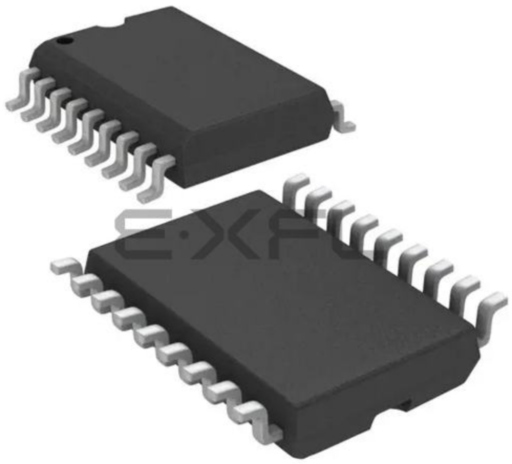

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	-
Peripherals	POR, WDT
Number of I/O	13
Program Memory Size	1.75KB (1K x 14)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	36 x 8
Voltage - Supply (Vcc/Vdd)	4V ~ 6V
Data Converters	A/D 4x8b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	18-SOIC (0.295", 7.50mm Width)
Supplier Device Package	18-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16c71-20i-so

TABLE 1-1: PIC16C71X FAMILY OF DEVICES

		PIC16C710	PIC16C71	PIC16C711	PIC16C715	PIC16C72	PIC16CR72[(1)]
Clock	Maximum Frequency of Operation (MHz)	20	20	20	20	20	20
Memory	EPROM Program Memory (x14 words)	512	1K	1K	2K	2K	—
	ROM Program Memory (14K words)	—	—	—	—	—	2K
	Data Memory (bytes)	36	36	68	128	128	128
Peripherals	Timer Module(s)	TMR0	TMR0	TMR0	TMR0	TMR0, TMR1, TMR2	TMR0, TMR1, TMR2
	Capture/Compare/PWM Module(s)	—	—	—	—	1	1
	Serial Port(s) (SPI/I^2C, USART)	—	—	—	—	SPI/I^2C	SPI/I^2C
	Parallel Slave Port	—	—	—	—	—	—
	A/D Converter (8-bit) Channels	4	4	4	4	5	5
Features	Interrupt Sources	4	4	4	4	8	8
	I/O Pins	13	13	13	13	22	22
	Voltage Range (Volts)	2.5-6.0	3.0-6.0	2.5-6.0	2.5-5.5	2.5-6.0	3.0-5.5
	In-Circuit Serial Programming	Yes	Yes	Yes	Yes	Yes	Yes
	Brown-out Reset	Yes	—	Yes	Yes	Yes	Yes
	Packages	18-pin DIP, SOIC; 20-pin SSOP	18-pin DIP, SOIC	18-pin DIP, SOIC; 20-pin SSOP	18-pin DIP, SOIC; 20-pin SSOP	28-pin SDIP, SOIC, SSOP	28-pin SDIP, SOIC, SSOP

		PIC16C73A	PIC16C74A	PIC16C76	PIC16C77
Clock	Maximum Frequency of Operation (MHz)	20	20	20	20
Memory	EPROM Program Memory (x14 words)	4K	4K	8K	8K
	Data Memory (bytes)	192	192	376	376
Peripherals	Timer Module(s)	TMR0, TMR1, TMR2	TMR0, TMR1, TMR2	TMR0, TMR1, TMR2	TMR0, TMR1, TMR2
	Capture/Compare/PWM Module(s)	2	2	2	2
	Serial Port(s) (SPI/I^2C, USART)	SPI/I^2C, USART	SPI/I^2C, USART	SPI/I^2C, USART	SPI/I^2C, USART
	Parallel Slave Port	—	Yes	—	Yes
	A/D Converter (8-bit) Channels	5	8	5	8
Features	Interrupt Sources	11	12	11	12
	I/O Pins	22	33	22	33
	Voltage Range (Volts)	2.5-6.0	2.5-6.0	2.5-6.0	2.5-6.0
	In-Circuit Serial Programming	Yes	Yes	Yes	Yes
	Brown-out Reset	Yes	Yes	Yes	Yes
	Packages	28-pin SDIP, SOIC	40-pin DIP; 44-pin PLCC, MQFP, TQFP	28-pin SDIP, SOIC	40-pin DIP; 44-pin PLCC, MQFP, TQFP

All PIC16/17 Family devices have Power-on Reset, selectable Watchdog Timer, selectable code protect and high I/O current capability. All PIC16C7XX Family devices use serial programming with clock pin RB6 and data pin RB7.

Note 1: Please contact your local Microchip sales office for availability of these devices.

TABLE 4-2: PIC16C715 SPECIAL FUNCTION REGISTER SUMMARY

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR, PER	Value on all other resets (3)
Bank 0											
00h(1)	INDF	Addressing this location uses contents of FSR to address data memory (not a physical register)								0000 0000	0000 0000
01h	TMR0	Timer0 module's register								xxxx xxxx	uuuu uuuu
02h(1)	PCL	Program Counter's (PC) Least Significant Byte								0000 0000	0000 0000
03h(1)	STATUS	IRP(4)	RP1(4)	RP0	$\overline{\text{TO}}$	$\overline{\text{PD}}$	Z	DC	C	0001 1xxx	000q quuu
04h(1)	FSR	Indirect data memory address pointer								xxxx xxxx	uuuu uuuu
05h	PORTA	—	—	—	PORTA Data Latch when written: PORTA pins when read					---x 0000	---u 0000
06h	PORTB	PORTB Data Latch when written: PORTB pins when read								xxxx xxxx	uuuu uuuu
07h	—	Unimplemented								—	—
08h	—	Unimplemented								—	—
09h	—	Unimplemented								—	—
0Ah(1,2)	PCLATH	—	—	—	Write Buffer for the upper 5 bits of the Program Counter					---0 0000	---0 0000
0Bh(1)	INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	—	ADIF	—	—	—	—	—	—	-0-- ----	-0-- ----
0Dh	—	Unimplemented								—	—
0Eh	—	Unimplemented								—	—
0Fh	—	Unimplemented								—	—
10h	—	Unimplemented								—	—
11h	—	Unimplemented								—	—
12h	—	Unimplemented								—	—
13h	—	Unimplemented								—	—
14h	—	Unimplemented								—	—
15h	—	Unimplemented								—	—
16h	—	Unimplemented								—	—
17h	—	Unimplemented								—	—
18h	—	Unimplemented								—	—
19h	—	Unimplemented								—	—
1Ah	—	Unimplemented								—	—
1Bh	—	Unimplemented								—	—
1Ch	—	Unimplemented								—	—
1Dh	—	Unimplemented								—	—
1Eh	ADRES	A/D Result Register								xxxx xxxx	uuuu uuuu
1Fh	ADCON0	ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/$\overline{\text{DONE}}$	—	ADON	0000 00-0	0000 00-0

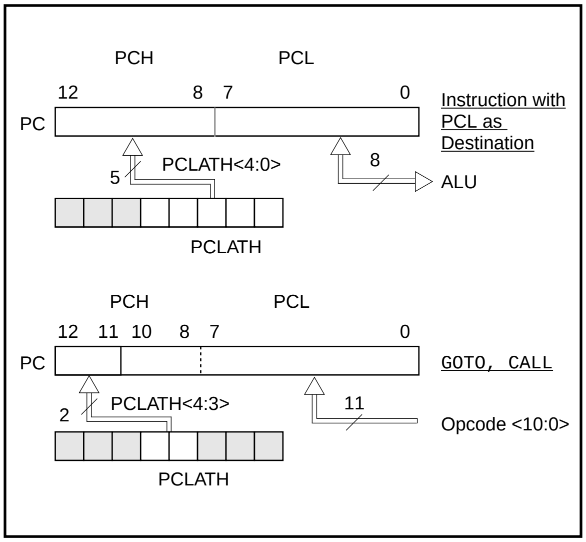
Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented read as '0'.
Shaded locations are unimplemented, read as '0'.

