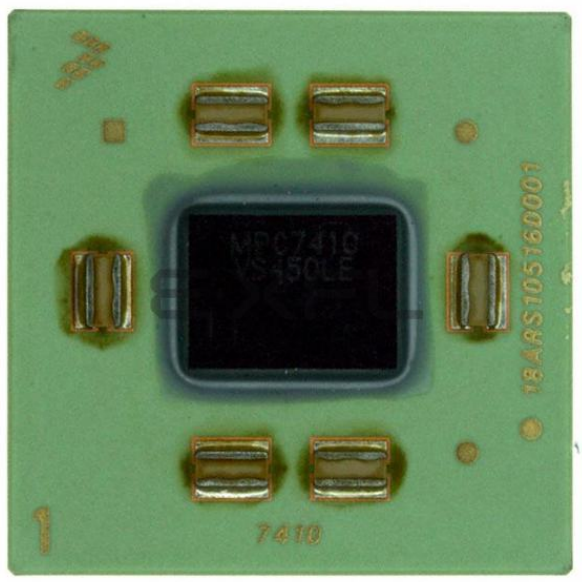

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC G4
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	400MHz
Co-Processors/DSP	-
RAM Controllers	-
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	-
SATA	-
USB	-
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	360-CLGA, FCCLGA
Supplier Device Package	360-FCCLGA (25x25)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc7410vs400ne

- The MPC7410 is implemented in a next generation process technology for core frequency improvement.
- The MPC7410 floating-point unit has been improved to make latency equal for double- and single-precision operations involving multiplication.
- The completion queue has been extended to eight slots.
- There are no other significant changes to scalar pipelines, decode/dispatch/completion mechanisms, or the branch unit. The MPC750 four-stage pipeline model is unchanged (fetch, decode/dispatch, execute, complete/writeback).

Some comments on the MPC7410 with respect to the MPC7400:

- The MPC7410 adds configurable direct-mapped SRAM capability to the L2 cache interface.
- The MPC7410 adds 32-bit interface support to the L2 cache interface. The MPC7410 implements a 19th L2 address pin (L2ASPARE on the MPC7400) in order to support additional address range.
- The MPC7410 removes support for 3.3-V I/O on the L2 cache interface.

Figure 1 shows a block diagram of the MPC7410.

2 Features

This section summarizes features of the MPC7410 implementation of the PowerPC architecture. Major features of the MPC7410 are as follows:

- Branch processing unit
  - Four instructions fetched per clock
  - One branch processed per cycle (plus resolving two speculations)
  - Up to one speculative stream in execution, one additional speculative stream in fetch
  - 512-entry branch history table (BHT) for dynamic prediction
  - 64-entry, four-way set-associative branch target instruction cache (BTIC) for eliminating branch delay slots
- Dispatch unit
  - Full hardware detection of dependencies (resolved in the execution units)
  - Dispatch two instructions to eight independent units (system, branch, load/store, fixed-point unit 1, fixed-point unit 2, floating-point, AltiVec permute, AltiVec ALU)
  - Serialization control (predispatch, postdispatch, execution serialization)

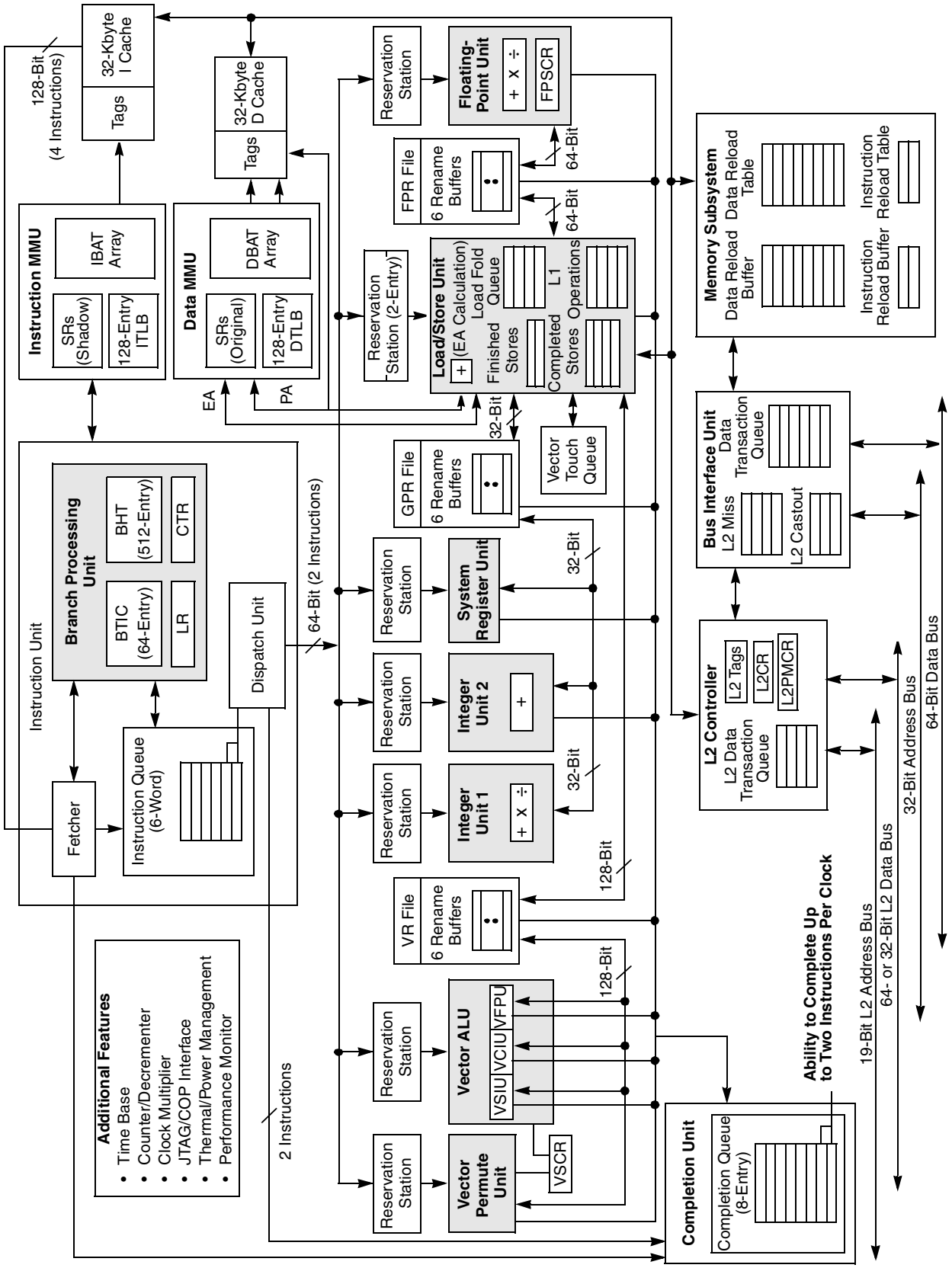


Figure 1. MPC7410 Block Diagram

Table 1. Absolute Maximum Ratings [1] (continued)

Characteristic	Symbol	Maximum Value	Unit	Notes
Rework temperature	T_{rwk}	260	°C	—

Notes:
1. Functional and tested operating conditions are given in Table 3. Absolute maximum ratings are stress ratings only, and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect device reliability or cause permanent damage to the device.
2. **Caution**: V_{in} must not exceed OV_{DD} or $L2OV_{DD}$ by more than 0.2 V at any time including during power-on reset.
3. **Caution**: $L2OV_{DD}/OV_{DD}$ must not exceed $V_{DD}/AV_{DD}/L2AV_{DD}$ by more than 2.0 V at any time including during power-on reset; this limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
4. **Caution**: $V_{DD}/AV_{DD}/L2AV_{DD}$ must not exceed $L2OV_{DD}/OV_{DD}$ by more than 0.4 V at any time including during power-on reset; this limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
5. V_{in} may overshoot/undershoot to a voltage and for a maximum duration as shown in Figure 2.
6. M*xx*7410*xxnnn*LE (Rev. 1.4) and later only. Previous revisions do not support 3.3 V OV_{DD} and have a maximum value OV_{DD} of –0.3 to 2.8 V.

