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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	64MHz
Connectivity	I²C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LVD, POR, PWM, WDT
Number of I/O	44
Program Memory Size	64KB (32K x 16)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	4K x 8
Voltage - Supply (Vcc/Vdd)	2.3V ~ 5.5V
Data Converters	A/D 43x12b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-TQFP Exposed Pad
Supplier Device Package	48-TQFP-EP (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18f56k42t-i-pt

7.2.2.6 Oscillator Status and Manual Enable

The Ready status of each oscillator (including the ADCRC oscillator) is displayed in OSCSTAT (Register 7-4). The oscillators (but not the PLL) may be explicitly enabled through OSCEN (Register 7-7).

7.2.2.7 HFOR and MFOR Bits

The HFOR and MFOR bits indicate that the HFINTOSC and MFINTOSC is ready. These clocks are always valid for use at all times, but only accurate after they are ready.

When a new value is loaded into the OSCFRQ register, the HFOR and MFOR bits will clear, and set again when the oscillator is ready. During pending OSCFRQ changes the MFINTOSC clock will stall at a high or a low state, until the HFINTOSC resumes operation.

7.3 Clock Switching

The system clock source can be switched between external and internal clock sources via software using the New Oscillator Source (NOSC) bits of the OSCCON1 register. The following clock sources can be selected using the following:

- External oscillator
- Internal Oscillator Block (INTOSC)

Note:	The Clock Switch Enable bit in Configuration Word 1 can be used to enable or disable the clock switching capability. When cleared, the NOSC and NDIV bits cannot be changed by user software. When set, writing to NOSC and NDIV is allowed and would switch the clock frequency.

7.3.1 NEW OSCILLATOR SOURCE (NOSC) AND NEW DIVIDER SELECTION REQUEST (NDIV) BITS

The New Oscillator Source (NOSC) and New Divider Selection Request (NDIV) bits of the OSCCON1 register select the system clock source and frequency that are used for the CPU and peripherals.

When new values of NOSC and NDIV are written to OSCCON1, the current oscillator selection will continue to operate while waiting for the new clock source to indicate that it is stable and ready. In some cases, the newly requested source may already be in use, and is ready immediately. In the case of a divider-only change, the new and old sources are the same, so the old source will be ready immediately. The device may enter Sleep while waiting for the switch as described in **Section 7.3.2 "Clock Switch and Sleep"**.

When the new oscillator is ready, the New Oscillator Ready (NOSCR) bit of OSCCON3 and the Clock Switch Interrupt Flag (CSWIF) bit of the respective PIR register are set. If Clock Switch Interrupts are enabled (CSWIE = 1), an interrupt will be generated at that time. The Oscillator Ready (ORDY) bit of OSCCON3 can also be polled to determine when the oscillator is ready in lieu of an interrupt.

Note:	The CSWIF interrupt will not wake the system from Sleep.

If the Clock Switch Hold (CSWHOLD) bit of OSCCON3 is clear, the oscillator switch will occur when the New Oscillator is Ready bit (NOSCR) is set, and the interrupt (if enabled) will be serviced at the new oscillator setting.

If CSWHOLD is set, the oscillator switch is suspended, while execution continues using the current (old) clock source. When the NOSCR bit is set, software should:

- Set CSWHOLD = 0 so the switch can complete, or
- Copy COSC into NOSC to abandon the switch.

If DOZE is in effect, the switch occurs on the next clock cycle, whether or not the CPU is operating during that cycle.

Changing the clock post-divider without changing the clock source (i.e., changing FOSC from 1 MHz to 2 MHz) is handled in the same manner as a clock source change, as described previously. The clock source will already be active, so the switch is relatively quick. CSWHOLD must be clear (CSWHOLD = 0) for the switch to complete.

The current COSC and CDIV are indicated in the OSCCON2 register up to the moment when the switch actually occurs, at which time OSCCON2 is updated and ORDY is set. NOSCR is cleared by hardware to indicate that the switch is complete.

TABLE 9-1: IVT ADDRESS CALCULATION SUMMARY

IVT Address Calculation		Interrupt Priority INTCON0 Register, IPEN bit	
		0	1
Multi-Vector Enable CONFIG 2L register MVECEN bit	0	IVTBASE	High Priority IVTBASE
			Low Priority IVTBASE + 8 words
	1	IVTBASE + 2*(Vector Number)	

9.2.4 ACCESS CONTROL FOR IVTBASE REGISTERS

The Interrupt controller has an IVTLOCKED bit which can be set to avoid inadvertent changes to the IVTBASE registers contents. Setting and clearing this bit requires a special sequence as an extra precaution against inadvertent changes.

To allow writes to IVTBASE registers, the interrupts must be disabled (GIEH = 0) and the IVTLOCKED bit must be cleared. The user must follow the sequence shown in Example 9-1 to clear the IVTLOCKED bit.

EXAMPLE 9-1: IVT UNLOCK SEQUENCE

```
; Disable Interrupts:
        BCF             INTCON0, GIE;
; Bank to IVTLOCK register
        BANKSEL         IVTLOCK;
        MOVLW           55h;

; Required sequence, next 4 instructions
        MOVWF           IVTLOCK;
        MOVLW           AAh;
        MOVWF           IVTLOCK;

; Clear IVTLOCKED bit to enable writes
        BCF             IVTLOCK, IVTLOCKED;

; Enable Interrupts
        BSF             INTCON0, GIE;
```

The user must follow the sequence shown in Example 9-2 to set the IVTLOCKED bit.

EXAMPLE 9-2: IVT LOCK SEQUENCE

```
; Disable Interrupts:
        BCF             INTCON0, GIE;
; Bank to IVTLOCK register
        BANKSEL         IVTLOCK;
        MOVLW           55h;

; Required sequence, next 4 instructions
        MOVWF           IVTLOCK;
        MOVLW           AAh;
        MOVWF           IVTLOCK;

; Set IVTLOCKED bit to enable writes
        BSF             IVTLOCK, IVTLOCKED;

; Enable Interrupts
        BSF             INTCON0, GIE;
```

When the IVT1WAY Configuration bit is set, the IVTLOCKED bit can be cleared and set only once after a device Reset. The unlock operation in Example 9-1 will have no effect after the lock sequence in Example 9-2 is used to set the IVTLOCK. Unlocking is inhibited until a system Reset occurs.

