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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	64MHz
Connectivity	I ² C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, DMA, HLVD, POR, PWM, WDT
Number of I/O	36
Program Memory Size	128KB (64K x 16)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	8K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 35x12b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	40-UQFN Exposed Pad
Supplier Device Package	40-UQFN (5x5)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf47k42-i-mv

3.1 System Arbitration

The System Arbiter resolves memory access between the System Level Selections (i.e., Main, Interrupt Service Routine) and Peripheral Selection (i.e., DMA and Scanner) based on user-assigned priorities. Each of the system level and peripheral selections has its own priority selection registers. Memory access priority is resolved using the number written to the corresponding Priority registers, 0 being the highest priority and 4 the lowest. The default priorities are listed in [Table 3-1](#).

In case the user wants to change priorities, ensure each Priority register is written with a unique value from 0 to 4.

TABLE 3-1: DEFAULT PRIORITIES

Selection		Priority register Reset value
System Level	ISR	0
	MAIN	1
Peripheral	DMA1	2
	DMA2	3
	SCANNER	4

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TABLE 4-1: PROGRAM AND DATA EEPROM MEMORY MAP

PIC18(L)F45/55K42		PIC18(L)F26/46/56K42		PIC18(L)F27/47/57K42		
PC<21:0>		PC<21:0>		PC<21:0>		
↕		↕		↕		
Stack (31 levels)		Stack (31 levels)		Stack (31 levels)		
↓		↓		↓		
00 0000h	Reset Vector	Reset Vector	Reset Vector	00 0000h		
...		
00 0008h	Interrupt Vector High ⁽²⁾	Interrupt Vector High ⁽²⁾	Interrupt Vector High ⁽²⁾	00 0008h		
...		
00 0018h	Interrupt Vector Low ⁽²⁾	Interrupt Vector Low ⁽²⁾	Interrupt Vector Low ⁽²⁾	00 0018h		
00 001Ah	Program Flash Memory (16 KW) ⁽³⁾	Program Flash Memory (32 KW) ⁽³⁾	Program Flash Memory (64 KW) ⁽³⁾	00 001Ah		
00 7FFFh				00 7FFFh		
00 8000h	Reserved ⁽⁴⁾			Reserved ⁽⁴⁾	00 8000h	
00 FFFFh					00 FFFFh	
01 0000h			01 0000h			
01 FFFFh			01 FFFFh			
02 0000h			02 0000h			
1F FFFFh			1F FFFFh			
20 0000	User IDs (8 Words) ⁽⁵⁾			20 0000h		
...				...		
20 000Fh				20 000Fh		
20 0010h	Reserved			20 0010h		
...				...		
2F FFFFh				2F FFFFh		
30 0000h	Configuration Words (5 Words) ⁽⁵⁾			30 0000h		
...				...		
30 0009h				30 0009h		
30 000Ah	Reserved			30 000Ah		
...				...		
30 FFFFh				30 FFFFh		
31 0000h	Data EEPROM (256 Bytes)	Data EEPROM (1024Bytes)		31 0000h		
...				...		
31 00FFh				31 00FFh		
31 0100h	Reserved	Reserved		31 0100h		
...				...		
31 03FFh			31 03FFh			
31 0400h			31 0400h			
...			...			
3E FFFFh			3E FFFFh			
3F 0000h	Device Information Area ^{(5),(7)}			3F 0000h		
...				...		
3F 003Fh				3F 003Fh		
3F0040h	Reserved			3F0040h		
...				...		
3F FEFFh				3F FEFFh		
3F FF00h	Device Configuration Information (5 Words) ^{(5),(6),(7)}			3F FF00h		
...				...		
3F FF09h				3F FF09h		
3F FF0Ah	Reserved			3F FF0Ah		
...				...		
3F FFFBh				3F FFFBh		
3F FFFCh	Revision ID (1 Word) ^{(5),(6),(7)}			3F FFFCh		
...				...		
3F FFFDh				3F FFFDh		
3F FFFEh	Device ID (1 Word) ^{(5),(6),(7)}			3F FFFEh		
...				...		
3F FFFFh				3F FFFFh		

Note

1: The stack is a separate SRAM panel, apart from all user memory panels.

2: 00 0008h location is used as the reset default for the IVTBASE register, the vector table can be relocated in the memory by programming the IVTBASE register.

3: Storage area Flash is implemented as the last 128 Words of user Flash.

4: The addresses do not roll over. The region is read as '0'.

5: Not code-protected.

6: Hard-coded in silicon.

7: This region cannot be written by the user and it's not affected by a Bulk Erase.

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REGISTER 7-5: OSCFRQ: HFINTOSC FREQUENCY SELECTION REGISTER

U-0	U-0	U-0	U-0	R/W-q/q	R/W-q/q	R/W-q/q	R/W-q/q
—	—	—	—	FRQ<3:0>			
bit 7				bit 0			

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

q = Reset value is determined by hardware

bit 7-4

Unimplemented: Read as '0'

bit 3-0

FRQ<3:0>: HFINTOSC Frequency Selection bits⁽¹⁾

FRQ<3:0>	Nominal Freq (MHz)
1001	Reserved
1010	
1111	
1110	
1101	
1100	
1011	
1000	64
0111	48
0110	32
0101	16
0100	12
0011	8
0010	4
0001	2
0000	1

Note 1: Refer to [Table 7-2](#) for more information.

