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**Understanding Embedded - CPLDs (Complex Programmable Logic Devices)** 

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

### **Applications of Embedded - CPLDs**

Details	
roduct Status	Active
rogrammable Type	In System Programmable
elay Time tpd(1) Max	12 ns
oltage Supply - Internal	3V ~ 3.6V
umber of Logic Elements/Blocks	32
umber of Macrocells	512
umber of Gates	10000
imber of I/O	176
perating Temperature	0°C ~ 70°C (TA)
ounting Type	Surface Mount
ckage / Case	208-BFQFP
upplier Device Package	208-PQFP (28x28)
ırchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=epm7512aegc208-12n

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

# Functional Description

The MAX 7000A architecture includes the following elements:

- Logic array blocks (LABs)
- Macrocells
- Expander product terms (shareable and parallel)
- Programmable interconnect array
- I/O control blocks

The MAX 7000A architecture includes four dedicated inputs that can be used as general-purpose inputs or as high-speed, global control signals (clock, clear, and two output enable signals) for each macrocell and I/O pin. Figure 1 shows the architecture of MAX 7000A devices.

For registered functions, each macrocell flipflop can be individually programmed to implement D, T, JK, or SR operation with programmable clock control. The flipflop can be bypassed for combinatorial operation. During design entry, the designer specifies the desired flipflop type; the Altera software then selects the most efficient flipflop operation for each registered function to optimize resource utilization.

Each programmable register can be clocked in three different modes:

- Global clock signal. This mode achieves the fastest clock-to-output performance.
- Global clock signal enabled by an active-high clock enable. A clock enable is generated by a product term. This mode provides an enable on each flipflop while still achieving the fast clock-to-output performance of the global clock.
- Array clock implemented with a product term. In this mode, the flipflop can be clocked by signals from buried macrocells or I/O pins.

Two global clock signals are available in MAX 7000A devices. As shown in Figure 1, these global clock signals can be the true or the complement of either of the global clock pins, GCLK1 or GCLK2.

Each register also supports asynchronous preset and clear functions. As shown in Figure 2, the product-term select matrix allocates product terms to control these operations. Although the product-term-driven preset and clear from the register are active high, active-low control can be obtained by inverting the signal within the logic array. In addition, each register clear function can be individually driven by the active-low dedicated global clear pin (GCLRn). Upon power-up, each register in a MAX 7000AE device may be set to either a high or low state. This power-up state is specified at design entry. Upon power-up, each register in EPM7128A and EPM7256A devices are set to a low state.

All MAX 7000A I/O pins have a fast input path to a macrocell register. This dedicated path allows a signal to bypass the PIA and combinatorial logic and be clocked to an input D flipflop with an extremely fast (as low as 2.5 ns) input setup time.

# In-System Programmability

MAX 7000A devices can be programmed in-system via an industry-standard 4-pin IEEE Std. 1149.1 (JTAG) interface. ISP offers quick, efficient iterations during design development and debugging cycles. The MAX 7000A architecture internally generates the high programming voltages required to program EEPROM cells, allowing in-system programming with only a single 3.3-V power supply. During in-system programming, the I/O pins are tri-stated and weakly pulled-up to eliminate board conflicts. The pull-up value is nominally 50 k $\Omega$ .

MAX 7000AE devices have an enhanced ISP algorithm for faster programming. These devices also offer an ISP\_Done bit that provides safe operation when in-system programming is interrupted. This ISP\_Done bit, which is the last bit programmed, prevents all I/O pins from driving until the bit is programmed. This feature is only available in EPM7032AE, EPM7064AE, EPM7128AE, EPM7256AE, and EPM7512AE devices.

