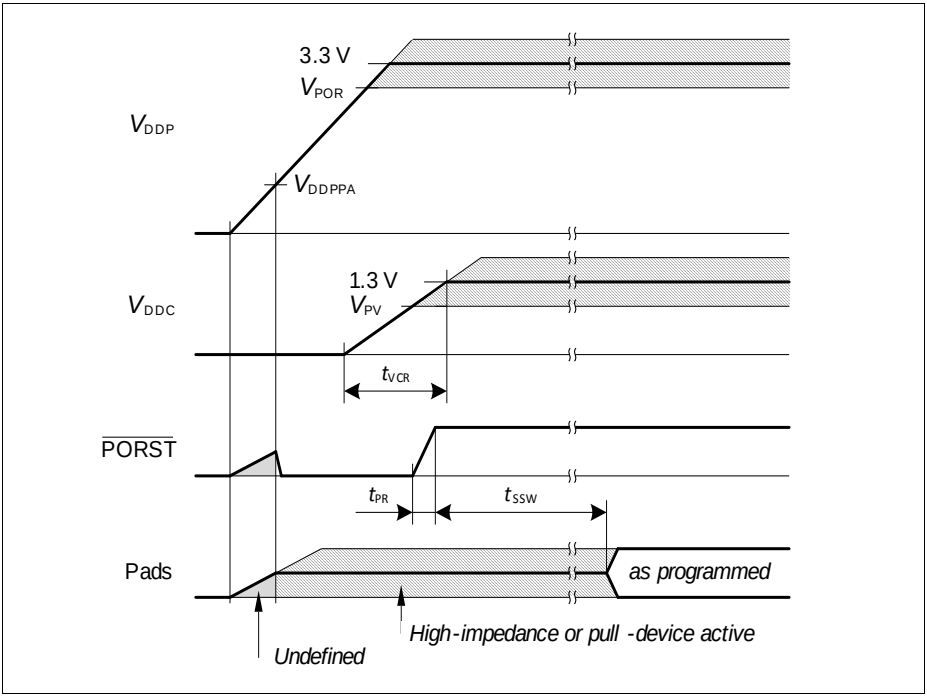

Figure 22 Power-Up Behavior

3.3.3 Power Sequencing

While starting up and shutting down as well as when switching power modes of the system it is important to limit the current load steps. A typical cause for such load steps is changing the CPU frequency f_{CPU}. Load steps exceeding the below defined values may cause a power on reset triggered by the supply monitor.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Table 46 USIC IIC Fast Mode Timing[1)]

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Fall time of both SDA and SCL	t_1 CC/SR	20 + 0.1*C_b [2)]	-	300	ns	
Rise time of both SDA and SCL	t_2 CC/SR	20 + 0.1*C_b [2)]	-	300	ns	
Data hold time	t_3 CC/SR	0	-	-	µs	
Data set-up time	t_4 CC/SR	100	-	-	ns	
LOW period of SCL clock	t_5 CC/SR	1.3	-	-	µs	
HIGH period of SCL clock	t_6 CC/SR	0.6	-	-	µs	
Hold time for (repeated) START condition	t_7 CC/SR	0.6	-	-	µs	
Set-up time for repeated START condition	t_8 CC/SR	0.6	-	-	µs	
Set-up time for STOP condition	t_9 CC/SR	0.6	-	-	µs	
Bus free time between a STOP and START condition	t_{10} CC/SR	1.3	-	-	µs	
Capacitive load for each bus line	C_b SR	-	-	400	pF	

1) Due to the wired-AND configuration of an IIC bus system, the port drivers of the SCL and SDA signal lines need to operate in open-drain mode. The high level on these lines must be held by an external pull-up device, approximalely 10 kOhm for operation at 100 kbit/s, approximately 2 kOhm for operation at 400 kbit/s.

2) C_b refers to the total capacitance of one bus line in pF.

Table 50 SD Card Bus Timing for Full-Speed Mode[1)] (cont'd)

Parameter	Symbol	Values		Unit	Note/ Test Condition
		Min.	Max.		
SD card output valid time	t_{ODLY}	–	14	ns	
SD card output hold time	t_{OH}	0	–	ns	

1) Reference card timing values for calculation examples. Not subject to production test and not characterized.