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FIGURE 15-3: DMA COUNTERS BLOCK DIAGRAM

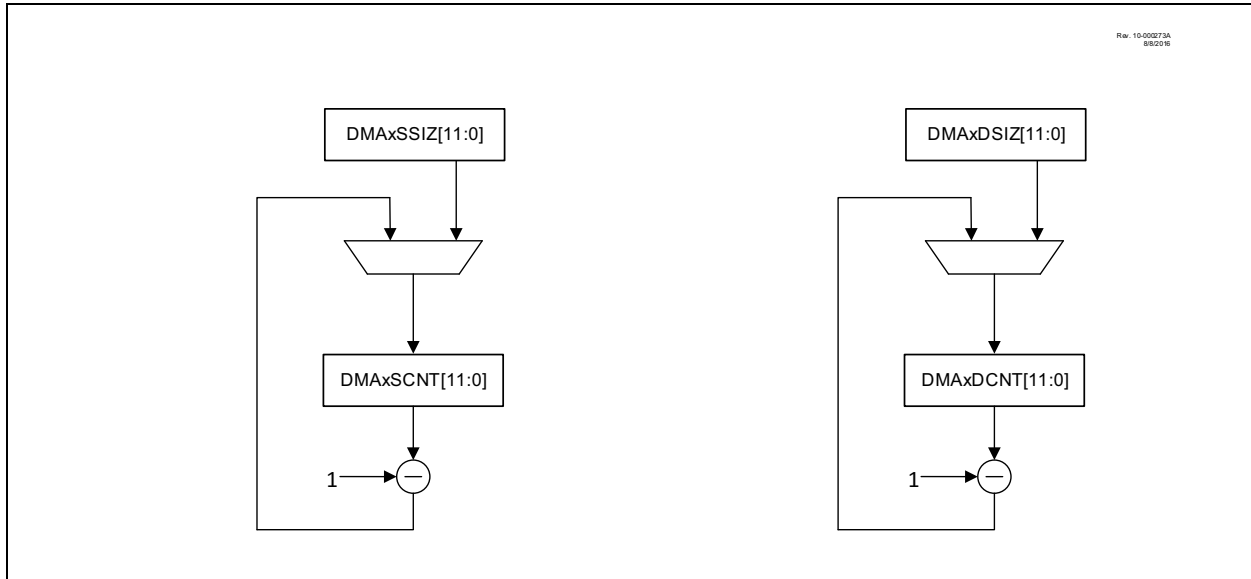


Table 15-2 has a few examples of configuring DMA Message sizes.

TABLE 15-2: EXAMPLE MESSAGE SIZE TABLE

Operation	Example	SCNT	DCNT	Comments
Read from single SFR location to RAM	U1RXB	1	N	N equals the number of bytes desired in the destination buffer. $N \geq 1$.
Write to single SFR location from RAM	U1TXB	N	1	N equals the number of bytes desired in the source buffer. $N \geq 1$.
Read from multiple SFR location	ADRES[H:L]	2	$2*N$	N equals the number of ADC results to be stored in memory. $N \geq 1$
	TMR1[H:L]	2	$2*N$	N equals the number of TMR1 Acquisition results to be stored in memory. $N \geq 1$
	SMT1CPR[U:H:L]	3	$3*N$	N equals the number of Capture Pulse Width measurements to be stored in memory. $N \geq 1$
Write to Multiple SFR registers	PWMDC[H:L]	$2*N$	2	N equals the number of PWM duty cycle values to be loaded from a memory table. $N \geq 1$
	All ADC registers	$N*31$	31	Using the DMA to transfer a complete ADC context from RAM to the ADC registers. $N \geq 1$

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17.8 Register Definitions: PPS Input Selection

REGISTER 17-1: xxxPPS: PERIPHERAL xxx INPUT SELECTION

U-0	U-0	R/W-m/u ^(1,3)	R/W-m/u ⁽¹⁾	R/W-m/u ⁽¹⁾	R/W-m/u ⁽¹⁾	R/W-m/u ⁽¹⁾	R/W-m/u ⁽¹⁾
—	—	xxxPPS<5:0>					
bit 7		bit 0					

Legend:

R = Readable bit	W = Writable bit	-n/n = Value at POR and BOR/Value at all other Resets
u = Bit is unchanged	x = Bit is unknown	q = value depends on peripheral
'1' = Bit is set	U = Unimplemented bit, read as '0'	m = value depends on default location for that input
'0' = Bit is cleared		

bit 7-6 **Unimplemented:** Read as '0'

bit 5-3 **xxxPPS<5:3>:** Peripheral xxx Input PORTx Pin Selection bits
See [Table 17-1](#) for the list of available ports and default pin locations.
101 = PORTF⁽²⁾
100 = PORTE⁽³⁾
011 = PORTD⁽³⁾
010 = PORTC
001 = PORTB
000 = PORTA

bit 2-0 **xxxPPS<2:0>:** Peripheral xxx Input PORTx Pin Selection bits
111 = Peripheral input is from PORTx Pin 7 (Rx7)
110 = Peripheral input is from PORTx Pin 6 (Rx6)
101 = Peripheral input is from PORTx Pin 5 (Rx5)
100 = Peripheral input is from PORTx Pin 4 (Rx4)
011 = Peripheral input is from PORTx Pin 3 (Rx3)
010 = Peripheral input is from PORTx Pin 2 (Rx2)
001 = Peripheral input is from PORTx Pin 1 (Rx1)
000 = Peripheral input is from PORTx Pin 0 (Rx0)

- Note 1:** The Reset value 'm' of this register is determined by device default locations for that input.
2: Reserved on PIC18LF26/27/45/46/57K42 parts.
3: Reserved on PIC18LF26/27K42 parts.