ISP simplifies the manufacturing flow by allowing devices to be mounted on a PCB with standard pick-and-place equipment before they are programmed. MAX 7000A devices can be programmed by downloading the information via in-circuit testers, embedded processors, the Altera MasterBlaster serial/USB communications cable, ByteBlasterMV parallel port download cable, and BitBlaster serial download cable. Programming the devices after they are placed on the board eliminates lead damage on high-pin-count packages (e.g., QFP packages) due to device handling. MAX 7000A devices can be reprogrammed after a system has already shipped to the field. For example, product upgrades can be performed in the field via software or modem.

In-system programming can be accomplished with either an adaptive or constant algorithm. An adaptive algorithm reads information from the unit and adapts subsequent programming steps to achieve the fastest possible programming time for that unit. A constant algorithm uses a predefined (non-adaptive) programming sequence that does not take advantage of adaptive algorithm programming time improvements. Some in-circuit testers cannot program using an adaptive algorithm. Therefore, a constant algorithm must be used. MAX 7000AE devices can be programmed with either an adaptive or constant (non-adaptive) algorithm. EPM7128A and EPM7256A device can only be programmed with an adaptive algorithm; users programming these two devices on platforms that cannot use an adaptive algorithm should use EPM7128AE and EPM7256AE devices.

The Jam Standard Test and Programming Language (STAPL), JEDEC standard JESD 71, can be used to program MAX 7000A devices with incircuit testers, PCs, or embedded processors.



For more information on using the Jam STAPL language, see *Application Note 88* (Using the Jam Language for ISP & ICR via an Embedded Processor) and *Application Note 122* (Using Jam STAPL for ISP & ICR via an Embedded Processor).

ISP circuitry in MAX 7000AE devices is compliant with the IEEE Std. 1532 specification. The IEEE Std. 1532 is a standard developed to allow concurrent ISP between multiple PLD vendors.

## **Programming Sequence**

During in-system programming, instructions, addresses, and data are shifted into the MAX 7000A device through the TDI input pin. Data is shifted out through the TDO output pin and compared against the expected data.

Programming a pattern into the device requires the following six ISP stages. A stand-alone verification of a programmed pattern involves only stages 1, 2, 5, and 6.

- Enter ISP. The enter ISP stage ensures that the I/O pins transition smoothly from user mode to ISP mode. The enter ISP stage requires 1 ms.
- 2. *Check ID*. Before any program or verify process, the silicon ID is checked. The time required to read this silicon ID is relatively small compared to the overall programming time.
- 3. *Bulk Erase*. Erasing the device in-system involves shifting in the instructions to erase the device and applying one erase pulse of 100 ms.
- Program. Programming the device in-system involves shifting in the address and data and then applying the programming pulse to program the EEPROM cells. This process is repeated for each EEPROM address.
- Verify. Verifying an Altera device in-system involves shifting in addresses, applying the read pulse to verify the EEPROM cells, and shifting out the data for comparison. This process is repeated for each EEPROM address.
- 6. Exit ISP. An exit ISP stage ensures that the I/O pins transition smoothly from ISP mode to user mode. The exit ISP stage requires 1 ms.

The programming times described in Tables 5 through 7 are associated with the worst-case method using the enhanced ISP algorithm.

Device	Progra	mming	Stand-Alone	Verification
	t <sub>PPULSE</sub> (s)	Cycle <sub>PTCK</sub>	t <sub>VPULSE</sub> (s)	Cycle <sub>VTCK</sub>
EPM7032AE	2.00	55,000	0.002	18,000
EPM7064AE	2.00	105,000	0.002	35,000
EPM7128AE	2.00	205,000	0.002	68,000
EPM7256AE	2.00	447,000	0.002	149,000
EPM7512AE	2.00	890,000	0.002	297,000
EPM7128A (1)	5.11	832,000	0.03	528,000
EPM7256A (1)	6.43	1,603,000	0.03	1,024,000

Tables 6 and 7 show the in-system programming and stand alone verification times for several common test clock frequencies.