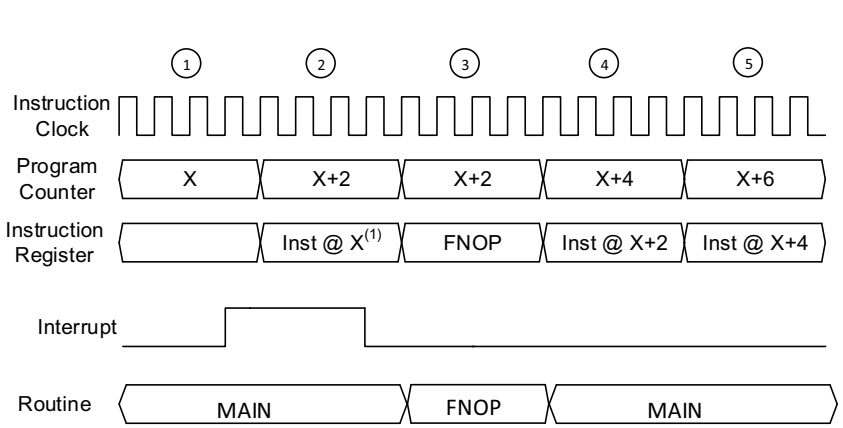
9.7.1 ABORTING INTERRUPTS

If the last instruction before the interrupt controller vectors to the ISR from main routine clears the GIE, PIE or PIR bit associated with the interrupt, the controller executes one force NOP cycle before it returns to the main routine.

Figure 9-10 illustrates the sequence of events when a peripheral interrupt is asserted and then cleared on the last executed instruction cycle.

If the GIE, PIE or PIR bit associated with the interrupt is cleared prior to vectoring to the ISR, then the controller continues executing the main routine.

FIGURE 9-10: INTERRUPT TIMING DIAGRAM - ABORTING INTERRUPTS



Note 1: Inst @ X clears the interrupt flag, Example BCF INTCON0, GIE.

TABLE 16-11: SUMMARY OF REGISTERS ASSOCIATED WITH I/O (CONTINUED)

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INLVLC	INLVLC7	INLVLC6	INLVLC5	INLVLC4(5)	INLVLC3(5)	INLVLC2	INLVLC1	INLVLC0	270
INLVLD(6)	INLVLD7	INLVLD6	INLVLD5	INLVLD4	INLVLD3	INLVLD2	INLVLD1(5)	INLVLD0(5)	270
INLVLF(7)	INLVLF7	INLVLF6	INLVLF5	INLVLF4	INLVLF3	INLVLF2	INLVLF1	INLVLF0	270
INLVLE	—	—	—	—	INLVLE3	—	—	—	270
RB1I2C	—	SLEW	PU<1:0>		—	—	TH<1:0>		271
RB2I2C	—	SLEW	PU<1:0>		—	—	TH<1:0>		271
RC3I2C	—	SLEW	PU<1:0>		—	—	TH<1:0>		271
RC4I2C	—	SLEW	PU<1:0>		—	—	TH<1:0>		271
RD0I2C(6)	—	SLEW	PU<1:0>		—	—	TH<1:0>		271
RD1I2C(6)	—	SLEW	PU<1:0>		—	—	TH<1:0>		271

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by I/O Ports.

Note 1: Bits RB6 and RB7 read '1' while in Debug mode.
2: Bit PORTE3 is read-only, and will read '1' when MCLRE = 1 (Master Clear enabled).
3: Bits RB6 and RB7 read '1' while in Debug mode.
4: If MCLRE = 1, the weak pull-up in RE3 is always enabled; bit WPUE3 is not affected.
5: Any peripheral using the I^2C pins read the I^2C ST inputs when enabled via RxyI2C.
6: Unimplemented in PIC18(L)F26/27K42.
7: Unimplemented in PIC18(L)F26/27/45/46/47K42 parts.

REGISTER 17-3: PPSLOCK: PPS LOCK REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	U-0	R/W-0/0
—	—	—	—	—	—	—	PPSLOCKED
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