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REGISTER 19-3: PMD2: PMD CONTROL REGISTER 2

U-0	R/W-0/0	R/W-0/0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0
—	DACMD	ADCMD	—	—	CMP2MD	CMP1MD	ZCDMD ⁽¹⁾
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

q = Value depends on condition

- bit 7 **Unimplemented:** Read as '0'
- bit 6 **DACMD:** Disable DAC bit
 1 = DAC module disabled
 0 = DAC module enabled
- bit 5 **ADCMD:** Disable ADCC bit
 1 = ADCC module disabled
 0 = ADCC module enabled
- bit 4-3 **Unimplemented:** Read as '0'
- bit 2 **CMP2MD:** Disable Comparator CMP2 bit
 1 = CMP2 module disabled
 0 = CMP2 module enabled
- bit 1 **CMP1MD:** Disable Comparator CMP1 bit
 1 = CMP1 module disabled
 0 = CMP1 module enabled
- bit 0 **ZCDMD:** Disable Zero-Cross Detect module bit⁽¹⁾
 1 = ZCD module disabled
 0 = ZCD module enabled

Note 1: Subject to $\overline{\text{ZCD}}$ bit in CONFIG2H.

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TABLE 21-3: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER1/3/5 AS A TIMER/COUNTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
TxCON	—	—	CKPS<1:0>		—	SYNC	RD16	ON	313
TxGCON	GE	GPOL	GTM	GSPM	GO/DONE	GVAL	—	—	314
TxCLK	—	—	—	CS<4:0>					315
TxGATE	—	—	—	GSS<4:0>					316
TMRxL	Least Significant Byte of the 16-bit TMR3 Register								317
TMRxH	Holding Register for the Most Significant Byte of the 16-bit TMR3 Register								317

Legend: — = Unimplemented location, read as '0'. Shaded cells are not used by TIMER1/3/5.

25.6.3 PERIOD AND DUTY CYCLE MODE

In Duty Cycle mode, either the duty cycle or period (depending on polarity) of the SMT1_signal can be acquired relative to the SMT clock. The CPW register is updated on a falling edge of the signal, and the CPR register is updated on a rising edge of the signal, along with the SMT1TMR resetting to 0x0001. In addition, the GO bit is reset on a rising edge when the SMT is in Single Acquisition mode. See [Figure 25-6](#) and [Figure 25-7](#).

26.3 Clock Source

The clock source is used to drive the dead-band timing circuits. The CWG module allows the following clock sources to be selected:

- FOSC (system clock)
- HFINTOSC

When the HFINTOSC is selected, the HFINTOSC will be kept running during Sleep. Therefore, CWG modes requiring dead band can operate in Sleep, provided that the CWG data input is also active during Sleep. The clock sources are selected using the CS bit of the CWGxCLKCON register (Register 26-3). The system clock FOSC, is disabled in Sleep and thus dead-band control cannot be used.

26.4 Selectable Input Sources

The CWG generates the output waveforms from the following input sources:

TABLE 26-1: SELECTABLE INPUT SOURCES

Source Peripheral	Signal Name	ISM<2:0>
CWGxPPS	Pin selected by CWGxPPS	000
CCP1	CCP1 Output	001
CCP2	CCP2 Output	010
PWM3	PWM3 Output	011
PWM4	PWM4 Output	100
CMP1	Comparator 1 Output	101
CMP2	Comparator 2 Output	110
DSM	Data signal modulator output	111

The input sources are selected using the IS<4:0> bits in the CWGxISM register (Register 26-4).

26.5 Output Control

26.5.1 CWG OUTPUTS

Each CWG output can be routed to a Peripheral Pin Select (PPS) output via the RxyPPS register (see Section 17.0 “Peripheral Pin Select (PPS) Module”).

26.5.2 POLARITY CONTROL

The polarity of each CWG output can be selected independently. When the output polarity bit is set, the corresponding output is active-high. Clearing the output polarity bit configures the corresponding output as active-low. However, polarity does not affect the override levels. Output polarity is selected with the POLY bits of the CWGxCON1. Auto-shutdown and steering options are unaffected by polarity.

26.6 Dead-Band Control

The dead-band control provides non-overlapping PWM signals to prevent shoot-through current in PWM switches. Dead-band operation is employed for Half-Bridge and Full-Bridge modes. The CWG contains two 6-bit dead-band counters. One is used for the rising edge of the input source control in Half-Bridge mode or for reverse dead-band Full-Bridge mode. The other is used for the falling edge of the input source control in Half-Bridge mode or for forward dead band in Full-Bridge mode.

Dead band is timed by counting CWG clock periods from zero up to the value in the rising or falling dead-band counter registers. See CWGxDBR and CWGxDBF registers, respectively.

26.6.1 DEAD-BAND FUNCTIONALITY IN HALF-BRIDGE MODE

In Half-Bridge mode, the dead-band counters dictate the delay between the falling edge of the normal output and the rising edge of the inverted output. This can be seen in Figure 26-2.

