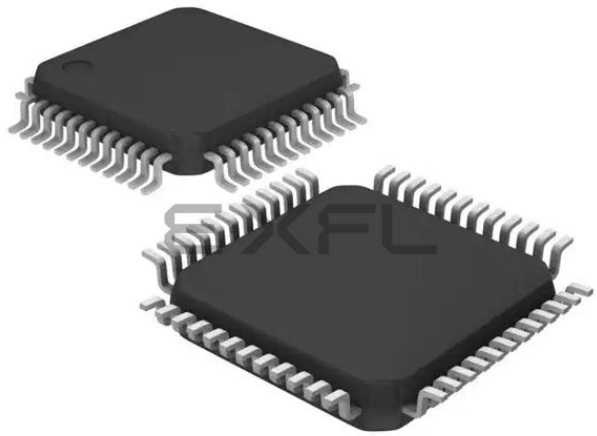

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	S12Z
Core Size	16-Bit
Speed	32MHz
Connectivity	CANbus, I²C, SCI, SPI
Peripherals	DMA, POR, PWM, WDT
Number of I/O	28
Program Memory Size	64KB (64K x 8)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	4K x 8
Voltage - Supply (Vcc/Vdd)	3.5V ~ 40V
Data Converters	A/D 10x10b; D/A 1x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	48-LQFP
Supplier Device Package	48-LQFP (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/s912zvc64f0vlfr

3.1.1 Glossary

Table 3-2. Glossary Of Terms

Term	Definition
DBG	On chip Debug Module
BDM	Active Background Debug Mode
CPU	S12Z CPU
SSC	Special Single Chip Mode (device operating mode
NSC	Normal Single Chip Mode (device operating mode)
BDCSI	Background Debug Controller Serial Interface. This refers to the single pin BKGD serial interface.
EWAIT	Optional S12 feature which allows external devices to delay external accesses until deassertion of EWAIT

3.1.2 Features

The BDC includes these distinctive features:

- Single-wire communication with host development system
- SYNC command to determine communication rate
- Genuine non-intrusive handshake protocol
- Enhanced handshake protocol for error detection and stop mode recognition
- Active out of reset in special single chip mode
- Most commands not requiring active BDM, for minimal CPU intervention
- Full global memory map access without paging
- Simple flash mass erase capability

3.1.3 Modes of Operation

S12 devices feature power modes (run, wait, and stop) and operating modes (normal single chip, special single chip). Furthermore, the operation of the BDC is dependent on the device security status.

3.1.3.1 BDC Modes

The BDC features module specific modes, namely disabled, enabled and active. These modes are dependent on the device security and operating mode. In active BDM the CPU ceases execution, to allow BDC system access to all internal resources including CPU internal registers.

3.1.3.2 Security and Operating mode Dependency

In device run mode the BDC dependency is as follows

- Normal modes, unsecure device
  General BDC operation available. The BDC is disabled out of reset.

3.4.3 Clock Source

The BDC clock source can be mapped to a constant frequency clock source or a PLL based fast clock. The clock source for the BDC is selected by the CLKSW bit as shown in Figure 3-5. The BDC internal clock is named BDCSI clock. If BDCSI clock is mapped to the BDCCLK by CLKSW then the serial interface communication is not affected by bus/core clock frequency changes. If the BDC is mapped to BDCFCLK then the clock is connected to a PLL derived source at device level (typically bus clock), thus can be subject to frequency changes in application. Debugging through frequency changes requires SYNC pulses to re-synchronize. The sources of BDCCLK and BDCFCLK are specified at device level.

BDC accesses of internal device resources always use the device core clock. Thus if the ACK handshake protocol is not enabled, the clock frequency relationship must be taken into account by the host.

When changing the clock source via the CLKSW bit a minimum delay of 150 cycles at the initial clock speed must elapse before a SYNC can be sent. This guarantees that the start of the next BDC command uses the new clock for timing subsequent BDC communications.

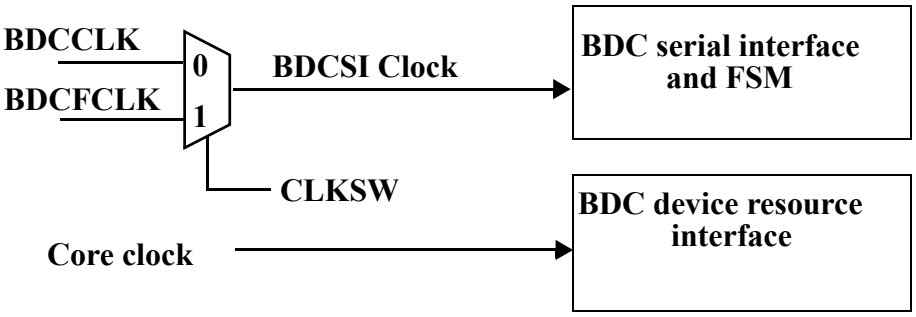


Figure 3-5. Clock Switch

3.4.4 BDC Commands

BDC commands can be classified into three types as shown in Table 3-7.