Note 1: These registers can be addressed from either bank.
2: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8> whose contents are transferred to the upper byte of the program counter.
3: Other (non power-up) resets include external reset through $\overline{\text{MCLR}}$ and Watchdog Timer Reset.
4: The IRP and RP1 bits are reserved on the PIC16C715, always maintain these bits clear.

4.3 PCL and PCLATH

The program counter (PC) is 13-bits wide. The low byte comes from the PCL register, which is a readable and writable register. The upper bits (PC<12:8>) are not readable, but are indirectly writable through the PCLATH register. On any reset, the upper bits of the PC will be cleared. Figure 4-14 shows the two situations for the loading of the PC. The upper example in the figure shows how the PC is loaded on a write to PCL (PCLATH<4:0> → PCH). The lower example in the figure shows how the PC is loaded during a `CALL` or `GOTO` instruction (PCLATH<4:3> → PCH).

FIGURE 4-14: LOADING OF PC IN DIFFERENT SITUATIONS



4.3.1 COMPUTED GOTO

A computed GOTO is accomplished by adding an offset to the program counter (`ADDWF PCL`). When doing a table read using a computed GOTO method, care should be exercised if the table location crosses a PCL memory boundary (each 256 byte block). Refer to the application note *"Implementing a Table Read"* (AN556).

4.3.2 STACK

The PIC16CXX family has an 8 level deep x 13-bit wide hardware stack. The stack space is not part of either program or data space and the stack pointer is not readable or writable. The PC is PUSHed onto the stack when a `CALL` instruction is executed or an interrupt causes a branch. The stack is POPed in the event of a `RETURN`, `RETLW` or a `RETFIE` instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer. This means that after the stack has been PUSHed eight times, the ninth push overwrites the value that was stored from the first push. The tenth push overwrites the second push (and so on).

Note 1: There are no status bits to indicate stack overflow or stack underflow conditions.

Note 2: There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the `CALL`, `RETURN`, `RETLW`, and `RETFIE` instructions, or the vectoring to an interrupt address.

4.4 Program Memory Paging

The PIC16C71X devices ignore both paging bits (PCLATH<4:3>, which are used to access program memory when more than one page is available. The use of PCLATH<4:3> as general purpose read/write bits for the PIC16C71X is not recommended since this may affect upward compatibility with future products.

6.3.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control, i.e., it can be changed “on the fly” during program execution.

Note:	To avoid an unintended device RESET, the following instruction sequence (shown in Example 6-1) must be executed when changing the prescaler assignment from Timer0 to the WDT. This sequence must be followed even if the WDT is disabled.

EXAMPLE 6-1: CHANGING PRESCALER (TIMER0→WDT)

```
BCF     STATUS, RP0  ;Bank 0
CLRF    TMR0         ;Clear TMR0 & Prescaler
BSF     STATUS, RP0  ;Bank 1
CLRWDT               ;Clears WDT
MOVLW   b'xxxx1xxx'  ;Selects new prescale value
MOVWF   OPTION_REG   ;and assigns the prescaler to the WDT
BCF     STATUS, RP0  ;Bank 0
```

To change prescaler from the WDT to the Timer0 module use the sequence shown in Example 6-2.

EXAMPLE 6-2: CHANGING PRESCALER (WDT→TIMER0)

```
CLRWDT                 ;Clear WDT and prescaler
BSF       STATUS, RP0 ;Bank 1
MOVLW     b'xxxx0xxx' ;Select TMR0, new prescale value and
MOVWF     OPTION_REG  ;clock source
BCF       STATUS, RP0 ;Bank 0
```

TABLE 6-1: REGISTERS ASSOCIATED WITH TIMER0

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets
01h	TMR0	Timer0 module’s register								xxxx xxxx	uuuu uuuu
0Bh,8Bh,	INTCON	GIE	ADIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000u
81h	OPTION	$\overline{\text{RBPU}}$	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
85h	TRISA	—	—	—	PORTA Data Direction Register					---1 1111	---1 1111

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by Timer0.

7.1 A/D Acquisition Requirements

For the A/D converter to meet its specified accuracy, the charge holding capacitor (C_{HOLD}) must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 7-5. The source impedance (R_S) and the internal sampling switch (R_{SS}) impedance directly affect the time required to charge the capacitor C_{HOLD}. The sampling switch (R_{SS}) impedance varies over the device voltage (V_{DD}), Figure 7-5. The source impedance affects the offset voltage at the analog input (due to pin leakage current). **The maximum recommended impedance for analog sources is 10 kΩ**. After the analog input channel is selected (changed) this acquisition must be done before the conversion can be started.

To calculate the minimum acquisition time, Equation 7-1 may be used. This equation calculates the acquisition time to within 1/2 LSb error is used (512 steps for the A/D). The 1/2 LSb error is the maximum error allowed for the A/D to meet its specified accuracy.

EQUATION 7-1: A/D MINIMUM CHARGING TIME

$$V_{HOLD} = (V_{REF} - (V_{REF}/512)) \cdot (1 - e^{(-T_{CAP}/C_{HOLD}(R_{IC} + R_{SS} + R_S))})$$

Given: $V_{HOLD} = (V_{REF}/512)$, for 1/2 LSb resolution

The above equation reduces to:

$$T_{CAP} = -(51.2\ pF)(1\ k\Omega + R_{SS} + R_S)\ \ln(1/511)$$

Example 7-1 shows the calculation of the minimum required acquisition time T_{ACQ}. This calculation is based on the following system assumptions.

C_{HOLD} = 51.2 pF

Rs = 10 kΩ

1/2 LSb error

V_{DD} = 5V → Rss = 7 kΩ

Temp (application system max.) = 50°C

V_{HOLD} = 0 @ t = 0

> **Note 1:** The reference voltage (V_{REF}) has no effect on the equation, since it cancels itself out.
>
> **Note 2:** The charge holding capacitor (C_{HOLD}) is not discharged after each conversion.
>
> **Note 3:** The maximum recommended impedance for analog sources is 10 kΩ. This is required to meet the pin leakage specification.
>
> **Note 4:** After a conversion has completed, a 2.0T_{AD} delay must complete before acquisition can begin again. During this time the holding capacitor is not connected to the selected A/D input channel.