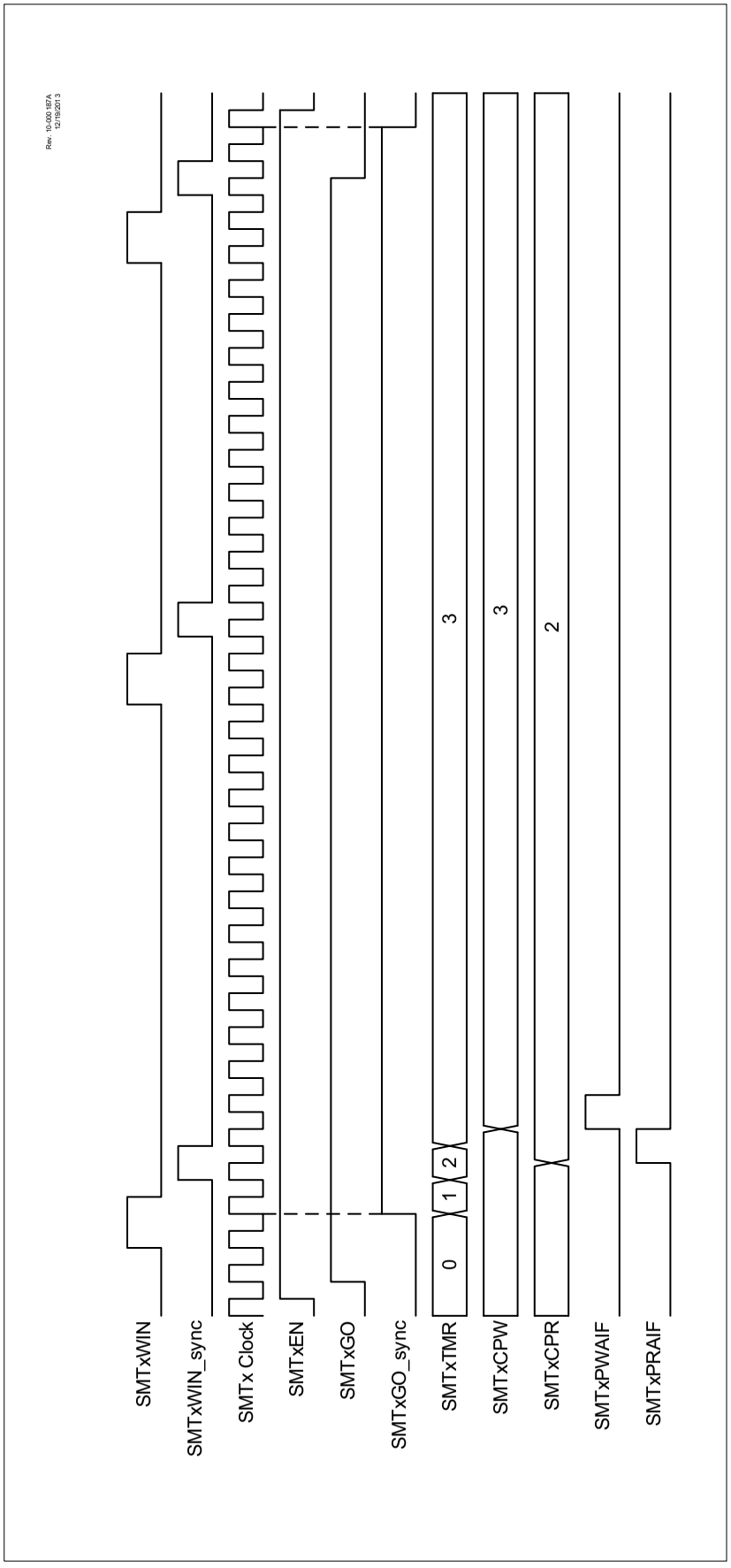
bit 7-1 **Unimplemented:** Read as ‘0’

bit 0 **PPSLOCKED:** PPS Locked bit

1 = PPS is locked.

0 = PPS is not locked. PPS selections can be changed.

FIGURE 25-17: CAPTURE MODE SINGLE ACQUISITION TIMING DIAGRAM



25.7 Interrupts

The SMT can trigger an interrupt under three different conditions:

- PW Acquisition Complete
- PR Acquisition Complete
- Counter Period Match

The interrupts are controlled by the PIR and PIE registers of the device.

25.7.1 PW AND PR ACQUISITION INTERRUPTS

The SMT can trigger interrupts whenever it updates the SMT1CPW and SMT1CPR registers, the circumstances for which are dependent on the SMT mode, and are discussed in each mode's specific section. The SMT1CPW interrupt is controlled by SMT1PWAIF and SMT1PWAIE bits in the respective PIR and PIE registers. The SMT1CPR interrupt is controlled by the SMT1PRAIF and SMT1PRAIE bits, also located in the respective PIR and PIE registers.

In synchronous SMT modes, the interrupt trigger is synchronized to the SMT1CLK. In Asynchronous modes, the interrupt trigger is asynchronous. In either mode, once triggered, the interrupt will be synchronized to the CPU clock.

25.7.2 COUNTER PERIOD MATCH INTERRUPT

As described in **Section 25.1.2 "Period Match interrupt"**, the SMT will also interrupt upon SMT1TMR, matching SMT1PR with its period match limit functionality described in **Section 25.3 "Halt Operation"**. The period match interrupt is controlled by SMT1IF and SMT1IE, located in the respective PIR and PIE registers.

Measuring VCPINV can be difficult, especially when the waveform is relative to VDD. However, by combining Equations 29-2 and 29-3, the resistor value can be determined from the time difference between the ZCD_output high and low intervals. Note that the time difference, ΔT, is 4*TOFFSET. The equation for determining the pull-up and pull-down resistor values from the high and low ZCD_output periods is shown in Equation 29-4.

EQUATION 29-4: PULL-UP/DOWN RESISTOR VALUES

$$R = R_{SERIES}\left(\frac{V_{BIAS}}{V_{PEAK}\left(\sin\left(\pi Freq\frac{(\Delta T)}{2}\right)\right)} - 1\right)$$

R is pull-up or pull-down resistor.

VBIAS is VPULLUP when R is pull-up or VDD when R is pull-down.

ΔT is the ZCDOUT high and low period difference.

29.6 Handling VPEAK Variations

If the peak amplitude of the external voltage is expected to vary, the series resistor must be selected to keep the ZCD current source and sink below the design maximum range of ± 600 μA and above a reasonable minimum range. A general rule of thumb is that the maximum peak voltage can be no more than six times the minimum peak voltage. To ensure that the maximum current does not exceed ± 600 μA and the minimum is at least ± 100 μA, compute the series resistance as shown in Equation 29-5. The compensating pull-up for this series resistance can be determined with Equation 29-3 because the pull-up value is not dependent to the peak voltage.

EQUATION 29-5: SERIES R FOR V RANGE

$$R_{SERIES} = \frac{V_{MAXPEAK} + V_{MINPEAK}}{7\times10^{-4}}$$

29.7 Operation During Sleep

The ZCD current sources and interrupts are unaffected by Sleep.