26.6.2 DEAD-BAND FUNCTIONALITY IN FULL-BRIDGE MODE

In Full-Bridge mode, the dead-band counters are used when undergoing a direction change. The MODE<0> bit of the CWGxCON0 register can be set or cleared while the CWG is running, allowing for changes from Forward to Reverse mode. The CWGxA and CWGxC signals will change immediately upon the first rising input edge following a direction change, but the modulated signals (CWGxB or CWGxD, depending on the direction of the change) will experience a delay dictated by the dead-band counters.

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REGISTER 26-5: CWGxSTR⁽¹⁾: CWG STEERING CONTROL REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
OVRD	OVRC	OVRB	OVRA	STRD ⁽²⁾	STRC ⁽²⁾	STRB ⁽²⁾	STRA ⁽²⁾
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

q = Value depends on condition

bit 7 **OVRD**: Steering Data D bit

bit 6 **OVRC**: Steering Data C bit

bit 5 **OVRB**: Steering Data B bit

bit 4 **OVRA**: Steering Data A bit

bit 3 **STRD**: Steering Enable bit D⁽²⁾

1 = CWGxD output has the CWG data input waveform with polarity control from POLD bit

0 = CWGxD output is assigned to value of OVRD bit

bit 2 **STRC**: Steering Enable bit C⁽²⁾

1 = CWGxC output has the CWG data input waveform with polarity control from POLC bit

0 = CWGxC output is assigned to value of OVRC bit

bit 1 **STRB**: Steering Enable bit B⁽²⁾

1 = CWGxB output has the CWG data input waveform with polarity control from POLB bit

0 = CWGxB output is assigned to value of OVRB bit

bit 0 **STRA**: Steering Enable bit A⁽²⁾

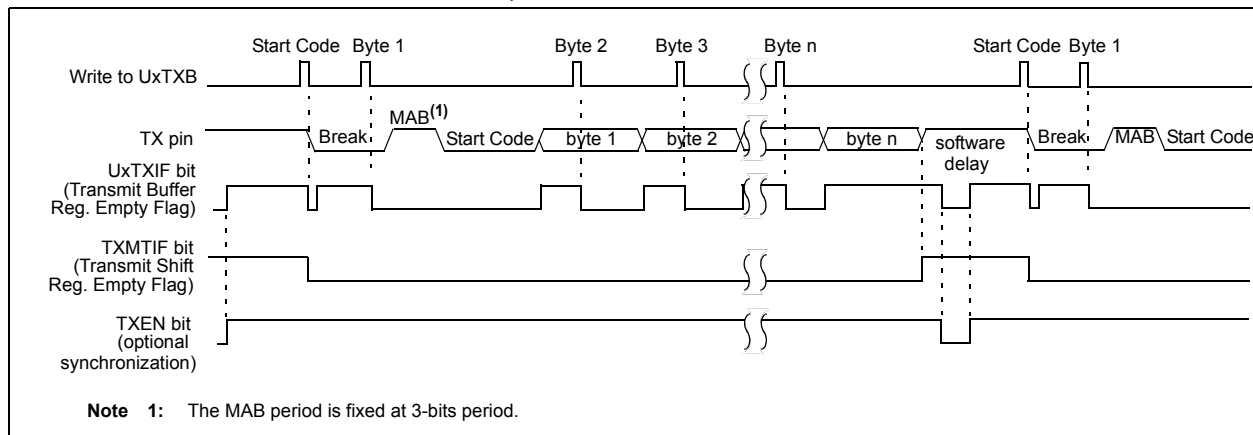
1 = CWGxA output has the CWG data input waveform with polarity control from POLA bit

0 = CWGxA output is assigned to value of OVRA bit

Note 1: The bits in this register apply only when MODE<2:0> = 00x ([Register 26-1](#), Steering modes).

2: This bit is double-buffered when MODE<2:0> = 001.

FIGURE 31-6: DMX TRANSMIT SEQUENCE



31.5 LIN Modes (UART1 only)

LIN is a protocol used primarily in automotive applications. The LIN network consists of two kinds of software processes: a Master process and a Slave process. Each network has only one Master process and one or more Slave processes.

From a physical layer point of view, the UART on one processor may be driven by both a Master and a Slave process, as long as only one Master process exists on the network.

A LIN transaction consists of a Master process followed by a Slave process. The Slave process may involve more than one Slave where one is transmitting and the other(s) are receiving. The transaction begins by the following Master process transmission sequence:

1. Break
2. Delimiter bit
3. Sync Field
4. PID byte

The PID determines which Slave processes are expected to respond to the Master. When the PID byte is complete, the TX output remains in the Idle state. One or more of the Slave processes may respond to the Master process. If no one responds within the inter-byte period, the Master is free to start another transmission. The inter-byte period is timed by software using a means other than the UART.

The Slave process follows the Master process. When the Slave software recognizes the PID then that Slave process responds by either transmitting the required response or by receiving the transmitted data. Only Slave processes send data. Therefore, Slave processes receiving data are receiving that of another Slave process.

When a Slave sends data, the Slave UART automatically calculates the checksum for the transmitted bytes as they are sent and appends the inverted checksum byte to the slave response.