Device				1	TCK				Units
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EPM7032AE	2.01	2.01	2.03	2.06	2.11	2.28	2.55	3.10	s
EPM7064AE	2.01	2.02	2.05	2.11	2.21	2.53	3.05	4.10	S
EPM7128AE	2.02	2.04	2.10	2.21	2.41	3.03	4.05	6.10	s
EPM7256AE	2.05	2.09	2.23	2.45	2.90	4.24	6.47	10.94	S
EPM7512AE	2.09	2.18	2.45	2.89	3.78	6.45	10.90	19.80	S
EPM7128A (1)	5.19	5.27	5.52	5.94	6.77	9.27	13.43	21.75	s
EPM7256A (1)	6.59	6.75	7.23	8.03	9.64	14.45	22.46	38.49	S

Table 8. MAX 7000A	JTAG Instructions
JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern output at the device pins
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation
IDCODE	Selects the IDCODE register and places it between the TDI and TDO pins, allowing the IDCODE to be serially shifted out of TDO
USERCODE	Selects the 32-bit USERCODE register and places it between the TDI and TDO pins, allowing the USERCODE value to be shifted out of TDO. The USERCODE instruction is available for MAX 7000AE devices only
UESCODE	These instructions select the user electronic signature (UESCODE) and allow the UESCODE to be shifted out of TDO. UESCODE instructions are available for EPM7128A and EPM7256A devices only.
ISP Instructions	These instructions are used when programming MAX 7000A devices via the JTAG ports with the MasterBlaster, ByteBlasterMV, or BitBlaster download cable, or using a Jam STAPL File, JBC File, or SVF File via an embedded processor or test equipment.

The instruction register length of MAX 7000A devices is 10 bits. The user electronic signature (UES) register length in MAX 7000A devices is 16 bits. The MAX 7000AE USERCODE register length is 32 bits. Tables 9 and 10 show the boundary-scan register length and device IDCODE information for MAX 7000A devices.

Table 9. MAX 7000A Boundary-So	can Register Length
Device	Boundary-Scan Register Length
EPM7032AE	96
EPM7064AE	192
EPM7128A	288
EPM7128AE	288
EPM7256A	480
EPM7256AE	480
EPM7512AE	624

Table 10. 32	Table 10. 32-Bit MAX 7000A Device IDCODE Note (1)												
Device		IDCODE (32 E	Bits)										
	Version (4 Bits)	Part Number (16 Bits)	Manufacturer's Identity (11 Bits)	1 (1 Bit) (2)									
EPM7032AE	0001	0111 0000 0011 0010	00001101110	1									
EPM7064AE	0001	0111 0000 0110 0100	00001101110	1									
EPM7128A	0000	0111 0001 0010 1000	00001101110	1									
EPM7128AE	0001	0111 0001 0010 1000	00001101110	1									
EPM7256A	0000	0111 0010 0101 0110	00001101110	1									
EPM7256AE	0001	0111 0010 0101 0110	00001101110	1									
EPM7512AE	0001	0111 0101 0001 0010	00001101110	1									

#### Notes:

- (1) The most significant bit (MSB) is on the left.
- (2) The least significant bit (LSB) for all JTAG IDCODEs is 1.



See Application Note 39 (IEEE 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices) for more information on JTAG BST.

Figure 10 shows the typical output drive characteristics of MAX 7000A devices.

3.3 V MAX 7000AE Devices 2.5 V MAX 7000AE Devices 150 150 100 100 V<sub>CCINT</sub> = 3.3 V Typical I<sub>O</sub> Typical I<sub>O</sub>  $V_{CCINT} = 3.3 V$ Output Output  $V_{CCIO} = 3.3 V$  $V_{CCIO} = 2.5 \text{ V}$ Current (mA) Current (mA) Temperature = 25 °C Temperature = 25 °C 50 50  $I_{\cap H}$ 0 VO Output Voltage (V) Vo Output Voltage (V) EPM7128A & EPM7256A Devices 3.3 V 2.5 V EPM7128A & EPM7256A Devices 120 120  $I_{OL}$  $I_{OL}$ 80 Typical I<sub>O</sub> V<sub>CCINT</sub> = 3.3 V Typical I<sub>O</sub> V<sub>CCINT</sub> = 3.3 V Output Output  $V_{CCIO} = 3.3 V$  $V_{CCIO} = 2.5 V$ Temperature = 25°C Current (mA) Current (mA) Temperature = 25 °C