Figure 2 shows the allowable undershoot and overshoot voltage for the MPC7410.

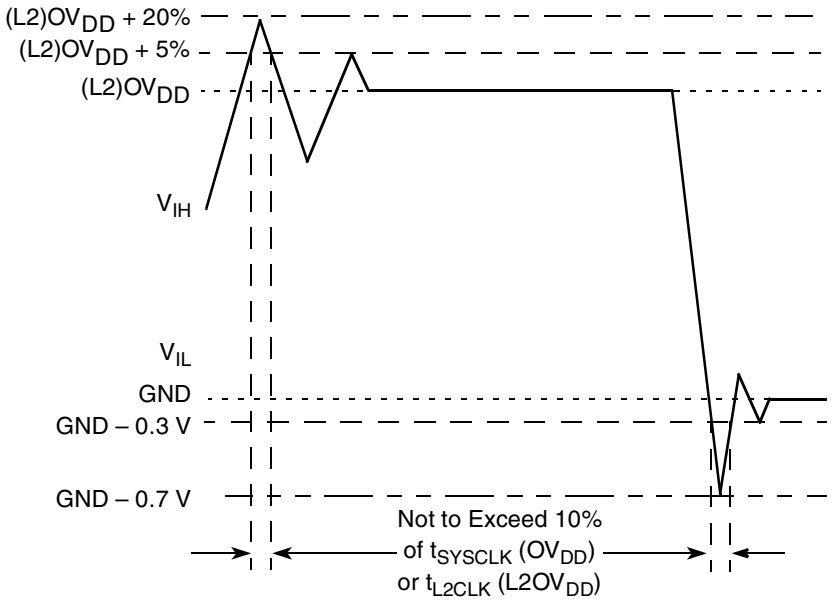


Figure 2. Overshoot/Undershoot Voltage

The MPC7410 provides several I/O voltages to support both compatibility with existing systems and migration to future systems. The MPC7410 core voltage must always be provided at nominal voltage (see Table 3 for actual recommended core voltage). Voltage to the L2 I/Os and processor interface I/Os are provided through separate sets of supply pins and may be provided at the voltages shown in Table 2. Voltage must be provided to the $L2OV_{DD}$ power pins even if the interface is not used. The input voltage threshold for each bus is selected by sampling the state of the voltage select pins BVSEL and L2VSEL at the negation of the signal $\overline{\text{HRESET}}$. These signals must remain stable during part operation and cannot change. The output voltage will swing from GND to the maximum voltage applied to the OV_{DD} or $L2OV_{DD}$ power pins.

4.2.1 Clock AC Specifications

Table 7 provides the clock AC timing specifications as defined in Figure 3.

Table 7. Clock AC Timing Specifications

At recommended operating conditions (see Table 3)

Characteristic	Symbol	Maximum Processor Core Frequency						Unit	Notes
		400 MHz		450 MHz		500 MHz			
		Min	Max	Min	Max	Min	Max		
Processor frequency	f_{core}	350	400	350	450	350	500	MHz	1
VCO frequency	f_{VCO}	700	800	700	900	700	1000	MHz	1
SYSCLK frequency	f_{SYSCLK}	33	133	33	133	33	133	MHz	1
SYSCLK cycle time	t_{SYSCLK}	7.5	30	7.5	30	7.5	30	ns	—
SYSCLK rise and fall time	t_{KR} and t_{KF}	—	0.5	—	0.5	—	0.5	ns/V	2
SYSCLK duty cycle measured at $OV_{DD}/2$	t_{KHKL}/t_{SYSCLK}	40	60	40	60	40	60	%	3
SYSCLK jitter	—	—	±150	—	±150	—	±150	ps	4
Internal PLL-relock time	—	—	100	—	100	—	100	µs	5

Notes:

1. **Caution**: The SYSCLK frequency and PLL_CFG[0:3] settings must be chosen such that the resulting SYSCLK (bus) frequency, CPU (core) frequency, and PLL (VCO) frequency do not exceed their respective maximum or minimum operating frequencies. Refer to the PLL_CFG[0:3] signal description in Section 8.1, "PLL Configuration," for valid PLL_CFG[0:3] settings.
2. Rise and fall times measurement are determined by the slew rates of the bus interface, rather than by time. As a result, the 0.5 ns rise/fall time spec of the 1.8- and 2.5-V bus interfaces is equivalent to the 1 ns rise/fall time of the 3.3-V bus interface. Both interfaces required a 2 V/ns slew rate. The slew rate is measured as a 1-V change (from 0.2 to 1.2 V) in 0.5 ns for the 1.8- and 2.5-V bus interfaces, whereas the 3.3-V bus interface required a 2-V change (from 0.4 to 2.4 V) in 1 ns.
3. Timing is guaranteed by design and characterization.
4. This represents total input jitter—short- and long-term combined—and is guaranteed by design.
5. Relock timing is guaranteed by design and characterization. PLL-relock time is the maximum amount of time required for PLL lock after a stable V_{DD} and SYSCLK are reached during the power-on reset sequence. This specification also applies when the PLL has been disabled and subsequently re-enabled during sleep mode. Also note that $\overline{\text{HRESET}}$ must be held asserted for a minimum of 255 bus clocks after the PLL-relock time during the power-on reset sequence.

Figure 3 provides the SYSCLK input timing diagram.

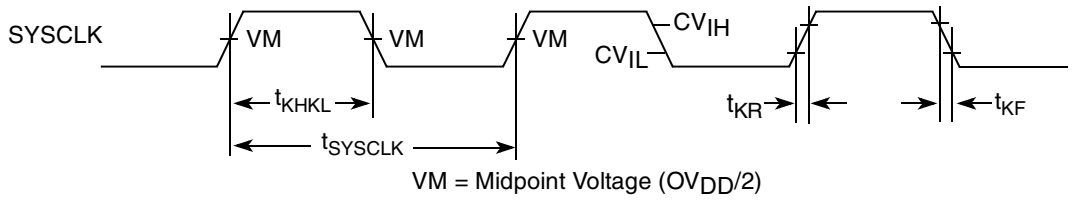


Figure 3. SYSCLK Input Timing Diagram

4.2.2 Processor Bus AC Specifications

Table 8 provides the processor bus AC timing specifications for the MPC7410 as defined in Figure 4 and Figure 5. Timing specifications for the L2 bus are provided in Section 4.2.3, "L2 Clock AC Specifications."

Table 8. Processor Bus AC Timing Specifications [1]

At recommended operating conditions (see Table 3)

Parameter	Symbol [2]	400, 450, 500 MHz		Unit	Notes
		Min	Max		
Input setup	t_{IVKH}	1.0	—	ns	4
Input hold	t_{IXKH}	0	—	ns	4
Output valid times: TS $\overline{ARTRY}$, $\overline{SHD0}$, $\overline{SHD1}$ All other outputs	 t_{KHTSV} t_{KHARV} t_{KHOV}	 — — —	 3.0 2.3 3.0	ns	5, 6
Output hold times: TS $\overline{ARTRY}$, $\overline{SHD0}$, $\overline{SHD1}$ All other outputs	 t_{KHTSX} t_{KHARX} t_{KHOX}	 0.5 0.5 0.5	 — — —	ns	5
SYSCLK to output enable	t_{KHOE}	0.5	—	ns	9
SYSCLK to output high impedance (all except $\overline{ABB}$/$\overline{AMON}$(0), $\overline{ARTRY}$/$\overline{SHD}$, $\overline{DBB}$/$\overline{DMON}$(0), $\overline{SHD0}$, $\overline{SHD1}$)	t_{KHOZ}	—	3.5	ns	
SYSCLK to $\overline{ABB}$/$\overline{AMON}$(0), $\overline{DBB}$/$\overline{DMON}$(0) high impedance after precharge	t_{KHABPZ}	—	1	t_{SYSCLK}	3, 7, 9
Maximum delay to $\overline{ARTRY}$, $\overline{SHD0}$, $\overline{SHD1}$ precharge	t_{KHARP}	—	1	t_{SYSCLK}	3, 8, 9