When a Slave receives data, the checksum is accumulated on each byte as it is received using the same algorithm as the sending process. The last byte, which is the inverted checksum value calculated by the sending process, is added to the locally calculated checksum by the UART. The check passes when the result is all '1's, otherwise the check fails and the CERIF bit is set.

Two methods for computing the checksum are available: legacy and enhanced. The legacy checksum includes only the data bytes. The enhanced checksum includes the PID and the data. The C0EN control bit in the UxCON2 register determines the checksum method. Setting C0EN to '1' selects the enhanced method. Software must select the appropriate method before the Start bit of the checksum byte is received.

31.5.1 LIN MASTER/SLAVE MODE

The LIN Master mode includes capabilities to generate Slave processes. The Master process stops at the PID transmission. Any data that is transmitted in Master/Slave mode is done as a Slave process. LIN Master/Slave mode is configured by the following settings:

- MODE<3:0> = 1100
- TXEN = 1
- RXEN = 1
- UxBRGH:L = Value to achieve desired baud rate
- TXPOL = 0 (for high Idle state)
- STP = desired Stop bits selection
- C0EN = desired checksum mode
- RxyPPS = TX pin selection code
- TX pin TRIS control = 0
- ON = 1

Note: The TXEN bit must be set before the Master process is received and remain set while in LIN mode whether or not the slave process is a transmitter.

31.9 Stop Bits

The number of Stop bits is user selectable with the STP bits in the UxCON2 register. The STP bits affect all modes of operation.

Stop bits selections include:

- 1 transmit with receive verify on first
- 1.5 transmit with receive verify on first
- 2 transmit with receive verify on both
- 2 transmit with receive verify on first only

In all modes, except DALI, the transmitter is idle for the number of Stop bit periods between each consecutively transmitted word. In DALI, the Stop bits are generated after the last bit in the transmitted data stream.

The input is checked for the idle level in the middle of the first Stop bit, when receive verify on first is selected, as well as in the middle of the second Stop bit, when verify on both is selected. If any Stop bit verification indicates a non-idle level, the framing error FERIF bit is set for the received word.

31.9.1 DELAYED UXRIF

When operating in Half-Duplex mode, where the microcontroller needs to reverse the transceiver direction after a reception, it may be more convenient to hold off the UXRIF interrupt until the end of the Stop bits to avoid line contention. The user selects when the UXRIF interrupt occurs with the STPMD bit in the UxFIFO register. When STPMD is '1', the UXRIF occurs at the end of the last Stop bit. When STPMD is '0', UXRIF occurs when the received byte is stored in the receive FIFO. When $STP<1:0> = 10$, the store operation is performed in the middle of the second Stop bit, otherwise, it is performed in the middle of the first Stop bit. The FERIF and PERIF interrupts are not delayed with STPMD. Only UXRIF is delayed when STPMD is set and should be the only indicator for reversing transceiver direction.

31.10 Operation after FIFO overflow

The Receive Shift Register (RSR) can be configured to stop or continue running during a receive FIFO overflow condition. Stopped operation is the Legacy mode.

When the RSR continues to run during an overflow condition, the first word received after clearing the overflow will always be valid.

When the RSR is stopped during an overflow condition, synchronization with the Start bits is lost. Therefore, the first word received after the overflow is cleared may start in the middle of a word.

Operation during overflow is selected with the RUNOVF bit in the UxCON2 register. Setting the RUNOVF bit selects the run during overflow method.

31.11 Receive and Transmit Buffers

The UART uses small buffer areas to transmit and receive data. These are sometimes referred to as FIFOs.

The receiver has a Receive Shift Register (RSR) and two buffer registers. The buffer at the top of the FIFO (earliest byte to enter the FIFO) is retrieved by reading the UxRXB register.

The transmitter has one Transmit Shift Register (TSR) and one buffer register. Writes to UxTXB go to the transmit buffer then immediately to the TSR, if it is empty. When the TSR is not empty, writes to UxTXB are held then transferred to the TSR when it becomes available.

31.11.1 FIFO STATUS

The UxFIFO register contains several status bits for determining the state of the receive and transmit buffers.

The RXBE bit indicates that the receive FIFO is empty. This bit is essentially the inverse of UxRXIF. The RXBF bit indicates that the receive FIFO is full.

The transmitter has only one buffer register so the status bits are essentially a copy and inverse of the UxTXIF bit. The TXBE bit indicates that the buffer is empty (same as UxTXIF) and the TXBF bit indicates that the buffer is full (UxTXIF inverse). A third transmitter status bit, TXWRE (transmit write error), is set whenever a UxTXB write is performed when the TXBF bit is set. This indicates that the write was unsuccessful.

31.11.2 FIFO RESET

All modes support resetting the receive and transmit buffers.

The receive buffer is flushed and all unread data discarded when the RXBE bit in the UxFIFO register is written to '1'. The MOVWF instruction with the TXBE bit cleared should be used to avoid inadvertently clearing a byte pending in the TSR when UxTXB is empty.

Data written to UxTXB when TXEN is low will be held in the Transmit Shift Register (TSR) then sent when TXEN is set. The transmit buffer and inactive TSR are flushed by setting the TXBE bit in the UxFIFO register. Setting TXBE while a character is actively transmitting from the TSR will complete the transmission without being flushed.

Clearing the ON bit will discard all received data and transmit data pending in the TSR and UxTXB.