Figure 10. Output Drive Characteristics of MAX 7000A Devices

V<sub>O</sub> Output Voltage (V)

# **Timing Model**

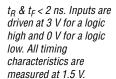
MAX 7000A device timing can be analyzed with the Altera software, a variety of popular industry-standard EDA simulators and timing analyzers, or with the timing model shown in Figure 11. MAX 7000A devices have predictable internal delays that enable the designer to determine the worst-case timing of any design. The software provides timing simulation, point-to-point delay prediction, and detailed timing analysis for device-wide performance evaluation.

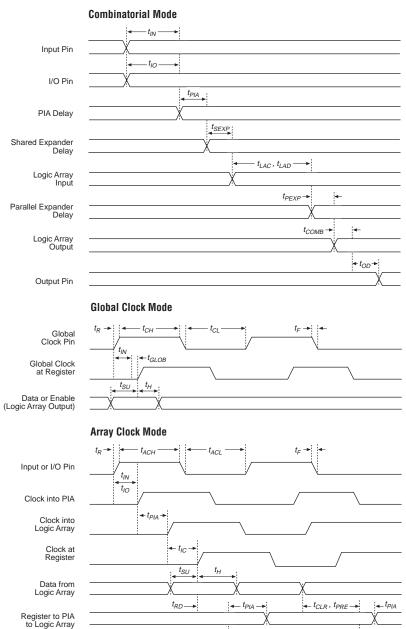
40

 $I_{OH}$ 

Vo Output Voltage (V)

## Figure 12. MAX 7000A Switching Waveforms





 $-t_{OD}$ 

**←** t<sub>OD</sub> -

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Register Output to Pin

Tables 17 through 30 show EPM7032AE, EPM7064AE, EPM7128AE, EPM7256AE, EPM7512AE, EPM7128A, and EPM7256A timing information.

Symbol	Parameter	Conditions			Speed	Grade			Unit
				4		7	-1	10	
			Min	Max	Min	Max	Min	Max	
t <sub>PD1</sub>	Input to non-registered output	C1 = 35 pF (2)		4.5		7.5		10	ns
t <sub>PD2</sub>	I/O input to non-registered output	C1 = 35 pF (2)		4.5		7.5		10	ns
t <sub>SU</sub>	Global clock setup time	(2)	2.9		4.7		6.3		ns
t <sub>H</sub>	Global clock hold time	(2)	0.0		0.0		0.0		ns
t <sub>FSU</sub>	Global clock setup time of fast input		2.5		3.0		3.0		ns
t <sub>FH</sub>	Global clock hold time of fast input		0.0		0.0		0.0		ns
t <sub>CO1</sub>	Global clock to output delay	C1 = 35 pF	1.0	3.0	1.0	5.0	1.0	6.7	ns
t <sub>CH</sub>	Global clock high time		2.0		3.0		4.0		ns
t <sub>CL</sub>	Global clock low time		2.0		3.0		4.0		ns
t <sub>ASU</sub>	Array clock setup time	(2)	1.6		2.5		3.6		ns
t <sub>AH</sub>	Array clock hold time	(2)	0.3		0.5		0.5		ns
t <sub>ACO1</sub>	Array clock to output delay	C1 = 35 pF (2)	1.0	4.3	1.0	7.2	1.0	9.4	ns
t <sub>ACH</sub>	Array clock high time		2.0		3.0		4.0		ns
t <sub>ACL</sub>	Array clock low time		2.0		3.0		4.0		ns
t <sub>CPPW</sub>	Minimum pulse width for clear and preset	(3)	2.0		3.0		4.0		ns
t <sub>CNT</sub>	Minimum global clock period	(2)		4.4		7.2		9.7	ns
f <sub>CNT</sub>	Maximum internal global clock frequency	(2), (4)	227.3		138.9		103.1		MHz
t <sub>ACNT</sub>	Minimum array clock period	(2)		4.4		7.2		9.7	ns
f <sub>ACNT</sub>	Maximum internal array clock frequency	(2), (4)	227.3		138.9		103.1		MHz