Full-Speed Output Path (Write)

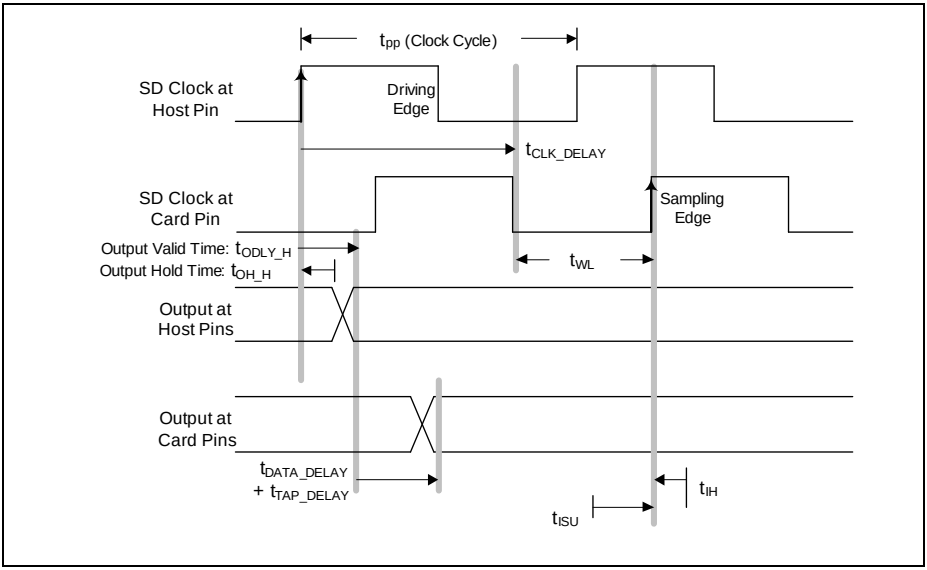

Figure 30 Full-Speed Output Path

Full-Speed Write Meeting Setup (Maximum Delay)

The following equations show how to calculate the allowed skew range between the SD_CLK and SD_DAT/CMD signals on the PCB.

No clock delay:

(1)

$$t_{ODLY\_F} + t_{DATA\_DELAY} + t_{TAP\_DELAY} + t_{ISU} < t_{WL}$$

Table 52 SD Card Bus Timing for High-Speed Mode[1)]

Parameter	Symbol	Values		Unit	Note/ Test Condition
		Min.	Max.		
SD card input setup time	t_{ISU}	6	–	ns	
SD card input hold time	t_{IH}	2	–	ns	
SD card output valid time	t_{ODLY}	–	14	ns	
SD card output hold time	t_{OH}	2.5	–	ns	

1) Reference card timing values for calculation examples. Not subject to production test and not characterized.

High-Speed Output Path (Write)

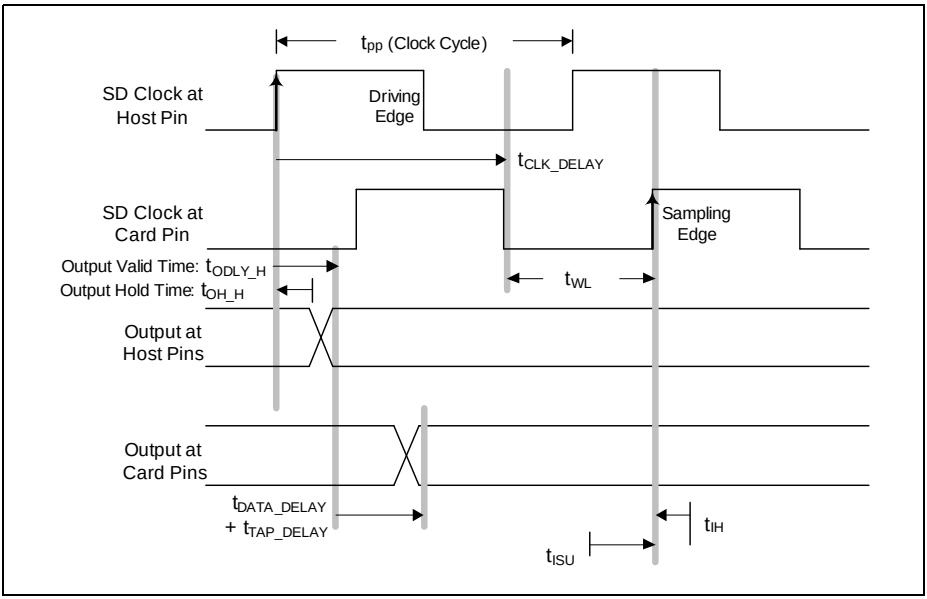

Figure 32 High-Speed Output Path

High-Speed Write Meeting Setup (Maximum Delay)

The following equations show how to calculate the allowed skew range between the SD_CLK and SD_DAT/CMD signals on the PCB.

