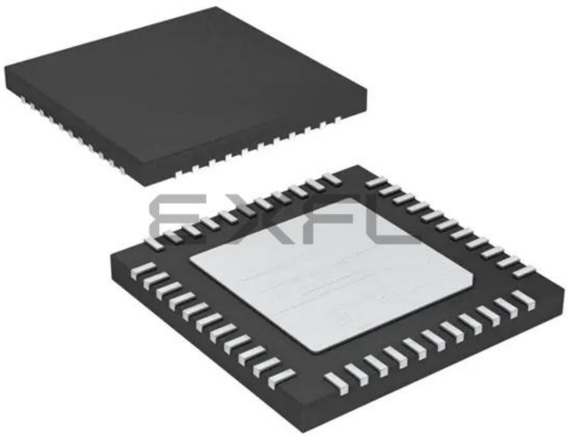


What is "Embedded - Microcontrollers"?

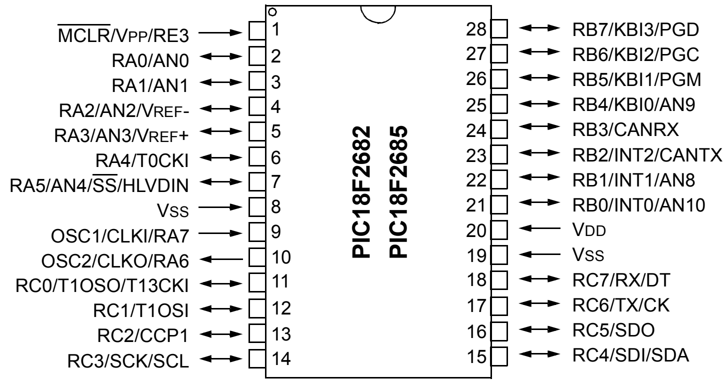
"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

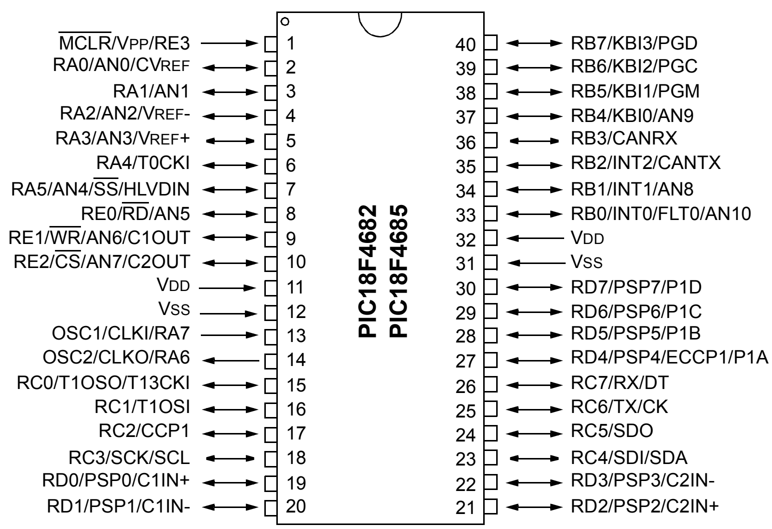
Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	40MHz
Connectivity	CANbus, I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, HLVD, POR, PWM, WDT
Number of I/O	36
Program Memory Size	80KB (40K x 16)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	3.25K x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 5.5V
Data Converters	A/D 11x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-VQFN Exposed Pad
Supplier Device Package	44-QFN (8x8)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf4682-i-ml

Pin Diagrams

28-Pin PDIP, SOIC



40-Pin PDIP



Note: Pinouts are subject to change.

TABLE 1-2: PIC18F2682/2685 PINOUT I/O DESCRIPTIONS (CONTINUED)

Pin Name	Pin Number PDIP, SOIC	Pin Type	Buffer Type	Description
				PORTA is a bidirectional I/O port.
RA0/AN0	2			
RA0		I/O	TTL	Digital I/O.
AN0		I	Analog	Analog input 0.
RA1/AN1	3			
RA1		I/O	TTL	Digital I/O.
AN1		I	Analog	Analog input 1.
RA2/AN2/VREF-	4			
RA2		I/O	TTL	Digital I/O.
AN2		I	Analog	Analog input 2.
VREF-		I	Analog	A/D reference voltage (low) input.
RA3/AN3/VREF+	5			
RA3		I/O	TTL	Digital I/O.
AN3		I	Analog	Analog input 3.
VREF+		I	Analog	A/D reference voltage (high) input.
RA4/T0CKI	6			
RA4		I/O	TTL	Digital I/O.
T0CKI		I	ST	Timer0 external clock input.
RA5/AN4/$\overline{SS}$/HLVDIN	7			
RA5		I/O	TTL	Digital I/O.
AN4		I	Analog	Analog input 4.
$\overline{SS}$		I	TTL	SPI slave select input.
HLVDIN		I	Analog	High/Low-Voltage Detect input.
RA6				See the OSC2/CLKO/RA6 pin.
RA7				See the OSC1/CLKI/RA7 pin.

Legend: TTL = TTL compatible input; CMOS = CMOS compatible input or output; ST = Schmitt Trigger input with CMOS levels; I = Input; O = Output; P = Power

REGISTER 2-2: OSCCON: OSCILLATOR CONTROL REGISTER

R/W-0	R/W-1	R/W-0	R/W-0	R[(1)]	R-0	R/W-0	R/W-0
IDLEN	IRCF2	IRCF1	IRCF0	OSTS	IOFS	SCS1	SCS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7 **IDLEN:** Idle Enable bit
- `1` = Device enters Idle mode on `SLEEP` instruction
- `0` = Device enters Sleep mode on `SLEEP` instruction

bit 6-4 **IRCF2:IRCF0:** Internal Oscillator Frequency Select bits
- `111` = 8 MHz (INTOSC drives clock directly)
- `110` = 4 MHz
- `101` = 2 MHz
- `100` = 1 MHz[(3)]
- `011` = 500 kHz
- `010` = 250 kHz
- `001` = 125 kHz
- `000` = 31 kHz (from either INTOSC/256 or INTRC directly)[(2)]

bit 3 **OSTS:** Oscillator Start-up Time-out Status bit[(1)]
- `1` = Oscillator Start-up Timer time-out has expired; primary oscillator is running
- `0` = Oscillator Start-up Timer time-out is running; primary oscillator is not ready

bit 2 **IOFS:** INTOSC Frequency Stable bit
- `1` = INTOSC frequency is stable and the frequency is provided by one of the RC modes
- `0` = INTOSC frequency is not stable

bit 1-0 **SCS1:SCS0:** System Clock Select bits
- `1x` = Internal oscillator block
- `01` = Timer1 oscillator
- `00` = Primary oscillator

Note 1: Depends on state of the IESO Configuration bit.
2: Source selected by the INTSRC bit (OSCTUNE<7>), see text.
3: Default output frequency of INTOSC on Reset.

