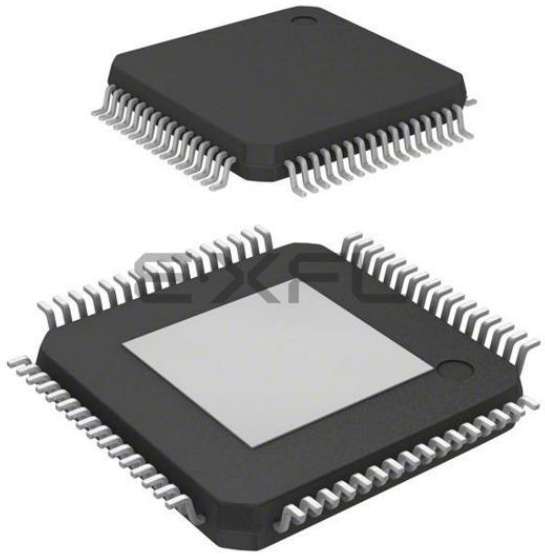

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Program Memory Type	FLASH
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RAM Size	8K x 8
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Data Converters	A/D 16x10b; D/A 1x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LQFP Exposed Pad
Supplier Device Package	64-LQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/s912zvc12f0mkh

Global Address	Register Name		Bit 7	6	5	4	3	2	1	Bit 0
0x028A–0x028B	Reserved	R	0	0	0	0	0	0	0	0
		W								
0x028C	PIEADH	R	PIEADH7	PIEADH6	PIEADH5	PIEADH4	PIEADH3	PIEADH2	PIEADH1	PIEADH0
		W								
0x028D	PIEADL	R	PIEADL7	PIEADL6	PIEADL5	PIEADL4	PIEADL3	PIEADL2	PIEADL1	PIEADL0
		W								
0x028E	PIFADH	R	PIFADH7	PIFADH6	PIFADH5	PIFADH4	PIFADH3	PIFADH2	PIFADH1	PIFADH0
		W								
0x028F	PIFADL	R	PIFADL7	PIFADL6	PIFADL5	PIFADL4	PIFADL3	PIFADL2	PIFADL1	PIFADL0
		W								
0x0290–0x0297	Reserved	R	0	0	0	0	0	0	0	0
		W								
0x0298	DIENADH	R	DIENADH7	DIENADH6	DIENADH5	DIENADH4	DIENADH3	DIENADH2	DIENADH1	DIENADH0
		W								
0x0299	DIENADL	R	DIENADL7	DIENADL6	DIENADL5	DIENADL4	DIENADL3	DIENADL2	DIENADL1	DIENADL0
		W								
0x029A–0x02BF	Reserved	R	0	0	0	0	0	0	0	0
		W								
0x02C0	PTT	R	PTT7	PTT6	PTT5	PTT4	PTT3	PTT2	PTT1	PTT0
		W								
0x02C1	PTIT	R	PTIT7	PTIT6	PTIT5	PTIT4	PTIT3	PTIT2	PTIT1	PTIT0
		W								
0x02C2	DDRT	R	DDRT7	DDRT6	DDRT5	DDRT4	DDRT3	DDRT2	DDRT1	DDRT0
		W								
0x02C3	PERT	R	PERT7	PERT6	PERT5	PERT4	PERT3	PERT2	PERT1	PERT0
		W								
0x02C4	PPST	R	PPST7	PPST6	PPST5	PPST4	PPST3	PPST2	PPST1	PPST0
		W								
0x02C5–0x02CE	Reserved	R	0	0	0	0	0	0	0	0
		W								
0x02CF	Reserved	R	0	0	0	0	0	0	0	0
		W								

2.3.4.4 Port L Input Register (PTIL)

Address 0x0331 Access: User read only[1]

	7	6	5	4	3	2	1	0
R	0	0	0	0	0	0	PTIL1	PTIL0
W								
Reset	0	0	0	0	0	0	0	0

Figure 2-24. Port L Input Register (PTIL)

[1] Read: Anytime
Write: No Write

Table 2-24. PTIL - Register Field Descriptions

Field	Description
1-0 PTIL1-0	**Port Input Data Register Port L —** A read returns the synchronized input state if the associated HVI pin is used in digital mode, that is the related DIENL bit is set to 1 and the pin is not used in analog mode (PTAENL=0). See Section 2.3.4.10, “Port L ADC Connection Enable Register (PTAENL)”. A one is read in any other case[1].

[1] Refer to PTTEL bit description in Section 2.3.4.12, “Port L Test Enable Register (PTTEL) for an override condition.

2.3.4.5 Port L Pull Select Register (PTPSL)

Address 0x0333 Access: User read/write[1]

	7	6	5	4	3	2	1	0
R	0	0	0	0	0	0	PTPSL1	PTPSL0
W								
Reset	0	0	0	0	0	0	0	0

Figure 2-25. Port L Pull Select Register (PTPSL)

[1] Read: Anytime
Write: Anytime

Table 2-25. PTPSL Register Field Descriptions

Field	Description
1-0 PTPSL1-0	**Port L Pull Select —** This bit selects a pull device on the HVI pin in analog mode for open input detection. By default a pulldown device is active as part of the input voltage divider. If this bit set to 1 and PTTEL=1 and not in stop mode a pullup to a level close to V_{DDX} takes effect and overrides the weak pulldown device. Refer to Section 2.5.5, “Open Input Detection on PL[1:0] (HVI)”). 1 Pullup enabled 0 Pulldown enabled

The BDC serial interface uses a clocking scheme in which the external host generates a falling edge on the BKGD pin to indicate the start of each bit time. This falling edge is sent for every bit whether data is transmitted or received. Data is transferred most significant bit (MSB) first at 16 target clock cycles per bit. The interface times out if during a command 512 clock cycles occur between falling edges from the host. The timeout forces the current command to be discarded.