29.8 Effects of a Reset

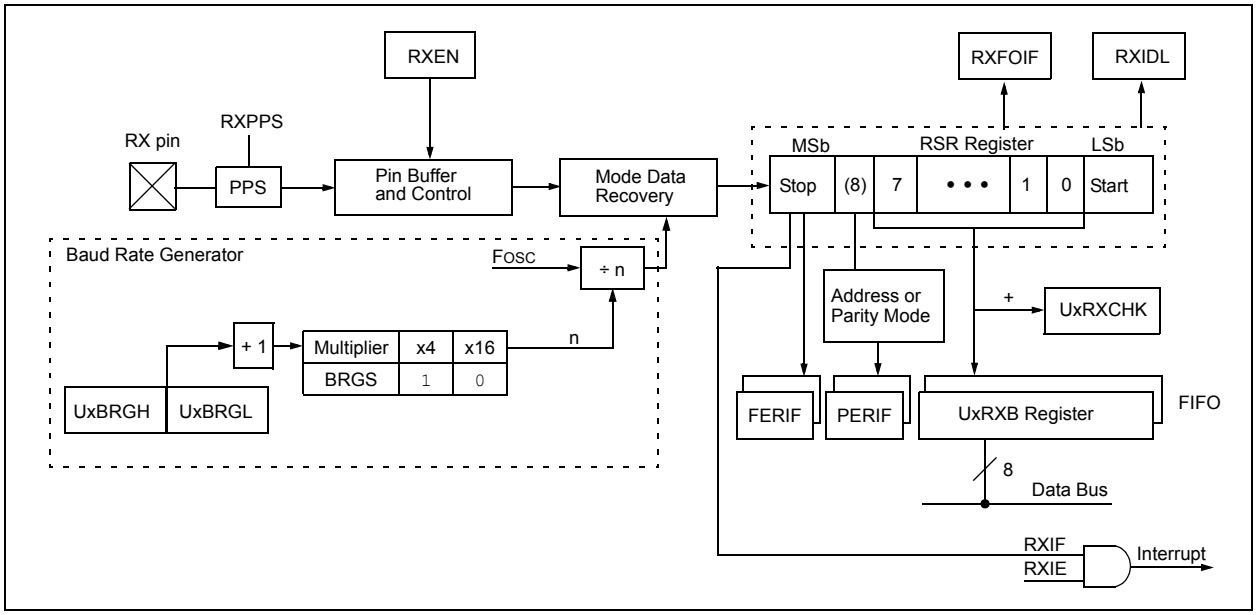
The ZCD circuit can be configured to default to the active or inactive state on Power-on-Reset (POR). When the $\overline{ZCD}$ Configuration bit is cleared, the ZCD circuit will be active at POR. When the $\overline{ZCD}$ Configuration bit is set, the SEN bit of the ZCDCON register must be set to enable the ZCD module.

29.9 Disabling the ZCD Module

The ZCD module can be disabled in two ways:

1. Configuration Word 2H has the $\overline{ZCD}$ bit which disables the ZCD module when set, but it can be enabled using the SEN bit of the ZCDCON register (Register 29-1). If the $\overline{ZCD}$ bit is clear, the ZCD is always enabled.
2. The ZCD can also be disabled using the ZCDMD bit of the respective PMD2 register (Register 19-3). This is subject to the status of the $\overline{ZCD}$ bit.

FIGURE 31-2: UART RECEIVE BLOCK DIAGRAM



The operation of the UART module is controlled through nineteen registers:

- Three control registers (UxCON0-UxCON2)
- Error enable and status (UxERRIE, UxERRIR, UxUIR)
- UART buffer status and control (UxFIFO)
- Three 9-bit protocol parameters (UxP1-UxP3)
- 16-bit baud rate generator (UxBRGH:L)
- Transmit buffer write (UxTXB)
- Receive buffer read (UxRXB)
- Receive checksum (UxRXCHK)
- Transmit checksum (UxTXCHK)

These registers are detailed in **Section 31.21 "Register Definitions: UART Control"**.

31.1 UART I/O Pin Configuration

The RX input pin is selected with the UxRPPS register. The TX output pin is selected with each pin's RxyPPS register. When the TRIS control for the pin corresponding to the TX output is cleared, then the UART will maintain control and the logic level on the TX pin. Changing the TXPOL bit in UxCON2 will immediately change the TX pin logic level regardless of the value of EN or TXEN.

31.2 UART Asynchronous Modes

The UART has five asynchronous modes:

- 7-bit
- 8-bit
- 8-bit with even parity in the 9th bit
- 8-bit with odd parity in the 9th bit
- 8-bit with address indicator in the 9th bit

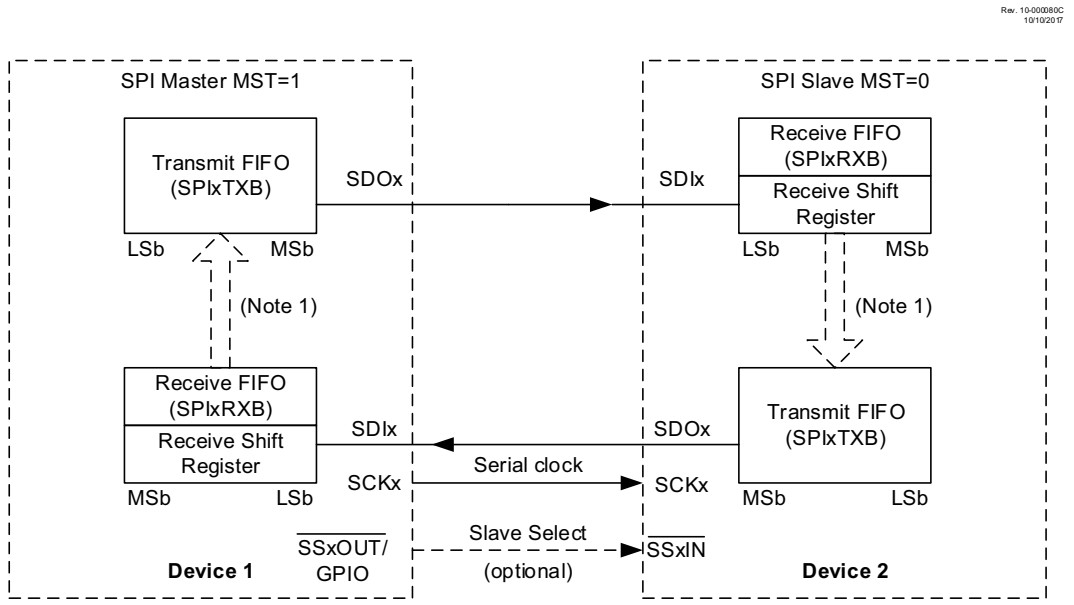
The UART transmits and receives data using the standard Non-Return-to-Zero (NRZ) format. NRZ is implemented with two levels: a VOH mark state, which represents a '1' data bit, and a VOL space state, which represents a '0' data bit. NRZ refers to the fact that consecutively transmitted data bits of the same value stay at the output level of that bit without returning to a neutral level between each bit transmission. An NRZ transmission port idles in the Mark state. Each character transmission consists of one Start bit followed by seven or eight data bits, one optional parity or address bit, and is always terminated by one or more Stop bits. The Start bit is always a space and the Stop bits are always marks. The most common data format is eight bits with no parity. Each transmitted bit persists for a period of 1/(Baud Rate). An on-chip dedicated 16-bit Baud Rate Generator is used to derive standard baud rate frequencies from the system oscillator. See **Section 31.17 "UART Baud Rate Generator (BRG)"** for more information.