9.6 INTx Pin Interrupts

External interrupts on the RB0/INT0, RB1/INT1 and RB2/INT2 pins are edge-triggered. If the corresponding INTEDGx bit in the INTCON2 register is set (= 1), the interrupt is triggered by a rising edge; if the bit is clear, the trigger is on the falling edge. When a valid edge appears on the RBx/INTx pin, the corresponding flag bit, INTxIF, is set. This interrupt can be disabled by clearing the corresponding enable bit INTxIE. Flag bit, INTxIF, must be cleared in software in the Interrupt Service Routine before re-enabling the interrupt.

All external interrupts (INT0, INT1 and INT2) can wake-up the processor from the power-managed modes if bit INTxIE was set prior to going into the power-managed modes. If the Global Interrupt Enable bit, GIE, is set, the processor will branch to the interrupt vector following wake-up.

Interrupt priority for INT1 and INT2 is determined by the value contained in the interrupt priority bits, INT1IP (INTCON3<6>) and INT2IP (INTCON3<7>). There is no priority bit associated with INT0. It is always a high priority interrupt source.

9.7 TMR0 Interrupt

In 8-bit mode (which is the default), an overflow in the TMR0 register (FFh → 00h) will set flag bit TMR0IF. In 16-bit mode, an overflow in the TMR0H:TMR0L register pair (FFFFh → 0000h) will set TMR0IF. The interrupt can be enabled/disabled by setting/clearing enable bit TMR0IE (INTCON<5>). Interrupt priority for Timer0 is determined by the value contained in the interrupt priority bit, TMR0IP (INTCON2<2>). See **Section 11.0 “Timer0 Module”** for further details on the Timer0 module.

9.8 PORTB Interrupt-on-Change

An input change on PORTB<7:4> sets flag bit, RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit, RBIE (INTCON<3>). Interrupt priority for PORTB interrupt-on-change is determined by the value contained in the interrupt priority bit, RBIP (INTCON2<0>).

9.9 Context Saving During Interrupts

During interrupts, the return PC address is saved on the stack. Additionally, the WREG, STATUS and BSR registers are saved on the fast return stack. If a fast return from interrupt is not used (See **Section 5.3 “Data Memory Organization”**), the user may need to save the WREG, STATUS and BSR registers on entry to the Interrupt Service Routine. Depending on the user’s application, other registers may also need to be saved. Example 9-1 saves and restores the WREG, STATUS and BSR registers during an Interrupt Service Routine.

EXAMPLE 9-1: SAVING STATUS, WREG AND BSR REGISTERS IN RAM

```
MOVWF   W_TEMP                          ; W_TEMP is in virtual bank
MOVFF   STATUS, STATUS_TEMP             ; STATUS_TEMP located anywhere
MOVFF   BSR, BSR_TEMP                   ; BSR_TMEP located anywhere
;
; USER ISR CODE
;
MOVFF   BSR_TEMP, BSR                   ; Restore BSR
MOVF    W_TEMP, W                       ; Restore WREG
MOVFF   STATUS_TEMP, STATUS             ; Restore STATUS
```

12.1 Timer1 Operation

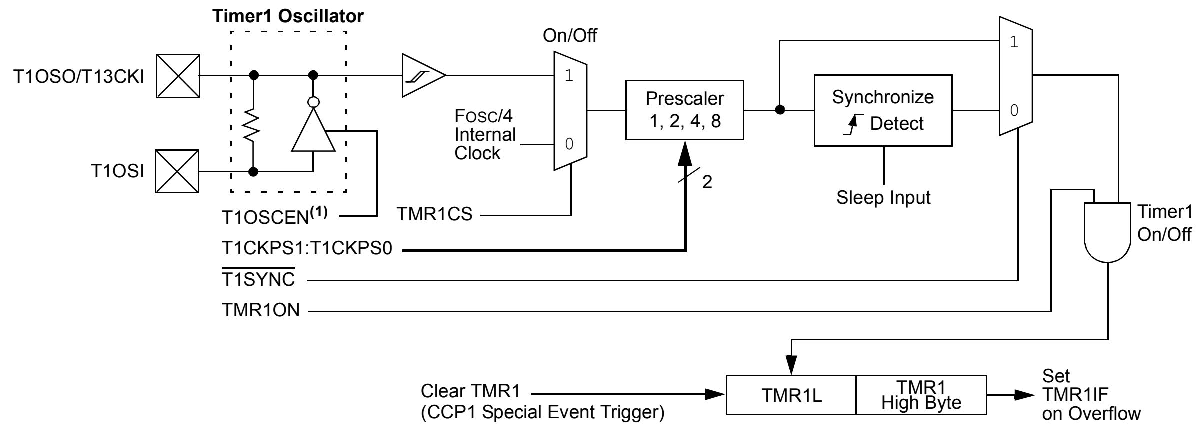
Timer1 can operate in one of these modes:

- Timer
- Synchronous Counter
- Asynchronous Counter

The operating mode is determined by the clock select bit, TMR3CS (T3CON<1>). When TMR3CS is cleared (= 0), Timer3 increments on every internal instruction cycle (Fosc/4). When the bit is set, Timer3 increments on every rising edge of the Timer1 external clock input or the Timer1 oscillator, if enabled.