Symbol	Parameter	Conditions			Speed	Grade			Unit
			-	5		-7		10	
			Min	Max	Min	Max	Min	Max	
t <sub>IN</sub>	Input pad and buffer delay			0.7		1.0		1.4	ns
$t_{IO}$	I/O input pad and buffer delay			0.7		1.0		1.4	ns
t <sub>FIN</sub>	Fast input delay			2.5		3.0		3.4	ns
t <sub>SEXP</sub>	Shared expander delay			2.0		2.9		3.8	ns
t <sub>PEXP</sub>	Parallel expander delay			0.4		0.7		0.9	ns
$t_{LAD}$	Logic array delay			1.6		2.4		3.1	ns
t <sub>LAC</sub>	Logic control array delay			0.7		1.0		1.3	ns
t <sub>IOE</sub>	Internal output enable delay			0.0		0.0		0.0	ns
t <sub>OD1</sub>	Output buffer and pad delay, slow slew rate = off V <sub>CCIO</sub> = 3.3 V	C1 = 35 pF		0.8		1.2		1.6	ns
t <sub>OD2</sub>	Output buffer and pad delay, slow slew rate = off V <sub>CCIO</sub> = 2.5 V	C1 = 35 pF (5)		1.3		1.7		2.1	ns
t <sub>OD3</sub>	Output buffer and pad delay, slow slew rate = on V <sub>CCIO</sub> = 2.5 V or 3.3 V	C1 = 35 pF		5.8		6.2		6.6	ns
t <sub>ZX1</sub>	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		4.0		4.0		5.0	ns
t <sub>ZX2</sub>	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 2.5 \text{ V}$	C1 = 35 pF (5)		4.5		4.5		5.5	ns
t <sub>ZX3</sub>	Output buffer enable delay, slow slew rate = on V <sub>CCIO</sub> = 3.3 V	C1 = 35 pF		9.0		9.0		10.0	ns
$t_{XZ}$	Output buffer disable delay	C1 = 5 pF		4.0		4.0		5.0	ns
$t_{SU}$	Register setup time		1.4		2.1		2.9		ns
$t_H$	Register hold time		0.6		1.0		1.3		ns
t <sub>FSU</sub>	Register setup time of fast input		1.1		1.6		1.6		ns
t <sub>FH</sub>	Register hold time of fast input		1.4		1.4		1.4		ns
$t_{RD}$	Register delay			0.8		1.2		1.6	ns
t <sub>COMB</sub>	Combinatorial delay			0.5		0.9		1.3	ns
$t_{IC}$	Array clock delay			1.2		1.7		2.2	ns

Symbol	Parameter	Conditions	Speed Grade							
			-5		-7		-10		-	
			Min	Max	Min	Max	Min	Max		
t <sub>EN</sub>	Register enable time			0.7		1.0		1.3	ns	
$t_{GLOB}$	Global control delay			1.1		1.6		2.0	ns	
t <sub>PRE</sub>	Register preset time			1.4		2.0		2.7	ns	
t <sub>CLR</sub>	Register clear time			1.4		2.0		2.7	ns	
$t_{PIA}$	PIA delay	(2)		1.4		2.0		2.6	ns	
$t_{LPA}$	Low-power adder	(6)		4.0		4.0		5.0	ns	

