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Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details	
Product Status	Obsolete
Programmable Type	In System Programmable
Delay Time tpd(1) Max	7.5 ns
Voltage Supply - Internal	3V ~ 3.6V
Number of Logic Elements/Blocks	4
Number of Macrocells	64
Number of Gates	1250
Number of I/O	68
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	100-TQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm7064aeti100-7

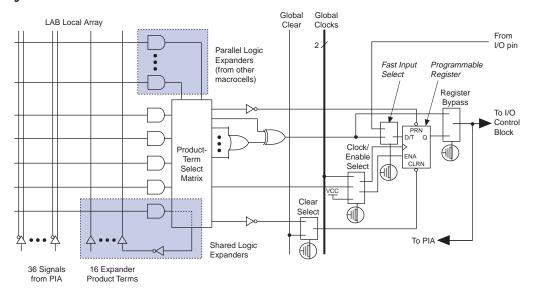
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Macrocells

MAX 7000A macrocells can be individually configured for either sequential or combinatorial logic operation. The macrocells consist of three functional blocks: the logic array, the product-term select matrix, and the programmable register. Figure 2 shows a MAX 7000A macrocell.

Figure 2. MAX 7000A Macrocell



Combinatorial logic is implemented in the logic array, which provides five product terms per macrocell. The product-term select matrix allocates these product terms for use as either primary logic inputs (to the OR and XOR gates) to implement combinatorial functions, or as secondary inputs to the macrocell's register preset, clock, and clock enable control functions.

Two kinds of expander product terms ("expanders") are available to supplement macrocell logic resources:

- Shareable expanders, which are inverted product terms that are fed back into the logic array
- Parallel expanders, which are product terms borrowed from adjacent macrocells

The Altera development system automatically optimizes product-term allocation according to the logic requirements of the design.

For registered functions, each macrocell flipflop can be individually programmed to implement D, T, JK, or SR operation with programmable clock control. The flipflop can be bypassed for combinatorial operation. During design entry, the designer specifies the desired flipflop type; the Altera software then selects the most efficient flipflop operation for each registered function to optimize resource utilization.

Each programmable register can be clocked in three different modes:

- Global clock signal. This mode achieves the fastest clock-to-output performance.
- Global clock signal enabled by an active-high clock enable. A clock enable is generated by a product term. This mode provides an enable on each flipflop while still achieving the fast clock-to-output performance of the global clock.
- Array clock implemented with a product term. In this mode, the flipflop can be clocked by signals from buried macrocells or I/O pins.

Two global clock signals are available in MAX 7000A devices. As shown in Figure 1, these global clock signals can be the true or the complement of either of the global clock pins, GCLK1 or GCLK2.

Each register also supports asynchronous preset and clear functions. As shown in Figure 2, the product-term select matrix allocates product terms to control these operations. Although the product-term-driven preset and clear from the register are active high, active-low control can be obtained by inverting the signal within the logic array. In addition, each register clear function can be individually driven by the active-low dedicated global clear pin (GCLRn). Upon power-up, each register in a MAX 7000AE device may be set to either a high or low state. This power-up state is specified at design entry. Upon power-up, each register in EPM7128A and EPM7256A devices are set to a low state.

All MAX 7000A I/O pins have a fast input path to a macrocell register. This dedicated path allows a signal to bypass the PIA and combinatorial logic and be clocked to an input D flipflop with an extremely fast (as low as 2.5 ns) input setup time.

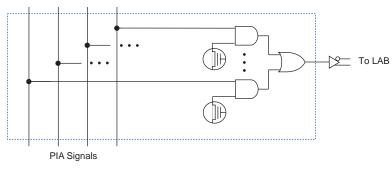
Parallel Expanders

Parallel expanders are unused product terms that can be allocated to a neighboring macrocell to implement fast, complex logic functions. Parallel expanders allow up to 20 product terms to directly feed the macrocell OR logic, with five product terms provided by the macrocell and 15 parallel expanders provided by neighboring macrocells in the LAB.

The compiler can allocate up to three sets of up to five parallel expanders to the macrocells that require additional product terms. Each set of five parallel expanders incurs a small, incremental timing delay (t_{PEXP}). For example, if a macrocell requires 14 product terms, the compiler uses the five dedicated product terms within the macrocell and allocates two sets of parallel expanders; the first set includes five product terms, and the second set includes four product terms, increasing the total delay by $2 \times t_{PEXP}$.