No clock delay:

(7)

$$t_{ODLY\_H} + t_{DATA\_DELAY} + t_{TAP\_DELAY} + t_{ISU} < t_{WL}$$

With clock delay:

(8)

$$t_{ODLY\_H} + t_{DATA\_DELAY} + t_{TAP\_DELAY} + t_{ISU} < t_{WL} + t_{CLK\_DELAY}$$

(9)

$$t_{DATA\_DELAY} + t_{TAP\_DELAY} - t_{CLK\_DELAY} < t_{WL} - t_{ISU} - t_{ODLY\_H}$$

$$t_{DATA\_DELAY} - t_{CLK\_DELAY} < t_{WL} - t_{ISU} - t_{ODLY\_H} - t_{TAP\_DELAY}$$

$$t_{DATA\_DELAY} - t_{CLK\_DELAY} < 10 - 6 - 14 - t_{TAP\_DELAY}$$

$$t_{DATA\_DELAY} - t_{CLK\_DELAY} < -10 - t_{TAP\_DELAY}$$

The data delay is less than the clock delay by at least 10 ns in the ideal case where t_{WL}= 10 ns.

High-Speed Write Meeting Hold (Minimum Delay)

The following equations show how to calculate the allowed skew range between the SD_CLK and SD_DAT/CMD signals on the PCB.

(10)

$$t_{CLK\_DELAY} < t_{WL} + t_{OH\_H} + t_{DATA\_DELAY} + t_{TAP\_DELAY} - t_{IH}$$

$$t_{CLK\_DELAY} - t_{DATA\_DELAY} < t_{WL} + t_{OH\_H} + t_{TAP\_DELAY} - t_{IH}$$

$$t_{CLK\_DELAY} - t_{DATA\_DELAY} < 10 + 2 + t_{TAP\_DELAY} - 2$$

$$t_{CLK\_DELAY} - t_{DATA\_DELAY} < 10 + t_{TAP\_DELAY}$$

The clock can be delayed versus data up to 13.2 ns (external delay line) in ideal case of t_{WL}= 10 ns, with maximum t_{TAP_DELAY} = 3.2 ns programmed.

High-Speed Input Path (Read)

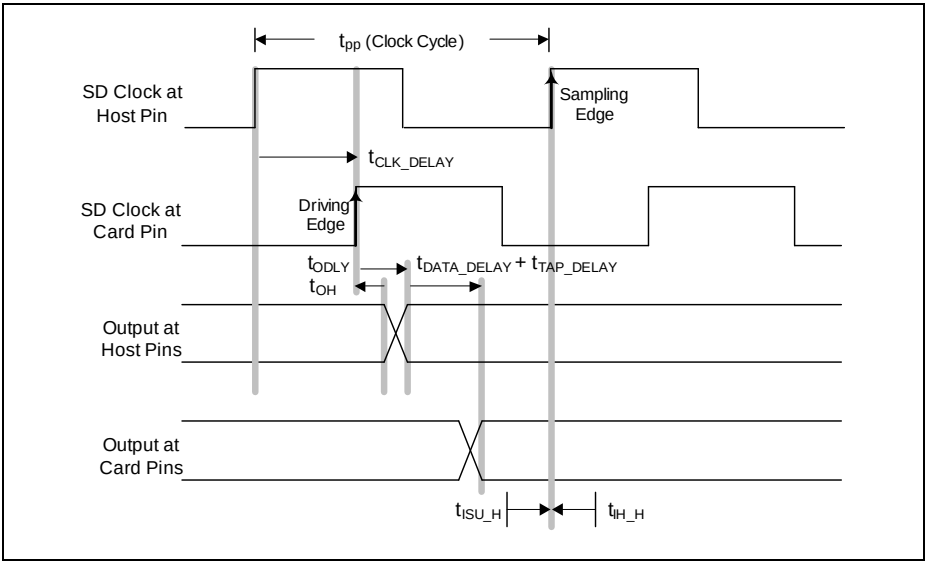

Figure 33 High-Speed Input Path

High-Speed Read Meeting Setup (Maximum Delay)

The following equations show how to calculate the allowed combined propagation delay range of the SD_CLK and SD_DAT/CMD signals on the PCB.

(11)

$$t_{CLK\_DELAY} + t_{DATA\_DELAY} + t_{TAP\_DELAY} + t_{ODLY} + t_{ISU\_H} < t_{pp}$$

$$t_{CLK\_DELAY} + t_{DATA\_DELAY} < t_{pp} - t_{ODLY} - t_{ISU\_H} - t_{TAP\_DELAY}$$

$$t_{CLK\_DELAY} + t_{DATA\_DELAY} < 20 - 14 - 2 - t_{TAP\_DELAY}$$

$$t_{CLK\_DELAY} + t_{DATA\_DELAY} < 4 - t_{TAP\_DELAY}$$

The data + clock delay can be up to 4 ns for a 20 ns clock cycle.