Symbol	Parameter	Conditions			Speed	Grade			Unit
		·	-!	5	-	7	-1	10	
			Min	Max	Min	Max	Min	Max	Ē
t <sub>PD1</sub>	Input to non- registered output	C1 = 35 pF (2)		5.5		7.5		10	ns
t <sub>PD2</sub>	I/O input to non- registered output	C1 = 35 pF (2)		5.5		7.5		10	ns
t <sub>SU</sub>	Global clock setup time	(2)	3.9		5.2		6.9		ns
t <sub>H</sub>	Global clock hold time	(2)	0.0		0.0		0.0		ns
t <sub>FSU</sub>	Global clock setup time of fast input		2.5		3.0		3.0		ns
t <sub>FH</sub>	Global clock hold time of fast input		0.0		0.0		0.0		ns
t <sub>CO1</sub>	Global clock to output delay	C1 = 35 pF	1.0	3.5	1.0	4.8	1.0	6.4	ns
t <sub>CH</sub>	Global clock high time		2.0		3.0		4.0		ns
t <sub>CL</sub>	Global clock low time		2.0		3.0		4.0		ns
t <sub>ASU</sub>	Array clock setup time	(2)	2.0		2.7		3.6		ns
t <sub>AH</sub>	Array clock hold time	(2)	0.2		0.3		0.5		ns
t <sub>ACO1</sub>	Array clock to output delay	C1 = 35 pF (2)	1.0	5.4	1.0	7.3	1.0	9.7	ns
t <sub>ACH</sub>	Array clock high time		2.0		3.0		4.0		ns
t <sub>ACL</sub>	Array clock low time		2.0		3.0		4.0		ns
t <sub>CPPW</sub>	Minimum pulse width for clear and preset	(3)	2.0		3.0		4.0		ns
t <sub>CNT</sub>	Minimum global clock period	(2)		5.8		7.9		10.5	ns
f <sub>CNT</sub>	Maximum internal global clock frequency	(2), (4)	172.4		126.6		95.2		MHz
t <sub>ACNT</sub>	Minimum array clock period	(2)		5.8		7.9		10.5	ns
f <sub>ACNT</sub>	Maximum internal array clock frequency	(2), (4)	172.4		126.6		95.2		MHz

Table 2	9. EPM7256A External Tir	ning Parame	ters	Note	(1)						
Symbol	Parameter	Conditions				Speed	Grade				Unit
			-	6		7	-1	10		12	Ė
			Min	Max	Min	Max	Min	Max	Min	Max	Ė
t <sub>PD1</sub>	Input to non-registered output	C1 = 35 pF (2)		6.0		7.5		10.0		12.0	ns
t <sub>PD2</sub>	I/O input to non- registered output	C1 = 35 pF (2)		6.0		7.5		10.0		12.0	ns
t <sub>SU</sub>	Global clock setup time	(2)	3.7		4.6		6.2		7.4		ns
t <sub>H</sub>	Global clock hold time	(2)	0.0		0.0		0.0		0.0		ns
t <sub>FSU</sub>	Global clock setup time of fast input		2.5		3.0		3.0		3.0		ns
t <sub>FH</sub>	Global clock hold time of fast input		0.0		0.0		0.0		0.0		ns
t <sub>CO1</sub>	Global clock to output delay	C1 = 35 pF	1.0	3.3	1.0	4.2	1.0	5.5	1.0	6.6	ns
t <sub>CH</sub>	Global clock high time		3.0		3.0		4.0		4.0		ns
t <sub>CL</sub>	Global clock low time		3.0		3.0		4.0		4.0		ns
t <sub>ASU</sub>	Array clock setup time	(2)	8.0		1.0		1.4		1.6		ns
t <sub>AH</sub>	Array clock hold time	(2)	1.9		2.7		4.0		5.1		ns
t <sub>ACO1</sub>	Array clock to output delay	C1 = 35 pF (2)	1.0	6.2	1.0	7.8	1.0	10.3	1.0	12.4	ns
t <sub>ACH</sub>	Array clock high time		3.0		3.0		4.0		4.0		ns
t <sub>ACL</sub>	Array clock low time		3.0		3.0		4.0		4.0		ns
t <sub>CPPW</sub>	Minimum pulse width for clear and preset	(3)	3.0		3.0		4.0		4.0		ns
t <sub>CNT</sub>	Minimum global clock period	(2)		6.4		8.0		10.7		12.8	ns
f <sub>CNT</sub>	Maximum internal global clock frequency	(2), (4)	156.3		125.0		93.5		78.1		MHz
t <sub>ACNT</sub>	Minimum array clock period	(2)		6.4		8.0		10.7		12.8	ns
f <sub>ACNT</sub>	Maximum internal array clock frequency	(2), (4)	156.3		125.0		93.5		78.1		MHz