EXAMPLE 7-1: CALCULATING THE MINIMUM REQUIRED AQUISITION TIME

T_{ACQ} = Amplifier Settling Time + Holding Capacitor Charging Time + Temperature Coefficient

T_{ACQ} = 5 µs + T_{CAP} + [(Temp - 25°C)(0.05 µs/°C)]

T_{CAP} = $-C_{HOLD}$ (R_{IC} + R_{SS} + R_S) ln(1/511)

-51.2 pF (1 kΩ + 7 kΩ + 10 kΩ) ln(0.0020)

-51.2 pF (18 kΩ) ln(0.0020)

-0.921 µs (-6.2364)

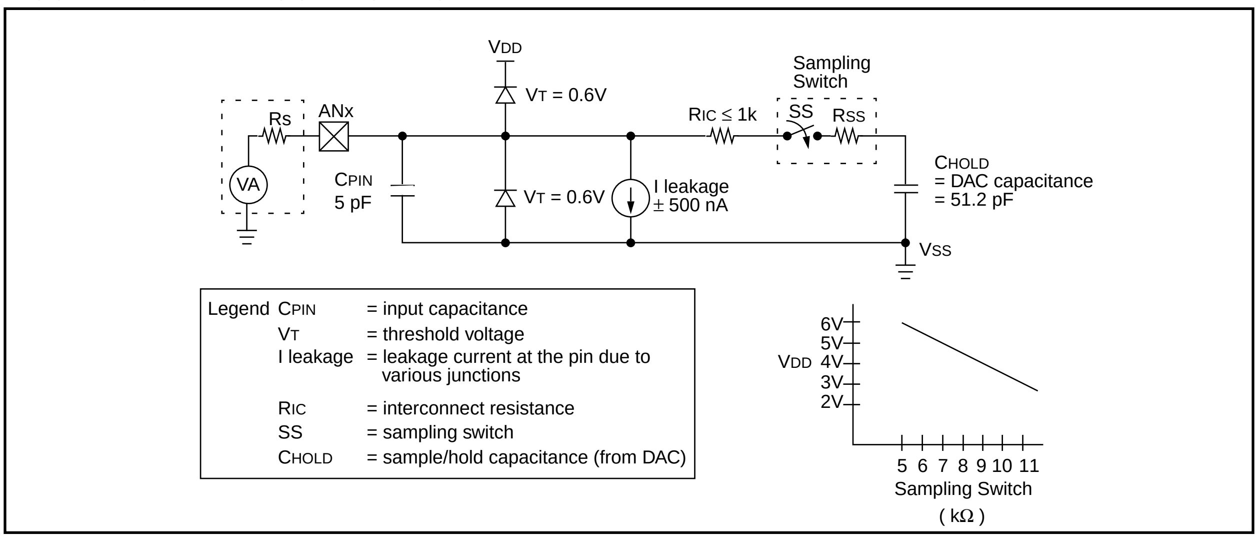
5.747 µs

T_{ACQ} = 5 µs + 5.747 µs + [(50°C - 25°C)(0.05 µs/°C)]

10.747 µs + 1.25 µs

11.997 µs

FIGURE 7-5: ANALOG INPUT MODEL



8.3 Reset

Applicable Devices	710	71	711	715

The PIC16CXX differentiates between various kinds of reset:

- Power-on Reset (POR)
- $\overline{\text{MCLR}}$ reset during normal operation
- $\overline{\text{MCLR}}$ reset during SLEEP
- WDT Reset (normal operation)
- Brown-out Reset (BOR) (PIC16C710/711/715)
- Parity Error Reset (PIC16C715)

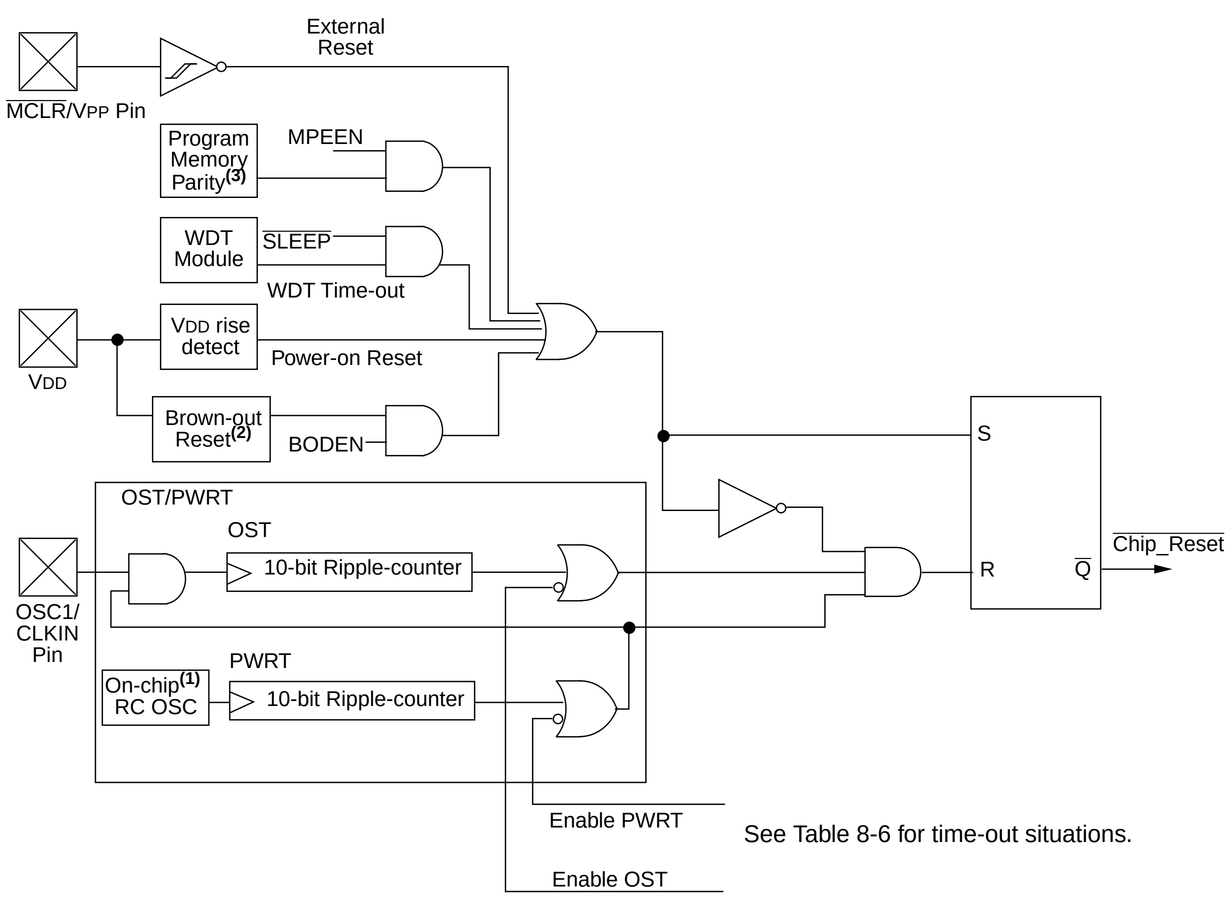
Some registers are not affected in any reset condition; their status is unknown on POR and unchanged in any other reset. Most other registers are reset to a “reset state” on Power-on Reset (POR), on the $\overline{\text{MCLR}}$ and WDT Reset, on $\overline{\text{MCLR}}$ reset during SLEEP, and Brown-out Reset (BOR). They are not affected by a WDT Wake-up, which is viewed as the resumption of normal operation. The $\overline{\text{TO}}$ and $\overline{\text{PD}}$ bits are set or cleared differently in different reset situations as indicated in Table 8-7, Table 8-8 and Table 8-9. These bits are used in software to determine the nature of the reset. See Table 8-10 and Table 8-11 for a full description of reset states of all registers.

A simplified block diagram of the on-chip reset circuit is shown in Figure 8-9.

The PIC16C710/711/715 have a $\overline{\text{MCLR}}$ noise filter in the $\overline{\text{MCLR}}$ reset path. The filter will detect and ignore small pulses.

It should be noted that a WDT Reset does not drive $\overline{\text{MCLR}}$ pin low.

FIGURE 8-9: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



Note 1: This is a separate oscillator from the RC oscillator of the CLKIN pin.
2: Brown-out Reset is implemented on the PIC16C710/711/715.
3: Parity Error Reset is implemented on the PIC16C715.