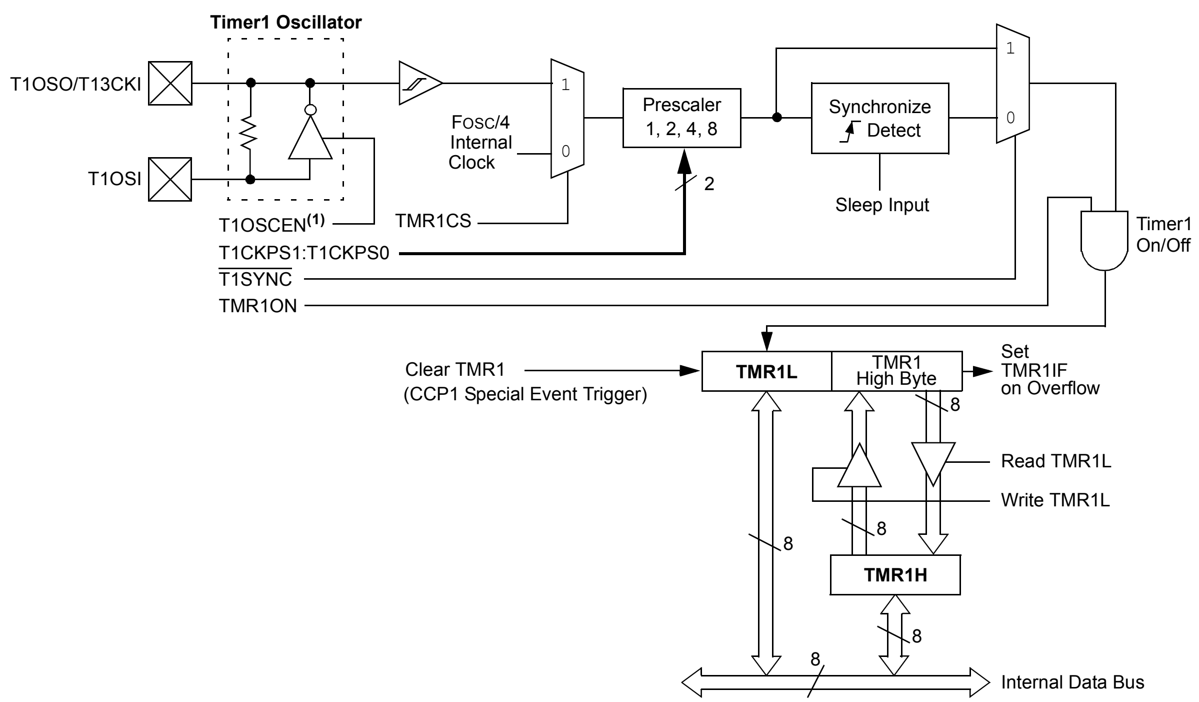
When Timer1 is enabled, the RC1/T1OSI and RC0/T1OSO/T13CKI pins become inputs. This means the values of TRISC<1:0> are ignored and the pins are read as '0'.

FIGURE 12-1: TIMER1 BLOCK DIAGRAM



Note 1: When enable bit, T1OSCEN, is cleared, the inverter and feedback resistor are turned off to eliminate power drain.

FIGURE 12-2: TIMER1 BLOCK DIAGRAM (16-BIT READ/WRITE MODE)



Note 1: When enable bit, T1OSCEN, is cleared, the inverter and feedback resistor are turned off to eliminate power drain.

15.2 Capture Mode

In Capture mode, the CCPR1H:CCPR1L (or ECCPR1H:ECCPR1L) register pair captures the 16-bit value of the TMR1 or TMR3 registers when an event occurs on the CCP1/ECCP1 pin (RC2 for 28/40/44-pin devices and RD4 for 40/44-pin devices). An event is defined as one of the following:

- every falling edge
- every rising edge
- every 4th rising edge
- every 16th rising edge

The event is selected by the mode select bits, CCP1M3:CCP1M0 (CCP1CON<3:0>). When a capture is made, the interrupt request flag bit, CCP1IF (PIR1<2>), is set; it must be cleared in software. If another capture occurs before the value in the CCPR1 register pair is read, the old captured value is overwritten by the new captured value.

15.2.1 CCP1 PIN CONFIGURATION

In Capture mode, the appropriate CCP1/ECCP1 pin should be configured as an input by setting the corresponding TRIS direction bit.

Note:	If RC2/CCP1 or RD4/PSP4/ECCP1/P1A is configured as an output, a write to the port can cause a capture condition.

15.2.2 TIMER1/TIMER3 MODE SELECTION

The timers that are to be used with the capture feature (Timer1 and/or Timer3) must be running in Timer mode or Synchronized Counter mode. In Asynchronous Counter mode, the capture operation may not work. The timer to be used with each CCP1 module is selected in the T3CON register (see **Section 15.1.1 "CCP1 Modules and Timer Resources"**).

15.2.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCP1IE or ECCP1IE interrupt enable bit clear to avoid false interrupts. The interrupt flag bit, CCP1IF or ECCP1IF, should also be cleared following any such change in operating mode.

15.2.4 CCP1 PRESCALER

There are four prescaler settings in Capture mode; they are specified as part of the operating mode selected by the mode select bits (CCP1M3:CCP1M0). Whenever the CCP1 module is turned off or the CCP1 module is not in Capture mode, the prescaler counter is cleared. This means that any Reset will clear the prescaler counter.

Switching from one capture prescaler to another may generate an interrupt. Also, the prescaler counter will not be cleared; therefore, the first capture may be from a non-zero prescaler. Example 15-1 shows the recommended method for switching between capture prescalers. This example also clears the prescaler counter and will not generate the "false" interrupt.

15.2.5 CAN MESSAGE TIME-STAMP

The CAN capture event occurs when a message is received in any of the receive buffers. When configured, the CAN module provides the trigger to the CCP1 module to cause a capture event. This feature is provided to "time-stamp" the received CAN messages.

This feature is enabled by setting the CANCAP bit of the CAN I/O Control register (CIOCON<4>). The message receive signal from the CAN module then takes the place of the events on the RC2/CCP1 pin.