Symbol	Parameter	Conditions	Speed Grade								
			-6		-7		-10		-1	12	
			Min	Max	Min	Max	Min	Max	Min	Max	
$t_{COMB}$	Combinatorial delay			1.6		2.0		2.7		3.2	ns
t <sub>IC</sub>	Array clock delay			2.7		3.4		4.5		5.4	ns
t <sub>EN</sub>	Register enable time			2.5		3.1		4.2		5.0	ns
t <sub>GLOB</sub>	Global control delay			1.1		1.4		1.8		2.2	ns
t <sub>PRE</sub>	Register preset time			2.3		2.9		3.8		4.6	ns
t <sub>CLR</sub>	Register clear time			2.3		2.9		3.8		4.6	ns
$t_{PIA}$	PIA delay	(2)		1.3		1.6		2.1		2.6	ns
$t_{LPA}$	Low-power adder	(6)		11.0		10.0		10.0		10.0	ns

#### Notes to tables:

- (1) These values are specified under the recommended operating conditions shown in Table 14 on page 28. See Figure 12 for more information on switching waveforms.
- (2) These values are specified for a PIA fan-out of one LAB (16 macrocells). For each additional LAB fan-out in these devices, add an additional 0.1 ns to the PIA timing value.
- (3) This minimum pulse width for preset and clear applies for both global clear and array controls. The t<sub>LPA</sub> parameter must be added to this minimum width if the clear or reset signal incorporates the t<sub>LAD</sub> parameter into the signal path.
- (4) This parameter is measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
- (5) Operating conditions:  $V_{CCIO} = 2.5 \pm 0.2 \text{ V}$  for commercial and industrial use.
- (6) The  $t_{LPA}$  parameter must be added to the  $t_{LAD}$ ,  $t_{LAC}$ ,  $t_{IC}$ ,  $t_{EN}$ ,  $t_{SEXP}$ ,  $t_{ACL}$ , and  $t_{CPPW}$  parameters for macrocells running in low-power mode.

# Power Consumption

Supply power (P) versus frequency ( $f_{MAX}$ , in MHz) for MAX 7000A devices is calculated with the following equation:

$$P = P_{INT} + P_{IO} = I_{CCINT} \times V_{CC} + P_{IO}$$

The  $P_{IO}$  value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note* 74 (*Evaluating Power for Altera Devices*).

The  $I_{CCINT}$  value depends on the switching frequency and the application logic. The  $I_{CCINT}$  value is calculated with the following equation:

$$I_{CCINT} =$$

$$(A \times MC_{TON}) + [B \times (MC_{DEV} - MC_{TON})] + (C \times MC_{USED} \times f_{\boldsymbol{MAX}} \times \boldsymbol{tog_{LC}})$$

The parameters in this equation are:

MC<sub>TON</sub> = Number of macrocells with the Turbo Bit option turned on, as reported in the MAX+PLUS II Report File (.rpt)

 $MC_{DEV}$  = Number of macrocells in the device

 $MC_{USED}$  = Total number of macrocells in the design, as reported in

the Report File

 $f_{MAX}$  = Highest clock frequency to the device

tog<sub>LC</sub> = Average percentage of logic cells toggling at each clock

(typically 12.5%)