Table 9. L2CLK Output AC Timing Specifications (continued)

At recommended operating conditions (see Table 3)

Parameter	Symbol	400 MHz		450 MHz		500 MHz		Unit	Notes
		Min	Max	Min	Max	Min	Max		
L2CLK_OUT output jitter	—	—	±150	—	±150	—	±150	ps	6

Notes:

1. L2CLK outputs are L2CLK_OUTA, L2CLK_OUTB, and L2SYNC_OUT pins. The L2CLK frequency to core frequency settings must be chosen such that the resulting L2CLK frequency and core frequency do not exceed their respective maximum or minimum operating frequencies. The maximum L2CLK frequency will be system dependent. L2CLK_OUTA and L2CLK_OUTB must have equal loading.
2. The nominal duty cycle of the L2CLK is 50% measured at midpoint voltage.
3. The DLL-relock time is specified in terms of L2CLKs. The number in the table must be multiplied by the period of L2CLK to compute the actual time duration in ns. Relock timing is guaranteed by design and characterization.
4. The L2CR[L2SL] bit should be set for L2CLK frequencies less than 150 MHz. This adds more delay to each tap of the DLL.
5. Allowable skew between L2SYNC_OUT and L2SYNC_IN.
6. Guaranteed by design and not tested. This output jitter number represents the maximum delay of one tap forward or one tap back from the current DLL tap as the phase comparator seeks to minimize the phase difference between L2SYNC_IN and the internal L2CLK. This number must be comprehended in the L2 timing analysis. The input jitter on SYSCLK affects L2CLK_OUT and the L2 address/data/control signals equally and, therefore, is already comprehended in the AC timing and does not have to be considered in the L2 timing analysis.

The L2CLK_OUT timing diagram is shown in Figure 7.

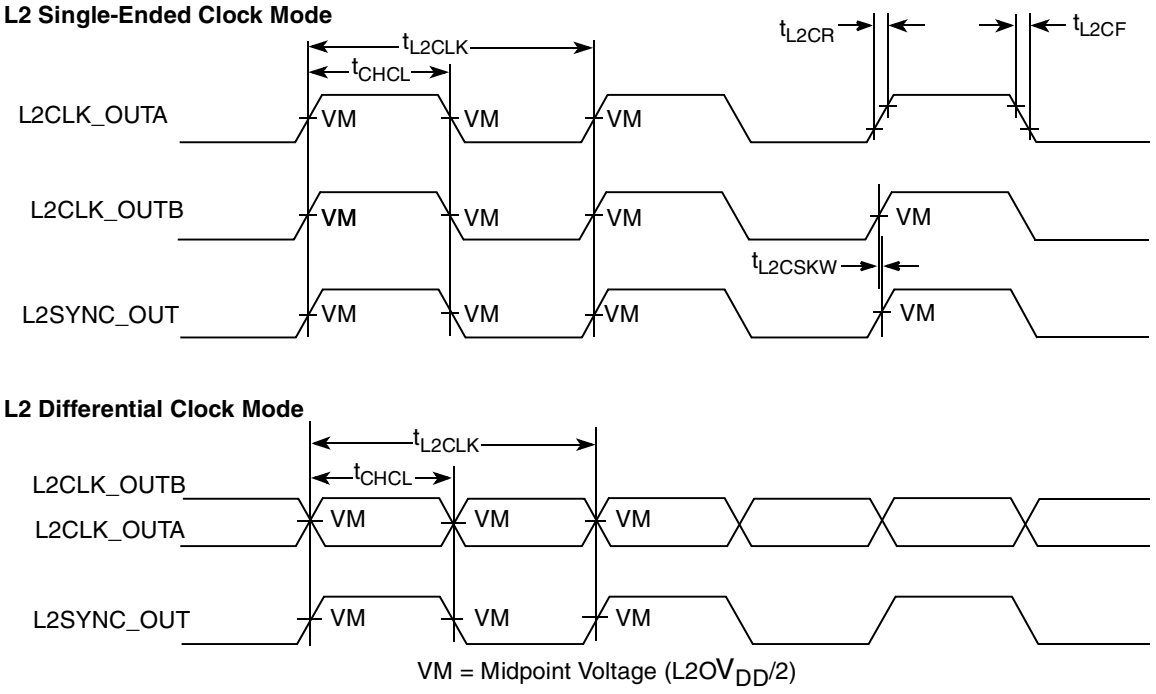


Figure 7. L2CLK_OUT Output Timing Diagram

Table 11. JTAG AC Timing Specifications (Independent of SYSCLK) [1] (continued)

At recommended operating conditions (see Table 3)

Parameter	Symbol	Min	Max	Unit	Notes
$\overline{\text{TRST}}$ assert time	t_{TRST}	25	—	ns	2
Input setup times: Boundary-scan data TMS, TDI	 t_{DVJH} t_{IVJH}	 4 0	 — —	ns	3
Input hold times: Boundary-scan data TMS, TDI	 t_{DXJH} t_{IXJH}	 20 25	 — —	ns	3
Valid times: Boundary-scan data TDO	 t_{JLDV} t_{JLOV}	 4 4	 20 25	ns	4
TCK to output high impedance: Boundary-scan data TDO	 t_{JLDZ} t_{JLOZ}	 3 3	 19 9	ns	 4, 5 5

Notes:

1. All outputs are measured from the midpoint voltage of the falling/rising edge of TCLK to the midpoint of the signal in question. The output timings are measured at the pins. All output timings assume a purely resistive 50-Ω load (see Figure 11). Time-of-flight delays must be added for trace lengths, vias, and connectors in the system.
2. $\overline{\text{TRST}}$ is an asynchronous level sensitive signal. The setup time is for test purposes only.
3. Non-JTAG signal input timing with respect to TCK.
4. Non-JTAG signal output timing with respect to TCK.
5. Guaranteed by design and characterization.

Figure 11 provides the AC test load for TDO and the boundary-scan outputs of the MPC7410.

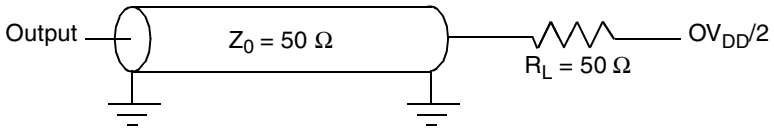


Figure 11. Alternate AC Test Load for the JTAG Interface

Figure 12 provides the JTAG clock input timing diagram.