TABLE 8-12: INITIALIZATION CONDITIONS FOR ALL REGISTERS, PIC16C710/71/711

Register	Power-on Reset, Brown-out Reset(5)	$\overline{MCLR}$ Resets WDT Reset	Wake-up via WDT or Interrupt
W	xxxx xxxx	uuuu uuuu	uuuu uuuu
INDF	N/A	N/A	N/A
TMR0	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCL	0000h	0000h	PC + 1(2)
STATUS	0001 1xxx	000q quuu(3)	uuuq quuu(3)
FSR	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTA	---x 0000	---u 0000	---u uuuu
PORTB	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCLATH	---0 0000	---0 0000	---u uuuu
INTCON	0000 000x	0000 000u	uuuu uuuu(1)
ADRES	xxxx xxxx	uuuu uuuu	uuuu uuuu
ADCON0	00-0 0000	00-0 0000	uu-u uuuu
OPTION	1111 1111	1111 1111	uuuu uuuu
TRISA	---1 1111	---1 1111	---u uuuu
TRISB	1111 1111	1111 1111	uuuu uuuu
PCON(4)	---- --0u	---- --uu	---- --uu
ADCON1	---- --00	---- --00	---- --uu

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition

Note 1: One or more bits in INTCON will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

3: See Table 8-10 for reset value for specific condition.

4: The PCON register is not implemented on the PIC16C71.

5: Brown-out reset is not implemented on the PIC16C71.

8.7 Watchdog Timer (WDT)

Applicable Devices	710	71	711	715

The Watchdog Timer is as a free running on-chip RC oscillator which does not require any external components. This RC oscillator is separate from the RC oscillator of the OSC1/CLKIN pin. That means that the WDT will run, even if the clock on the OSC1/CLKIN and OSC2/CLKOUT pins of the device has been stopped, for example, by execution of a SLEEP instruction. During normal operation, a WDT time-out generates a device RESET (Watchdog Timer Reset). If the device is in SLEEP mode, a WDT time-out causes the device to wake-up and continue with normal operation (Watchdog Timer Wake-up). The WDT can be permanently disabled by clearing configuration bit WDTE (Section 8.1).

8.7.1 WDT PERIOD

The WDT has a nominal time-out period of 18 ms, (with no prescaler). The time-out periods vary with temperature, V_{DD} and process variations from part to part (see DC specs). If longer time-out periods are desired, a prescaler with a division ratio of up to 1:128 can be assigned to the WDT under software control by writing to the OPTION register. Thus, time-out periods up to 2.3 seconds can be realized.

The CLRWDT and SLEEP instructions clear the WDT and the postscaler, if assigned to the WDT, and prevent it from timing out and generating a device RESET condition.

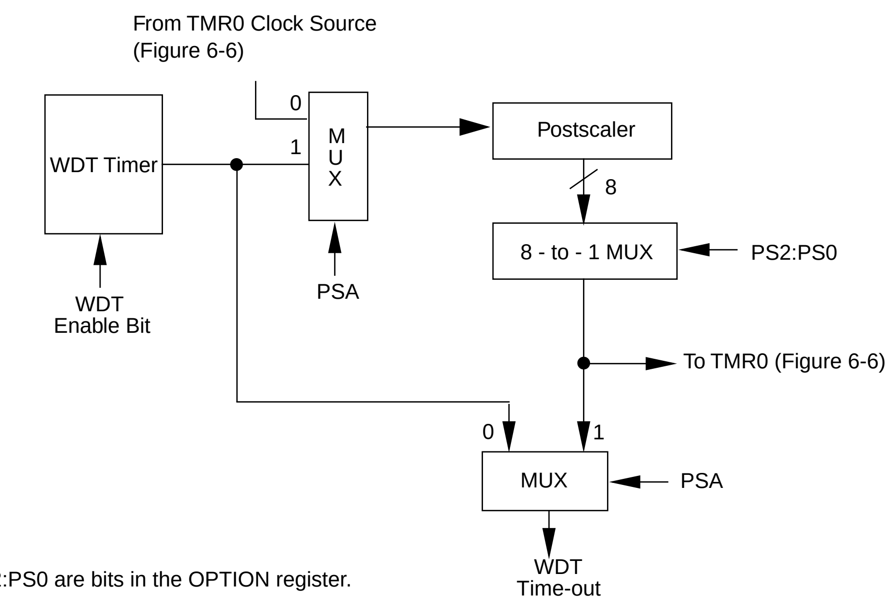
The $\overline{TO}$ bit in the STATUS register will be cleared upon a Watchdog Timer time-out.

8.7.2 WDT PROGRAMMING CONSIDERATIONS

It should also be taken into account that under worst case conditions (V_{DD} = Min., Temperature = Max., and max. WDT prescaler) it may take several seconds before a WDT time-out occurs.

> **Note:** When a CLRWDT instruction is executed and the prescaler is assigned to the WDT, the prescaler count will be cleared, but the prescaler assignment is not changed.

FIGURE 8-20: WATCHDOG TIMER BLOCK DIAGRAM



Note: PSA and PS2:PS0 are bits in the OPTION register.

FIGURE 8-21: SUMMARY OF WATCHDOG TIMER REGISTERS

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
2007h	Config. bits	**(1)**	BODEN(1)	CP1	CP0	PWRTE(1)	WDTE	FOSC1	FOSC0
81h,181h	OPTION	$\overline{RBPU}$	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0

Legend: Shaded cells are not used by the Watchdog Timer.
Note 1: See Figure 8-1, Figure 8-2 and Figure 8-3 for operation of these bits.

SUBWF — Subtract W from f

Syntax:	[*label*] SUBWF f,d
Operands:	0 ≤ f ≤ 127 d ∈ [0,1]
Operation:	(f) - (W) → (dest)
Status Affected:	C, DC, Z
Encoding:	`00 0010 dfff ffff`
Description:	Subtract (2's complement method) W register from register 'f'. If 'd' is 0 the result is stored in the W register. If 'd' is 1 the result is stored back in register 'f'.
Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process data	Write to dest

Example 1: `SUBWF REG1,1`

Before Instruction

REG1	=	3
W	=	2
C	=	?
Z	=	?

After Instruction

REG1	=	1
W	=	2
C	=	1; result is positive
Z	=	0

Example 2: Before Instruction

REG1	=	2
W	=	2
C	=	?
Z	=	?

After Instruction

REG1	=	0
W	=	2
C	=	1; result is zero
Z	=	1

Example 3: Before Instruction

REG1	=	1
W	=	2
C	=	?
Z	=	?

After Instruction

REG1	=	0xFF
W	=	2
C	=	0; result is negative
Z	=	0

SWAPF — Swap Nibbles in f

Syntax:	[*label*] SWAPF f,d
Operands:	0 ≤ f ≤ 127 d ∈ [0,1]
Operation:	(f<3:0>) → (dest<7:4>), (f<7:4>) → (dest<3:0>)
Status Affected:	None
Encoding:	`00 1110 dfff ffff`
Description:	The upper and lower nibbles of register 'f' are exchanged. If 'd' is 0 the result is placed in W register. If 'd' is 1 the result is placed in register 'f'.
Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process data	Write to dest