The BKGD pin is a pseudo open-drain pin and has a weak on-chip active pull-up that is enabled at all times. It is assumed that there is an external pull-up and that drivers connected to BKGD do not typically drive the high level. Since R-C rise time could be unacceptably long, the target system and host provide brief drive-high (speedup) pulses to drive BKGD to a logic 1. The source of this speedup pulse is the host for transmit cases and the target for receive cases.

The timing for host-to-target is shown in Figure 3-6 and that of target-to-host in Figure 3-7 and Figure 3-8. All cases begin when the host drives the BKGD pin low to generate a falling edge. Since the host and target operate from separate clocks, it can take the target up to one full clock cycle to recognize this edge; this synchronization uncertainty is illustrated in Figure 3-6. The target measures delays from this perceived start of the bit time while the host measures delays from the point it actually drove BKGD low to start the bit up to one target clock cycle earlier. Synchronization between the host and target is established in this manner at the start of every bit time.

Figure 3-6 shows an external host transmitting a logic 1 and transmitting a logic 0 to the BKGD pin of a target system. The host is asynchronous to the target, so there is up to a one clock-cycle delay from the host-generated falling edge to where the target recognizes this edge as the beginning of the bit time. Ten target clock cycles later, the target senses the bit level on the BKGD pin. Internal glitch detect logic requires the pin be driven high no later than eight target clock cycles after the falling edge for a logic 1 transmission.

Since the host drives the high speedup pulses in these two cases, the rising edges look like digitally driven signals.

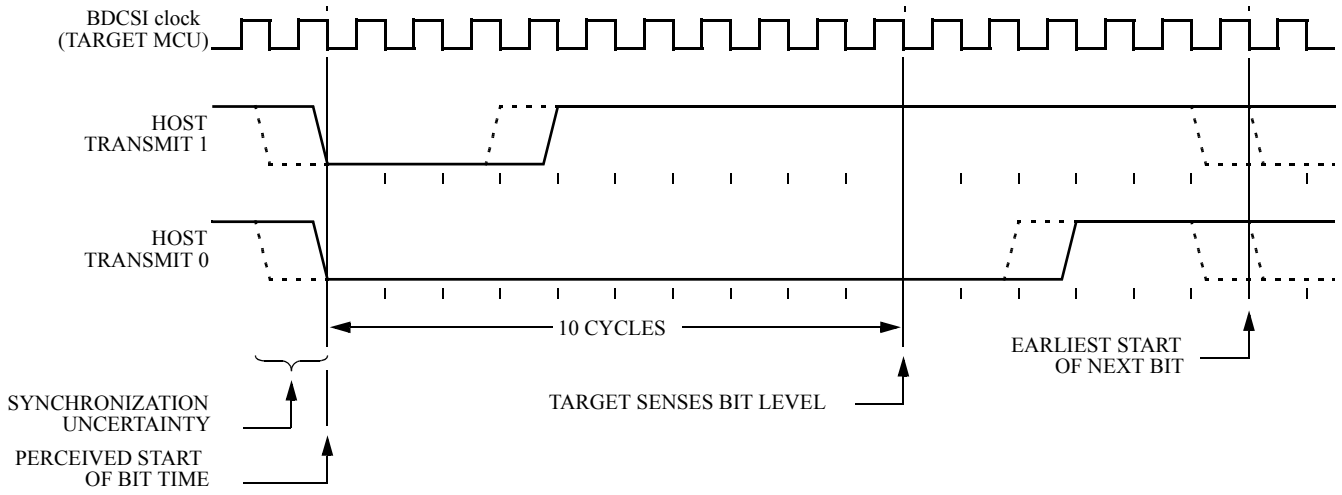

Figure 3-6. BDC Host-to-Target Serial Bit Timing

Figure 3-7 shows the host receiving a logic 1 from the target system. The host holds the BKGD pin low long enough for the target to recognize it (at least two target clock cycles). The host must release the low

Address	Register Name		Bit 7	6	5	4	3	2	1	Bit 0
0x00001D	INT_CFDATA5	R	0	0	0	0	0	PRIOLVL[2:0]		
		W								
0x00001E	INT_CFDATA6	R	0	0	0	0	0	PRIOLVL[2:0]		
		W								
0x00001F	INT_CFDATA7	R	0	0	0	0	0	PRIOLVL[2:0]		
		W								

= Unimplemented or Reserved

Figure 5-2. INT Register Summary

5.3.2.1 Interrupt Vector Base Register (IVBR)

Address: 0x000010

	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
R	IVB_ADDR[15:1]															0
W																
Reset	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	0

Figure 5-3. Interrupt Vector Base Register (IVBR)

Read: Anytime

Write: Anytime

Table 5-4. IVBR Field Descriptions

Field	Description
15–1 IVB_ADDR [15:1]	**Interrupt Vector Base Address Bits** — These bits represent the upper 15 bits of all vector addresses. Out of reset these bits are set to 0xFFFE (i.e., vectors are located at 0xFFFE00–0xFFFFFF). **Note:** A system reset will initialize the interrupt vector base register with “0xFFFE” before it is used to determine the reset vector address. Therefore, changing the IVBR has no effect on the location of the reset vector (0xFFFFFC–0xFFFFFF).