Table 3-7. BDC Command Types

Command Type	Secure Status	BDC Status	CPU Status	Command Set
Always-available	Secure or Unsecure	Enabled or Disabled	—	• Read/write access to BDCCSR • Mass erase flash memory using ERASE_FLASH • SYNC • ACK enable/disable

6.4.3.1 Comparator Match Events

6.4.3.1.1 Opcode Address Comparator Match

The comparator is loaded with the address of the selected instruction and the comparator control register INST bit is set. When the opcode reaches the execution stage of the instruction queue a match occurs just before the instruction executes, allowing a breakpoint immediately before the instruction boundary. The comparator address register must contain the address of the first opcode byte for the match to occur. Opcode address matches are data independent thus the RWE and RW bits are ignored. CPU compares are disabled when BDM becomes active.

6.4.3.1.2 Data Access Comparator Match

Data access matches are generated when an access occurs at the address contained in the comparator address register. The match can be qualified by the access data and by the access type (read/write). The breakpoint occurs a maximum of 2 instructions after the access in the CPU flow. Note, if a COF occurs between access and breakpoint, the opcode address of the breakpoint can be elsewhere in the memory map.

Opcode fetches are not classed as data accesses. Thus data access matches are not possible on opcode fetches.

6.4.3.2 External Event

The DBGEEV input signal can force a state sequencer transition, independent of internal comparator matches. The DBGEEV is an input signal mapped directly to a device pin and configured by the EEVE field in DBGC1. The external events can change the state sequencer state.

If configured to change the state sequencer state, then the external match is mapped to DBGSCRx bits C3SC[1:0]. The DBGEFR bit EEVF is set when an external event occurs.

6.4.3.3 Setting The TRIG Bit

Independent of comparator matches it is possible to initiate a breakpoint by writing the TRIG bit in DBGC1 to a logic “1”. This forces the state sequencer into the Final State. the transition to Final State is followed immediately by a transition to State0.

Breakpoints, if enabled, are issued on the transition to State0.

6.4.3.4 Event Priorities

If simultaneous events occur, the priority is resolved according to Table 6-31. Lower priority events are suppressed. It is thus possible to miss a lower priority event if it occurs simultaneously with an event of a higher priority. The event priorities dictate that in the case of simultaneous matches, the match on the higher comparator channel number (3,1,0) has priority.

If a write access to DBGC1 with the ARM bit position set occurs simultaneously to a hardware disarm from an internal event, then the ARM bit is cleared due to the hardware disarm.

Several examples of PLL divider settings are shown in Table 8-32. The following rules help to achieve optimum stability and shortest lock time:

- Use lowest possible f_{VCO} / f_{REF} ratio (SYNDIV value).
- Use highest possible REFCLK frequency f_{REF}.

Table 8-32. Examples of PLL Divider Settings

f_{osc}	REFDIV[3:0]	f_{REF}	REFFRQ[1:0]	SYNDIV[5:0]	f_{VCO}	VCOFRQ[1:0]	POSTDIV[4:0]	f_{PLL}	f_{bus}
off	$00	1MHz	00	$18	50MHz	01	$03	12.5MHz	6.25MHz
off	$00	1MHz	00	$18	50MHz	01	$00	50MHz	25MHz
4MHz	$00	4MHz	01	$05	48MHz	00	$00	48MHz	24MHz

The phase detector inside the PLL compares the feedback clock (FBCLK = VCOCLK/(SYNDIV+1)) with the reference clock (REFCLK = (IRC1M or OSCCLK)/(REFDIV+1)). Correction pulses are generated based on the phase difference between the two signals. The loop filter alters the DC voltage on the internal filter capacitor, based on the width and direction of the correction pulse which leads to a higher or lower VCO frequency.

The user must select the range of the REFCLK frequency (REFFRQ[1:0] bits) and the range of the VCOCLK frequency (VCOFRQ[1:0] bits) to ensure that the correct PLL loop bandwidth is set.

The lock detector compares the frequencies of the FBCLK and the REFCLK. Therefore the speed of the lock detector is directly proportional to the reference clock frequency. The circuit determines the lock condition based on this comparison. So e.g. a failure in the reference clock will cause the PLL not to lock.

If PLL LOCK interrupt requests are enabled, the software can wait for an interrupt request and for instance check the LOCK bit. If interrupt requests are disabled, software can poll the LOCK bit continuously (during PLL start-up) or at periodic intervals. In either case, only when the LOCK bit is set, the VCOCLK will have stabilized to the programmed frequency.

- The LOCK bit is a read-only indicator of the locked state of the PLL.
- The LOCK bit is set when the VCO frequency is within the tolerance, Δ_{Lock}, and is cleared when the VCO frequency is out of the tolerance, Δ_{unl}.
- Interrupt requests can occur if enabled (LOCKIE = 1) when the lock condition changes, toggling the LOCK bit.

In case of loss of reference clock (e.g. IRCCLK) the PLL will not lock or if already locked, then it will unlock. The frequency of the VCOCLK will be very low and will depend on the value of the VCOFRQ[1:0] bits.