Example `SWAPF REG, 0`

Before Instruction

REG1	=	0xA5

After Instruction

REG1	=	0xA5
W	=	0x5A

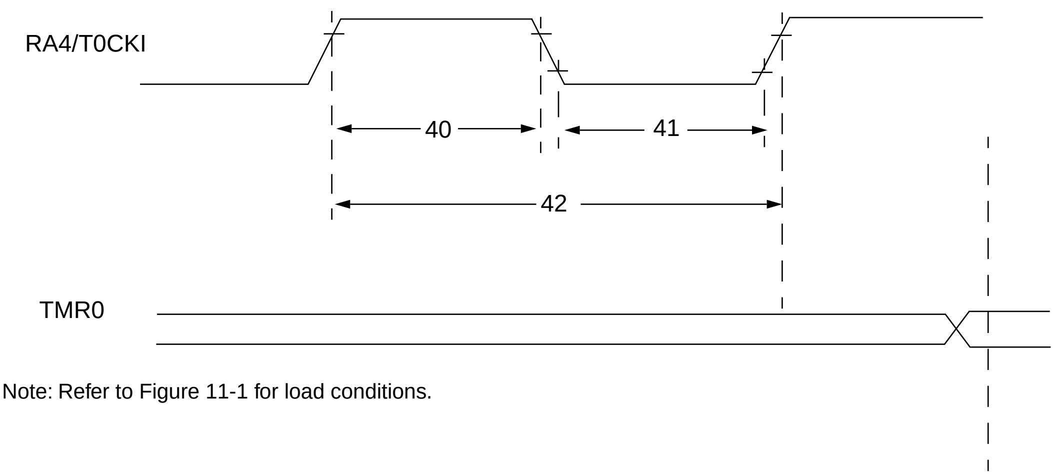
TRIS — Load TRIS Register

Syntax:	[*label*] TRIS f
Operands:	5 ≤ f ≤ 7
Operation:	(W) → TRIS register f;
Status Affected:	None
Encoding:	`00 0000 0110 0fff`
Description:	The instruction is supported for code compatibility with the PIC16C5X products. Since TRIS registers are readable and writable, the user can directly address them.
Words:	1
Cycles:	1
Example	

To maintain upward compatibility with future PIC16CXX products, do not use this instruction.

Applicable Devices	710	71	711	715

FIGURE 11-6: TIMER0 EXTERNAL CLOCK TIMINGS



Note: Refer to Figure 11-1 for load conditions.

TABLE 11-5: TIMER0 EXTERNAL CLOCK REQUIREMENTS

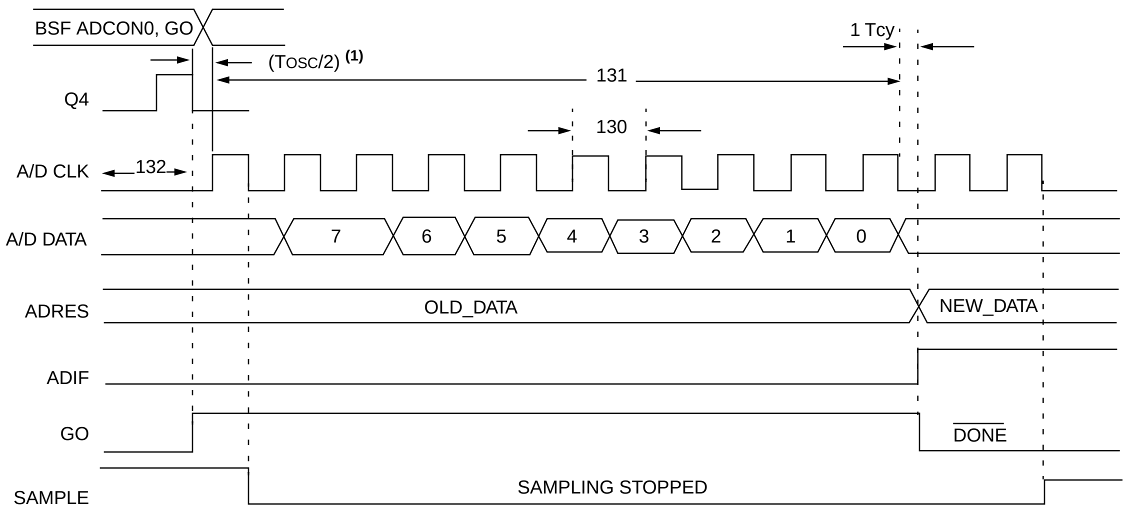
Param No.	Sym	Characteristic		Min	Typ†	Max	Units	Conditions
40	Tt0H	T0CKI High Pulse Width	No Prescaler	0.5TCY + 20*	—	—	ns	Must also meet parameter 42
			With Prescaler	10*	—	—	ns	
41	Tt0L	T0CKI Low Pulse Width	No Prescaler	0.5TCY + 20*	—	—	ns	Must also meet parameter 42
			With Prescaler	10*	—	—	ns	
42	Tt0P	T0CKI Period		Greater of: 20 ns or $\frac{TCY + 40}{N}$*	—	—	ns	N = prescale value (2, 4,..., 256)
48	Tcke2tmrI	Delay from external clock edge to timer increment		2Tosc	—	7Tosc	—	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25˚C unless otherwise stated. These parameters are for design guidance only and are not tested.

Applicable Devices	710	71	711	715

FIGURE 11-7: A/D CONVERSION TIMING



Note 1: If the A/D clock source is selected as RC, a time of TCY is added before the A/D clock starts. This allows the SLEEP instruction to be executed.

TABLE 11-7: A/D CONVERSION REQUIREMENTS

Param No.	Sym	Characteristic		Min	Typ†	Max	Units	Conditions
130	TAD	A/D clock period	PIC16**C**710/711	1.6	—	—	µs	TOSC based, VREF ≥ 3.0V
			PIC16**LC**710/711	2.0	—	—	µs	TOSC based, VREF full range
			PIC16**C**710/711	2.0*	4.0	6.0	µs	A/D RC mode
			PIC16**LC**710/711	3.0*	6.0	9.0	µs	A/D RC mode
131	TCNV	Conversion time (not including S/H time). (Note 1)		—	9.5	—	TAD	
132	TACQ	Acquisition time		Note 2	20	—	µs	
				5*	—	—	µs	The minimum time is the amplifier settling time. This may be used if the "new" input voltage has not changed by more than 1 LSb (i.e., 19.5 mV @ 5.12V) from the last sampled voltage (as stated on CHOLD).
134	TGO	Q4 to AD clock start		—	TOSC/2§	—	—	If the A/D clock source is selected as RC, a time of TCY is added before the A/D clock starts. This allows the SLEEP instruction to be executed.
135	TSWC	Switching from convert → sample time		1.5§	—	—	TAD	

* These parameters are characterized but not tested.
† Data in “Typ” column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.
§ This specification ensured by design.
Note 1: ADRES register may be read on the following TCY cycle.
2: See Section 7.1 for min conditions.

Applicable Devices	710	71	711	715

12.0 DC AND AC CHARACTERISTICS GRAPHS AND TABLES FOR PIC16C710 AND PIC16C711

The graphs and tables provided in this section are for design guidance and are not tested or guaranteed.

In some graphs or tables the data presented are outside specified operating range (i.e., outside specified VDD range). This is for information only and devices are guaranteed to operate properly only within the specified range.

Note:	The data presented in this section is a statistical summary of data collected on units from different lots over a period of time and matrix samples. 'Typical' represents the mean of the distribution at, 25°C, while 'max' or 'min' represents (mean +3σ) and (mean -3σ) respectively where σ is standard deviation.