If this feature is selected, then four different capture options for CCP1M<3:0> are available:

- `0100` – every time a CAN message is received
- `0101` – every time a CAN message is received
- `0110` – every 4th time a CAN message is received
- `0111` – Capture mode, every 16th time a CAN message is received

EXAMPLE 15-1: CHANGING BETWEEN CAPTURE PRESCALERS

```
CLRF   CCP1CON      ; Turn CCP1 module off
MOVLW  NEW_CAPT_PS  ; Load WREG with the
                    ; new prescaler mode
                    ; value and CCP1 ON
MOVWF  CCP1CON      ; Load CCP1CON with
                    ; this value
```

16.4.9 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the ECCP1 module for PWM operation:

1. Configure the PWM pins, P1A and P1B (and P1C and P1D, if used), as inputs by setting the corresponding TRIS bits.
2. Set the PWM period by loading the PR2 register.
3. Configure the ECCP1 module for the desired PWM mode and configuration by loading the ECCP1CON register with the appropriate values:
   - Select one of the available output configurations and direction with the EPWM1M1:EPWM1M0 bits.
   - Select the polarities of the PWM output signals with the ECCP1M3:ECCP1M0 bits.
4. Set the PWM duty cycle by loading the ECCPR1L register and ECCP1CON<5:4> bits.
5. For Half-Bridge Output mode, set the dead-band delay by loading ECCP1DEL<6:0> with the appropriate value.
6. If auto-shutdown operation is required, load the ECCP1AS register:
   - Select the auto-shutdown sources using the ECCPAS2:ECCPAS0 bits.
   - Select the shutdown states of the PWM output pins using PSSAC1:PSSAC0 and PSSBD1:PSSBD0 bits.
   - Set the ECCPASE bit (ECCP1AS<7>).
   - Configure the comparators using the CMCON register.
   - Configure the comparator inputs as analog inputs.
7. If auto-restart operation is required, set the PRSEN bit (ECCP1DEL<7>).
8. Configure and start TMR2:
   - Clear the TMR2 interrupt flag bit by clearing the TMR2IF bit (PIR1<1>).
   - Set the TMR2 prescale value by loading the T2CKPS bits (T2CON<1:0>).
   - Enable Timer2 by setting the TMR2ON bit (T2CON<2>).
9. Enable PWM outputs after a new PWM cycle has started:
   - Wait until TMRx overflows (TMRxIF bit is set).
   - Enable the ECCP1/P1A, P1B, P1C and/or P1D pin outputs by clearing the respective TRIS bits.
   - Clear the ECCPASE bit (ECCP1AS<7>).

16.4.10 EFFECTS OF A RESET

Both Power-on Reset and subsequent Resets will force all ports to Input mode and the ECCP1 registers to their Reset states.

This forces the Enhanced CCP1 module to reset to a state compatible with the standard CCP1 module.

NOTES:

REGISTER 23-54: BRGCON3: BAUD RATE CONTROL REGISTER 3

R/W-0	R/W-0	U-0	U-0	U-0	R/W-0	R/W-0	R/W-0
WAKDIS	WAKFIL	—	—	—	SEG2PH2[(1)]	SEG2PH1[(1)]	SEG2PH0[(1)]
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7 **WAKDIS:** Wake-up Disable bit
1 = Disable CAN bus activity wake-up feature
0 = Enable CAN bus activity wake-up feature

bit 6 **WAKFIL:** Selects CAN bus Line Filter for Wake-up bit
1 = Use CAN bus line filter for wake-up
0 = CAN bus line filter is not used for wake-up

bit 5-3 **Unimplemented:** Read as ‘0’

bit 2-0 **SEG2PH2:SEG2PH0:** Phase Segment 2 Time Select bits[(1)]
111 = Phase Segment 2 time = 8 x TQ
110 = Phase Segment 2 time = 7 x TQ
101 = Phase Segment 2 time = 6 x TQ
100 = Phase Segment 2 time = 5 x TQ
011 = Phase Segment 2 time = 4 x TQ
010 = Phase Segment 2 time = 3 x TQ
001 = Phase Segment 2 time = 2 x TQ
000 = Phase Segment 2 time = 1 x TQ

Note 1: Ignored if SEG2PHTS bit (BRGCON2<7>) is ‘0’.

REGISTER 23-58: IPR3: PERIPHERAL INTERRUPT PRIORITY REGISTER 3

Mode 0	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
	IRXIP	WAKIP	ERRIP	TXB2IP	TXB1IP(1)	TXB0IP(1)	RXB1IP	RXB0IP

Mode 1,2	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
	IRXIP	WAKIP	ERRIP	TXBnIP	TXB1IP(1)	TXB0IP(1)	RXBnIP	FIFOWMIP
	bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7 **IRXIP:** CAN Invalid Received Message Interrupt Priority bit
1 = High priority
0 = Low priority

bit 6 **WAKIP:** CAN bus Activity Wake-up Interrupt Priority bit
1 = High priority
0 = Low priority

bit 5 **ERRIP:** CAN bus Error Interrupt Priority bit
1 = High priority
0 = Low priority

bit 4 When CAN is in Mode 0:
TXB2IP: CAN Transmit Buffer 2 Interrupt Priority bit
1 = High priority
0 = Low priority

When CAN is in Mode 1 or 2:
TXBnIP: CAN Transmit Buffer Interrupt Priority bit
1 = High priority
0 = Low priority

bit 3 **TXB1IP:** CAN Transmit Buffer 1 Interrupt Priority bit(1)
1 = High priority
0 = Low priority

bit 2 **TXB0IP:** CAN Transmit Buffer 0 Interrupt Priority bit(1)
1 = High priority
0 = Low priority

bit 1 When CAN is in Mode 0:
RXB1IP: CAN Receive Buffer 1 Interrupt Priority bit
1 = High priority
0 = Low priority

When CAN is in Mode 1 or 2:
RXBnIP: CAN Receive Buffer Interrupts Priority bit
1 = High priority
0 = Low priority

bit 0 When CAN is in Mode 0:
RXB0IP: CAN Receive Buffer 0 Interrupt Priority bit
1 = High priority
0 = Low priority

When CAN is in Mode 1:
Unimplemented: Read as ‘0’

When CAN is in Mode 2:
FIFOWMIP: FIFO Watermark Interrupt Priority bit
1 = High priority
0 = Low priority

Note 1: In CAN Mode 1 and 2, these bits are forced to ‘0’.