Address	Name		Bit 7	6	5	4	3	2	1	Bit 0
0x0111-0x0114	Reserved	R	0	0	0	0	0	0	0	0
		W								
0x0115	DBGAAH	R W	DBGAA[23:16]							
0x0116	DBGAAM	R W	DBGAA[15:8]							
0x0117	DBGAAL	R W	DBGAA[7:0]							
0x0118	DBGAD0	R W	Bit 31	30	29	28	27	26	25	Bit 24
0x0119	DBGAD1	R W	Bit 23	22	21	20	19	18	17	Bit 16
0x011A	DBGAD2	R W	Bit 15	14	13	12	11	10	9	Bit 8
0x011B	DBGAD3	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x011C	DBGADM0	R W	Bit 31	30	29	28	27	26	25	Bit 24
0x011D	DBGADM1	R W	Bit 23	22	21	20	19	18	17	Bit 16
0x011E	DBGADM2	R W	Bit 15	14	13	12	11	10	9	Bit 8
0x011F	DBGADM3	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x0120	DBGBCTL	R W	0	0	INST	0	RW	RWE	reserved	COMPE
0x0121-0x0124	Reserved	R W	0	0	0	0	0	0	0	0
0x0125	DBGBAH	R W	DBGBA[23:16]							
0x0126	DBGBAM	R W	DBGBA[15:8]							
0x0127	DBGBAL	R W	DBGBA[7:0]							
0x0128-0x012F	Reserved	R W	0	0	0	0	0	0	0	0
0x0130-0x013F	Reserved	R W	0	0	0	0	0	0	0	0

Figure 6-2. Quick Reference to DBG Registers

6.3.2.5 Debug State Control Register 3 (DBGSCR3)

Address: 0x0109

	7	6	5	4	3	2	1	0
R	C3SC1	C3SC0	0	0	C1SC1	C1SC0	C0SC1	C0SC0
W								
Reset	0	0	0	0	0	0	0	0

Figure 6-8. Debug State Control Register 3 (DBGSCR3)

Read: Anytime.

Write: If DBG is not armed.

The state control register three selects the targeted next state whilst in State3. The matches refer to the outputs of the comparator match control logic as depicted in Figure 6-1 and described in Section 6.3.2.8, “Debug Comparator A Control Register (DBGACTL)”. Comparators must be enabled by setting the comparator enable bit in the associated DBGxCTL control register.

Table 6-11. DBGSCR3 Field Descriptions

Field	Description
1–0 C0SC[1:0]	Channel 0 State Control. These bits select the targeted next state whilst in State3 following a match0.
3–2 C1SC[1:0]	Channel 1 State Control. These bits select the targeted next state whilst in State3 following a match1.
7–6 C3SC[1:0]	Channel 3 State Control. If EEVE !=10, these bits select the targeted next state whilst in State3 following a match3. If EEVE =10, these bits select the targeted next state whilst in State3 following an external event.

Table 6-12. State3 Match State Sequencer Transitions

CxSC[1:0]	Function
00	Match has no effect
01	Match forces sequencer to State1
10	Match forces sequencer to State2
11	Match forces sequencer to Final State

In the case of simultaneous matches, the match on the higher channel number (3....0) has priority.

6.3.2.6 Debug Event Flag Register (DBGEFR)

Address: 0x010A

	7	6	5	4	3	2	1	0
R	0	TRIGF	0	EEVF	ME3	0	ME1	ME0
W								
Reset	0	0	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 6-9. Debug Event Flag Register (DBGEFR)

This supply domain is monitored by the Low Voltage Reset circuit.

8.2.6 .VDDC, VSSC — CANPHY Supply Pin

VDDC is the supply domain for the CANPHY.

An off-chip decoupling capacitor (10µF plus 220 nF(X7R ceramic)) between VDDC and VSSC is required.

This supply domain is monitored by the Low Voltage Reset circuit.

8.2.7 BCTL — Base Control Pin for external PNP

BCTL is the ballast connection for the on chip voltage regulator. It provides the base current of an external

BJT (PNP) of the VDDX and VDDA supplies. An additional 1KΩ resistor between emitter and base of the BJT is required.

8.2.8 BCTLC — Base Control Pin for external PNP for VDDC

BCTLC is the ballast connection for the on chip voltage regulator. It provides the base current of an external BJT (PNP) of the VDDC supply. An additional 1KΩ resistor between emitter and base of the BJT is required.

8.2.9 VSS — Core Logic Ground Pin

VSS is the core logic supply return pins. It must be grounded.

8.2.10 VDD — Internal Regulator Output Supply (Core Logic)

Node VDD is a device internal supply output of the voltage regulator that provides the power supply for the internal core logic.

This supply domain is monitored by the Low Voltage Reset circuit and The Power On Reset circuit.

8.2.11 VDDF — Internal Regulator Output Supply (NVM Logic)

Node VDDF is a device internal supply output of the voltage regulator that provides the power supply for the NVM logic.

This supply domain is monitored by the Low Voltage Reset circuit.

8.2.12 API_EXTCLK — API external clock output pin

This pin provides the signal selected via APIES and is enabled with APIEA bit. See the device specification if this clock output is available on this device and to which pin it might be connected.