Two groups of eight macrocells within each LAB (e.g., macrocells 1 through 8 and 9 through 16) form two chains to lend or borrow parallel expanders. A macrocell borrows parallel expanders from lower-numbered macrocells. For example, macrocell 8 can borrow parallel expanders from macrocell 7, from macrocells 7 and 6, or from macrocells 7, 6, and 5. Within each group of eight, the lowest-numbered macrocell can only lend parallel expanders, and the highest-numbered macrocell can only borrow them. Figure 4 shows how parallel expanders can be borrowed from a neighboring macrocell.

Figure 5. MAX 7000A PIA Routing



While the routing delays of channel-based routing schemes in masked or FPGAs are cumulative, variable, and path-dependent, the MAX 7000A PIA has a predictable delay. The PIA makes a design's timing performance easy to predict.

I/O Control Blocks

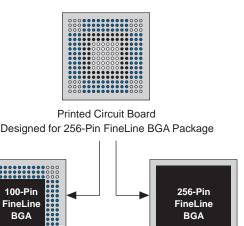
The I/O control block allows each I/O pin to be individually configured for input, output, or bidirectional operation. All I/O pins have a tri-state buffer that is individually controlled by one of the global output enable signals or directly connected to ground or $V_{CC}.$ Figure 6 shows the I/O control block for MAX 7000A devices. The I/O control block has 6 or 10 global output enable signals that are driven by the true or complement of two output enable signals, a subset of the I/O pins, or a subset of the I/O macrocells.

SameFrame Pin-Outs

MAX 7000A devices support the SameFrame pin-out feature for FineLine BGA packages. The SameFrame pin-out feature is the arrangement of balls on FineLine BGA packages such that the lower-ball-count packages form a subset of the higher-ball-count packages. SameFrame pin-outs provide the flexibility to migrate not only from device to device within the same package, but also from one package to another. A given printed circuit board (PCB) layout can support multiple device density/package combinations. For example, a single board layout can support a range of devices from an EPM7128AE device in a 100-pin FineLine BGA package to an EPM7512AE device in a 256-pin FineLine BGA package.

The Altera design software provides support to design PCBs with SameFrame pin-out devices. Devices can be defined for present and future use. The software generates pin-outs describing how to lay out a board to take advantage of this migration (see Figure 7).

Figure 7. SameFrame Pin-Out Example



100-Pin FineLine BGA Package (Reduced I/O Count or Logic Requirements) 256-Pin FineLine BGA Package (Increased I/O Count or Logic Requirements)

Programming Times

The time required to implement each of the six programming stages can be broken into the following two elements:

- A pulse time to erase, program, or read the EEPROM cells.
- A shifting time based on the test clock (TCK) frequency and the number of TCK cycles to shift instructions, address, and data into the device.

By combining the pulse and shift times for each of the programming stages, the program or verify time can be derived as a function of the TCK frequency, the number of devices, and specific target device(s). Because different ISP-capable devices have a different number of EEPROM cells, both the total fixed and total variable times are unique for a single device.

Programming a Single MAX 7000A Device

The time required to program a single MAX 7000A device in-system can be calculated from the following formula:

$$t_{PROG} = t_{PPULSE} + \frac{Cycle_{PTCK}}{f_{TCK}}$$

where: t_{PROG} = Programming time t_{PPULSE} = Sum of the fixed times to erase, program, and

verify the EEPROM cells

 $Cycle_{PTCK}$ = Number of TCK cycles to program a device

= TCK frequency

The ISP times for a stand-alone verification of a single MAX 7000A device can be calculated from the following formula:

$$t_{VER} = t_{VPULSE} + \frac{Cycle_{VTCK}}{f_{TCK}}$$

where: t_{VER} = Verify time

 t_{VPULSE} = Sum of the fixed times to verify the EEPROM cells

 $Cycle_{VTCK}$ = Number of TCK cycles to verify a device

Table 8. MAX 7000A	JTAG Instructions
JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern output at the device pins
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation
IDCODE	Selects the IDCODE register and places it between the TDI and TDO pins, allowing the IDCODE to be serially shifted out of TDO
USERCODE	Selects the 32-bit USERCODE register and places it between the TDI and TDO pins, allowing the USERCODE value to be shifted out of TDO. The USERCODE instruction is available for MAX 7000AE devices only
UESCODE	These instructions select the user electronic signature (UESCODE) and allow the UESCODE to be shifted out of TDO. UESCODE instructions are available for EPM7128A and EPM7256A devices only.
ISP Instructions	These instructions are used when programming MAX 7000A devices via the JTAG ports with the MasterBlaster, ByteBlasterMV, or BitBlaster download cable, or using a Jam STAPL File, JBC File, or SVF File via an embedded processor or test equipment.

Power Sequencing & Hot-Socketing

Because MAX 7000A devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The V_{CCIO} and V_{CCINT} power planes can be powered in any order.

Signals can be driven into MAX 7000AE devices before and during power-up (and power-down) without damaging the device. Additionally, MAX 7000AE devices do not drive out during power-up. Once operating conditions are reached, MAX 7000AE devices operate as specified by the user.

MAX 7000AE device I/O pins will not source or sink more than 300 μA of DC current during power-up. All pins can be driven up to 5.75 V during hot-socketing, except the OE1 and GLCRn pins. The OE1 and GLCRn pins can be driven up to 3.6 V during hot-socketing. After V_{CCINT} and V_{CCIO} reach the recommended operating conditions, these two pins are 5.0-V tolerant.

EPM7128A and EPM7256A devices do not support hot-socketing and may drive out during power-up.

Design Security

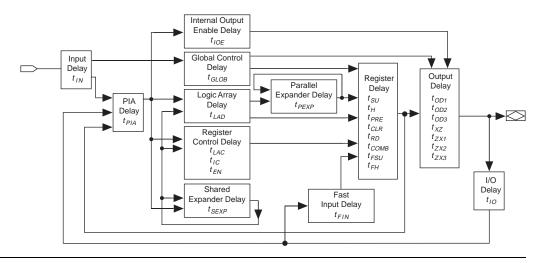
All MAX 7000A devices contain a programmable security bit that controls access to the data programmed into the device. When this bit is programmed, a design implemented in the device cannot be copied or retrieved. This feature provides a high level of design security because programmed data within EEPROM cells is invisible. The security bit that controls this function, as well as all other programmed data, is reset only when the device is reprogrammed.

Generic Testing

MAX 7000A devices are fully tested. Complete testing of each programmable EEPROM bit and all internal logic elements ensures 100% programming yield. AC test measurements are taken under conditions equivalent to those shown in Figure 9. Test patterns can be used and then erased during early stages of the production flow.

Table 1	4. MAX 7000A Device Recomm	ended Operating Conditions			
Symbol	Parameter	Conditions	Min	Max	Unit
V _{CCINT}	Supply voltage for internal logic and input buffers	(3), (13)	3.0	3.6	V
V _{CCIO}	Supply voltage for output drivers, 3.3-V operation	(3)	3.0	3.6	V
	Supply voltage for output drivers, 2.5-V operation	(3)	2.3	2.7	V
V _{CCISP}	Supply voltage during in- system programming		3.0	3.6	V
V _I	Input voltage	(4)	-0.5	5.75	V
Vo	Output voltage		0	V _{CCIO}	V
T _A	Ambient temperature	Commercial range	0	70	° C
		Industrial range (5)	-40	85	° C
TJ	Junction temperature	Commercial range	0	90	° C
		Industrial range (5)	-40	105	° C
		Extended range (5)	-40	130	° C
t _R	Input rise time			40	ns
t _F	Input fall time			40	ns

Figure 11. MAX 7000A Timing Model



The timing characteristics of any signal path can be derived from the timing model and parameters of a particular device. External timing parameters, which represent pin-to-pin timing delays, can be calculated as the sum of internal parameters. Figure 12 shows the timing relationship between internal and external delay parameters.



See Application Note 94 (Understanding MAX 7000 Timing) for more information.