9.2.1 Modes of Operation

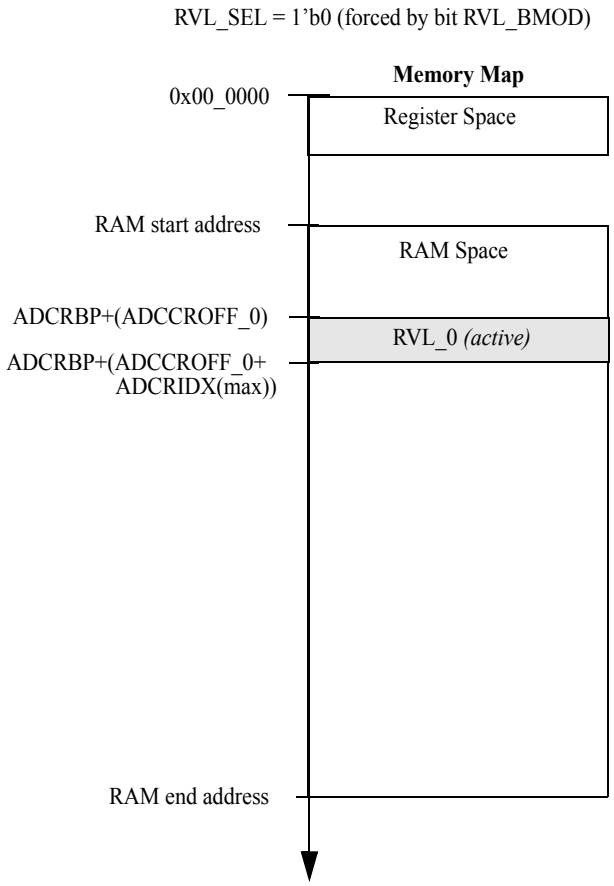
9.2.1.1 Conversion Modes

This architecture provides **single**, **multiple,** or **continuous conversion** on a **single channel** or on **multiple channels based on the Command Sequence List**.

9.2.1.2 MCU Operating Modes

- **MCU Stop Mode**
  Before issuing an MCU Stop Mode request the ADC should be idle (no conversion or conversion sequence or Command Sequence List ongoing).
  If a conversion, conversion sequence, or CSL is in progress when an MCU Stop Mode request is issued, a Sequence Abort Event occurs automatically and any ongoing conversion finish. After the Sequence Abort Event finishes, if the STR_SEQA bit is set (STR_SEQA=1), then the conversion result is stored and the corresponding flags are set. If the STR_SEQA bit is cleared (STR_SEQA=0), then the conversion result is not stored and the corresponding flags are not set. The microcontroller then enters MCU Stop Mode without SEQAD_IF being set.
  Alternatively, the Sequence Abort Event can be issued by software before an MCU Stop Mode request. As soon as flag SEQAD_IF is set the MCU Stop Mode request can be is issued.
  With the occurrence of the MCU Stop Mode Request until exit from Stop Mode all flow control signals (RSTA, SEQA, LDOK, TRIG) are cleared.

  After exiting MCU Stop Mode, the following happens in the order given with expected event(s) depending on the conversion flow control mode:
  — In ADC conversion flow control mode “Trigger Mode” a Restart Event is expected to simultaneously set bits TRIG and RSTA, causing the ADC to execute the Restart Event (CMD_IDX and RVL_IDX cleared) followed by the Trigger Event. The Restart Event can be generated automatically after exit from MCU Stop Mode if bit AUT_RSTA is set.
  — In ADC conversion flow control mode “Restart Mode”, a Restart Event is expected to set bit RSTA only (ADC already aborted at MCU Stop Mode entry hence bit SEQA must not be set simultaneously) causing the ADC to execute the Restart Event (CDM_IDX and RVL_IDX cleared). The Restart Event can be generated automatically after exit from MCU Stop Mode if bit AUT_RSTA is set.
  — The RVL buffer select (RVL_SEL) is not changed if a CSL is in process at MCU Stop Mode request. Hence the same buffer will be used after exit from Stop Mode that was used when the Stop Mode request occurred.



Note: Address register names in () are not absolute addresses instead they are a sample offset or sample index

Figure 9-34. Result Value List Schema in Single Buffer Mode

While ADC is enabled, one Result Value List is active (indicated by bit RVL_SEL). The conversion Result Value List can be read anytime. When the ADC is enabled the conversion result address registers (ADCRBP, ADCCROFF_0/1, ADCRIDX) are read only and register ADCRIDX is under control of the ADC.

A conversion result is always stored as 16bit entity in unsigned data representation. Left and right justification inside the entity is selected via the DJM control bit. Unused bits inside an entity are stored zero.

Table 9-32. Conversion Result Justification Overview

Conversion Resolution (SRES[1:0])	Left Justified Result (DJM = 1'b0)	Right Justified Result (DJM = 1'b1)
8 bit	{Result[7:0],8'b00000000}	{8'b00000000,Result[7:0]}
10 bit	{Result[9:0],6'b000000}	{6'b000000,Result[9:0]}
12 bit	{Result[11:0],4'b0000}	{4'b0000,Result[11:0]}

9.6 Resets

At reset the ADC12B_LBA is disabled and in a power down state. The reset state of each individual bit is listed within Section 9.4.2, “Register Descriptions” which details the registers and their bit-fields.

9.7 Interrupts

The ADC supports three types of interrupts:

- Conversion Interrupt
- Sequence Abort Interrupt
- Error and Conversion Flow Control Issue Interrupt

Each of the interrupt types is associated with individual interrupt enable bits and interrupt flags.