Table 9-9. ADCFLWCTL Field Descriptions

Field	Description
7 SEQA	**Conversion Sequence Abort Event** — This bit indicates that a conversion sequence abort event is in progress. When this bit is set the ongoing conversion sequence and current CSL will be aborted at the next conversion boundary. This bit gets cleared when the ongoing conversion sequence is aborted and ADC is idle. This bit can only be set if bit ADC_EN is set. This bit is cleared if bit ADC_EN is clear. *Data Bus Control:* This bit can be controlled via the data bus if access control is configured accordingly via ACC_CFG[1:0]. Writing a value of 1'b0 does not clear the flag. Writing a one to this bit does not clear it but causes an overrun if the bit has already been set. See Section 9.5.3.2.6, "Conversion flow control in case of conversion sequence control bit overrun scenarios for more details. *Internal Interface Control:* This bit can be controlled via the internal interface Signal "Seq_Abort" if access control is configured accordingly via ACC_CFG[1:0]. After being set an additional request via the internal interface Signal "Seq_Abort" causes an overrun. See also conversion flow control in case of overrun situations. *General:* In both conversion flow control modes (Restart Mode and Trigger Mode) when bit RSTA gets set automatically bit SEQA gets set when the ADC has not reached one of the following scenarios: - A Sequence Abort request is about to be executed or has been executed. - "End Of List" command type has been executed or is about to be executed In case bit SEQA is set automatically the Restart error flag RSTA_EIF is set to indicate an unexpected Restart Request. 0 No conversion sequence abort request. 1 Conversion sequence abort request.
6 TRIG	**Conversion Sequence Trigger Bit** — This bit starts a conversion sequence if set and no conversion or conversion sequence is ongoing. This bit is cleared when the first conversion of a sequence starts to sample. This bit can only be set if bit ADC_EN is set. This bit is cleared if bit ADC_EN is clear. *Data Bus Control:* This bit can be controlled via the data bus if access control is configured accordingly via ACC_CFG[1:0]. Writing a value of 1'b0 does not clear the flag. After being set this bit can not be cleared by writing a value of 1'b1 instead the error flag TRIG_EIF is set. See also Section 9.5.3.2.6, "Conversion flow control in case of conversion sequence control bit overrun scenarios for more details. *Internal Interface Control:* This bit can be controlled via the internal interface Signal "Trigger" if access control is configured accordingly via ACC_CFG[1:0]. After being set an additional request via internal interface Signal "Trigger" causes the flag TRIG_EIF to be set. 0 No conversion sequence trigger. 1 Trigger to start conversion sequence.

In freeze mode there is a software programmable option to disable the input clock to the prescaler. This is useful for emulation.

Wait: The prescaler keeps on running, unless PSWAI in PWMCTL is set to 1.

Freeze: The prescaler keeps on running, unless PFRZ in PWMCTL is set to 1.

13.1.3 Block Diagram

Figure 13-1 shows the block diagram for the 8-bit up to 8-channel scalable PWM block.

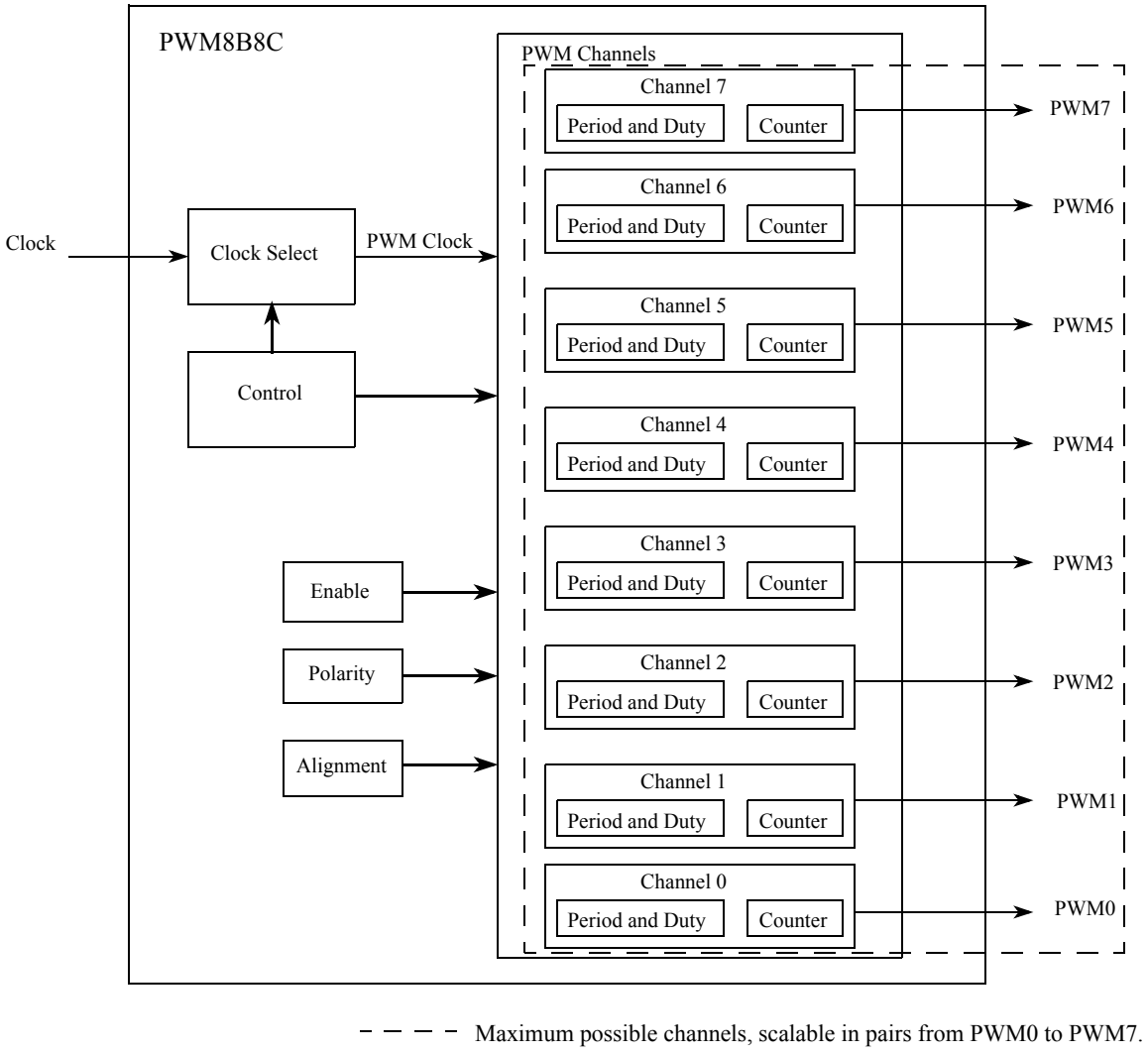

Figure 13-1. Scalable PWM Block Diagram

13.2 External Signal Description

The scalable PWM module has a selected number of external pins. Refer to device specification for exact number.