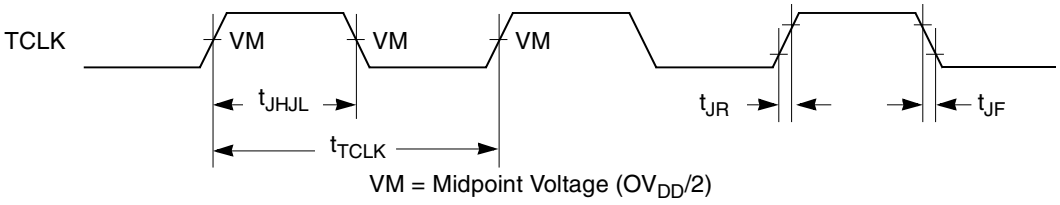
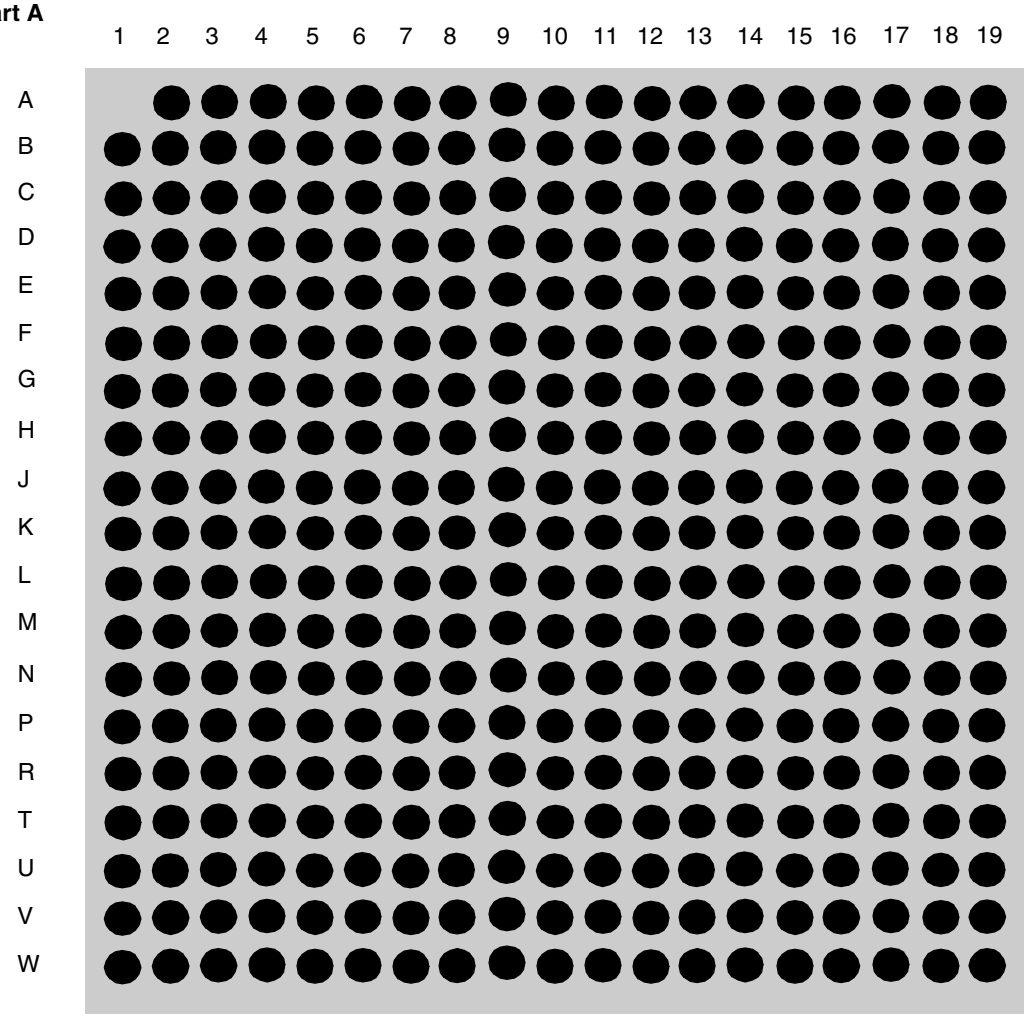


Figure 12. JTAG Clock Input Timing Diagram

5 Pin Assignments

Figure 16, part A shows the pinout for the MPC7410, 360 CBGA, 360 HCTE, and 360 HCTE Lead Free C5 Spheres packages as viewed from the top surface. Figure 16, part B shows the side profile of the CBGA and HCTE_CBGA packages to indicate the direction of the top surface view. Figure 16, part C shows the side profile of the HCTE_LGA package to indicate the direction of the top surface view.

Part A



Not to Scale

Part B

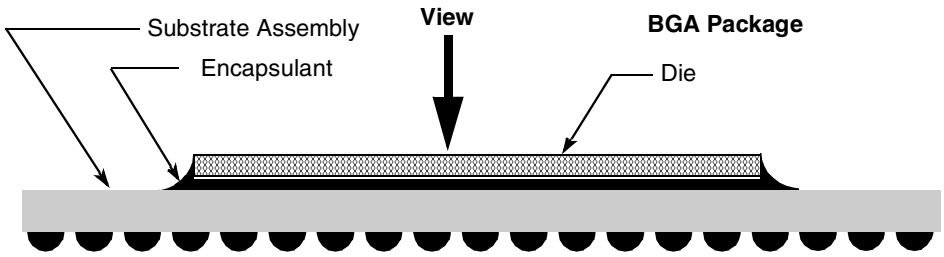


Table 12. Pinout Listing for the MPC7410, 360 CBGA and 360 HCTE Packages (continued)

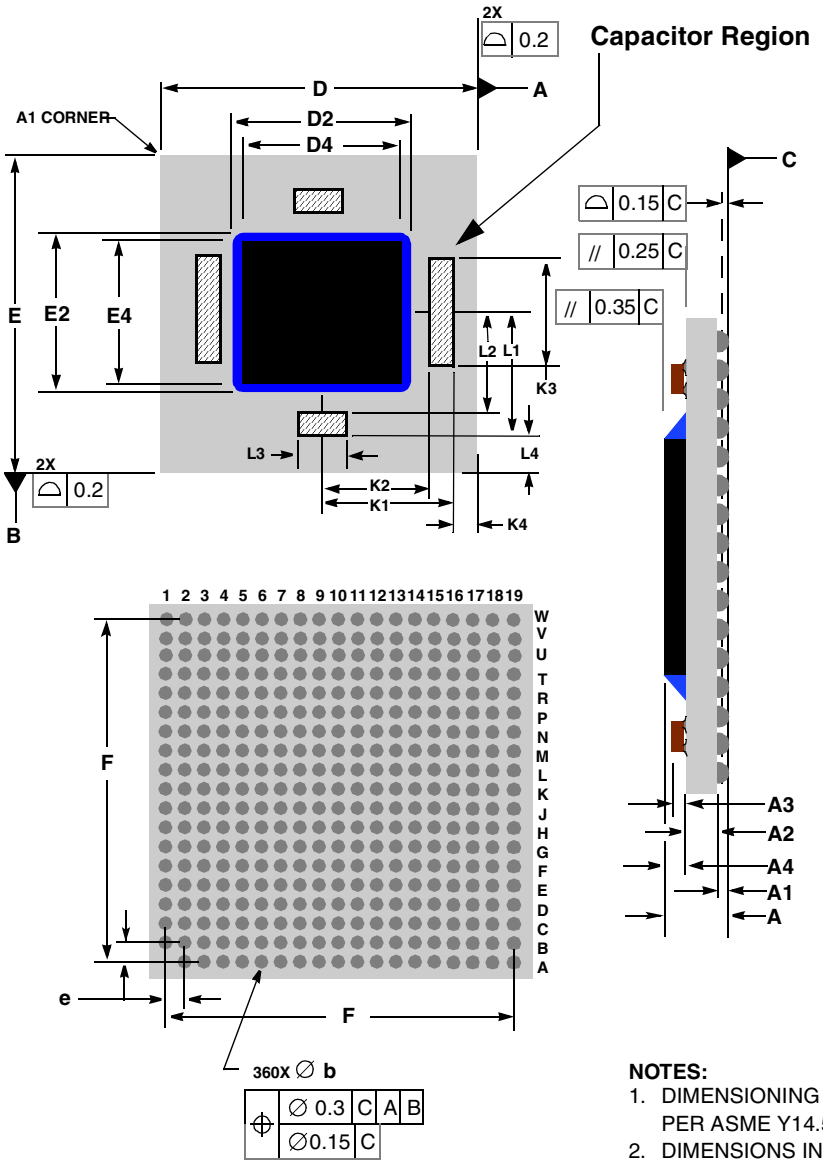
Signal Name	Pin Number	Active	I/O	I/F Select [1]	Notes
DL[0:31]	M6, P3, N4, N5, R3, M7, T2, N6, U2, N7, P11, V13, U12, P12, T13, W13, U13, V10, W8, T11, U11, V12, V8, T1, P1, V1, U1, N1, R2, V3, U3, W2	High	I/O	BVSEL	—
DP[0:7]	L1, P2, M2, V2, M1, N2, T3, R1	High	I/O	BVSEL	—
$\overline{\text{DRDY}}$	K9	Low	Output	BVSEL	6, 8, 13
$\overline{\text{DBWO}}$ DTI[0]	D1	Low	Input	BVSEL	—
DTI[1:2]	H6, G1	High	Input	BVSEL	5, 10, 13
$\overline{\text{EMODE}}$	A3	Low	Input	BVSEL	7, 10
$\overline{\text{GBL}}$	B1	Low	I/O	BVSEL	—
GND	D10, D14, D16, D4, D6, E12, E8, F4, F6, F10, F14, F16, G9, G11, H5, H8, H10, H12, H15, J9, J11, K4, K6, K8, K10, K12, K14, K16, L9, L11, M5, M8, M10, M12, M15, N9, N11, P4, P6, P10, P14, P16, R8, R12, T4, T6, T10, T14, T16	—	—	N/A	—
$\overline{\text{HIT}}$	B5	Low	Output	BVSEL	6, 8
$\overline{\text{HRESET}}$	B6	Low	Input	BVSEL	—
$\overline{\text{INT}}$	C11	Low	Input	BVSEL	—
L1_TSTCLK	F8	High	Input	BVSEL	2
L2ADDR[0:16]	L17, L18, L19, M19, K18, K17, K15, J19, J18, J17, J16, H18, H17, J14, J13, H19, G18	High	Output	L2VSEL	—
L2ADDR[17:18]	K19,W19	High	Output	L2VSEL	8
$L2AV_{DD}$	L13	—	Input	V_{DD}	—
$\overline{\text{L2CE}}$	P17	Low	Output	L2VSEL	—
L2CLK_OUTA	N15	High	Output	L2VSEL	—
L2CLK_OUTB	L16	High	Output	L2VSEL	—
L2DATA[0:63]	U14, R13, W14, W15, V15, U15, W16, V16, W17, V17, U17, W18, V18, U18, V19, U19, T18, T17, R19, R18, R17, R15, P19, P18, P13, N14, N13, N19, N17, M17, M13, M18, H13, G19, G16, G15, G14, G13, F19, F18, F13, E19, E18, E17, E15, D19, D18, D17, C18, C17, B19, B18, B17, A18, A17, A16, B16, C16, A14, A15, C15, B14, C14, E13	High	I/O	L2VSEL	—
L2DP[0:7]	V14, U16, T19, N18, H14, F17, C19, B15	High	I/O	L2VSEL	—
$L2OV_{DD}$	D15, E14, E16, H16, J15, L15, M16, K13, P15, R14, R16, T15, F15	—	—	N/A	11
L2SYNC_IN	L14	High	Input	L2VSEL	—
L2SYNC_OUT	M14	High	Output	L2VSEL	—
L2_TSTCLK	F7	High	Input	BVSEL	2