9.7.1 ADC Conversion Interrupt

The ADC provides one conversion interrupt associated to 16 interrupt enable bits with dedicated interrupt flags. The 16 interrupt flags consist of:

- 15 conversion interrupt flags which can be associated to any conversion completion.
- One additional interrupt flag which is fixed to the “End Of List” conversion command type within the active CSL.

The association of the conversion number with the interrupt flag number is done in the conversion command.

9.7.2 ADC Sequence Abort Done Interrupt

The ADC provides one sequence abort done interrupt associated with the sequence abort request for conversion flow control. Hence, there is only one dedicated interrupt flag and interrupt enable bit for conversion sequence abort and it occurs when the sequence abort is done.

Table 10-5. BATIF Register Field Descriptions

Field	Description
1 BVHIF	**BATS Interrupt Flag High Detect** — The flag is set to 1 when BVHC status bit changes. 0 No change of the BVHC status bit since the last clearing of the flag. 1 BVHC status bit has changed since the last clearing of the flag.
0 BVLIF	**BATS Interrupt Flag Low Detect** — The flag is set to 1 when BVLC status bit changes. 0 No change of the BVLC status bit since the last clearing of the flag. 1 BVLC status bit has changed since the last clearing of the flag.

10.3.2.5 Reserved Register

Module Base + 0x0006
Module Base + 0x0007

Access: User read/write[1]

	7	6	5	4	3	2	1	0
R W	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved
Reset	x	x	x	x	x	x	x	x

Figure 10-8. Reserved Register

[1] Read: Anytime
Write: Only in special mode

NOTE

These reserved registers are designed for factory test purposes only and are not intended for general user access. Writing to these registers when in special mode can alter the module's functionality.

10.4 Functional Description

10.4.1 General

The BATS module allows measuring the voltage on the VSUP pin. The voltage at the VSUP pin can be routed via an internal voltage divider to an internal Analog to Digital Converter Channel. Also the BATS module can be configured to generate a low and high voltage interrupt based on VSUP. The trigger level of the high and low interrupt are selectable.

10.4.2 Interrupts

This section describes the interrupt generated by the BATS module. The interrupt is only available in CPU run mode. Entering and exiting CPU stop mode has no effect on the interrupt flags.

To make sure the interrupt generation works properly the bus clock frequency must be higher than the Voltage Warning Low Pass Filter frequency (f_{VWLP_filter}).

Table 11-19. Pin Action

PAMOD	PEDGE	Pin Action
0	0	Falling edge
0	1	Rising edge
1	0	Div. by 64 clock enabled with pin high level
1	1	Div. by 64 clock enabled with pin low level

NOTE

If the timer is not active (TEN = 0 in TSCR), there is no divide-by-64 because the ÷64 clock is generated by the timer prescaler.

Table 11-20. Timer Clock Selection

CLK1	CLK0	Timer Clock
0	0	Use timer prescaler clock as timer counter clock
0	1	Use PACLK as input to timer counter clock
1	0	Use PACLK/256 as timer counter clock frequency
1	1	Use PACLK/65536 as timer counter clock frequency

For the description of PACLK please refer Figure 11-30**.**

If the pulse accumulator is disabled (PAEN = 0), the prescaler clock from the timer is always used as an input clock to the timer counter. The change from one selected clock to the other happens immediately after these bits are written.

11.3.2.16 Pulse Accumulator Flag Register (PAFLG)

[1] .

Module Base + 0x0021

	7	6	5	4	3	2	1	0
R	0	0	0	0	0	0	PAOVF	PAIF
W								
Reset	0	0	0	0	0	0	0	0

Unimplemented or Reserved

Figure 11-25. Pulse Accumulator Flag Register (PAFLG)

Read: Anytime

Write: Anytime

When the TFFCA bit in the TSCR register is set, any access to the PACNT register will clear all the flags in the PAFLG register. Timer module or Pulse Accumulator must stay enabled (TEN=1 or PAEN=1) while clearing these bits.

- The counter is written (counter resets to $00)
- The channel is disabled

In this way, the output of the PWM will always be either the old waveform or the new waveform, not some variation in between. If the channel is not enabled, then writes to the period register will go directly to the latches as well as the buffer.

NOTE

Reads of this register return the most recent value written. Reads do not necessarily return the value of the currently active period due to the double buffering scheme.

See Section 13.4.2.3, “PWM Period and Duty” for more information.

To calculate the output period, take the selected clock source period for the channel of interest (A, B, SA, or SB) and multiply it by the value in the period register for that channel:

- Left aligned output (CAEx = 0)

  PWMx Period = Channel Clock Period * PWMPERx
- Center Aligned Output (CAEx = 1)

  PWMx Period = Channel Clock Period * (2 * PWMPERx)

For boundary case programming values, please refer to Section 13.4.2.8, “PWM Boundary Cases”.

Module Base + 0x0014 = PWMPER0, 0x0015 = PWMPER1, 0x0016 = PWMPER2, 0x0017 = PWMPER3
Module Base + 0x0018 = PWMPER4, 0x0019 = PWMPER5, 0x001A = PWMPER6, 0x001B = PWMPER7

	7	6	5	4	3	2	1	0
R W	Bit 7	6	5	4	3	2	1	Bit 0
Reset	1	1	1	1	1	1	1	1

Figure 13-13. PWM Channel Period Registers (PWMPERx)

[1] This register is available only when the corresponding channel exists and is reserved if that channel does not exist. Writes to a reserved register have no functional effect. Reads from a reserved register return zeroes.