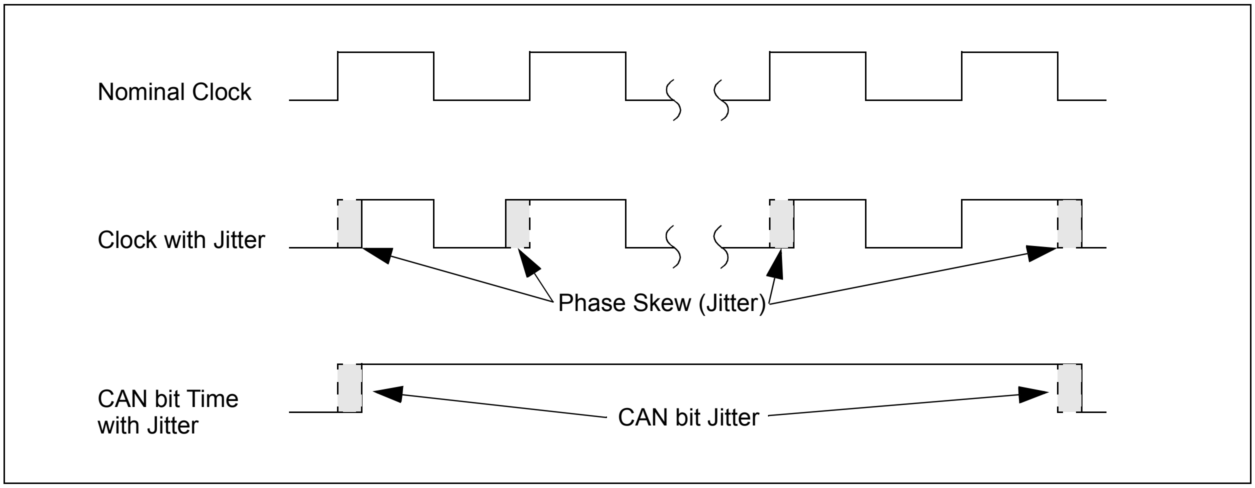
23.9.1 EXTERNAL CLOCK, INTERNAL CLOCK AND MEASURABLE JITTER IN HSPLL-BASED OSCILLATORS

The microcontroller clock frequency generated from a PLL circuit is subject to a jitter, also defined as Phase Jitter or Phase Skew. For its PIC18 Enhanced microcontrollers, Microchip specifies phase jitter (P_{jitter}) as being 2% (Gaussian distribution, within 3 standard deviations, see parameter F13 in Table 27-7) and Total Jitter (T_{jitter}) as being 2 * P_{jitter}.

The CAN protocol uses a bit-stuffing technique that inserts a bit of a given polarity following five bits with the opposite polarity. This gives a total of 10 bits transmitted without re-synchronization (compensation for jitter or phase error).

Given the random nature of the jitter error added, it can be shown that the total error caused by the jitter tends to cancel itself over time. For a period of 10 bits, it is necessary to add only two jitter intervals to correct for jitter-induced error: one interval in the beginning of the 10-bit period and another at the end. The overall effect is shown in Figure 23-5.

FIGURE 23-5: EFFECTS OF PHASE JITTER ON THE MICROCONTROLLER CLOCK AND CAN BIT TIME



Once these considerations are taken into account, it is possible to show that the relation between the jitter and the total frequency error can be defined as:

EQUATION 23-4:

$$\Delta f = \frac{T_{jitter}}{10 \times NBT} = \frac{2 \times P_{jitter}}{10 \times NBT}$$

where jitter is expressed in terms of time and NBT is the Nominal Bit Time.

For example, assume a CAN bit rate of 125 Kb/s, which gives an NBT of 8 μs. For a 16 MHz clock generated from a 4x PLL, the jitter at this clock frequency is:

EQUATION 23-5:

$$2\% \times \frac{1}{16\ MHz} = \frac{0.02}{16 \times 10^{6}} = 1.25ns$$

The resultant frequency error is:

EQUATION 23-6:

$$\frac{2 \times (1.25 \times 10^{-9})}{10 \times (8 \times 10^{-6})} = 3.125 \times 10^{-5} = 0.0031\%$$

23.10 Synchronization

To compensate for phase shifts between the oscillator frequencies of each of the nodes on the bus, each CAN controller must be able to synchronize to the relevant signal edge of the incoming signal. When an edge in the transmitted data is detected, the logic will compare the location of the edge to the expected time (Sync_Seg). The circuit will then adjust the values of Phase Segment 1 and Phase Segment 2 as necessary. There are two mechanisms used for synchronization.

23.10.1 HARD SYNCHRONIZATION

Hard synchronization is only done when there is a recessive to dominant edge during a bus Idle condition, indicating the start of a message. After hard synchronization, the bit time counters are restarted with Sync_Seg. Hard synchronization forces the edge which has occurred to lie within the synchronization segment of the restarted bit time. Due to the rules of synchronization, if a hard synchronization occurs, there will not be a resynchronization within that bit time.

23.10.2 RESYNCHRONIZATION

As a result of resynchronization, Phase Segment 1 may be lengthened or Phase Segment 2 may be shortened. The amount of lengthening or shortening of the phase buffer segments has an upper bound given by the Synchronization Jump Width (SJW). The value of the SJW will be added to Phase Segment 1 (see Figure 23-6) or subtracted from Phase Segment 2 (see Figure 23-7). The SJW is programmable between 1 TQ and 4 TQ.

Clocking information will only be derived from recessive to dominant transitions. The property, that only a fixed maximum number of successive bits have the same value, ensures resynchronization to the bit stream during a frame.

The phase error of an edge is given by the position of the edge relative to Sync_Seg, measured in TQ. The phase error is defined in magnitude of TQ as follows:

- e = 0 if the edge lies within Sync_Seg.
- e > 0 if the edge lies before the sample point.
- e < 0 if the edge lies after the sample point of the previous bit.

If the magnitude of the phase error is less than, or equal to, the programmed value of the Synchronization Jump Width, the effect of a resynchronization is the same as that of a hard synchronization.