Symbol	Parameter	Conditions	Speed Grade						
			-	5	-7		-10		
			Min	Max	Min	Max	Min	Max	
t _{IN}	Input pad and buffer delay			0.7		1.0		1.4	ns
t_{IO}	I/O input pad and buffer delay			0.7		1.0		1.4	ns
t _{FIN}	Fast input delay			2.5		3.0		3.4	ns
t _{SEXP}	Shared expander delay			2.0		2.9		3.8	ns
t _{PEXP}	Parallel expander delay			0.4		0.7		0.9	ns
t_{LAD}	Logic array delay			1.6		2.4		3.1	ns
t _{LAC}	Logic control array delay			0.7		1.0		1.3	ns
t _{IOE}	Internal output enable delay			0.0		0.0		0.0	ns
t _{OD1}	Output buffer and pad delay, slow slew rate = off V _{CCIO} = 3.3 V	C1 = 35 pF		0.8		1.2		1.6	ns
t _{OD2}	Output buffer and pad delay, slow slew rate = off V _{CCIO} = 2.5 V	C1 = 35 pF (5)		1.3		1.7		2.1	ns
t _{OD3}	Output buffer and pad delay, slow slew rate = on V _{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		5.8		6.2		6.6	ns
t _{ZX1}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		4.0		4.0		5.0	ns
t _{ZX2}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 2.5 \text{ V}$	C1 = 35 pF (5)		4.5		4.5		5.5	ns
t _{ZX3}	Output buffer enable delay, slow slew rate = on V _{CCIO} = 3.3 V	C1 = 35 pF		9.0		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	C1 = 5 pF		4.0		4.0		5.0	ns
t_{SU}	Register setup time		1.4		2.1		2.9		ns
t_H	Register hold time		0.6		1.0		1.3		ns
t _{FSU}	Register setup time of fast input		1.1		1.6		1.6		ns
t _{FH}	Register hold time of fast input		1.4		1.4		1.4		ns
t _{RD}	Register delay			8.0		1.2		1.6	ns
t _{COMB}	Combinatorial delay			0.5		0.9		1.3	ns
t _{IC}	Array clock delay			1.2		1.7		2.2	ns

Symbol	Parameter	Conditions	Speed Grade							
			-5			7	-1	10		
			Min	Max	Min	Max	Min	Max	1	
t _{PD1}	Input to non- registered output	C1 = 35 pF (2)		5.5		7.5		10	ns	
t _{PD2}	I/O input to non- registered output	C1 = 35 pF (2)		5.5		7.5		10	ns	
t _{SU}	Global clock setup time	(2)	3.9		5.2		6.9		ns	
t _H	Global clock hold time	(2)	0.0		0.0		0.0		ns	
t _{FSU}	Global clock setup time of fast input		2.5		3.0		3.0		ns	
t _{FH}	Global clock hold time of fast input		0.0		0.0		0.0		ns	
t _{CO1}	Global clock to output delay	C1 = 35 pF	1.0	3.5	1.0	4.8	1.0	6.4	ns	
t _{CH}	Global clock high time		2.0		3.0		4.0		ns	
t _{CL}	Global clock low time		2.0		3.0		4.0		ns	
t _{ASU}	Array clock setup time	(2)	2.0		2.7		3.6		ns	
t _{AH}	Array clock hold time	(2)	0.2		0.3		0.5		ns	
t _{ACO1}	Array clock to output delay	C1 = 35 pF (2)	1.0	5.4	1.0	7.3	1.0	9.7	ns	
t _{ACH}	Array clock high time		2.0		3.0		4.0		ns	
t _{ACL}	Array clock low time		2.0		3.0		4.0		ns	
t _{CPPW}	Minimum pulse width for clear and preset	(3)	2.0		3.0		4.0		ns	
t _{CNT}	Minimum global clock period	(2)		5.8		7.9		10.5	ns	
f _{CNT}	Maximum internal global clock frequency	(2), (4)	172.4		126.6		95.2		MHz	
t _{ACNT}	Minimum array clock period	(2)		5.8		7.9		10.5	ns	
f _{ACNT}	Maximum internal array clock frequency	(2), (4)	172.4		126.6		95.2		MHz	