Read: Anytime

Write: Anytime

13.3.2.12 PWM Channel Duty Registers (PWMDTYx)

There is a dedicated duty register for each channel. The value in this register determines the duty of the associated PWM channel. The duty value is compared to the counter and if it is equal to the counter value a match occurs and the output changes state.

The duty registers for each channel are double buffered so that if they change while the channel is enabled, the change will NOT take effect until one of the following occurs:

- The effective period ends
- The counter is written (counter resets to $00)

To determine the value of a data bit and to detect noise, recovery logic takes samples at RT8, RT9, and RT10. Table 14-18 summarizes the results of the data bit samples.

Table 14-18. Data Bit Recovery

RT8, RT9, and RT10 Samples	Data Bit Determination	Noise Flag
000	0	0
001	0	1
010	0	1
011	1	1
100	0	1
101	1	1
110	1	1
111	1	0

NOTE

The RT8, RT9, and RT10 samples do not affect start bit verification. If any or all of the RT8, RT9, and RT10 start bit samples are logic 1s following a successful start bit verification, the noise flag (NF) is set and the receiver assumes that the bit is a start bit (logic 0).

To verify a stop bit and to detect noise, recovery logic takes samples at RT8, RT9, and RT10. Table 14-19 summarizes the results of the stop bit samples.

Table 14-19. Stop Bit Recovery

RT8, RT9, and RT10 Samples	Framing Error Flag	Noise Flag
000	1	0
001	1	1
010	1	1
011	0	1
100	1	1
101	0	1
110	0	1
111	0	0

14.5.3.1 Description of Interrupt Operation

The SCI only originates interrupt requests. The following is a description of how the SCI makes a request and how the MCU should acknowledge that request. The interrupt vector offset and interrupt number are chip dependent. The SCI only has a single interrupt line (SCI Interrupt Signal, active high operation) and all the following interrupts, when generated, are ORed together and issued through that port.

14.5.3.1.1 TDRE Description

The TDRE interrupt is set high by the SCI when the transmit shift register receives a byte from the SCI data register. A TDRE interrupt indicates that the transmit data register (SCIDRH/L) is empty and that a new byte can be written to the SCIDRH/L for transmission.Clear TDRE by reading SCI status register 1 with TDRE set and then writing to SCI data register low (SCIDRL).

14.5.3.1.2 TC Description

The TC interrupt is set by the SCI when a transmission has been completed. Transmission is completed when all bits including the stop bit (if transmitted) have been shifted out and no data is queued to be transmitted. No stop bit is transmitted when sending a break character and the TC flag is set (providing there is no more data queued for transmission) when the break character has been shifted out. A TC interrupt indicates that there is no transmission in progress. TC is set high when the TDRE flag is set and no data, preamble, or break character is being transmitted. When TC is set, the TXD pin becomes idle (logic 1). Clear TC by reading SCI status register 1 (SCISR1) with TC set and then writing to SCI data register low (SCIDRL).TC is cleared automatically when data, preamble, or break is queued and ready to be sent.

14.5.3.1.3 RDRF Description

The RDRF interrupt is set when the data in the receive shift register transfers to the SCI data register. A RDRF interrupt indicates that the received data has been transferred to the SCI data register and that the byte can now be read by the MCU. The RDRF interrupt is cleared by reading the SCI status register one (SCISR1) and then reading SCI data register low (SCIDRL).

14.5.3.1.4 OR Description

The OR interrupt is set when software fails to read the SCI data register before the receive shift register receives the next frame. The newly acquired data in the shift register will be lost in this case, but the data already in the SCI data registers is not affected. The OR interrupt is cleared by reading the SCI status register one (SCISR1) and then reading SCI data register low (SCIDRL).

14.5.3.1.5 IDLE Description

The IDLE interrupt is set when 10 consecutive logic 1s (if M = 0) or 11 consecutive logic 1s (if M = 1) appear on the receiver input. Once the IDLE is cleared, a valid frame must again set the RDRF flag before an idle condition can set the IDLE flag. Clear IDLE by reading SCI status register 1 (SCISR1) with IDLE set and then reading SCI data register low (SCIDRL).

NOTE

When peripherals with duplex capability are used, take care not to simultaneously enable two receivers whose serial outputs drive the same system slave's serial data output line.

As long as no more than one slave device drives the system slave's serial data output line, it is possible for several slaves to receive the same transmission from a master, although the master would not receive return information from all of the receiving slaves.

If the CPHA bit in SPI control register 1 is clear, odd numbered edges on the SCK input cause the data at the serial data input pin to be latched. Even numbered edges cause the value previously latched from the serial data input pin to shift into the LSB or MSB of the SPI shift register, depending on the LSBFE bit.

If the CPHA bit is set, even numbered edges on the SCK input cause the data at the serial data input pin to be latched. Odd numbered edges cause the value previously latched from the serial data input pin to shift into the LSB or MSB of the SPI shift register, depending on the LSBFE bit.