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REGISTER 31-8: UxBRGL: UART BAUD RATE GENERATOR LOW REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
BRG<7:0>							
bit 7				bit 0			

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

bit 7-0 **BRG<7:0>**: Least Significant Byte of Baud Rate Generator

REGISTER 31-9: UxBRGH: UART BAUD RATE GENERATOR HIGH REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
BRG<15:8>							
bit 7				bit 0			

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

bit 7-0 **BRG<15:8>**: Most Significant Byte of Baud Rate Generator

Note 1: The UxBRG registers should only be written when ON = 0.

2: Maximum BRG value when MODE = '100x' and BRGS = 1 is 0x7FFE.

3: Maximum BRG value when MODE = '100x' and BRGS = 0 is 0x1FFE.

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REGISTER 32-9: SPIxCON2: SPI CONFIGURATION REGISTER 2

R-0/0	R-0/0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0
BUSY	SSFLT	—	—	—	SSET	TXR ⁽¹⁾	RXR ⁽¹⁾
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

bit 7 **BUSY:** SPI Module Busy Status bit

1 = Data exchange is busy

0 = Data exchange is not taking place

bit 6 **SSFLT:** SS(in) Fault Status bit

If SSET = 0

1 = SS(in) ended the transaction unexpectedly, and the data byte being received was lost

0 = SS(in) ended normally

If SSET = 1

This bit is unchanged.

bit 5-3 **Unimplemented:** Read as '0'

bit 2 **SSET:** Slave Select Enable bit

Master mode:

1 = SS(out) is driven to the active state continuously

0 = SS(out) is driven to the active state while the transmit counter is not zero

Slave mode:

1 = SS(in) is ignored and data is clocked on all SCK(in) (as though SS = TRUE at all times)

0 = SS(in) enables/disables data input and tri-states SDO if the TRIS bit associated with the SDO pin is set (see [Table 32-2](#) for details)

bit 1 **TXR:** Transmit Data-Required Control bit⁽¹⁾

1 = TxFIFO data is required for a transfer

0 = TxFIFO data is not required for a transfer

bit 0 **RXR:** Receive FIFO Space-Required Control bit⁽¹⁾

1 = Data transfers are suspended if the RxFIFO is full

0 = Received data is not stored in the FIFO

Note 1: See [Table 32-1](#) as well as [Section 32.5 “Master mode”](#) and [Section 32.6 “Slave Mode”](#) for more details pertaining to TXR and RXR function.

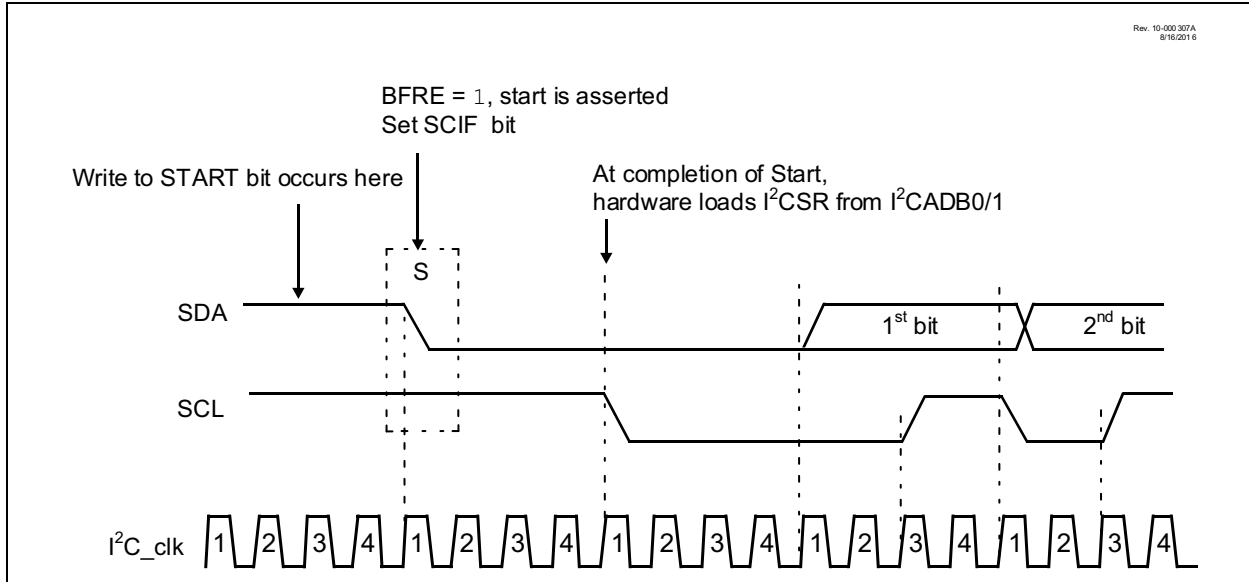
2: This register should not be written to while a transfer is in progress (BUSY bit of SPIxCON2 is set).

33.5.5 I²C MASTER MODE START CONDITION TIMING

The user can initiate a Start condition by either writing to the Start bit (S) of the I2CxCON0 register or by writing to the I2CxTXB register based on the ABD bit setting. Master hardware waits for BFRE = 1, before

asserting the Start condition. The action of the SDA being driven low while SCL is high is the Start condition, causing the SCIF bit to be set. One Tscl later the SCL is asserted low, ending the start sequence. [Figure 33-15](#) shows the Start condition timing.