When the transmit shift register is not transmitting a frame, the TXD pin goes to the idle condition, logic 1. If at any time software clears the TE bit in SCI control register 2 (SCICR2), the transmitter enable signal goes low and the transmit signal goes idle.

If software clears TE while a transmission is in progress (TC = 0), the frame in the transmit shift register continues to shift out. To avoid accidentally cutting off the last frame in a message, always wait for TDRE to go high after the last frame before clearing TE.

To separate messages with preambles with minimum idle line time, use this sequence between messages:

1. Write the last byte of the first message to SCIDRH/L.
2. Wait for the TDRE flag to go high, indicating the transfer of the last frame to the transmit shift register.
3. Queue a preamble by clearing and then setting the TE bit.
4. Write the first byte of the second message to SCIDRH/L.

14.4.5.3 Break Characters

Writing a logic 1 to the send break bit, SBK, in SCI control register 2 (SCICR2) loads the transmit shift register with a break character. A break character contains all logic 0s and has no start, stop, or parity bit. Break character length depends on the M bit in SCI control register 1 (SCICR1). As long as SBK is at logic 1, transmitter logic continuously loads break characters into the transmit shift register. After software clears the SBK bit, the shift register finishes transmitting the last break character and then transmits at least one logic 1. The automatic logic 1 at the end of a break character guarantees the recognition of the start bit of the next frame.

The SCI recognizes a break character when there are 10 or 11(M = 0 or M = 1) consecutive zero received. Depending if the break detect feature is enabled or not receiving a break character has these effects on SCI registers.

If the break detect feature is disabled (BKDFE = 0):

- Sets the framing error flag, FE
- Sets the receive data register full flag, RDRF
- Clears the SCI data registers (SCIDRH/L)
- May set the overrun flag, OR, noise flag, NF, parity error flag, PE, or the receiver active flag, RAF (see 3.4.4 and 3.4.5 SCI Status Register 1 and 2)

If the break detect feature is enabled (BKDFE = 1) there are two scenarios[1]

The break is detected right from a start bit or is detected during a byte reception.

- Sets the break detect interrupt flag, BKDIF
- Does not change the data register full flag, RDRF or overrun flag OR
- Does not change the framing error flag FE, parity error flag PE.
- Does not clear the SCI data registers (SCIDRH/L)
- May set noise flag NF, or receiver active flag RAF.

1. A Break character in this context are either 10 or 11 consecutive zero received bits

Figure 14-17 shows two cases of break detect. In trace RXD_1 the break symbol starts with the start bit, while in RXD_2 the break starts in the middle of a transmission. If BRKDFE = 1, in RXD_1 case there will be no byte transferred to the receive buffer and the RDRF flag will not be modified. Also no framing error or parity error will be flagged from this transfer. In RXD_2 case, however the break signal starts later during the transmission. At the expected stop bit position the byte received so far will be transferred to the receive buffer, the receive data register full flag will be set, a framing error and if enabled and appropriate a parity error will be set. Once the break is detected the BRKDIF flag will be set.

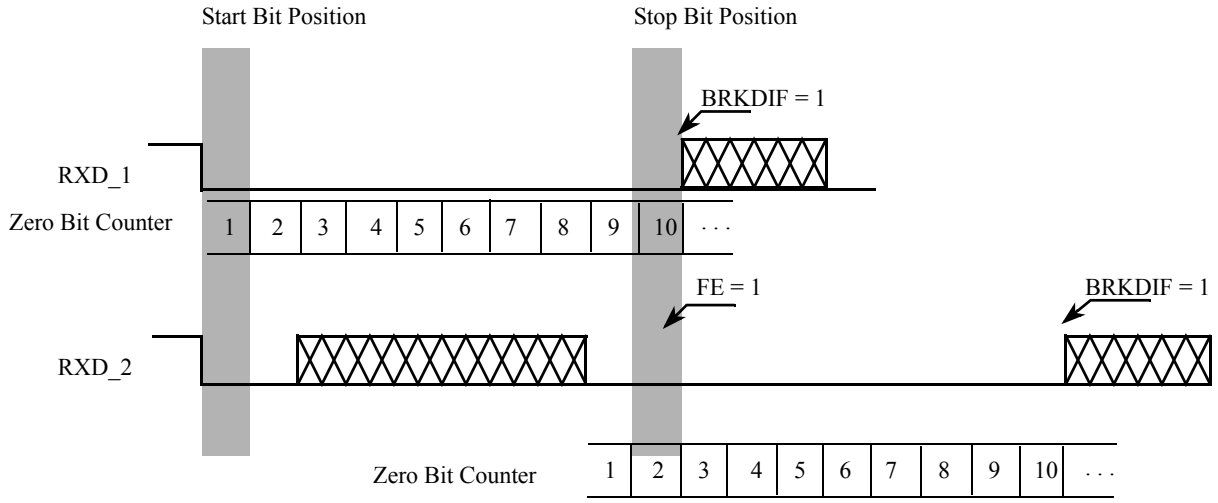

Figure 14-17. Break Detection if BRKDFE = 1 (M = 0)

14.4.5.4 Idle Characters

An idle character (or preamble) contains all logic 1s and has no start, stop, or parity bit. Idle character length depends on the M bit in SCI control register 1 (SCICR1). The preamble is a synchronizing idle character that begins the first transmission initiated after writing the TE bit from 0 to 1.