In all the asynchronous modes, the UART transmits and receives the LSb first. The UART's transmitter and receiver are functionally independent, but share the same data format and baud rate. Parity is supported by the hardware by even and odd parity modes.

31.2.1 UART ASYNCHRONOUS TRANSMITTER

The UART transmitter block diagram is shown in Figure 31-1. The heart of the transmitter is the serial Transmit Shift Register (TSR), which is not directly accessible by software. The TSR obtains its data from the transmit buffer, which is the UxTXB register.

FIGURE 32-2: SPI MASTER/SLAVE CONNECTION WITH FIFOs



Note 1: In some modes, if the Transmit FIFO is empty, the most recently received byte of data will be transmitted.

2: This diagram assumes that the LSBF bit is cleared (communications are MSb-first). If LSBF is set, the communications will be LSb-first.

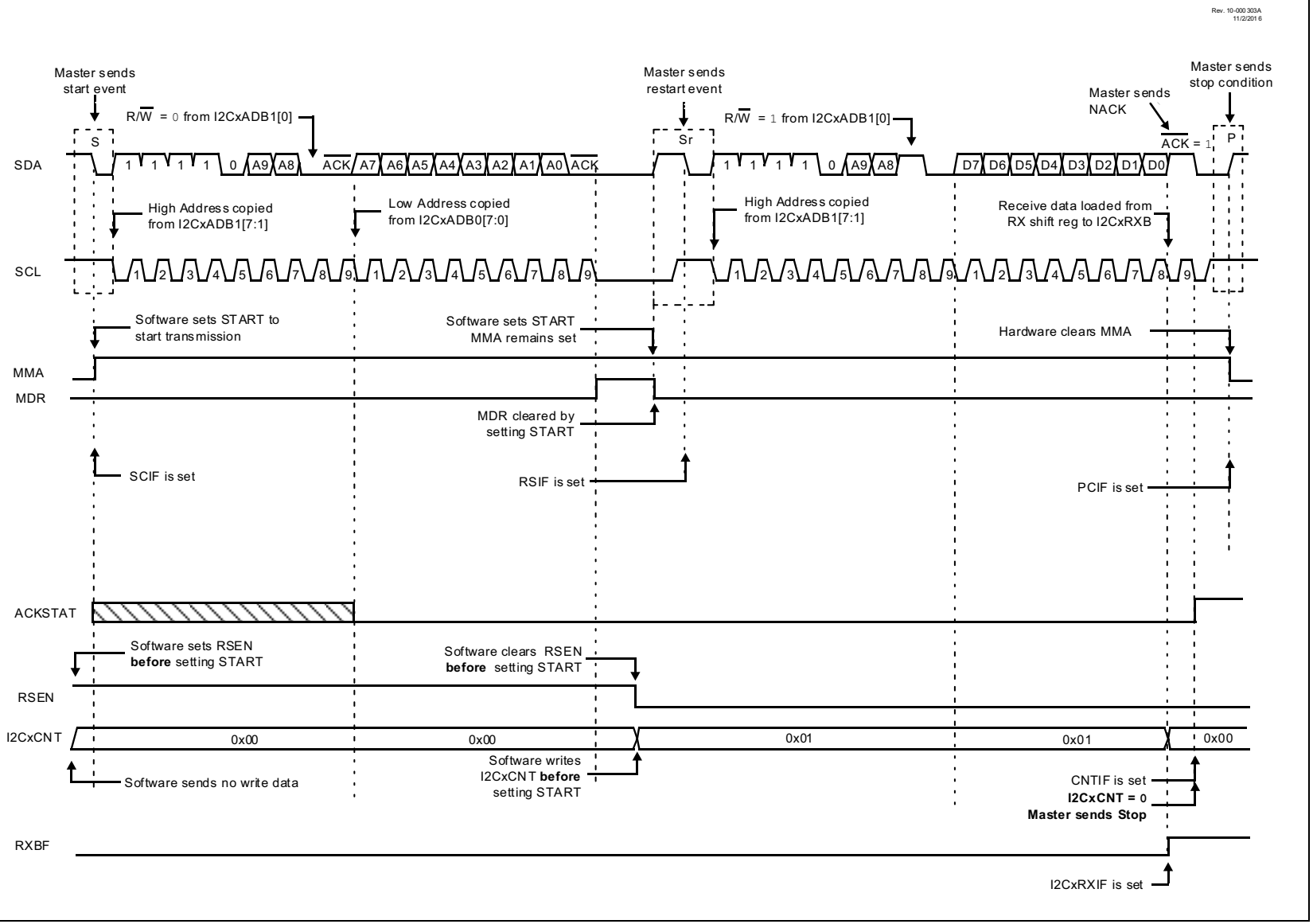
32.8.3.4 Receiver Overflow and Transmitter Underflow Interrupts

The receiver overflow interrupt triggers if data is received when the RXFIFO is already full and RXR = 1. In this case, the data will be discarded and the RXOIF bit will be set. The receiver overflow interrupt flag is the RXOIF bit of SPIxINTF. The receiver overflow interrupt enable bit is the RXOIE bit of SPIxINTE.