FIGURE 33-15: START CONDITION TIMING



33.5.6 I²C MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition occurs when the Start bit of the I2CxCON0 register is set and the master module is waiting from a Restart clock stretch event (RSEN = 1 and I2CxCNT = 0).

When the Start bit is set, the SDA pin is released high for Tscl/2. Then the SCL pin is released floated high) for Tscl/2. If the SDA pin is detected low, bus collision flag (BCLIF) is set and the master goes idle. If SDA is detected high, the SDA pin will be pulled low (Start condition) for Tscl. Last, SCL is asserted low and I2CxADB0/1 is loaded into the shift register. As soon as a Restart condition is detected on the SDA and SCL pins, the RSCIF bit is set. [Figure 33-16](#) shows the timings for repeated Start Condition.

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REGISTER 33-4: I2CxCLK: I²C CLOCK SELECTION REGISTER

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	—	CLK<3:0>			
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

HS = Hardware set

HC = Hardware clear

bit 7-4

Unimplemented: Read as '0'

bit 3-0

CLK<3:0>: I²C Clock Selection Bits

CLK<3:0>	I ² Cx Clock Selection
1010-1111	Reserved
1001	SMT1 overflow
1000	TMR6 post scaled output
0111	TMR4 post scaled output
0110	TMR2 post scaled output
0101	TMR0 overflow
0100	Clock Reference output
0011	MFINTOSC (500 kHz)
0010	HFINTOSC
0001	Fosc
0000	Fosc/4

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REGISTER 33-5: I2Cx BTO: I²C BUS TIMEOUT SELECTION REGISTER

U-0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	—	—	BTO<2:0>		
bit 7						bit 0	

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

HS = Hardware set

HC = Hardware clear

bit 7-3

Unimplemented: Read as '0'

bit 2-0

BTO<2:0>: I²C Bus Timeout Selection bits

BTO<2:0>	I ² Cx Bus Timeout Selection
111	CLC4OUT
110	CLC3OUT
101	CLC2OUT
100	CLC1OUT
011	TMR6 post scaled output
010	TMR4 post scaled output
001	TMR2 post scaled output
000	Reserved

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36.1 ADC Configuration

When configuring and using the ADC the following functions must be considered:

- Port configuration
- Channel selection
- ADC voltage reference selection
- ADC conversion clock source
- Interrupt control
- Result formatting
- Conversion Trigger Selection
- ADC Acquisition Time
- ADC Precharge Time
- Additional Sample and Hold Capacitor
- Single/Double Sample Conversion
- Guard Ring Outputs

36.1.1 PORT CONFIGURATION

The ADC can be used to convert both analog and digital signals. When converting analog signals, the I/O pin should be configured for analog by setting the associated TRIS and ANSEL bits. Refer to [Section 16.0 “I/O Ports”](#) for more information.

Note: Analog voltages on any pin that is defined as a digital input may cause the input buffer to conduct excess current.

36.1.2 CHANNEL SELECTION

There are several channel selections available:

- Eight PORTA pins (RA<7:0>)
- Eight PORTB pins (RB<7:0>)
- Eight PORTC pins (RC<7:0>)
- Eight PORTD pins (RD<7:0>, PIC18(L)F45/46/47/55/56/57K42 only)
- Three PORTE pins (RE<2:0>, PIC18(L)F45/46/47/55/56/57K42 only)
- Eight PORTF pins (RD<7:0>, PIC18(L)F55/56/57K42 only)
- Temperature Indicator
- DAC output
- Fixed Voltage Reference (FVR)
- Vss (ground)

The ADPCH register determines which channel is connected to the sample and hold circuit.

When changing channels, a delay is required before starting the next conversion.

Refer to [Section 36.2 “ADC Operation”](#) for more information.

36.1.3 ADC VOLTAGE REFERENCE

The PREF<1:0> bits of the ADREF register provide control of the positive voltage reference. The positive voltage reference can be:

- VREF+ pin
- VDD
- FVR outputs

The NREF bit of the ADREF register provides control of the negative voltage reference. The negative voltage reference can be:

- VREF- pin
- VSS

See [Section 34.0 “Fixed Voltage Reference \(FVR\)”](#) for more details on the Fixed Voltage Reference.

36.1.4 CONVERSION CLOCK

The source of the conversion clock is software selectable via the ADCLK register and the CS bits of the ADCON0 register. If Fosc is selected as the ADC clock, there is a prescaler available to divide the clock so that it meets the ADC clock period specification. The ADC clock source options are the following:

- Fosc/(2*n)(where n is from 1 to 128)
- FRC (dedicated RC oscillator)

The time to complete one bit conversion is defined as TAD. Refer [Figure 36-2](#) for the complete timing details of the ADC conversion.

For correct conversion, the appropriate TAD specification must be met. Refer to [Table 44-16](#) for more information. [Table 36-1](#) gives examples of appropriate ADC clock selections.

Note 1: Unless using the FRC, any changes in the system clock frequency will change the ADC clock frequency, which may adversely affect the ADC result.

2: The internal control logic of the ADC runs off of the clock selected by the CS bit of ADCON0. What this can mean is when the CS bit of ADCON0 is set to '1' (ADC runs on FRC), there may be unexpected delays in operation when setting ADC control bits.