If the TE bit is cleared during a transmission, the TXD pin becomes idle after completion of the transmission in progress. Clearing and then setting the TE bit during a transmission queues an idle character to be sent after the frame currently being transmitted.

NOTE

When queueing an idle character, return the TE bit to logic 1 before the stop bit of the current frame shifts out through the TXD pin. Setting TE after the stop bit appears on TXD causes data previously written to the SCI data register to be lost. Toggle the TE bit for a queued idle character while the TDRE flag is set and immediately before writing the next byte to the SCI data register.

If the TE bit is clear and the transmission is complete, the SCI is not the master of the TXD pin

The transmitting device can address messages to selected receivers by including addressing information in the initial frame or frames of each message.

The WAKE bit in SCI control register 1 (SCICR1) determines how the SCI is brought out of the standby state to process an incoming message. The WAKE bit enables either idle line wakeup or address mark wakeup.

14.4.6.6.1 Idle Input line Wakeup (WAKE = 0)

In this wakeup method, an idle condition on the RXD pin clears the RWU bit and wakes up the SCI. The initial frame or frames of every message contain addressing information. All receivers evaluate the addressing information, and receivers for which the message is addressed process the frames that follow. Any receiver for which a message is not addressed can set its RWU bit and return to the standby state. The RWU bit remains set and the receiver remains on standby until another idle character appears on the RXD pin.

Idle line wakeup requires that messages be separated by at least one idle character and that no message contains idle characters.

The idle character that wakes a receiver does not set the receiver idle bit, IDLE, or the receive data register full flag, RDRF.

The idle line type bit, ILT, determines whether the receiver begins counting logic 1s as idle character bits after the start bit or after the stop bit. ILT is in SCI control register 1 (SCICR1).

14.4.6.6.2 Address Mark Wakeup (WAKE = 1)

In this wakeup method, a logic 1 in the most significant bit (MSB) position of a frame clears the RWU bit and wakes up the SCI. The logic 1 in the MSB position marks a frame as an address frame that contains addressing information. All receivers evaluate the addressing information, and the receivers for which the message is addressed process the frames that follow.Any receiver for which a message is not addressed can set its RWU bit and return to the standby state. The RWU bit remains set and the receiver remains on standby until another address frame appears on the RXD pin.

The logic 1 MSB of an address frame clears the receiver’s RWU bit before the stop bit is received and sets the RDRF flag.

Address mark wakeup allows messages to contain idle characters but requires that the MSB be reserved for use in address frames.

NOTE

With the WAKE bit clear, setting the RWU bit after the RXD pin has been idle can cause the receiver to wake up immediately.

14.4.7 Single-Wire Operation

Normally, the SCI uses two pins for transmitting and receiving. In single-wire operation, the RXD pin is disconnected from the SCI. The SCI uses the TXD pin for both receiving and transmitting.

Table 16-8. IBCR Field Descriptions

Field	Description
7 IBEN	**I-Bus Enable** — This bit controls the software reset of the entire IIC bus module. 0 The module is reset and disabled. This is the power-on reset situation. When low the interface is held in reset but registers can be accessed 1 The IIC bus module is enabled.This bit must be set before any other IBCR bits have any effect If the IIC bus module is enabled in the middle of a byte transfer the interface behaves as follows: slave mode ignores the current transfer on the bus and starts operating whenever a subsequent start condition is detected. Master mode will not be aware that the bus is busy, hence if a start cycle is initiated then the current bus cycle may become corrupt. This would ultimately result in either the current bus master or the IIC bus module losing arbitration, after which bus operation would return to normal.
6 IBIE	**I-Bus Interrupt Enable** 0 Interrupts from the IIC bus module are disabled. Note that this does not clear any currently pending interrupt condition 1 Interrupts from the IIC bus module are enabled. An IIC bus interrupt occurs provided the IBIF bit in the status register is also set.
5 MS/SL	**Master/Slave Mode Select Bit** — Upon reset, this bit is cleared. When this bit is changed from 0 to 1, a START signal is generated on the bus, and the master mode is selected. When this bit is changed from 1 to 0, a STOP signal is generated and the operation mode changes from master to slave.A STOP signal should only be generated if the IBIF flag is set. MS/$\overline{SL}$ is cleared without generating a STOP signal when the master loses arbitration. 0 Slave Mode 1 Master Mode
4 Tx/Rx	**Transmit/Receive Mode Select Bit** — This bit selects the direction of master and slave transfers. When addressed as a slave this bit should be set by software according to the SRW bit in the status register. In master mode this bit should be set according to the type of transfer required. Therefore, for address cycles, this bit will always be high. 0 Receive 1 Transmit
3 TXAK	**Transmit Acknowledge Enable** — This bit specifies the value driven onto SDA during data acknowledge cycles for both master and slave receivers. The IIC module will always acknowledge address matches, provided it is enabled, regardless of the value of TXAK. Note that values written to this bit are only used when the IIC bus is a receiver, not a transmitter. 0 An acknowledge signal will be sent out to the bus at the 9th clock bit after receiving one byte data 1 No acknowledge signal response is sent (i.e., acknowledge bit = 1)
2 RSTA	**Repeat Start** — Writing a 1 to this bit will generate a repeated START condition on the bus, provided it is the current bus master. This bit will always be read as a low. Attempting a repeated start at the wrong time, if the bus is owned by another master, will result in loss of arbitration. 1 Generate repeat start cycle
1 RESERVED	**Reserved** — Bit 1 of the IBCR is reserved for future compatibility. This bit will always read 0.
0 IBSWAI	**I Bus Interface Stop in Wait Mode** 0 IIC bus module clock operates normally 1 Halt IIC bus module clock generation in wait mode

Wait mode is entered via execution of a CPU WAI instruction. In the event that the IBSWAI bit is set, all clocks internal to the IIC will be stopped and any transmission currently in progress will halt.If the CPU were woken up by a source other than the IIC module, then clocks would restart and the IIC would resume from where was during the previous transmission. It is not possible for the IIC to wake up the CPU when its internal clocks are stopped.