Table 12. Pinout Listing for the MPC7410, 360 CBGA and 360 HCTE Packages (continued)

Signal Name	Pin Number	Active	I/O	I/F Select [1]	Notes
L2VSEL	A19	High	Input	N/A	1, 3, 8, 9, 14
$\overline{\text{L2WE}}$	N16	Low	Output	L2VSEL	—
L2ZZ	G17	High	Output	L2VSEL	—
$\overline{\text{LSSD\_MODE}}$	F9	Low	Input	BVSEL	2
$\overline{\text{MCP}}$	B11	Low	Input	BVSEL	15
OV_{DD}	D5, D8, D12, E4, E6, E9, E11, F5, H4, J5, L5, M4, P5, R4, R6, R9, R11, T5, T8, T12	—	—	N/A	—
PLL_CFG[0:3]	A4, A5, A6, A7	High	Input	BVSEL	4
$\overline{\text{QACK}}$	B2	Low	Input	BVSEL	—
$\overline{\text{QREQ}}$	J3	Low	Output	BVSEL	—
$\overline{\text{RSRV}}$	D3	Low	Output	BVSEL	—
$\overline{\text{SHD0}}$	B3	Low	I/O	BVSEL	8
$\overline{\text{SHD1}}$	B4	Low	I/O	BVSEL	5, 8
$\overline{\text{SMI}}$	A12	Low	Input	BVSEL	—
$\overline{\text{SRESET}}$	E10	Low	Input	BVSEL	—
SYSCLK	H9	—	Input	BVSEL	—
$\overline{\text{TA}}$	F1	Low	Input	BVSEL	—
TBEN	A2	High	Input	BVSEL	—
$\overline{\text{TBST}}$	A11	Low	Output	BVSEL	—
TCK	B10	High	Input	BVSEL	—
TDI	B7	High	Input	BVSEL	9
TDO	D9	High	Output	BVSEL	—
$\overline{\text{TEA}}$	J1	Low	Input	BVSEL	—
TMS	C8	High	Input	BVSEL	9
$\overline{\text{TRST}}$	A10	Low	Input	BVSEL	9
$\overline{\text{TS}}$	K7	Low	I/O	BVSEL	—
TSIZ[0:2]	A9, B9, C9	High	Output	BVSEL	—
TT[0:4]	C10, D11, B12, C12, F11	High	I/O	BVSEL	—
$\overline{\text{WT}}$	C3	Low	I/O	BVSEL	—

7.3 Mechanical Dimensions for the MPC7410, 360 CBGA and 360 HCTE_CBGA

Figure 17 provides the mechanical dimensions and bottom surface nomenclature of the MPC7410, 360 CBGA and 360 HCTE_CBGA packages.



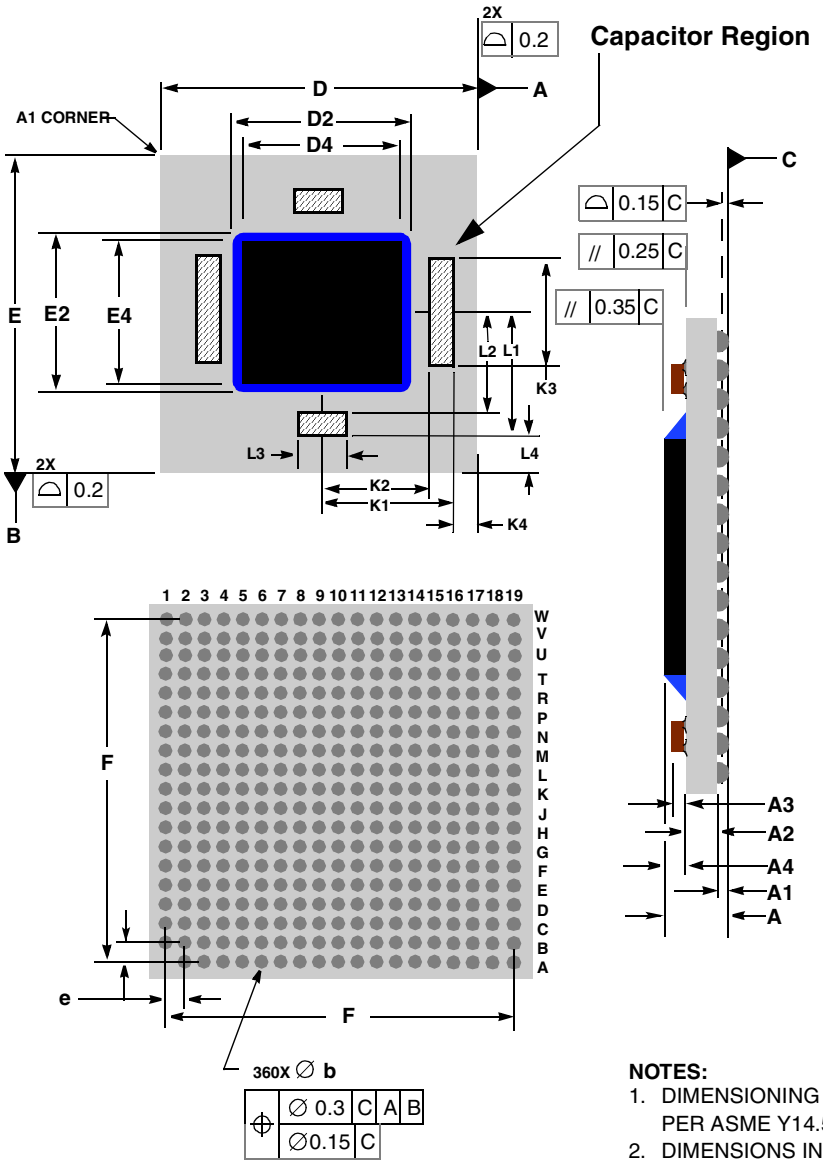
	Millimeters	
DIM	MIN	MAX
A	2.72	3.20
A1	0.80	1.00
A2	1.10	1.30
A3	—	0.60
A4	0.82	0.90
b	0.82	0.93
D	25.00 BSC	
D2	—	12.50
D4	6.00	9.00
e	1.27 BSC	
E	25.00 BSC	
E2	—	14.30
E4	8.00	11.00
F	22.86 BSC	
K1	—	9.75
K2	6.46	—
K3	8.20	8.60
K4	2.75	—
L1	—	9.50
L2	6.94	—
L3	3.10	3.30
L4	3.00	—