The Transmitter Underflow interrupt flag triggers if a data transfer begins when the TXFIFO is empty and TXR = 1. In this case, the most recently received data will be transmitted and the TXUIF bit will be set. The transmitter underflow interrupt flag is the TXUIF bit of SPIxINTF. The transmitter underflow interrupt enable bit is the TXUIE bit of SPIxINTE.

Both of these interrupts will only occur in Slave mode, as Master mode will not allow the RXFIFO to overflow or the TXFIFO to underflow.

FIGURE 33-22: I²C MASTER, 10-BIT ADDRESS, RECEPTION (USING RSTEN BIT)



36.5.2 PRECHARGE CONTROL

The precharge stage is an optional period of time that brings the external channel and internal sample and hold capacitor to known voltage levels. Precharge is enabled by writing a non-zero value to the ADPRE register. This stage is initiated when an ADC conversion begins, either from setting the GO bit, a special event trigger, or a conversion restart from the computation functionality. If the ADPRE register is cleared when an ADC conversion begins, this stage is skipped.

During the precharge time, CHOLD is disconnected from the outer portion of the sample path that leads to the external capacitive sensor and is connected to either VDD or VSS, depending on the value of the PPOL bit of ADCON1. At the same time, the port pin logic of the selected analog channel is overridden to drive a digital high or low out, in order to precharge the outer portion of the ADC's sample path, which includes the external sensor. The output polarity of this override is also determined by the PPOL bit of ADCON1. The amount of time that this charging receives is controlled by the ADPRE register.

Note 1:	The external charging overrides the TRIS setting of the respective I/O pin.
2:	If there is a device attached to this pin, Precharge should not be used.

36.5.3 ACQUISITION CONTROL

The Acquisition stage is an optional time for the voltage on the internal sample and hold capacitor to charge or discharge from the selected analog channel.This acquisition time is controlled by the ADACQ register. If PRE = 0, acquisition starts at the beginning of conversion. When PRE = 1, the acquisition stage begins when precharge ends.

At the start of the acquisition stage, the port pin logic of the selected analog channel is overridden to turn off the digital high/low output drivers so they do not affect the final result of the charge averaging. Also, the selected ADC channel is connected to CHOLD. This allows charge averaging to proceed between the precharged channel and the CHOLD capacitor.

Note:	When PRE! = 0, acquisition time cannot be '0'. In this case, setting ADACQ to '0' will set a maximum acquisition time (8191 ADC clock cycles). When precharge is disabled, setting ADACQ to '0' will disable hardware acquisition time control.

36.5.4 GUARD RING OUTPUTS

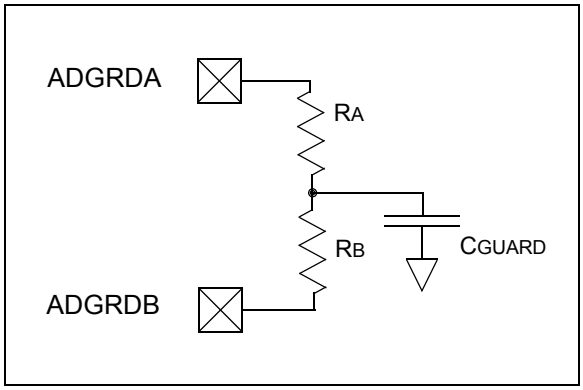
Figure 36-8 shows a typical guard ring circuit. CGUARD represents the capacitance of the guard ring trace placed on the PCB board. The user selects values for RA and RB that will create a voltage profile on CGUARD, which will match the selected acquisition channel.

The purpose of the guard ring is to generate a signal in phase with the CVD sensing signal to minimize the effects of the parasitic capacitance on sensing electrodes. It also can be used as a mutual drive for mutual capacitive sensing. For more information about active guard and mutual drive, see Application Note AN1478, *"mTouch™ Sensing Solution Acquisition Methods Capacitive Voltage Divider"* (DS01478).

The ADC has two guard ring drive outputs, ADGRDA and ADGRDB. These outputs can be routed through PPS controls to I/O pins (see **Section 17.0 "Peripheral Pin Select (PPS) Module"** for details) and the polarity of these outputs are controlled by the ADGPOL and ADIPEN bits of ADCON1.

At the start of the first precharge stage, both outputs are set to match the ADGPOL bit of ADCON1. Once the acquisition stage begins, ADGRDA changes polarity, while ADGRDB remains unchanged. When performing a double sample conversion, setting the ADIPEN bit of ADCON1 causes both guard ring outputs to transition to the opposite polarity of ADGPOL at the start of the second precharge stage, and ADGRDA toggles again for the second acquisition. For more information on the timing of the guard ring output, refer to Figure 36-8 and Figure 36-9.

FIGURE 36-8: GUARD RING CIRCUIT



38.13 Register Definitions: Comparator Control

Long bit name prefixes for the Comparators are shown in Table 38-2. Refer to **Section 1.3.2.2 "Long Bit Names"** for more information.

TABLE 38-2:

Peripheral	Bit Name Prefix
C1	C1
C2	C2

REGISTER 38-1: CMxCON0: COMPARATOR x CONTROL REGISTER 0

R/W-0/0	R-0/0	U-0	R/W-0/0	U-0	U-1	R/W-0/0	R/W-0/0
EN	OUT	—	POL	—	—	HYS	SYNC
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **EN:** Comparator Enable bit
- 1 = Comparator is enabled
- 0 = Comparator is disabled and consumes no active power

bit 6 **OUT:** Comparator Output bit

If POL = 0 (noninverted polarity):
- 1 = CxVP > CxVN
- 0 = CxVP < CxVN

If POL = 1 (inverted polarity):
- 1 = CxVP < CxVN
- 0 = CxVP > CxVN

bit 5 **Unimplemented:** Read as '0'

bit 4 **POL:** Comparator Output Polarity Select bit
- 1 = Comparator output is inverted
- 0 = Comparator output is not inverted

bit 3 **Unimplemented:** Read as '0'

bit 2 **Unimplemented:** Read as '1'

bit 1 **HYS:** Comparator Hysteresis Enable bit
- 1 = Comparator hysteresis enabled
- 0 = Comparator hysteresis disabled

bit 0 **SYNC:** Comparator Output Synchronous Mode bit
- 1 = Comparator output to Timer1/3/5 and I/O pin is synchronous to changes on Timer1 clock source.
- 0 = Comparator output to Timer1/3/5 and I/O pin is asynchronous
  Output updated on the falling edge of Timer1/3/5 clock source.