When CPHA is set, the first edge is used to get the first data bit onto the serial data output pin. When CPHA is clear and the $\overline{SS}$ input is low (slave selected), the first bit of the SPI data is driven out of the serial data output pin. After the nth[1] shift, the transfer is considered complete and the received data is transferred into the SPI data register. To indicate transfer is complete, the SPIF flag in the SPI status register is set.

NOTE

A change of the bits CPOL, CPHA, SSOE, LSBFE, MODFEN, SPC0, or BIDIROE with SPC0 set in slave mode will corrupt a transmission in progress and must be avoided.

15.4.3 Transmission Formats

During an SPI transmission, data is transmitted (shifted out serially) and received (shifted in serially) simultaneously. The serial clock (SCK) synchronizes shifting and sampling of the information on the two serial data lines. A slave select line allows selection of an individual slave SPI device; slave devices that are not selected do not interfere with SPI bus activities. Optionally, on a master SPI device, the slave select line can be used to indicate multiple-master bus contention.

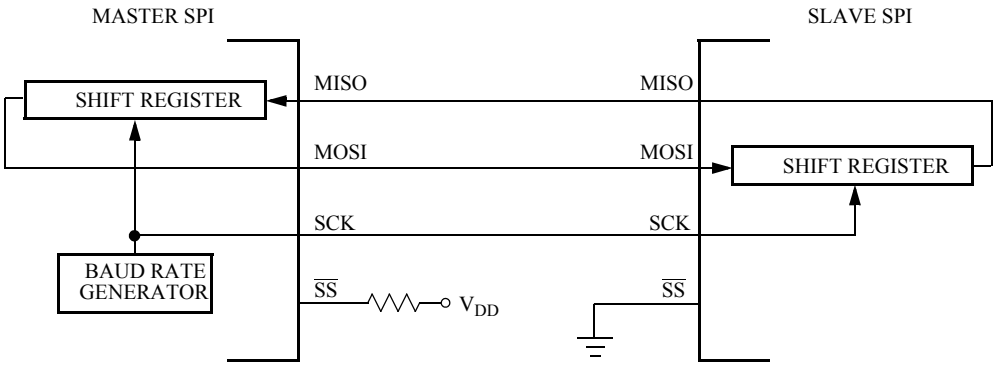


Figure 15-11. Master/Slave Transfer Block Diagram

1. n depends on the selected transfer width, please refer to Section 15.3.2.2, "SPI Control Register 2 (SPICR2)

15.4.7.5.2 SPIF

SPIF occurs when new data has been received and copied to the SPI data register. After SPIF is set, it does not clear until it is serviced. SPIF has an automatic clearing process, which is described in Section 15.3.2.4, “SPI Status Register (SPISR)”.

15.4.7.5.3 SPTEF

SPTEF occurs when the SPI data register is ready to accept new data. After SPTEF is set, it does not clear until it is serviced. SPTEF has an automatic clearing process, which is described in Section 15.3.2.4, “SPI Status Register (SPISR)”.

16.4.1.5 Repeated START Signal

As shown in Figure 16-10, a repeated START signal is a START signal generated without first generating a STOP signal to terminate the communication. This is used by the master to communicate with another slave or with the same slave in different mode (transmit/receive mode) without releasing the bus.

16.4.1.6 Arbitration Procedure

The Inter-IC bus is a true multi-master bus that allows more than one master to be connected on it. If two or more masters try to control the bus at the same time, a clock synchronization procedure determines the bus clock, for which the low period is equal to the longest clock low period and the high is equal to the shortest one among the masters. The relative priority of the contending masters is determined by a data arbitration procedure, a bus master loses arbitration if it transmits logic 1 while another master transmits logic 0. The losing masters immediately switch over to slave receive mode and stop driving SDA output. In this case the transition from master to slave mode does not generate a STOP condition. Meanwhile, a status bit is set by hardware to indicate loss of arbitration.

16.4.1.7 Clock Synchronization

Because wire-AND logic is performed on SCL line, a high-to-low transition on SCL line affects all the devices connected on the bus. The devices start counting their low period and as soon as a device's clock has gone low, it holds the SCL line low until the clock high state is reached.However, the change of low to high in this device clock may not change the state of the SCL line if another device clock is within its low period. Therefore, synchronized clock SCL is held low by the device with the longest low period. Devices with shorter low periods enter a high wait state during this time (see Figure 16-11). When all devices concerned have counted off their low period, the synchronized clock SCL line is released and pulled high. There is then no difference between the device clocks and the state of the SCL line and all the devices start counting their high periods.The first device to complete its high period pulls the SCL line low again.