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TABLE 44-23: SPI MODE REQUIREMENTS (MASTER MODE)

Standard Operating Conditions (unless otherwise stated)							
Param No.	Symbol	Characteristic	Min.	Typ†	Max.	Units	Conditions
	T _{SCK}	SCK Cycle Time (2x Prescaled)	61	—	—	ns	Transmit only mode
			—	16 ^(†)	—	MHz	
			95	—	—	ns	Full duplex mode
			—	10 ^(†)	—	MHz	
SP70*	T _{ssL2scH} , T _{ssL2scL}	SDO to SCK↓ or SCK↑ input	T _{SCK}	—	—	ns	FST = 0
			0	—	—	ns	FST = 1
SP71*	T _{scH}	SCK output high time	0.5 T _{SCK} - 12	—	0.5 T _{SCK} + 12	ns	
SP72*	T _{scL}	SCK output low time	0.5 T _{SCK} - 12	—	0.5 T _{SCK} + 12	ns	
SP73*	T _{dIV2scH} , T _{dIV2scL}	Setup time of SDI data input to SCK edge	85	—	—	ns	
SP74*	T _{sch2dIL} , T _{scL2dIL}	Hold time of SDI data input to SCK edge	0	—	—	ns	
		Hold time of SDI data input to final SCK	0.5 T _{SCK}	—	—	ns	CKE = 0, SMP = 1
SP75*	T _{doR}	SDO data output rise time	—	10	25	ns	CL = 50 pF
SP76*	T _{doF}	SDO data output fall time	—	10	25	ns	CL = 50 pF
SP78*	T _{scR}	SCK output rise time	—	10	25	ns	CL = 50 pF
SP79*	T _{scF}	SCK output fall time	—	10	25	ns	CL = 50 pF
SP80*	T _{sch2doV} , T _{scL2doV}	SDO data output valid after SCK edge	- 15	—	15	ns	CL = 20 pF
SP81*	T _{doV2scH} , T _{doV2scL}	SDO data output valid to first SCK edge	T _{SCK} - 10	—	—	ns	CL = 20 pF CKE = 1
SP82*	T _{ssL2doV}	SDO data output valid after SS↓ edge	—	—	50	ns	CL = 20 pF
SP83*	T _{sch2ssH} , T _{scL2ssH}	SS↑ after last SCK edge	0.5 T _{SCK} - 10	—	—	ns	
SP84*	T _{ssH2ssL}	SS↑ to SS↓ edge	0.5 T _{SCK} - 10	—	—	ns	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: SPIxCON1.SMP bit must be set and the slew rate control must be disabled on the clock and data pins (clear the corresponding bits in SLRCONx register) for SPI to operate over 4 MHz.

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Note: Unless otherwise noted, $V_{IN} = 5V$, $F_{OSC} = 300\text{ kHz}$, $C_{IN} = 0.1\text{ }\mu F$, $T_A = 25^\circ C$.

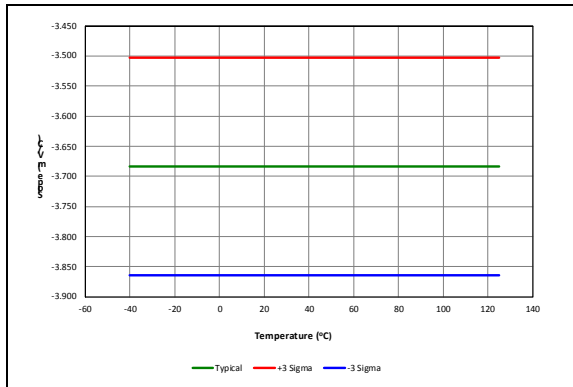


FIGURE 45-1: High Range Temperature Indicator Voltage Sensitivity Across Temperature

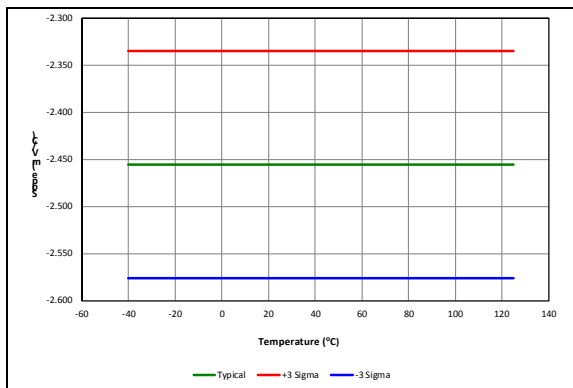


FIGURE 45-2: Low Range Temperature Indicator Voltage Sensitivity Across Temperature

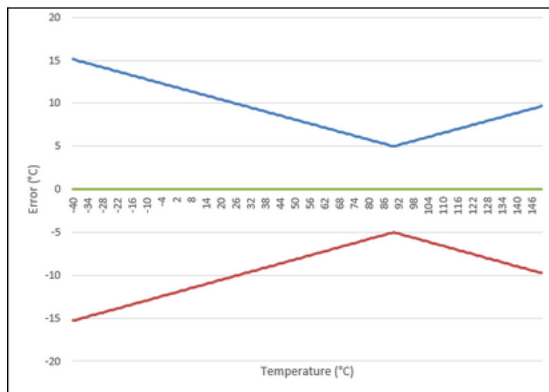


FIGURE 45-3: Temperature Indicator Performance Over Temperature