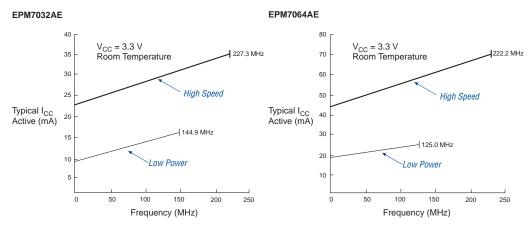
A, B, C = Constants, shown in Table 31

Table 31. MAX 7000A I <sub>CC</sub> Equation Constants			
Device	A	В	C
EPM7032AE	0.71	0.30	0.014
EPM7064AE	0.71	0.30	0.014
EPM7128A	0.71	0.30	0.014
EPM7128AE	0.71	0.30	0.014
EPM7256A	0.71	0.30	0.014
EPM7256AE	0.71	0.30	0.014
EPM7512AE	0.71	0.30	0.014

This calculation provides an  $I_{CC}$  estimate based on typical conditions using a pattern of a 16-bit, loadable, enabled, up/down counter in each LAB with no output load. Actual  $I_{CC}$  should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

Figure 13 shows the typical supply current versus frequency for MAX 7000A devices.

Figure 13. I<sub>CC</sub> vs. Frequency for MAX 7000A Devices (Part 1 of 2)



#### EPM7128A & EPM7128AE

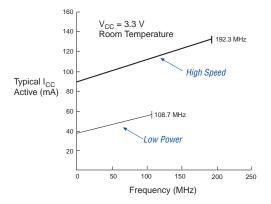


Figure 15. 49-Pin Ultra FineLine BGA Package Pin-Out Diagram

Package outlines not drawn to scale.

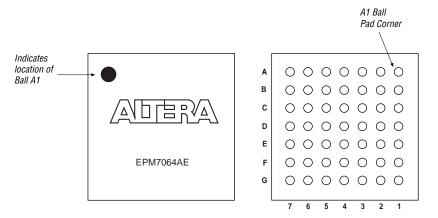


Figure 16. 84-Pin PLCC Package Pin-Out Diagram

Package outline not drawn to scale.

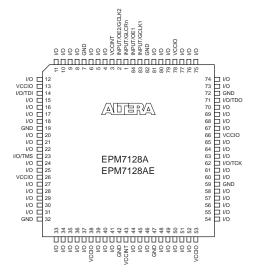
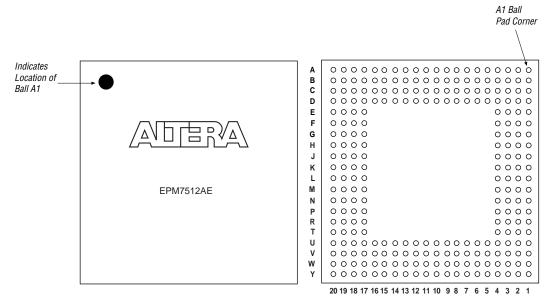


Figure 22. 256-Pin BGA Package Pin-Out Diagram

Package outline not drawn to scale.



#### Version 4.3

The following changes were made in the MAX 7000A Programmable Logic Device Data Sheet version 4.3:

- Added extended temperature devices to document
- Updated Table 14.

### Version 4.2

The following changes were made in the MAX 7000A Programmable Logic Device Data Sheet version 4.2:

- Removed *Note* (1) from Table 2.
- Removed *Note* (4) from Tables 3 and 4.

#### Version 4.1

The following changes were made in the MAX 7000A Programmable Logic Device Data Sheet version 4.1:

- Updated leakage current information in Table 15.
- Updated Note (9) of Table 15.
- Updated *Note* (1) of Tables 17 through 30.



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Altera customers are advised to obtain the latest version of device specifications before relying on any published information and before placing orders for products or services.

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