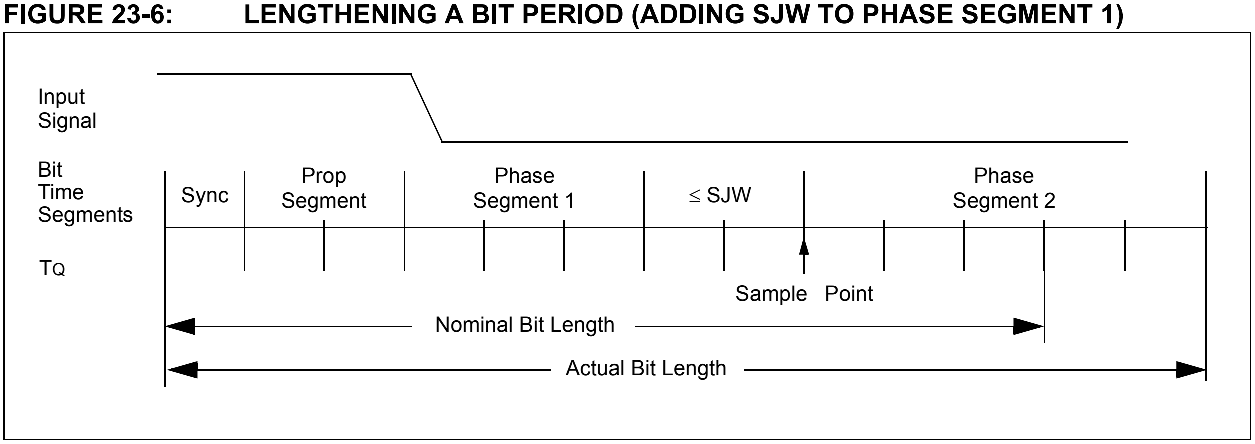
If the magnitude of the phase error is larger than the Synchronization Jump Width and if the phase error is positive, then Phase Segment 1 is lengthened by an amount equal to the Synchronization Jump Width.

If the magnitude of the phase error is larger than the resynchronization jump width and if the phase error is negative, then Phase Segment 2 is shortened by an amount equal to the Synchronization Jump Width.

23.10.3 SYNCHRONIZATION RULES

- Only one synchronization within one bit time is allowed.
- An edge will be used for synchronization only if the value detected at the previous sample point (previously read bus value) differs from the bus value immediately after the edge.
- All other recessive to dominant edges fulfilling rules 1 and 2 will be used for resynchronization, with the exception that a node transmitting a dominant bit will not perform a resynchronization as a result of a recessive to dominant edge with a positive phase error.

FIGURE 23-6: LENGTHENING A BIT PERIOD (ADDING SJW TO PHASE SEGMENT 1)



23.14 Error Detection

The CAN protocol provides sophisticated error detection mechanisms. The following errors can be detected.

23.14.1 CRC ERROR

With the Cyclic Redundancy Check (CRC), the transmitter calculates special check bits for the bit sequence, from the start of a frame until the end of the data field. This CRC sequence is transmitted in the CRC field. The receiving node also calculates the CRC sequence using the same formula and performs a comparison to the received sequence. If a mismatch is detected, a CRC error has occurred and an error frame is generated. The message is repeated.

23.14.2 ACKNOWLEDGE ERROR

In the Acknowledge field of a message, the transmitter checks if the Acknowledge slot (which was sent out as a recessive bit) contains a dominant bit. If not, no other node has received the frame correctly. An Acknowledge error has occurred, an error frame is generated and the message will have to be repeated.

23.14.3 FORM ERROR

If a node detects a dominant bit in one of the four segments, including End-of-Frame, interframe space, Acknowledge delimiter or CRC delimiter, then a form error has occurred and an error frame is generated. The message is repeated.

23.14.4 BIT ERROR

A bit error occurs if a transmitter sends a dominant bit and detects a recessive bit, or if it sends a recessive bit and detects a dominant bit, when monitoring the actual bus level and comparing it to the just transmitted bit. In the case where the transmitter sends a recessive bit and a dominant bit is detected during the arbitration field and the Acknowledge slot, no bit error is generated because normal arbitration is occurring.

23.14.5 STUFF BIT ERROR

If, between the Start-of-Frame and the CRC delimiter, six consecutive bits with the same polarity are detected, the bit stuffing rule has been violated. A stuff bit error occurs and an error frame is generated. The message is repeated.

23.14.6 ERROR STATES

Detected errors are made public to all other nodes via error frames. The transmission of the erroneous message is aborted and the frame is repeated as soon as possible. Furthermore, each CAN node is in one of the three error states: “error-active”, “error-passive” or “bus-off”, according to the value of the internal error counters. The error-active state is the usual state where the bus node can transmit messages and activate error frames (made of dominant bits) without any restrictions. In the error-passive state, messages and passive error frames (made of recessive bits) may be transmitted. The bus-off state makes it temporarily impossible for the station to participate in the bus communication. During this state, messages can neither be received nor transmitted.

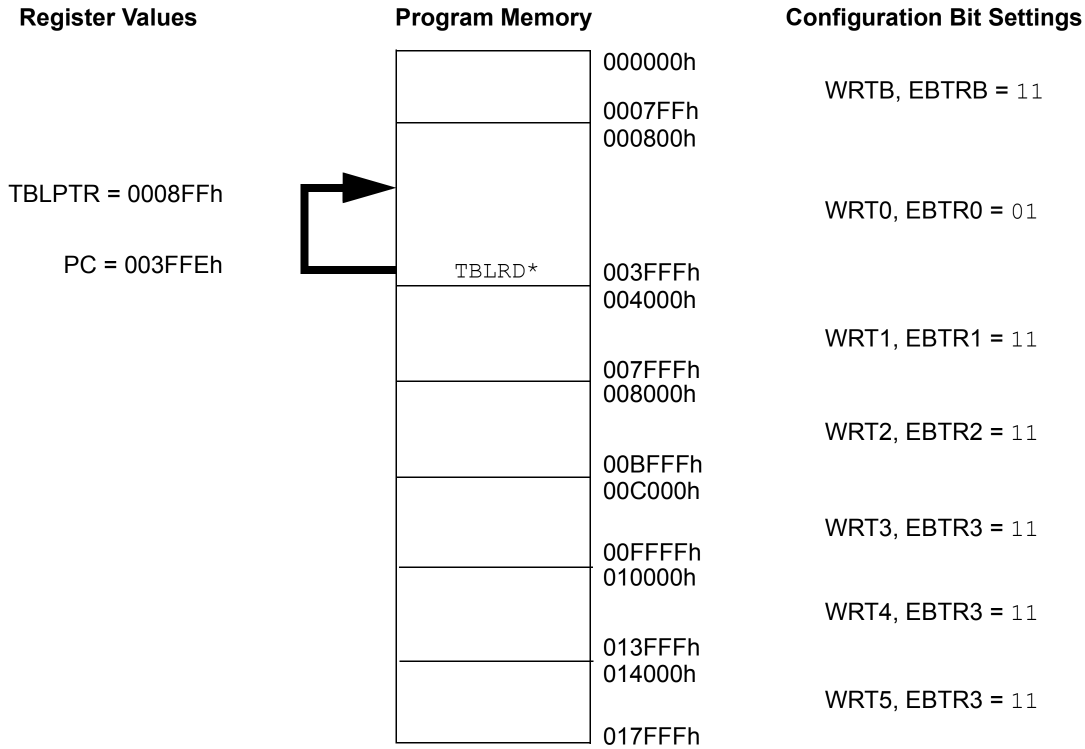
23.14.7 ERROR MODES AND ERROR COUNTERS

The PIC18F2682/2685/4682/4685 devices contain two error counters: the Receive Error Counter (RXERRCNT) and the Transmit Error Counter (TXERRCNT). The values of both counters can be read by the MCU. These counters are incremented or decremented in accordance with the CAN bus specification.