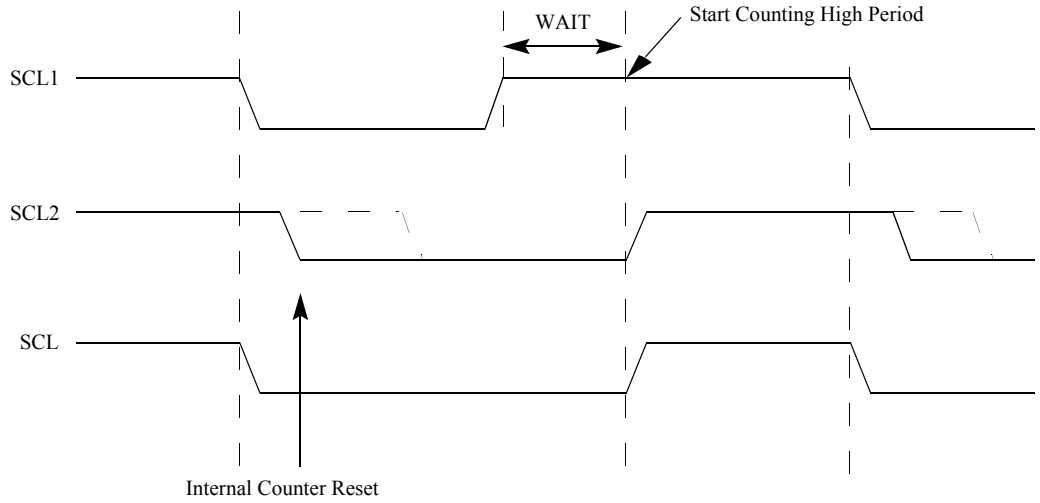


Figure 16-12. IIC-Bus Clock Synchronization

If a master receiver wants to terminate a data transfer, it must inform the slave transmitter by not acknowledging the last byte of data which can be done by setting the transmit acknowledge bit (TXAK) before reading the 2nd last byte of data. Before reading the last byte of data, a STOP signal must be generated first. The following is an example showing how a STOP signal is generated by a master receiver.

```
MASR    DEC     RXCNT       ;DECREASE THE RXCNT
        BEQ     ENMASR      ;LAST BYTE TO BE READ
        MOVB    RXCNT,D1    ;CHECK SECOND LAST BYTE
        DEC     D1          ;TO BE READ
        BNE     NXMAR       ;NOT LAST OR SECOND LAST
LAMAR   BSET    IBCR,#$08   ;SECOND LAST, DISABLE ACK
                            ;TRANSMITTING
        BRA     NXMAR
ENMASR  BCLR    IBCR,#$20   ;LAST ONE, GENERATE 'STOP' SIGNAL
NXMAR   MOVB    IBDR,RXBUF  ;READ DATA AND STORE
        RTI
```

16.7.1.5 Generation of Repeated START

At the end of data transfer, if the master continues to want to communicate on the bus, it can generate another START signal followed by another slave address without first generating a STOP signal. A program example is as shown.

```
RESTART BSET    IBCR,#$04       ;ANOTHER START (RESTART)
        MOVB    CALLING,IBDR    ;TRANSMIT THE CALLING ADDRESS;D0=R/W
```

16.7.1.6 Slave Mode

In the slave interrupt service routine, the module addressed as slave bit (IAAS) should be tested to check if a calling of its own address has just been received. If IAAS is set, software should set the transmit/receive mode select bit (Tx/Rx bit of IBCR) according to the R/W command bit (SRW). Writing to the IBCR clears the IAAS automatically. Note that the only time IAAS is read as set is from the interrupt at the end of the address cycle where an address match occurred, interrupts resulting from subsequent data transfers will have IAAS cleared. A data transfer may now be initiated by writing information to IBDR, for slave transmits, or dummy reading from IBDR, in slave receive mode. The slave will drive SCL low in-between byte transfers, SCL is released when the IBDR is accessed in the required mode.

In slave transmitter routine, the received acknowledge bit (RXAK) must be tested before transmitting the next byte of data. Setting RXAK means an 'end of data' signal from the master receiver, after which it must be switched from transmitter mode to receiver mode by software. A dummy read then releases the SCL line so that the master can generate a STOP signal.

16.7.1.7 Arbitration Lost

If several masters try to engage the bus simultaneously, only one master wins and the others lose arbitration. The devices which lost arbitration are immediately switched to slave receive mode by the hardware. Their data output to the SDA line is stopped, but SCL continues to be generated until the end of the byte during which arbitration was lost. An interrupt occurs at the falling edge of the ninth clock of this transfer with IBAL=1 and MS/SL=0. If one master attempts to start transmission while the bus is being engaged by another master, the hardware will inhibit the transmission; switch the MS/SL bit from 1 to 0

Table 17-7. CPIE Register Field Descriptions

Field	Description
4 CPVFIE	**CAN Physical Layer Voltage-Failure Interrupt Enable** If enabled, the CAN Physical Layer generates an interrupt if any of the CAN Physical Layer voltage failure interrupt flags assert. 0 Voltage failure interrupt is disabled 1 Voltage failure interrupt is enabled
3 CPDTIE	**CPTXD-Dominant Timeout Interrupt Enable** If enabled, the CAN Physical Layer generates an interrupt if the CPTXD-dominant timeout interrupt flag asserts. 0 CPTXD-dominant timeout interrupt is disabled 1 CPTXD-dominant timeout interrupt is enabled
0 CPOCIE	**CAN Physical Layer Over-current Interrupt Enable** If enabled, the CAN Physical Layer generates an interrupt if any of the CAN Physical Layer over-current interrupt flags assert. 0 Over-current interrupt is disabled 1 Over-current interrupt is enabled

17.4.2.8 CAN Physical Layer Interrupt Flag Register (CPIF)

Module Base + 0x0007 Access: User read/write[1]

	7	6	5	4	3	2	1	0
R	CHVHIF	CHVLIF	CLVHIF	CLVLIF	CPDTIF	0	CHOCIF	CLOCIF
W								
Reset	0	0	0	0	0	0	0	0

Figure 17-9. CAN Physical Layer Interrupt Flag Register (CPIF)

[1] Read: Anytime
Write: Anytime, write 1 to clear

If any of the flags is asserted an error interrupt is pending if enabled. A flag can be cleared by writing a logic level 1 to the corresponding bit location. Writing a 0 has no effect.