ANDWF **AND W with f**

Syntax: ANDWF f {,d {,a}}

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

Operation: (W) .AND. (f) → dest

Status Affected: N, Z

Encoding:

0001	01da	ffff	ffff

Description: The contents of W are AND'ed with register 'f'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f' (default).
If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See **Section 41.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: `ANDWF REG, 0, 0`

Before Instruction
W = 17h
REG = C2h
After Instruction
W = 02h
REG = C2h

BC **Branch if Carry**

Syntax: BC n

Operands: $-128 \le n \le 127$

Operation: if CARRY bit is '1'
(PC) + 2 + 2n → PC

Status Affected: None

Encoding:

1110	0010	nnnn	nnnn

Description: If the CARRY bit is '1', then the program will branch.
The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a 2-cycle instruction.

Words: 1

Cycles: 1(2)

Q Cycle Activity:
If Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation

Example:
```
HERE    BC   5
```
Before Instruction
PC = address (HERE)
After Instruction
If CARRY = 1;
PC = address (HERE + 12)
If CARRY = 0;
PC = address (HERE + 2)

BNOV **Branch if Not Overflow**

Syntax: BNOV n

Operands: -128 ≤ n ≤ 127

Operation: if OVERFLOW bit is '0'
(PC) + 2 + 2n → PC

Status Affected: None

Encoding:

1110	0101	nnnn	nnnn

Description: If the OVERFLOW bit is '0', then the program will branch.
The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a 2-cycle instruction.

Words: 1

Cycles: 1(2)

Q Cycle Activity:
If Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation

Example: `HERE BNOV Jump`

Before Instruction
PC = address `(HERE)`
After Instruction
If OVERFLOW = 0;
PC = address `(Jump)`
If OVERFLOW = 1;
PC = address `(HERE + 2)`

BNZ **Branch if Not Zero**

Syntax: BNZ n

Operands: -128 ≤ n ≤ 127

Operation: if ZERO bit is '0'
(PC) + 2 + 2n → PC

Status Affected: None

Encoding:

1110	0001	nnnn	nnnn

Description: If the ZERO bit is '0', then the program will branch.
The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a 2-cycle instruction.

Words: 1

Cycles: 1(2)

Q Cycle Activity:
If Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation

Example: `HERE BNZ Jump`

Before Instruction
PC = address `(HERE)`
After Instruction
If ZERO = 0;
PC = address `(Jump)`
If ZERO = 1;
PC = address `(HERE + 2)`

SUBFWB **Subtract f from W with borrow**

Syntax: SUBFWB f {,d {,a}}

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

Operation: (W) – (f) – ($\overline{C}$) → dest

Status Affected: N, OV, C, DC, Z

Encoding:

0101	01da	ffff	ffff

Description: Subtract register 'f' and CARRY flag (borrow) from W (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored in register 'f' (default).
If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See **Section 41.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example 1:
```
SUBFWB   REG, 1, 0
```
Before Instruction
REG = 3
W = 2
C = 1
After Instruction
REG = FF
W = 2
C = 0
Z = 0
N = 1 ; result is negative

Example 2:
```
SUBFWB   REG, 0, 0
```
Before Instruction
REG = 2
W = 5
C = 1
After Instruction
REG = 2
W = 3
C = 1
Z = 0
N = 0 ; result is positive

Example 3:
```
SUBFWB   REG, 1, 0
```
Before Instruction
REG = 1
W = 2
C = 0
After Instruction
REG = 0
W = 2
C = 1
Z = 1 ; result is zero
N = 0

SUBLW **Subtract W from literal**

Syntax: SUBLW k

Operands: $0 \le k \le 255$

Operation: k – (W) → W

Status Affected: N, OV, C, DC, Z

Encoding:

0000	1000	kkkk	kkkk

Description: W is subtracted from the 8-bit literal 'k'. The result is placed in W.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'	Process Data	Write to W

Example 1:
```
SUBLW   02h
```
Before Instruction
W = 01h
C = ?
After Instruction
W = 01h
C = 1 ; result is positive
Z = 0
N = 0

Example 2:
```
SUBLW   02h
```
Before Instruction
W = 02h
C = ?
After Instruction
W = 00h
C = 1 ; result is zero
Z = 1
N = 0

Example 3:
```
SUBLW   02h
```
Before Instruction
W = 03h
C = ?
After Instruction
W = FFh ; (2's complement)
C = 0 ; result is negative
Z = 0
N = 1

43.11 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM™ and dsPICDEM™ demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ® security ICs, CAN, IrDA®, PowerSmart battery management, SEEVAL® evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

43.12 Third-Party Development Tools

Microchip also offers a great collection of tools from third-party vendors. These tools are carefully selected to offer good value and unique functionality.