FIGURE 12-1: TYPICAL IPD vs. VDD (WDT DISABLED, RC MODE)

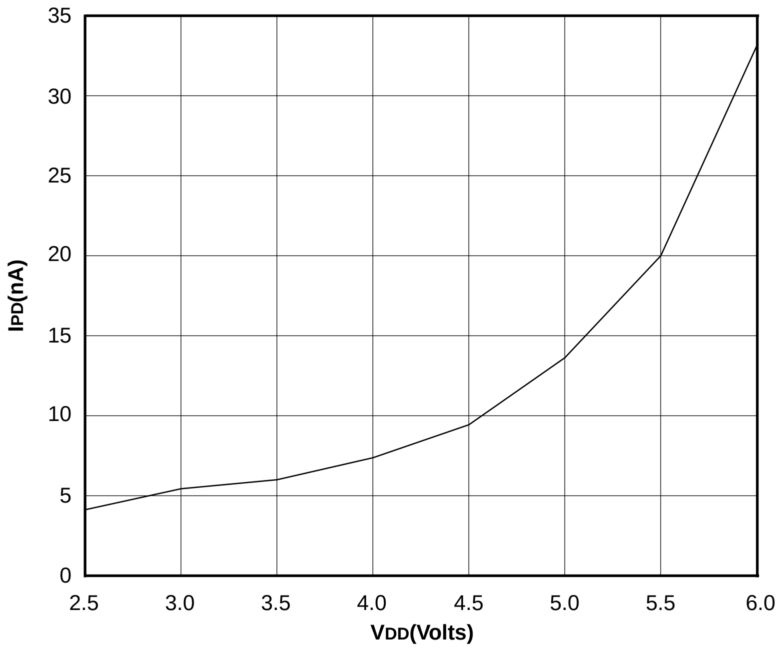
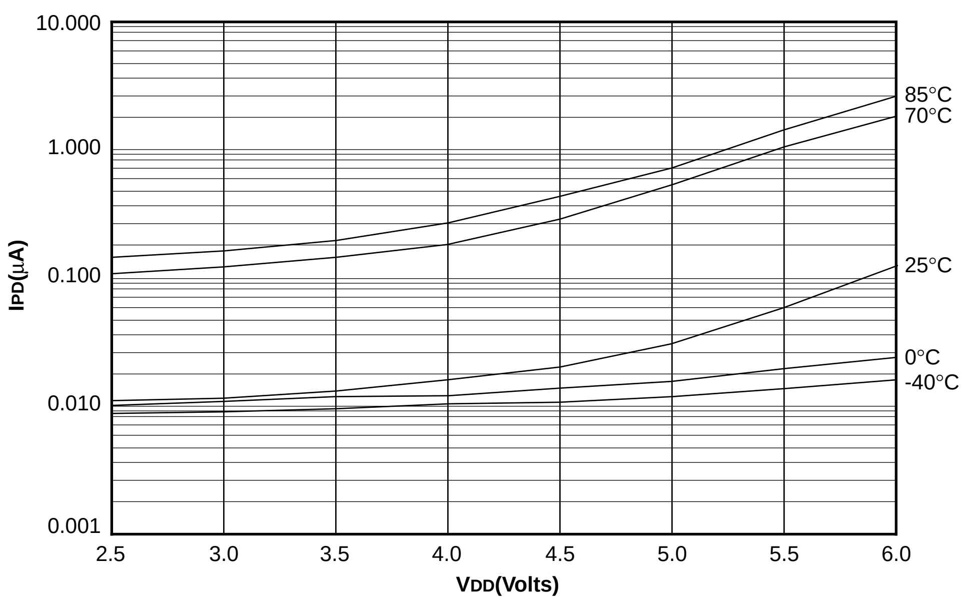


FIGURE 12-2: MAXIMUM IPD vs. VDD (WDT DISABLED, RC MODE)



Applicable Devices	710	71	711	715

13.1 DC Characteristics: PIC16C715-04 (Commercial, Industrial, Extended) PIC16C715-10 (Commercial, Industrial, Extended) PIC16C715-20 (Commercial, Industrial, Extended))

DC CHARACTERISTICS							**Standard Operating Conditions (unless otherwise stated)** Operating temperature 0˚C ≤ T_A ≤ +70˚C (commercial) -40˚C ≤ T_A ≤ +85˚C (industrial) -40˚C ≤ T_A ≤ +125˚C (extended)
Param. No.	**Characteristic**	**Sym**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
D001 D001A	Supply Voltage	V_{DD}	4.0 4.5	- -	5.5 5.5	V V	XT, RC and LP osc configuration HS osc configuration
D002*	RAM Data Retention Voltage (Note 1)	V_{DR}	-	1.5	-	V	Device in SLEEP mode
D003	V_{DD} start voltage to ensure internal Power-on Reset signal	V_{POR}	-	V_{SS}	-	V	See section on Power-on Reset for details
D004*	V_{DD} rise rate to ensure internal Power-on Reset signal	S_{VDD}	0.05	-	-	V/ms	See section on Power-on Reset for details
D005	Brown-out Reset Voltage	B_{VDD}	3.7	4.0	4.3	V	BODEN configuration bit is enabled
D010	Supply Current (Note 2)	I_{DD}	-	2.7	5	mA	XT, RC osc configuration (PIC16C715-04) F_{OSC} = 4 MHz, V_{DD} = 5.5V (Note 4)
D013			-	13.5	30	mA	HS osc configuration (PIC16C715-20) F_{OSC} = 20 MHz, V_{DD} = 5.5V
D015	Brown-out Reset Current (Note 5)	ΔI_{BOR}	-	300*	500	µA	BOR enabled V_{DD} = 5.0V
D020 D021 D021A D021B	Power-down Current (Note 3)	I_{PD}	- - - -	10.5 1.5 1.5 1.5	42 21 24 30	µA µA µA µA	V_{DD} = 4.0V, WDT enabled, -40°C to +85°C V_{DD} = 4.0V, WDT disabled, -0°C to +70°C V_{DD} = 4.0V, WDT disabled, -40°C to +85°C V_{DD} = 4.0V, WDT disabled, -40°C to +125°C
D023	Brown-out Reset Current (Note 5)	ΔI_{BOR}	-	300*	500	µA	BOR enabled V_{DD} = 5.0V

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25˚C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which V_{DD} can be lowered in SLEEP mode without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.
The test conditions for all I_{DD} measurements in active operation mode are:
OSC1 = external square wave, from rail to rail; all I/O pins tristated, pulled to V_{DD}
$\overline{MCLR}$ = V_{DD}; WDT enabled/disabled as specified.

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to V_{DD} and V_{SS}.

4: For RC osc configuration, current through Rext is not included. The current through the resistor can be estimated by the formula Ir = V_{DD}/2Rext (mA) with Rext in kOhm.

5: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base I_{DD} or I_{PD} measurement.

Applicable Devices	710	71	711	715

**TABLE 13-6: A/D CONVERTER CHARACTERISTICS:
PIC16C715-04 (COMMERCIAL, INDUSTRIAL, EXTENDED)
PIC16C715-10 (COMMERCIAL, INDUSTRIAL, EXTENDED)
PIC16C715-20 (COMMERCIAL, INDUSTRIAL, EXTENDED)**

Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
	NR	Resolution	—	—	8-bits	—	VREF = VDD, VSS ≤ AIN ≤ VREF
	NINT	Integral error	—	—	less than ±1 LSb	—	VREF = VDD, VSS ≤ AIN ≤ VREF
	NDIF	Differential error	—	—	less than ±1 LSb	—	VREF = VDD, VSS ≤ AIN ≤ VREF
	NFS	Full scale error	—	—	less than ±1 LSb	—	VREF = VDD, VSS ≤ AIN ≤ VREF
	NOFF	Offset error	—	—	less than ±1 LSb	—	VREF = VDD, VSS ≤ AIN ≤ VREF
	—	Monotonicity	—	guaranteed	—	—	VSS ≤ AIN ≤ VREF
	VREF	Reference voltage	2.5V	—	VDD + 0.3	V	
	VAIN	Analog input voltage	VSS - 0.3	—	VREF + 0.3	V	
	ZAIN	Recommended impedance of analog voltage source	—	—	10.0	kΩ	
	IAD	A/D conversion current (VDD)	—	180	—	μA	Average current consumption when A/D is on. (Note 1)
	IREF	VREF input current (Note 2)	—	—	1 10	mA μA	During sampling All other times

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: When A/D is off, it will not consume any current other than minor leakage current. The power-down current spec includes any such leakage from the A/D module.