Table 17-8. CPIF Register Field Descriptions

Field	Description
7 CHVHIF	**CANH Voltage Failure High Interrupt Flag** This flag is set to 1 when the CPCHVH bit in the CAN Physical Layer Status Register (CPSR) changes. 0 No change in CPCHVH 1 CPCHVH has changed
6 CHVLIF	**CANH Voltage Failure Low Interrupt Flag** This flag is set to 1 when the CPCHVL bit in the CAN Physical Layer Status Register (CPSR) changes. 0 No change in CPCHVL 1 CPCHVL has changed

Table 18-35. Time Segment Syntax

Syntax	Description
SYNC_SEG	System expects transitions to occur on the CAN bus during this period.
Transmit Point	A node in transmit mode transfers a new value to the CAN bus at this point.
Sample Point	A node in receive mode samples the CAN bus at this point. If the three samples per bit option is selected, then this point marks the position of the third sample.

The synchronization jump width (see the Bosch CAN 2.0A/B specification for details) can be programmed in a range of 1 to 4 time quanta by setting the SJW parameter.

The SYNC_SEG, TSEG1, TSEG2, and SJW parameters are set by programming the MSCAN bus timing registers (CANBTR0, CANBTR1) (see Section 18.3.2.3, "MSCAN Bus Timing Register 0 (CANBTR0)" and Section 18.3.2.4, "MSCAN Bus Timing Register 1 (CANBTR1)").

Table 18-36 gives an overview of the Bosch CAN 2.0A/B specification compliant segment settings and the related parameter values.

NOTE

It is the user's responsibility to ensure the bit time settings are in compliance with the CAN standard.

Table 18-36. Bosch CAN 2.0A/B Compliant Bit Time Segment Settings

Time Segment 1	TSEG1	Time Segment 2	TSEG2	Synchronization Jump Width	SJW
5 .. 10	4 .. 9	2	1	1 .. 2	0 .. 1
4 .. 11	3 .. 10	3	2	1 .. 3	0 .. 2
5 .. 12	4 .. 11	4	3	1 .. 4	0 .. 3
6 .. 13	5 .. 12	5	4	1 .. 4	0 .. 3
7 .. 14	6 .. 13	6	5	1 .. 4	0 .. 3
8 .. 15	7 .. 14	7	6	1 .. 4	0 .. 3
9 .. 16	8 .. 15	8	7	1 .. 4	0 .. 3

18.4.4 Modes of Operation

18.4.4.1 Normal System Operating Modes

The MSCAN module behaves as described within this specification in all normal system operating modes. Write restrictions exist for some registers.

N.8 0x0400-0x042F TIM1

Register Name		Bit 7	6	5	4	3	2	1	Bit 0
0x0400 TIOS	R					IOS3	IOS2	IOS1	IOS0
	W								
0x0401 CFORC	R	0	0	0	0	0	0	0	0
	W					FOC3	FOC2	FOC1	FOC0
0x0402 Reserved	R								
	W								
0x0403 Reserved	R								
	W								
0x0404 TCNTH	R	TCNT15	TCNT14	TCNT13	TCNT12	TCNT11	TCNT10	TCNT9	TCNT8
	W								
0x0405 TCNTL	R	TCNT7	TCNT6	TCNT5	TCNT4	TCNT3	TCNT2	TCNT1	TCNT0
	W								
0x0406 TSCR1	R	TEN	TSWAI	TSFRZ	TFFCA	PRNT	0	0	0
	W								
0x0407 TTOV	R					TOV3	TOV2	TOV1	TOV0
	W								
0x0408 TCTL1	R								
	W								
0x0409 TCTL2	R	OM3	OL3	OM2	OL2	OM1	OL1	OM0	OL0
	W								
0x040A TCTL3	R								
	W								
0x040B TCTL4	R	EDG3B	EDG3A	EDG2B	EDG2A	EDG1B	EDG1A	EDG0B	EDG0A
	W								
0x040C TIE	R					C3I	C2I	C1I	C0I
	W								
0x040D TSCR2	R	TOI	0	0	0		PR2	PR1	PR0
	W								
0x040E TFLG1	R					C3F	C2F	C1F	C0F
	W								
0x040F TFLG2	R	TOF	0	0	0	0	0	0	0
	W								
0x0410–0x041F TCxH–TCxL[1]	R	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8
	W								
	R	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	W								
0x0420–0x042B Reserved	R								
	W								
0x042C OCPD	R					OCPD3	OCPD2	OCPD1	OCPD0
	W								
0x042D Reserved	R								
0x042E PTPSR	R	PTPS7	PTPS6	PTPS5	PTPS4	PTPS3	PTPS2	PTPS1	PTPS0
	W								

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