Symbol	Parameter	Conditions	Speed Grade								
			-6		-7		-10		-12		
			Min	Max	Min	Max	Min	Max	Min	Max	
t _{RD}	Register delay			1.7		2.1		2.8		3.3	ns
t _{COMB}	Combinatorial delay			1.7		2.1		2.8		3.3	ns
t _{IC}	Array clock delay			2.4		3.0		4.1		4.9	ns
t _{EN}	Register enable time			2.4		3.0		4.1		4.9	ns
t _{GLOB}	Global control delay			1.0		1.2		1.7		2.0	ns
t _{PRE}	Register preset time			3.1		3.9		5.2		6.2	ns
t _{CLR}	Register clear time			3.1		3.9		5.2		6.2	ns
t _{PIA}	PIA delay	(2)		0.9		1.1		1.5		1.8	ns
t_{LPA}	Low-power adder	(6)		11.0		10.0		10.0		10.0	ns

Symbol	Parameter	Conditions	Speed Grade								
			-6		-7		-10		-12		
			Min	Max	Min	Max	Min	Max	Min	Max	1
t_{COMB}	Combinatorial delay			1.6		2.0		2.7		3.2	ns
t _{IC}	Array clock delay			2.7		3.4		4.5		5.4	ns
t _{EN}	Register enable time			2.5		3.1		4.2		5.0	ns
t _{GLOB}	Global control delay			1.1		1.4		1.8		2.2	ns
t _{PRE}	Register preset time			2.3		2.9		3.8		4.6	ns
t _{CLR}	Register clear time			2.3		2.9		3.8		4.6	ns
t_{PIA}	PIA delay	(2)		1.3		1.6		2.1		2.6	ns
t_{LPA}	Low-power adder	(6)		11.0		10.0		10.0		10.0	ns

Notes to tables:

- (1) These values are specified under the recommended operating conditions shown in Table 14 on page 28. See Figure 12 for more information on switching waveforms.
- (2) These values are specified for a PIA fan-out of one LAB (16 macrocells). For each additional LAB fan-out in these devices, add an additional 0.1 ns to the PIA timing value.
- (3) This minimum pulse width for preset and clear applies for both global clear and array controls. The t_{LPA} parameter must be added to this minimum width if the clear or reset signal incorporates the t_{LAD} parameter into the signal path.
- (4) This parameter is measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
- (5) Operating conditions: $V_{CCIO} = 2.5 \pm 0.2 \text{ V}$ for commercial and industrial use.
- (6) The t_{LPA} parameter must be added to the t_{LAD} , t_{LAC} , t_{IC} , t_{EN} , t_{SEXP} , t_{ACL} , and t_{CPPW} parameters for macrocells running in low-power mode.

Power Consumption

Supply power (P) versus frequency (f_{MAX} , in MHz) for MAX 7000A devices is calculated with the following equation:

$$P = P_{INT} + P_{IO} = I_{CCINT} \times V_{CC} + P_{IO}$$

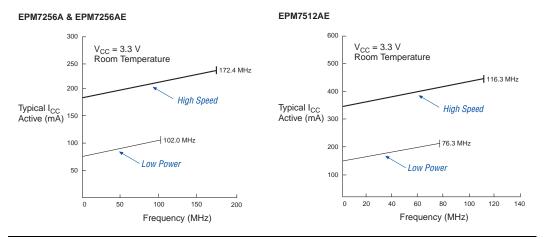
The $P_{\rm IO}$ value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

The I_{CCINT} value depends on the switching frequency and the application logic. The I_{CCINT} value is calculated with the following equation:

$$I_{CCINT} =$$

$$(A \times MC_{TON}) + [B \times (MC_{DEV} - MC_{TON})] + (C \times MC_{USED} \times f_{\boldsymbol{MAX}} \times \boldsymbol{tog_{LC}})$$

Figure 13. I_{CC} vs. Frequency for MAX 7000A Devices (Part 2 of 2)



Device Pin-Outs

See the Altera web site (http://www.altera.com) or the *Altera Digital Library* for pin-out information.

Figures 14 through 23 show the package pin-out diagrams for MAX 7000A devices.

Figure 14. 44-Pin PLCC/TQFP Package Pin-Out Diagram

Package outlines not drawn to scale.