If it were the case that the IBSWAI bit was cleared when the WAI instruction was executed, the IIC internal clocks and interface would remain alive, continuing the operation which was currently underway. It is also

Table 18-25. Message Buffer Organization

Offset Address	Register	Access
0x00X0	IDR0 — Identifier Register 0	R/W
0x00X1	IDR1 — Identifier Register 1	R/W
0x00X2	IDR2 — Identifier Register 2	R/W
0x00X3	IDR3 — Identifier Register 3	R/W
0x00X4	DSR0 — Data Segment Register 0	R/W
0x00X5	DSR1 — Data Segment Register 1	R/W
0x00X6	DSR2 — Data Segment Register 2	R/W
0x00X7	DSR3 — Data Segment Register 3	R/W
0x00X8	DSR4 — Data Segment Register 4	R/W
0x00X9	DSR5 — Data Segment Register 5	R/W
0x00XA	DSR6 — Data Segment Register 6	R/W
0x00XB	DSR7 — Data Segment Register 7	R/W
0x00XC	DLR — Data Length Register	R/W
0x00XD	TBPR — Transmit Buffer Priority Register[1]	R/W
0x00XE	TSRH — Time Stamp Register (High Byte)	R
0x00XF	TSRL — Time Stamp Register (Low Byte)	R

[1] Not applicable for receive buffers

Figure 18-24 shows the common 13-byte data structure of receive and transmit buffers for extended identifiers. The mapping of standard identifiers into the IDR registers is shown in Figure 18-25.

All bits of the receive and transmit buffers are ‘x’ out of reset because of RAM-based implementation[1]. All reserved or unused bits of the receive and transmit buffers always read ‘x’.

1. Exception: The transmit buffer priority registers are 0 out of reset.

19.4.2.1 Control Register (DACCTL)

Module Base + 0x0000 Access: User read/write[1]

	7	6	5	4	3	2	1	0
R	FVR	DRIVE	0	0	0	DACM[2:0]		
W								
Reset	1	0	0	0	0	0	0	0

= Unimplemented

Figure 19-3. Control Register (DACCTL)

[1] Read: Anytime
Write: Anytime

Table 19-3. DACCTL Field Description

Field	Description
7 FVR	**Full Voltage Range** — This bit defines the voltage range of the DAC. 0 DAC resistor network operates with the reduced voltage range 1 DAC resistor network operates with the full voltage range **Note:** For more details see Section 19.5.8, “Analog output voltage calculation”.
6 DRIVE	**Drive Select** — This bit selects the output drive capability of the operational amplifier, see electrical Spec. for more details. 0 Low output drive for high resistive loads 1 High output drive for low resistive loads
2:0 DACM[2:0]	**Mode Select** — These bits define the mode of the DAC. A write access with an unsupported mode will be ignored. 000 Off 001 Operational Amplifier 010 Internal DAC only 100 Unbuffered DAC 101 Unbuffered DAC with Operational Amplifier 111 Buffered DAC other Reserved

Table 22-58. Set Field Margin Level Command FCCOB Requirements

Register	FCCOB Parameters	
FCCOB0	0x0E	Global address [23:16] to identify Flash block
FCCOB1	Global address [15:0] to identify Flash block	
FCCOB2	Margin level setting.	

Upon clearing CCIF to launch the Set Field Margin Level command, the Memory Controller will set the field margin level for the targeted block and then set the CCIF flag.

NOTE

When the EEPROM block is targeted, the EEPROM field margin levels are applied only to the EEPROM reads. However, when the P-Flash block is targeted, the P-Flash field margin levels are applied to both P-Flash and EEPROM reads. It is not possible to apply field margin levels to the P-Flash block only.

Valid margin level settings for the Set Field Margin Level command are defined in Table 22-59.

Table 22-59. Valid Set Field Margin Level Settings

FCCOB2	Level Description
0x0000	Return to Normal Level
0x0001	User Margin-1 Level[1]
0x0002	User Margin-0 Level[2]
0x0003	Field Margin-1 Level[1]
0x0004	Field Margin-0 Level[2]

1 Read margin to the erased state

2 Read margin to the programmed state

Table 22-60. Set Field Margin Level Command Error Handling

Register	Error Bit	Error Condition
FSTAT	ACCERR	Set if CCOBIX[2:0] != 010 at command launch
		Set if command not available in current mode (see Table 22-28)
		Set if an invalid global address [23:0] is supplied see Table 22-2)
		Set if an invalid margin level setting is supplied
	FPVIOL	None
	MGSTAT1	None
	MGSTAT0	None

Table 22-65. Erase EEPROM Sector Command FCCOB Requirements

Register	FCCOB Parameters	
FCCOB0	0x12	Global address [23:16] to identify EEPROM block
FCCOB1	Global address [15:0] anywhere within the sector to be erased. See <st-blue>Section 22.1.2.2 EEPROM Features for EEPROM sector size.	