The PIC18F2682/2685/4682/4685 devices are error-active if both error counters are below the error-passive limit of 128. They are error-passive if at least one of the error counters equals or exceeds 128. They go to bus-off if the transmit error counter equals or exceeds the bus-off limit of 256. The devices remain in this state until the bus-off recovery sequence is received. The bus-off recovery sequence consists of 128 occurrences of 11 consecutive recessive bits (see Figure 23-8). Note that the CAN module, after going bus-off, will recover back to error-active without any intervention by the MCU if the bus remains Idle for 128 x 11 bit times. If this is not desired, the error Interrupt Service Routine should address this. The current Error mode of the CAN module can be read by the MCU via the COMSTAT register.

Additionally, there is an Error State Warning flag bit, EWARN, which is set if at least one of the error counters equals or exceeds the error warning limit of 96. EWARN is reset if both error counters are less than the error warning limit.

FIGURE 24-8: EXTERNAL BLOCK TABLE READ (EBTRn) ALLOWED



Results: Table reads permitted within Blockn, even when EBTRBn = 0.
TABLAT register returns the value of the data at the location TBLPTR.

BRA — Unconditional Branch

Syntax: BRA n

Operands: $-1024 \leq n \leq 1023$

Operation: (PC) + 2 + 2n → PC

Status Affected: None

Encoding:

1101	0nnn	nnnn	nnnn

Description: Add the 2's complement number '2n' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is a two-cycle instruction.

Words: 1

Cycles: 2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

Example: `HERE BRA Jump`

Before Instruction
PC = address `(HERE)`
After Instruction
PC = address `(Jump)`

BSF — Bit Set f

Syntax: BSF f, b {,a}

Operands: $0 \leq f \leq 255$
$0 \leq b \leq 7$
$a \in [0,1]$

Operation: 1 → f<b>

Status Affected: None

Encoding:

1000	bbba	ffff	ffff

Description: Bit 'b' in register 'f' is set.

If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See **Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write register 'f'

Example: `BSF FLAG_REG, 7, 1`

Before Instruction
FLAG_REG = 0Ah
After Instruction
FLAG_REG = 8Ah

DAW — Decimal Adjust W Register

Syntax:	DAW
Operands:	None

Operation:	If [W<3:0> >9] or [DC = 1] then
	(W<3:0>) + 6 → W<3:0>;
	else
	(W<3:0>) → W<3:0>
	
	If [W<7:4> >9] or [C = 1] then
	(W<7:4>) + 6 → W<7:4>;
	C = 1;
	else
	(W<7:4>) → W<7:4>

Status Affected:	C
Encoding:	0000 0000 0000 0111
Description:	DAW adjusts the eight-bit value in W, resulting from the earlier addition of two variables (each in packed BCD format) and produces a correct packed BCD result.
Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register W	Process Data	Write W

Example 1:

`DAW`

Before Instruction
W = A5h
C = 0
DC = 0
After Instruction
W = 05h
C = 1
DC = 0

Example 2:

Before Instruction
W = CEh
C = 0
DC = 0
After Instruction
W = 34h
C = 1
DC = 0

DECF — Decrement f

Syntax:	DECF f {,d {,a}}
Operands:	0 ≤ f ≤ 255 d ∈ [0,1] a ∈ [0,1]
Operation:	(f) – 1 → dest
Status Affected:	C, DC, N, OV, Z
Encoding:	0000 01da ffff ffff
Description:	Decrement register 'f'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f' (default). If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default). If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See **Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.
Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: `DECF CNT, 1, 0`

Before Instruction
CNT = 01h
Z = 0
After Instruction
CNT = 00h
Z = 1

RCALL — Relative Call

Syntax:	RCALL n
Operands:	$-1024 \leq n \leq 1023$
Operation:	(PC) + 2 → TOS, (PC) + 2 + 2n → PC
Status Affected:	None
Encoding:	`1101` `1nnn` `nnnn` `nnnn`
Description:	Subroutine call with a jump up to 1K from the current location. First, return address (PC + 2) is pushed onto the stack. Then, add the 2's complement number '2n' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is a two-cycle instruction.
Words:	1
Cycles:	2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'n' Push PC to stack	Process Data	Write to PC
No operation	No operation	No operation	No operation

Example: `HERE RCALL Jump`

Before Instruction
PC = Address `(HERE)`
After Instruction
PC = Address `(Jump)`
TOS = Address `(HERE + 2)`

RESET — Reset

Syntax:	RESET
Operands:	None
Operation:	Reset all registers and flags that are affected by a $\overline{MCLR}$ Reset.
Status Affected:	All
Encoding:	`0000` `0000` `1111` `1111`
Description:	This instruction provides a way to execute a $\overline{MCLR}$ Reset in software.
Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Start Reset	No operation	No operation

Example: `RESET`

After Instruction
Registers = Reset Value
Flags* = Reset Value

27.2 DC Characteristics: Power-Down and Supply Current PIC18F2682/2685/4682/4685 (Industrial) PIC18LF2682/2685/4682/4685 (Industrial) (Continued)

PIC18LF2682/2685/4682/4685 (Industrial)	**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial
PIC18F2682/2685/4682/4685 (Industrial, Extended)	**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended

Param No.	Device	Typ	Max	Units	Conditions		
	Supply Current (IDD)(2,3)						
	PIC18LF268X/468X	65	220	μA	-40°C	VDD = 2.0V	FOSC = 1 MHz (**PRI_IDLE** mode, EC oscillator)
		65	220	μA	+25°C		
		70	220	μA	+85°C		
	PIC18LF268X/468X	120	330	μA	-40°C	VDD = 3.0V	
		120	330	μA	+25°C		
		130	330	μA	+85°C		
	All devices	300	600	μA	-40°C	VDD = 5.0V	
		240	600	μA	+25°C		
		300	600	μA	+85°C		
	Extended devices only	320	600	μA	+125°C		
	PIC18LF268X/468X	260	760	μA	-40°C	VDD = 2.0V	FOSC = 4 MHz (**PRI_IDLE** mode, EC oscillator)
		255	760	μA	+25°C		
		270	760	μA	+85°C		
	PIC18LF268X/468X	420	1.4	μA	-40°C	VDD = 3.0V	
		430	1.4	μA	+25°C		
		450	1.4	μA	+85°C		
	All devices	0.9	2.2	mA	-40°C	VDD = 5.0V	
		0.9	2.2	mA	+25°C		
		0.9	2.2	mA	+85°C		
	Extended devices only	1	3	mA	+125°C		
	Extended devices only	2.8	7	mA	+125°C	VDD = 4.2V	FOSC = 25 MHz (**PRI_IDLE** mode, EC oscillator)
		4.3	11	mA	+125°C	VDD = 5.0V	
	All devices	6	18	mA	-40°C	VDD = 4.2 V	FOSC = 40 MHz (**PRI_IDLE** mode, EC oscillator)
		6.2	18	mA	+25°C		
		6.6	18	mA	+85°C		
	All devices	8.1	22	mA	-40°C	VDD = 5.0V	
		9.1	22	mA	+25°C		
		8.3	22	mA	+85°C		

Legend: Shading of rows is to assist in readability of the table.

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or VSS and all features that add delta current disabled (such as WDT, Timer1 Oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.
The test conditions for all IDD measurements in active operation mode are:
OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;
MCLR = VDD; WDT enabled/disabled as specified.

3: For RC oscillator configurations, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/2REXT (mA) with REXT in kΩ.

4: Standard low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

FIGURE 27-12: EXAMPLE SPI MASTER MODE TIMING (CKE = 0)

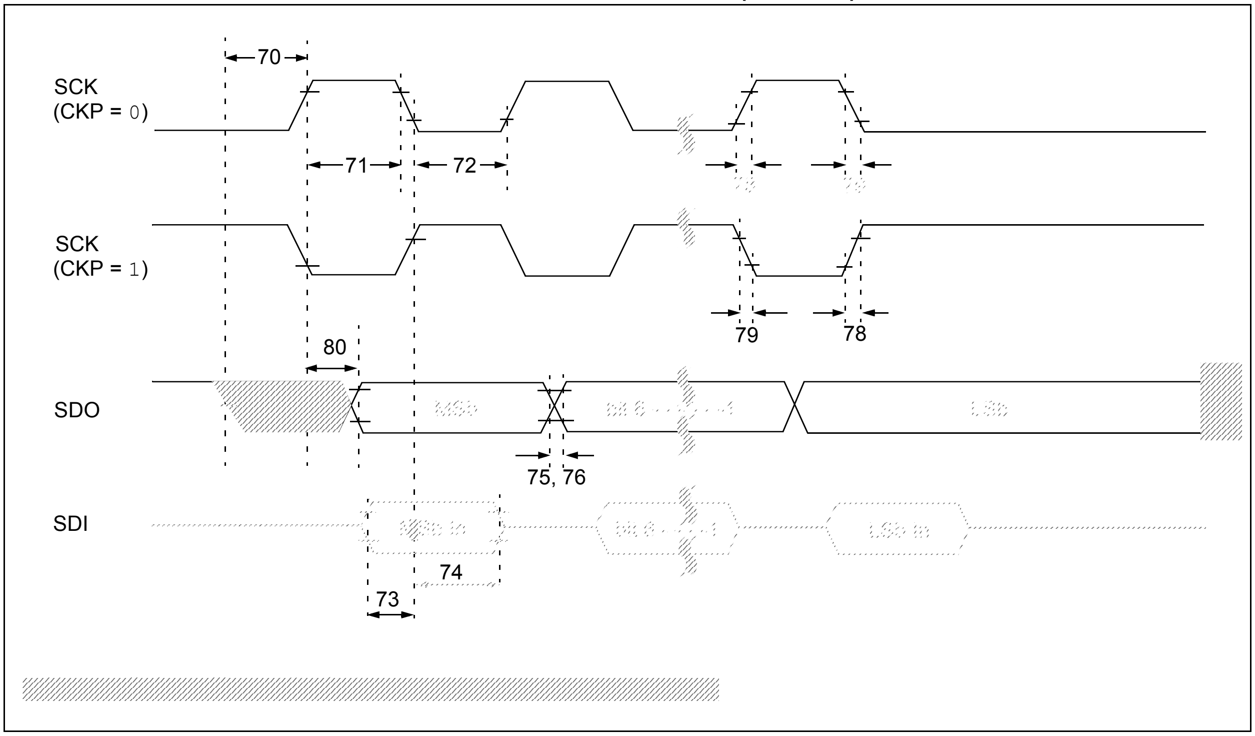


TABLE 27-14: EXAMPLE SPI MODE REQUIREMENTS (MASTER MODE, CKE = 0)

Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
73	TDIV2SCH, TDIV2SCL	Setup Time of SDI Data Input to SCK Edge		100	—	ns	
74	TSCH2DIL, TSCL2DIL	Hold Time of SDI Data Input to SCK Edge		100	—	ns	
75	TDOR	SDO Data Output Rise Time	PIC18**F**XXXX	—	25	ns	
			PIC18**LF**XXXX	—	45	ns	VDD = 2.0V
76	TDOF	SDO Data Output Fall Time		—	25	ns	
78	TSCR	SCK Output Rise Time	PIC18**F**XXXX	—	25	ns	
			PIC18**LF**XXXX	—	45	ns	VDD = 2.0V
79	TSCF	SCK Output Fall Time		—	25	ns	
80	TSCH2DOV, TSCL2DOV	SDO Data Output Valid after SCK Edge	PIC18**F**XXXX	—	50	ns	
			PIC18**LF**XXXX	—	100	ns	VDD = 2.0V

V

W

X