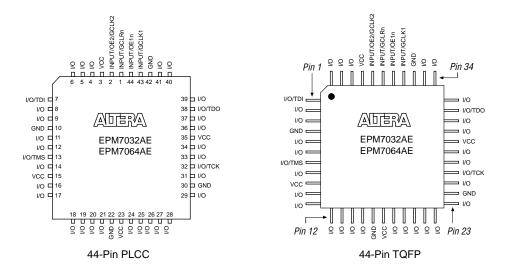


Figure 15. 49-Pin Ultra FineLine BGA Package Pin-Out Diagram

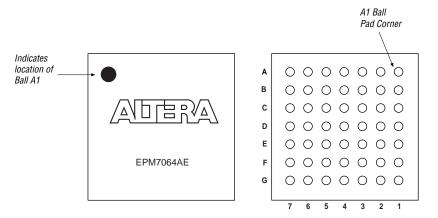


Figure 16. 84-Pin PLCC Package Pin-Out Diagram

Package outline not drawn to scale.

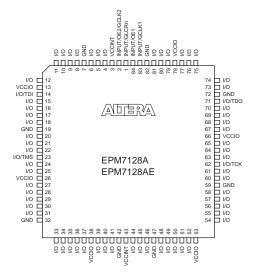


Figure 19. 144-Pin TQFP Package Pin-Out Diagram

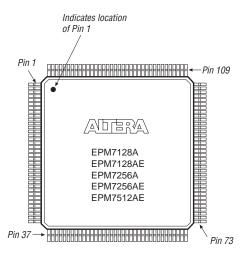


Figure 20. 169-Pin Ultra FineLine BGA Package Pin-Out Diagram

Package outline not drawn to scale.

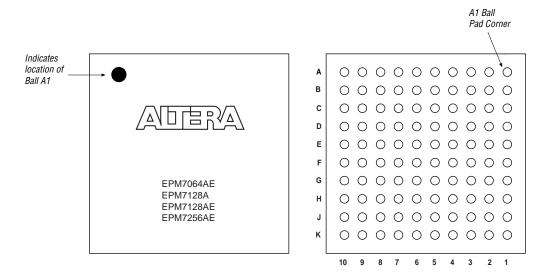


Figure 21. 208-Pin PQFP Package Pin-Out Diagram

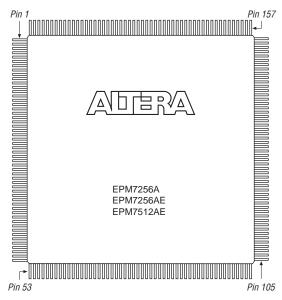


Figure 22. 256-Pin BGA Package Pin-Out Diagram

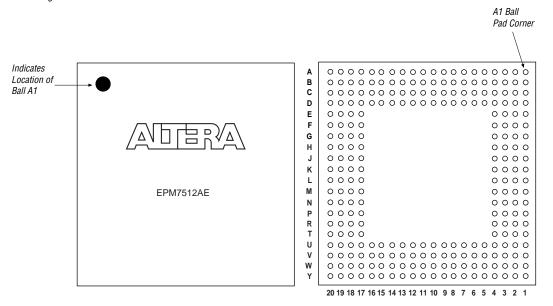
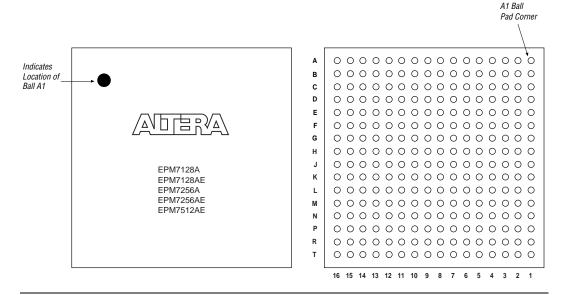


Figure 23. 256-Pin FineLine BGA Package Pin-Out Diagram



Revision History

The information contained in the *MAX 7000A Programmable Logic Device Data Sheet* version 4.5 supersedes information published in previous versions.

Version 4.5

The following changes were made in the MAX 7000A Programmable Logic Device Data Sheet version 4.5:

Updated text in the "Power Sequencing & Hot-Socketing" section.

Version 4.4

The following changes were made in the MAX 7000A Programmable Logic Device Data Sheet version 4.4:

- Added Tables 5 through 7.
- Added "Programming Sequence" on page 17 and "Programming Times" on page 18.