Upon clearing CCIF to launch the Erase EEPROM Sector command, the Memory Controller will erase the selected Flash sector and verify that it is erased. The CCIF flag will set after the Erase EEPROM Sector operation has completed.

Table 22-66. Erase EEPROM Sector Command Error Handling

Register	Error Bit	Error Condition
FSTAT	ACCERR	Set if CCOBIX[2:0] != 001 at command launch
		Set if command not available in current mode (see Table 22-28)
		Set if an invalid global address [23:0] is suppliedsee Table 22-2
		Set if a misaligned word address is supplied (global address [0] != 0)
	FPVIOL	Set if the selected area of the EEPROM memory is protected
	MGSTAT1	Set if any errors have been encountered during the verify operation
	MGSTAT0	Set if any non-correctable errors have been encountered during the verify operation

22.4.7.17 Protection Override Command

The Protection Override command allows the user to temporarily override the protection limits, either decreasing, increasing or disabling protection limits, on P-Flash and/or EEPROM, if the comparison key provided as a parameter loaded on FCCOB matches the value of the key previously programmed on the Flash Configuration Field (see Table 22-3.). The value of the Protection Override Comparison Key must not be 16'hFFFF, that is considered invalid and if used as argument will cause the Protection Override feature to be disabled. Any valid key value that does not match the value programmed in the Flash Configuration Field will cause the Protection Override feature to be disabled. Current status of the Protection Override feature can be observed on FPSTAT FPOVRD bit (see Section 22.3.2.4, "Flash Protection Status Register (FPSTAT)).

Table 22-67. Protection Override Command FCCOB Requirements

Register	FCCOB Parameters	
FCCOB0	0x13	Protection Update Selection [1:0] See Table 22-68.
FCCOB1	Comparison Key	
FCCOB2	reserved	New FPROT value
FCCOB3	reserved	New DFPROT value

I.3 Dynamic Electrical Characteristics

Table I-3. Dynamic Electrical Characteristics

Characteristics noted under conditions 5.5V <= VSUP <=18 V >, -40°C <= Tj <= 150°C > unless otherwise noted. Typical values noted reflect the approximate parameter mean at T_A = 25°C under nominal conditions unless otherwise noted.						
Num	**Ratings**	**Symbol**	**Min**	**Typ**	**Max**	**Unit**
	SIGNAL EDGE RISE AND FALL TIMES (CANH, CANL)					
1	Propagation Loop Delay TXD to RXD (Recessive to Dominant) Slew Rate 6 Slew Rate 5 Slew Rate 4 Slew Rate 2 Slew Rate 1 Slew Rate 0	t_{LRD}		 146 112 89 83 72 64	 (255)	ns
2	Propagation Delay TXD to CAN (Recessive to Dominant) Slew Rate 6 Slew Rate 5 Slew Rate 4 Slew Rate 2 Slew Rate 1 Slew Rate 0	t_{TRD}		 98 63 43 38 28 23		ns
3	Propagation Delay CAN to RXD (Recessive to Dominant, using slew rate 0)	t_{RRD}		42		ns
4	Propagation Loop Delay TXD to RXD (Dominant to Recessive) Slew Rate 6 Slew Rate 5 Slew Rate 4 Slew Rate 2 Slew Rate 1 Slew Rate 0	t_{LDR}		 366 224 153 139 114 102	 (255)	ns
5	Propagation Delay TXD to CAN (Dominant to Recessive) Slew Rate 6 Slew Rate 5 Slew Rate 4 Slew Rate 2 Slew Rate 1 Slew Rate 0	t_{TDR}		 280 152 90 81 56 46		ns
6	Propagation Delay CAN to RXD (Dominant to Recessive, using slew rate 0)	t_{RDR}		56		ns

Table K-3. NVM Reliability Characteristics (Junction Temperature From –40°C To +175°C)

NUM	C	Rating	Symbol	Min	Typ	Max	Unit
		Program Flash Arrays					
1	C	Data retention at an average junction temperature of T_{Javg} = 85°C[1] after up to 10,000 program/erase cycles	t_{NVMRET}	20	100[2]	—	Years
2	C	Program Flash number of program/erase cycles (-40°C ≤ Tj ≤ 150°C)	n_{FLPE}	10K	100K[3]	—	Cycles
		EEPROM Array					
3	C	Data retention at an average junction temperature of T_{Javg} = 85°C[1] after up to 100,000 program/erase cycles	t_{NVMRET}	5	100[2]	—	Years
4	C	Data retention at an average junction temperature of T_{Javg} = 85°C[1] after up to 10,000 program/erase cycles	t_{NVMRET}	10	100[2]	—	Years
5	C	Data retention at an average junction temperature of T_{Javg} = 85°C[1] after less than 100 program/erase cycles	t_{NVMRET}	20	100[2]	—	Years
6	C	EEPROM number of program/erase cycles (-40°C ≤ Tj ≤ 150°C)	n_{FLPE}	100K	500K[3]	—	Cycles

1 T_{Javg} does not exceed 85°C in a typical temperature profile over the lifetime of a consumer, industrial or automotive application.

2 Typical data retention values are based on intrinsic capability of the technology measured at high temperature and de-rated to 25°C using the Arrhenius equation. For additional information on how NXP defines Typical Data Retention, please refer to Engineering Bulletin EB618

3 Spec table quotes typical endurance evaluated at 25°C for this product family. For additional information on how NXP defines Typical Endurance, please refer to Engineering Bulletin EB619.

K.3 NVM Factory Shipping Condition

Devices are shipped from the factory with flash and EEPROM in the erased state. Data retention specifications begin at time of this erase operation. For additional information on how NXP defines Typical Data Retention, please refer to Engineering Bulletin EB618.