NOTES:
1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. DIMENSIONS IN MILLIMETERS.
3. TOP SIDE A1 CORNER INDEX IS A METALIZED FEATURE WITH VARIOUS SHAPES. BOTTOM SIDE A1 CORNER IS DESIGNATED WITH A BALL MISSING FROM THE ARRAY.

Figure 17. Mechanical Dimensions and Bottom Surface Nomenclature for the MPC7410, 360 CBGA and 360 HCTE_CBGA Packages

7.4 Mechanical Dimensions for the MPC7410, 360 HCTE_CBGA (Lead Free C5 Spheres)

Figure 18 provides the mechanical dimensions and bottom surface nomenclature of the MPC7410, 360 HCTE_CBGA (lead-free C5 spheres) package.



	Millimeters	
DIM	MIN	MAX
A	2.32	2.80
A1	0.40	0.60
A2	1.10	1.30
A3	—	0.60
A4	0.82	0.90
b	0.60	0.90
D	25.00 BSC	
D2	—	12.50
D4	6.00	9.00
e	1.27 BSC	
E	25.00 BSC	
E2	—	14.30
E4	8.00	11.00
F	22.86 BSC	
K1	—	9.75
K2	6.46	—
K3	8.20	8.60
K4	2.75	—
L1	—	9.50
L2	6.94	—
L3	3.10	3.30
L4	3.00	—

NOTES:
1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. DIMENSIONS IN MILLIMETERS.
3. TOP SIDE A1 CORNER INDEX IS A METALIZED FEATURE WITH VARIOUS SHAPES. BOTTOM SIDE A1 CORNER IS DESIGNATED WITH A BALL MISSING FROM THE ARRAY.

Figure 18. Mechanical Dimensions and Bottom Surface Nomenclature for the MPC7410 360 HCTE_CBGA (Lead-Free C5 Spheres) Package

8.4 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level through a resistor. Unused active low inputs should be tied to OV_{DD}. Unused active high inputs should be connected to GND. All NC (no connect) signals must remain unconnected.

Power and ground connections must be made to all external V_{DD}, OV_{DD}, $L2OV_{DD}$, and GND pins of the MPC7410. Note that power must be supplied to $L2OV_{DD}$ even if the L2 interface of the MPC7410 will not be used; the remainder of the L2 interface may be left unterminated.

8.5 Output Buffer DC Impedance

The MPC7410 60x and L2 I/O drivers are characterized over process, voltage, and temperature. To measure Z_0, an external resistor is connected from the chip pad to OV_{DD} or GND. Then, the value of each resistor is varied until the pad voltage is $OV_{DD}/2$ (see Figure 23).

The output impedance is the average of two components, the resistances of the pull-up and pull-down devices. When data is held low, SW2 is closed (SW1 is open), and R_N is trimmed until the voltage at the pad equals $(L2)OV_{DD}/2$. R_N then becomes the resistance of the pull-down devices. When data is held high, SW1 is closed (SW2 is open), and R_P is trimmed until the voltage at the pad equals $(L2)OV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

Figure 23 describes the driver impedance measurement circuit described above.

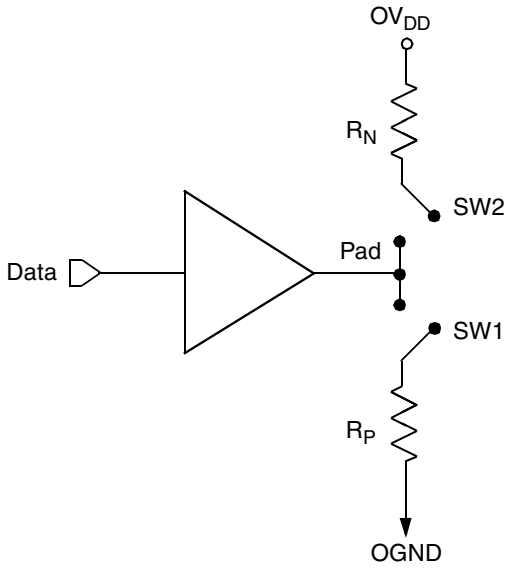


Figure 23. Driver Impedance Measurement Circuit

Alternately, the following is another method to determine the output impedance of the MPC7410. A voltage source, V_{force}, is connected to the output of the MPC7410, as in Figure 24. Data is held low, the voltage source is set to a value that is equal to $(L2)OV_{DD}/2$, and the current sourced by V_{force} is measured. The voltage drop across the pull-down device, which is equal to $(L2)OV_{DD}/2$, is divided by the measured current to determine the output impedance of the pull-down device, R_N. Similarly, the impedance of the pull-up device is determined by dividing the voltage drop of the pull-up, $(L2)OV_{DD}/2$, by the current sank by the pull-up when the data is high and V_{force} is equal to $(L2)OV_{DD}/2$. This method can be employed with either empirical data from a test setup or with data from simulation models, such as IBIS.

the MPC7410 or by other receivers in the system. These signals can be pulled up through weak (10-kΩ) pull-up resistors by the system, address bus driven mode can be enabled (see the *MPC7410 RISC Microprocessor Family Users' Manual* for more information on this mode), or these signals may be otherwise driven by the system during inactive periods of the bus to avoid this additional power draw. The snooped address and transfer attribute inputs are: A[0:31], AP[0:3], TT[0:4], $\overline{\text{CI}}$, $\overline{\text{WT}}$, and $\overline{\text{GBL}}$.

In systems where $\overline{\text{GBL}}$ is not connected and other devices may be asserting $\overline{\text{TS}}$ for a snoopable transaction while not driving $\overline{\text{GBL}}$ to the processor, we recommend that a strong (1 kΩ) pull-up resistor be used on $\overline{\text{GBL}}$. Note that the MPC7410 will only snoop transactions when $\overline{\text{GBL}}$ is asserted.

The data bus input receivers are normally turned off when no read operation is in progress and, therefore, do not require pull-up resistors on the bus. Other data bus receivers in the system, however, may require pull-ups, or that those signals be otherwise driven by the system during inactive periods by the system. The data bus signals are: DH[0:31], DL[0:31], and DP[0:7].

If address or data parity is not used by the system, and the respective parity checking is disabled through HID0, the input receivers for those pins are disabled, and those pins do not require pull-up resistors and should be left unconnected by the system. If parity checking is disabled through HID0, and parity generation is not required by the MPC7410 (note that the MPC7410 always generates parity), then all parity pins may be left unconnected by the system.

The L2 interface does not normally require pull-up resistors.