2: VREF current is from RA3 pin or VDD pin, whichever is selected as reference input.

Applicable Devices	710	71	711	715

14.0 DC AND AC CHARACTERISTICS GRAPHS AND TABLES FOR PIC16C715

The graphs and tables provided in this section are for design guidance and are not tested or guaranteed.

In some graphs or tables the data presented are outside specified operating range (i.e., outside specified VDD range). This is for information only and devices are guaranteed to operate properly only within the specified range.

Note: The data presented in this section is a statistical summary of data collected on units from different lots over a period of time and matrix samples. 'Typical' represents the mean of the distribution at, 25°C, while 'max' or 'min' represents (mean +3σ) and (mean -3σ) respectively where σ is standard deviation.

FIGURE 14-1: TYPICAL IPD vs. VDD (WDT DISABLED, RC MODE)

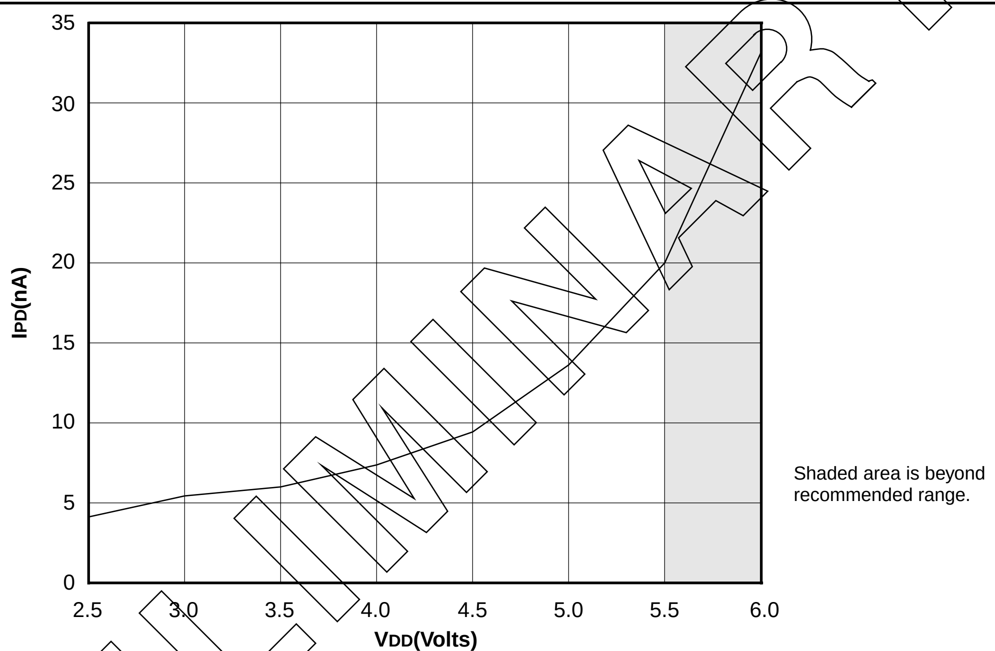
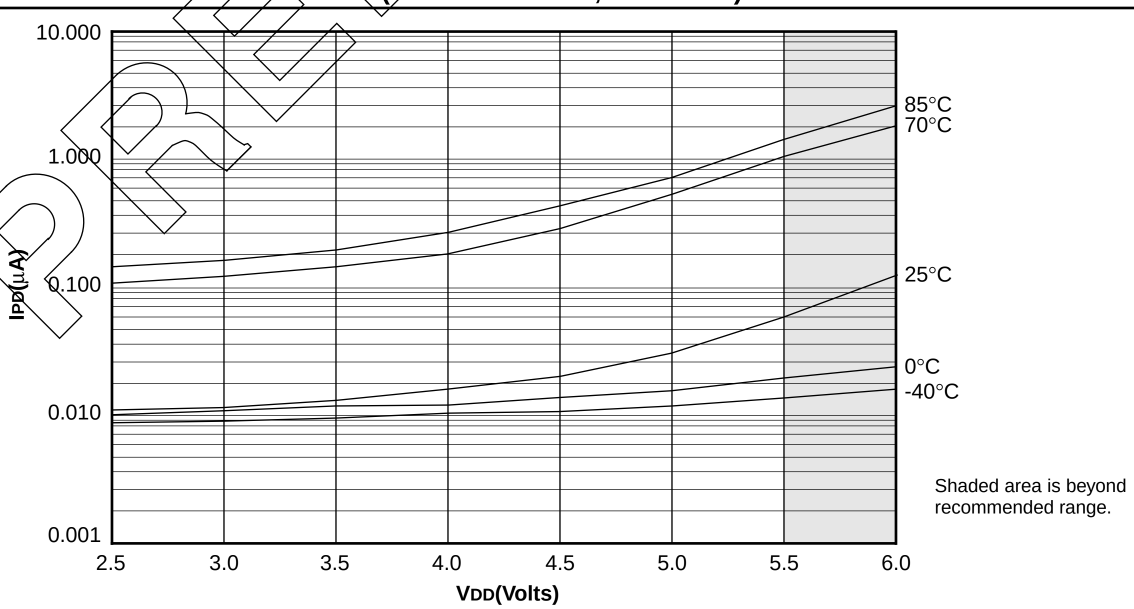


FIGURE 14-2: MAXIMUM IPD vs. VDD (WDT DISABLED, RC MODE)



Applicable Devices	710	71	711	715

15.5 Timing Diagrams and Specifications

FIGURE 15-2: EXTERNAL CLOCK TIMING

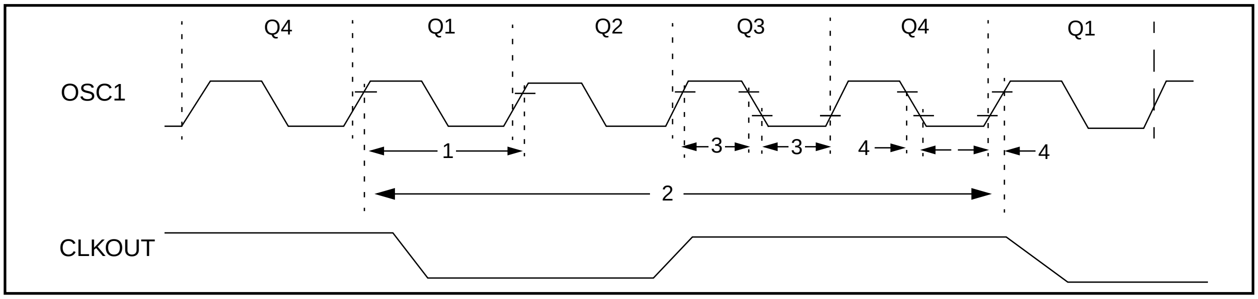


TABLE 15-2: EXTERNAL CLOCK TIMING REQUIREMENTS

Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
	Fosc	**External CLKIN Frequency (Note 1)**	DC	—	4	MHz	XT osc mode
			DC	—	4	MHz	HS osc mode (-04)
			DC	—	20	MHz	HS osc mode (-20)
			DC	—	200	kHz	LP osc mode
		Oscillator Frequency (Note 1)	DC	—	4	MHz	RC osc mode
			0.1	—	4	MHz	XT osc mode
			1	—	4	MHz	HS osc mode
			1	—	20	MHz	HS osc mode
1	Tosc	**External CLKIN Period (Note 1)**	250	—	—	ns	XT osc mode
			250	—	—	ns	HS osc mode (-04)
			50	—	—	ns	HS osc mode (-20)
			5	—	—	µs	LP osc mode
		Oscillator Period (Note 1)	250	—	—	ns	RC osc mode
			250	—	10,000	ns	XT osc mode
			250	—	1,000	ns	HS osc mode (-04)
			50	—	1,000	ns	HS osc mode (-20)
			5	—	—	µs	LP osc mode
2	TCY	**Instruction Cycle Time (Note 1)**	1.0	TCY	DC	µs	TCY = 4/Fosc
3	TosL, TosH	**External Clock in (OSC1) High or Low Time**	50	—	—	ns	XT oscillator
			2.5	—	—	µs	LP oscillator
			10	—	—	ns	HS oscillator
4	TosR, TosF	**External Clock in (OSC1) Rise or Fall Time**	25	—	—	ns	XT oscillator
			50	—	—	ns	LP oscillator
			15	—	—	ns	HS oscillator