- Device Programmers and Gang Programmers from companies, such as SoftLog and CCS
- Software Tools from companies, such as Gimpel and Trace Systems
- Protocol Analyzers from companies, such as Saleae and Total Phase
- Demonstration Boards from companies, such as MikroElektronika, Digilent® and Olimex
- Embedded Ethernet Solutions from companies, such as EZ Web Lynx, WIZnet and IPLogika®

FIGURE 44-7: CLKOUT AND I/O TIMING

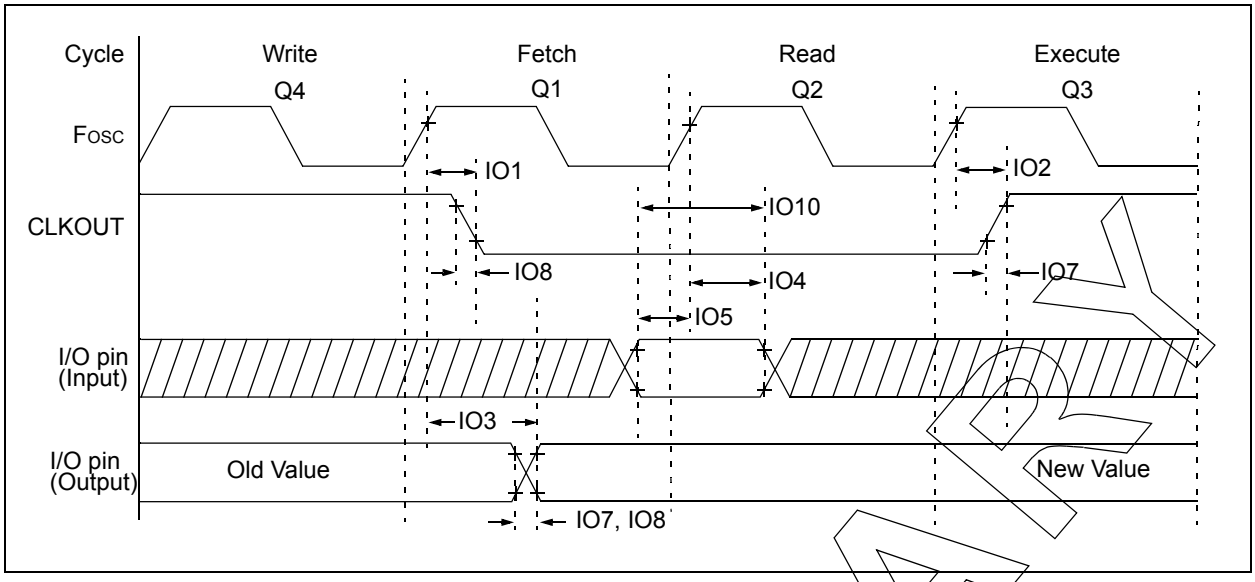


TABLE 44-12: I/O AND CLKOUT TIMING SPECIFICATIONS

Standard Operating Conditions (unless otherwise stated)							
Param No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
IO1*	$T_{CLKOUTH}$	CLKOUT rising edge delay (rising edge Fosc (Q1 cycle) to falling edge CLKOUT	—	—	70	ns	
IO2*	$T_{CLKOUTL}$	CLKOUT falling edge delay (rising edge Fosc (Q3 cycle) to rising edge CLKOUT	—	—	72	ns	
IO3*	T_{IO_VALID}	Port output valid time (rising edge Fosc (Q1 cycle) to port valid)	—	50	70	ns	
IO4*	T_{IO_SETUP}	Port input setup time (Setup time before rising edge Fosc – Q2 cycle)	20	—	—	ns	
IO5*	T_{IO_HOLD}	Port input hold time (Hold time after rising edge Fosc – Q2 cycle)	50	—	—	ns	
IO6*	T_{IOR_SLREN}	Port I/O rise time, slew rate enabled	—	25	—	ns	V_{DD} = 3.0V
IO7*	T_{IOR_SLRDIS}	Port I/O rise time, slew rate disabled	—	5	—	ns	V_{DD} = 3.0V
IO8*	T_{IOF_SLREN}	Port I/O fall time, slew rate enabled	—	25	—	ns	V_{DD} = 3.0V
IO9*	T_{IOF_SLRDIS}	Port I/O fall time, slew rate disabled	—	5	—	ns	V_{DD} = 3.0V
IO10*	T_{INT}	INT pin high or low time to trigger an interrupt	25	—	—	ns	
IO11*	T_{IOC}	Interrupt-on-Change minimum high or low time to trigger interrupt	25	—	—	ns	

*These parameters are characterized but not tested.

TABLE 44-24: SPI MODE REQUIREMENTS (SLAVE MODE)

Standard Operating Conditions (unless otherwise stated)							
Param No.	**Symbol**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
	TSCK	SCK Total Cycle Time	47	—	—	ns	Receive only mode
			—	20(1)	—	MHz	
			95	—	—	ns	Full duplex mode
			—	10(1)	—	MHz	
SP70*	TSSL2SCH, TSSL2SCL	$\overline{SS}$↓ to SCK↓ or SCK↑ input	0	—	—	ns	CKE = 0
			25	—	—	ns	CKE = 1
SP71*	TSCH	SCK input high time	20	—	—	ns	
SP72*	TSCL	SCK input low time	20	—	—	ns	
SP73*	TDIV2SCH, TDIV2SCL	Setup time of SDI data input to SCK edge	10	—	—	ns	
SP74*	TSCH2DIL, TSCL2DIL	Hold time of SDI data input to SCK edge	0	—	—	ns	
SP75*	TDOR	SDO data output rise time	—	10	25	ns	CL = 50 pF
SP76*	TDOF	SDO data output fall time	—	10	25	ns	CL = 50 pF
SP77*	TSSH2DOZ	$\overline{SS}$↑ to SDO output high-impedance	—	—	85	ns	
SP80*	TSCH2DOV, TSCL2DOV	SDO data output valid after SCK edge	—	—	85	ns	
SP82*	TSSL2DOV	SDO data output valid after $\overline{SS}$↓ edge	—	—	85	ns	
SP83*	TSCH2SSH, TSCL2SSH	$\overline{SS}$ ↑ after SCK edge	20	—	—	ns	
SP84*	TSSH2SSL	SS↑ to SS↓ edge	47	—	—	ns	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: SPIxCON1.SMP bit must be set and the slew rate control must be disabled on the clock and data pins (clear the corresponding bits in SLRCONx register) for SPI to operate over 4 MHz.