8.7 JTAG Configuration Signals

Boundary-scan testing is enabled through the JTAG interface signals. The $\overline{\text{TRST}}$ signal is optional in the IEEE 1149.1 specification, but is provided on all processors that implement the PowerPC architecture. While it is possible to force the TAP controller to the reset state using only the TCK and TMS signals, more reliable power-on reset performance will be obtained if the $\overline{\text{TRST}}$ signal is asserted during power-on reset. Because the JTAG interface is also used for accessing the common on-chip processor (COP) function, simply tying $\overline{\text{TRST}}$ to $\overline{\text{HRESET}}$ is not practical.

The COP function of these processors allows a remote computer system (typically, a PC with dedicated hardware and debugging software) to access and control the internal operations of the processor. The COP interface connects primarily through the JTAG port of the processor, with some additional status monitoring signals. The COP port requires the ability to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$ in order to fully control the processor. If the target system has independent reset sources, such as voltage monitors, watchdog timers, power supply failures, or push-button switches, then the COP reset signals must be merged into these signals with logic.

The arrangement shown in Figure 25 allows the COP port to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$, while ensuring that the target can drive $\overline{\text{HRESET}}$ as well. If the JTAG interface and COP header will not be used, $\overline{\text{TRST}}$ should be tied to $\overline{\text{HRESET}}$ through a 0-Ω isolation resistor so that it is asserted when the system reset signal ($\overline{\text{HRESET}}$) is asserted, ensuring that the JTAG scan chain is initialized during power-on. While Freescale recommends that the COP header be designed into the system as shown in Figure 25, if this is not possible, the isolation resistor will allow future access to $\overline{\text{TRST}}$ in the case where a JTAG interface may need to be wired onto the system in debug situations.

The COP header shown in Figure 25 adds many benefits—breakpoints, watchpoints, register and memory examination/modification, and other standard debugger features are possible through this interface—and can be as inexpensive as an unpopulated footprint for a header to be added when needed.

The COP interface has a standard header for connection to the target system, based on the 0.025" square-post 0.100" centered header assembly (often called a Berg header). The connector typically has pin 14 removed as a connector key.

There is no standardized way to number the COP header shown in Figure 25; consequently, many different pin numbers have been observed from emulator vendors. Some are numbered top-to-bottom then left-to-right, while others use left-to-right then top-to-bottom, while still others number the pins counter clockwise from pin 1 (as with an IC). Regardless of the numbering, the signal placement recommended in Figure 25 is common to all known emulators.

The $\overline{\text{QACK}}$ signal shown in Figure 25 is usually connected to the PCI bridge chip in a system and is an input to the MPC7410 informing it that it can go into the quiescent state. Under normal operation this occurs during a low-power mode selection. In order for COP to work, the MPC7410 must see this signal asserted (pulled down). While shown on the COP header, not all emulator products drive this signal. If the product does not, a pull-down resistor can be populated to assert this signal. Additionally, some emulator products implement open-drain type outputs and can only drive $\overline{\text{QACK}}$ asserted; for these tools, a pull-up resistor can be implemented to ensure this signal is negated when it is not being driven by the tool. Note that the pull-up and pull-down resistors on the $\overline{\text{QACK}}$ signal are mutually exclusive and it is never necessary to populate both in a system. To preserve correct power-down operation, $\overline{\text{QACK}}$ should be merged via logic so that it also can be driven by the PCI bridge.

8.8 Thermal Management Information

This section provides thermal management information for the MPC7410 for air-cooled applications. Proper thermal control design is primarily dependent on the system-level design—the heat sink, airflow, and thermal interface material. To reduce the die-junction temperature, heat sinks may be attached to the package by several methods such as spring clip to holes in the printed circuit board or with screws and springs to the printed circuit board; see Figure 26 for the BGA package and Figure 27 for the LGA package. This spring force should not exceed 5.5 pounds of force. Note that care should be taken to avoid focused forces being applied to die corners and/or edges when mounting heat sinks.

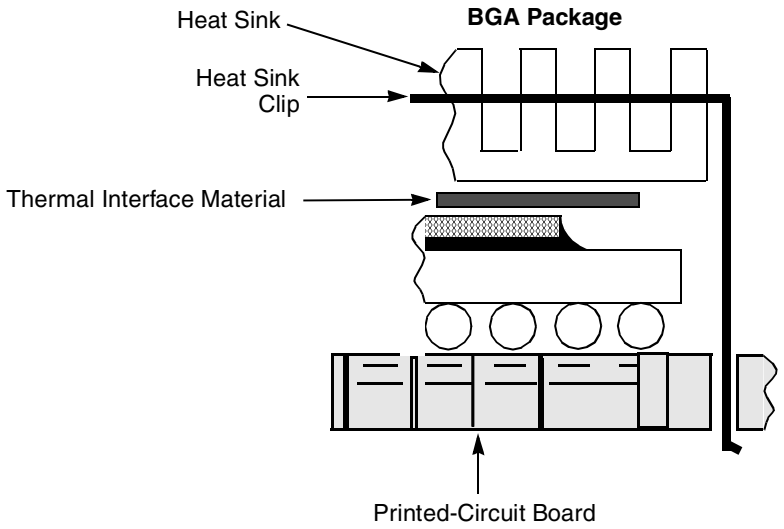


Figure 26. BGA Package Exploded Cross-Sectional View with Heat Sink Clip to PCB Option

Figure 29 describes the thermal performance of selected thermal interface materials.

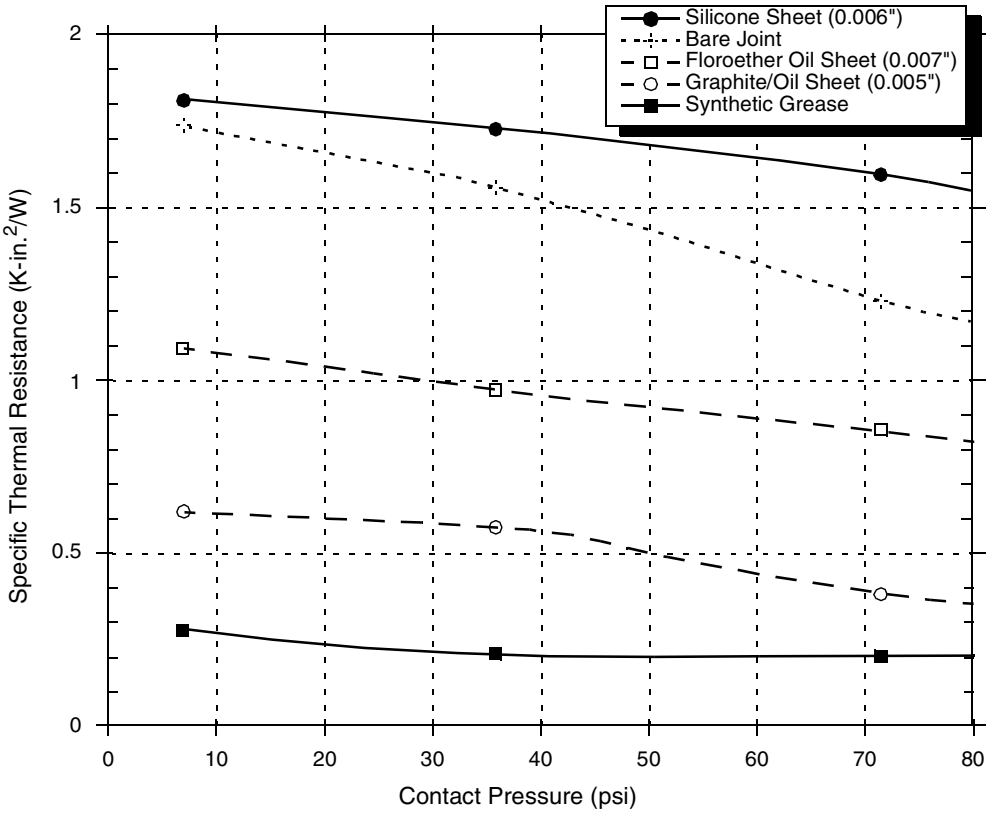


Figure 29. Thermal Performance of Select Thermal Interface Material

The board designer can choose between several types of thermal interface. Heat sink adhesive materials should be selected based on high conductivity, yet adequate mechanical strength to meet equipment shock/vibration requirements. There are several commercially-available thermal interfaces and adhesive materials provided by the following vendors:

Chomerics, Inc. 77 Dragon Court Woburn, MA 01888-4014 Internet: www.chomerics.com	781-935-4850
Dow-Corning Corporation Dow-Corning Electronic Materials 2200 W. Salzburg Rd. Midland, MI 48686-0997 Internet: www.dow.com	800-248-2481
Shin-Etsu MicroSi, Inc. 10028 S. 51st St. Phoenix, AZ 85044 Internet: www.microsi.com	888-642-7674

Thermagon Inc. 888-246-9050
4707 Detroit Ave.
Cleveland, OH 44102
Internet: www.thermagon.com

8.8.3 Heat Sink Selection Example

For preliminary heat sink sizing, the die-junction temperature can be expressed as follows:

$$T_j = T_a + T_r + (\theta_{jc} + \theta_{int} + \theta_{sa}) \times P_d$$

where:

T_j is the die-junction temperature
T_a is the inlet cabinet ambient temperature
T_r is the air temperature rise within the computer cabinet
θ_{jc} is the junction-to-case thermal resistance
θ_{int} is the adhesive or interface material thermal resistance
θ_{sa} is the heat sink base-to-ambient thermal resistance
P_d is the power dissipated by the device

During operation the die-junction temperatures (T_j) should be maintained less than the value specified in Table 3. The temperature of the air cooling the component greatly depends upon the ambient inlet air temperature and the air temperature rise within the electronic cabinet. An electronic cabinet inlet-air temperature (T_a) may range from 30° to 40°C. The air temperature rise within a cabinet (T_r) may be in the range of 5° to 10°C. The thermal resistance of the thermal interface material (θ_{int}) is typically about 1°C/W. Assuming a T_a of 30°C, a T_r of 5°C, a CBGA package θ_{jc} = 0.03, and a power consumption (P_d) of 5.0 W, the following expression for T_j is obtained:

Die-junction temperature: $T_j = 30°C + 5°C + (0.03°C/W + 1.0°C/W + \theta_{sa}) \times 5.0\ W$

For a Thermalloy heat sink #2328B, the heat sink-to-ambient thermal resistance (θ_{sa}) versus airflow velocity is shown in Figure 30.

Assuming an air velocity of 0.5 m/s, we have an effective R_{sa} of 7°C/W, thus

$$T_j = 30°C + 5°C + (0.03°C/W + 1.0°C/W + 7°C/W) \times 5.0\ W,$$

resulting in a die-junction temperature of approximately 75°C which is well within the maximum operating temperature of the component.

Other heat sinks offered by Aavid Thermalloy, Alpha Novatech, The Bergquist Company, IERC, and Wakefield Engineering offer different heat sink-to-ambient thermal resistances, and may or may not need airflow.

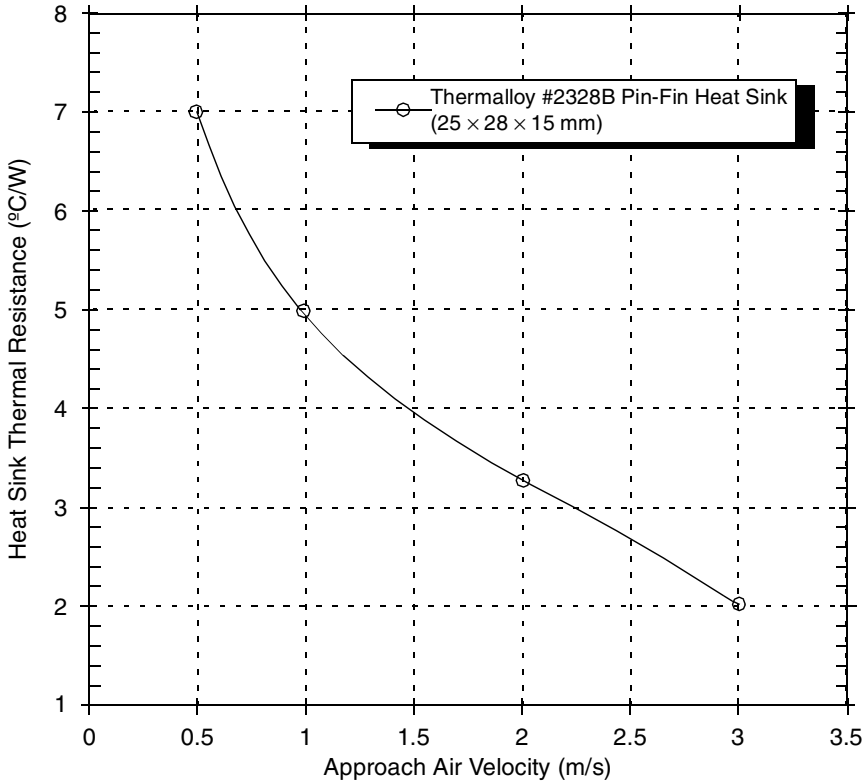


Figure 30. Thermalloy #2328B Heat Sink-to-Ambient Thermal Resistance Versus Airflow Velocity

Though the die junction-to-ambient and the heat sink-to-ambient thermal resistances are a common figure-of-merit used for comparing the thermal performance of various microelectronic packaging technologies, one should exercise caution when only using this metric in determining thermal management because no single parameter can adequately describe three-dimensional heat flow. The final die-junction operating temperature, is not only a function of the component-level thermal resistance, but the system-level design and its operating conditions. In addition to the component's power consumption, a number of factors affect the final operating die-junction temperature—airflow, board population (local heat flux of adjacent components), heat sink efficiency, heat sink attach, heat sink placement, next-level interconnect technology, system air temperature rise, altitude, and so on.

Due to the complexity and the many variations of system-level boundary conditions for today's microelectronic equipment, the combined effects of the heat transfer mechanisms (radiation, convection, and conduction) may vary widely. For these reasons, we recommend using conjugate heat transfer models for the board, as well as, system-level designs.

Table 16. Document Revision History (continued)

Revision	Date	Substantive Change(s)
1.1	—	Internal release.
		Table 12—added note 16 for $\overline{\text{ABB}}$/$\overline{\text{AMON}}$ and $\overline{\text{DBB}}$/$\overline{\text{DMON}}$ signal clarification.
		Table 12—changed $\overline{\text{CHK}}$ note 4 reference to note 2, signal is for factory test only. Changed previous note 4 ($\overline{\text{CHK}}$ related) to now provide additional PLL info.
		Table 1—modified maximum value for OV_{DD} from –0.3 to 3.465 to now be –0.3 to 3.6 and $L2OV_{DD}$ from –0.3 to 2.6 to now be –0.3 to 2.8. Modified note 6, OV_{DD} for revisions prior to Rev. 1.4 have maximum value for OV_{DD} of –0.3 to 2.8.
		Table 8—removed note 12. L2_TSTCLK is for factory use only (see Table 12, note 2).
		Section 1.10.2—revised section to include nomenclature tables for part markings not covered by this spec.
		Figure 2—added that under/overshoot for $L2OV_{DD}$ references t_{L2CLK} while OV_{DD} references t_{SYSCLK}.
		Table 4—added HCTE package (HX package descriptor) thermal characteristics.
		Section 1.5—added HCTE package (HX package descriptor). Both the CBGA and HCTE packages have the same pin assignments.
		Section 1.6—added HCTE package (HX package descriptor). Both the CBGA and HCTE packages have the same pinout listings.
		Section 1.7—added HCTE package (HX package descriptor). Both the CBGA and HCTE packages have the same package parameters and dimensions.
		Table 17—added HCTE package (HX package descriptor) to part numbering nomenclature.
		Table 21—added MPC7410THXnnnLE extended temperature HCTE package part numbers and part number specification document reference.

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