† Data in "Typ" column is at 5V, 25˚C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (TCY) equals four times the input oscillator time-base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min." values with an external clock applied to the OSC1/CLKIN pin. When an external clock input is used, the "Max." cycle time limit is "DC" (no clock) for all devices. OSC2 is disconnected (has no loading) for the PIC16C71.

Applicable Devices	710	71	711	715

FIGURE 16-17: TRANSCONDUCTANCE (gm) OF LP OSCILLATOR VS. VDD

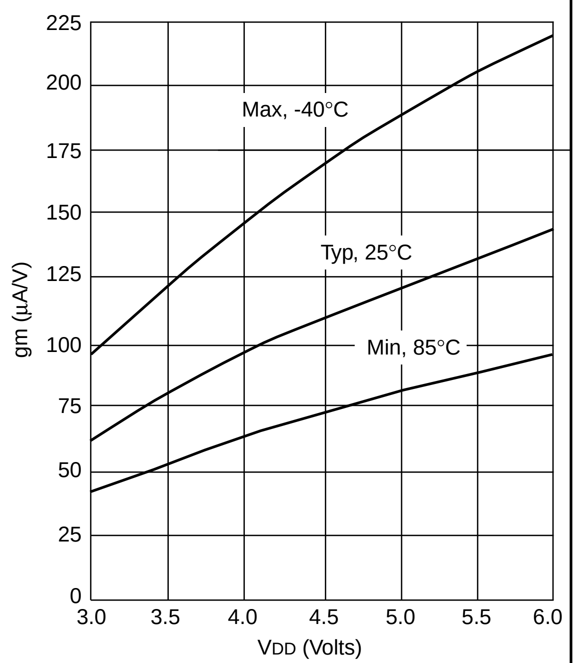


FIGURE 16-18: TRANSCONDUCTANCE (gm) OF XT OSCILLATOR VS. VDD

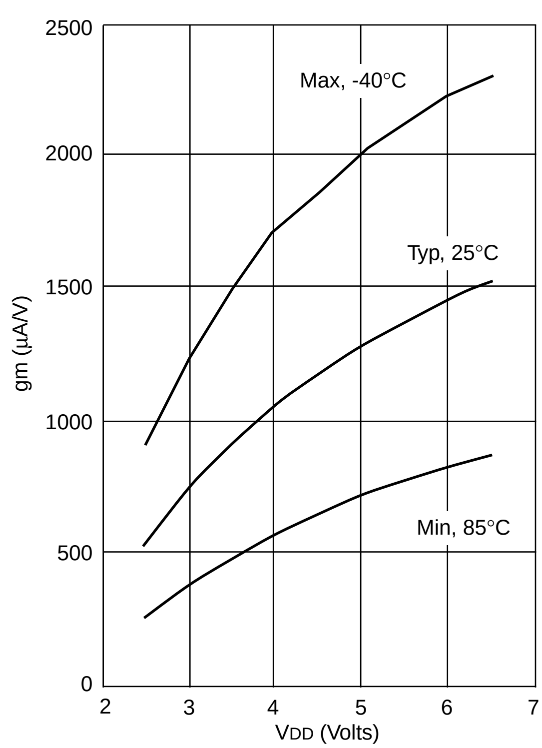


FIGURE 16-19: IOH VS. VOH, VDD = 3V

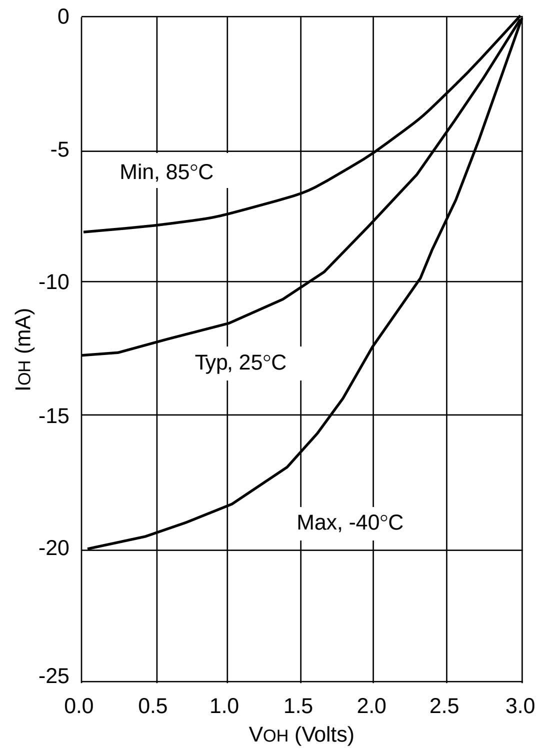
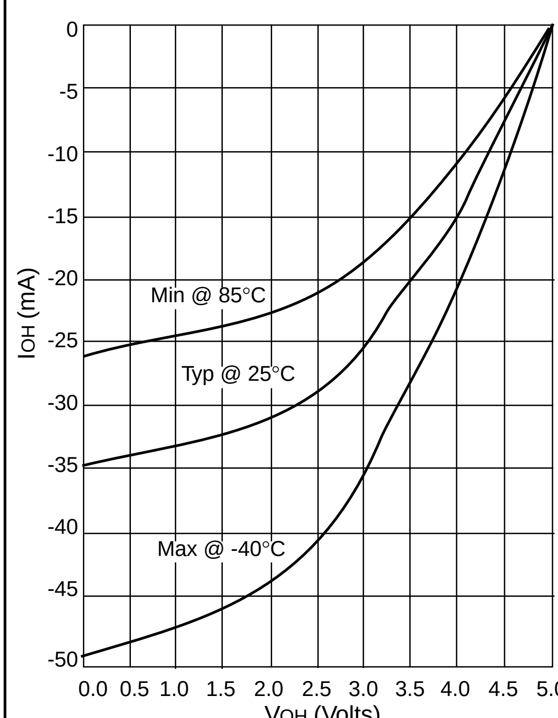


FIGURE 16-20: IOH VS. VOH, VDD = 5V



Data based on matrix samples. See first page of this section for details.

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Note the following details of the code protection feature on PICmicro® MCUs.

- The PICmicro family meets the specifications contained in the Microchip Data Sheet.
- Microchip believes that its family of PICmicro microcontrollers is one of the most secure products of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the PICmicro microcontroller in a manner outside the operating specifications contained in the data sheet. The person doing so may be engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as “unbreakable”.
- Code protection is constantly evolving. We at Microchip are committed to continuously improving the code protection features of our product.

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