4 Package and Reliability

The XMC4300 is a member of the XMC4000 Family of microcontrollers. It is also compatible to a certain extent with members of similar families or subfamilies.

Each package is optimized for the device it houses. Therefore, there may be slight differences between packages of the same pin-count but for different device types. In particular, the size of the Exposed Die Pad may vary.

If different device types are considered or planned for an application, it must be ensured that the board layout fits all packages under consideration.

4.1 Package Parameters

Table 60 provides the thermal characteristics of the packages used in XMC4300.

Table 60 Thermal Characteristics of the Packages

Parameter	Symbol	Limit Values		Unit	Package Types
		Min.	Max.		
Exposed Die Pad dimensions including U-Groove	Ex × Ey CC	-	7.0 × 7.0	mm	PG-LQFP-100-25
Exposed Die Pad dimensions excluding U-Groove	Ax × Ay CC	-	6.2 × 6.2	mm	PG-LQFP-100-25
Thermal resistance Junction-Ambient $T_J \leq 150$ °C	$R_{\Theta JA}$ CC	-	22.5	K/W	PG-LQFP-100-25[1)]

1) Device mounted on a 4-layer JEDEC board (JESD 51-7) with thermal vias; exposed pad soldered.

Note: For electrical reasons, it is required to connect the exposed pad to the board ground V_{SS}, independent of EMC and thermal requirements.

4.1.1 Thermal Considerations

When operating the XMC4300 in a system, the total heat generated in the chip must be dissipated to the ambient environment to prevent overheating and the resulting thermal damage.

The maximum heat that can be dissipated depends on the package and its integration into the target board. The “Thermal resistance $R_{\Theta JA}$” quantifies these parameters. The power dissipation must be limited so that the average junction temperature does not exceed 150 °C.

4.2 Package Outlines

The exposed die pad dimensions are listed in **Table 60**.

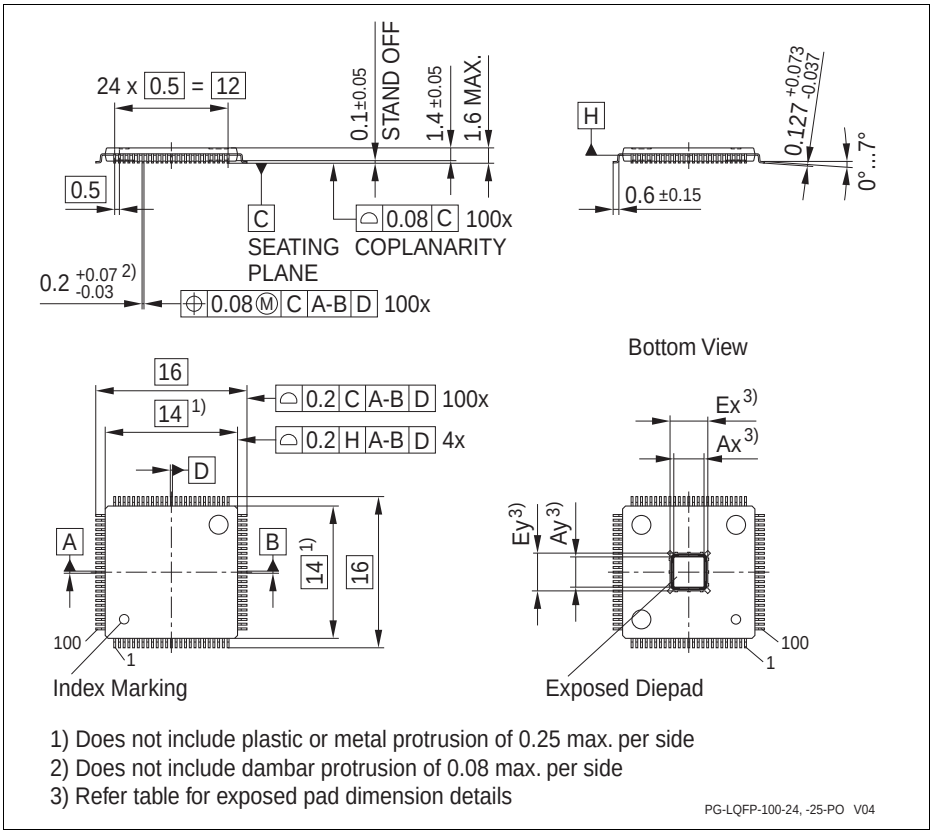

Figure 43 PG-LQFP-100-25 (Plastic Green Low Profile Quad Flat Package)

All dimensions in mm.

You can find complete information about Infineon packages, packing and marking in our Infineon Internet Page “Packages”: **http://www.infineon.com/packages**