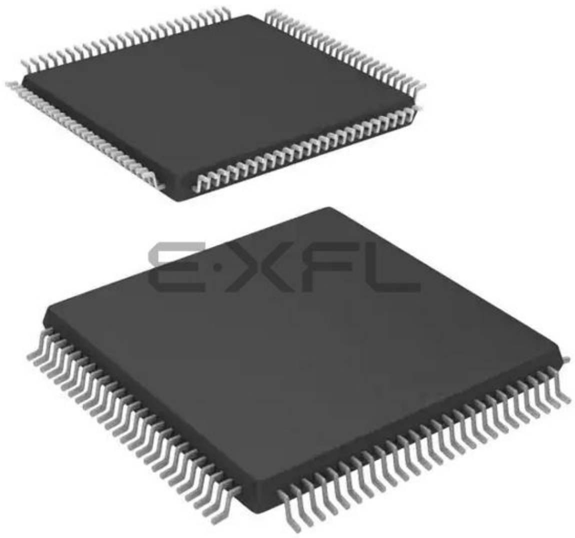
Intel - EP1K10TI100-2N Datasheet



Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	72
Number of Logic Elements/Cells	576
Total RAM Bits	12288
Number of I/O	66
Number of Gates	56000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	100-TQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1k10ti100-2n

…and More Features

  - -1 speed grade devices are compliant with ***PCI Local Bus Specification, Revision 2.2*** for 5.0-V operation
  - Built-in Joint Test Action Group (JTAG) boundary-scan test (BST) circuitry compliant with IEEE Std. 1149.1-1990, available without consuming additional device logic.
  - Operate with a 2.5-V internal supply voltage
  - In-circuit reconfigurability (ICR) via external configuration devices, intelligent controller, or JTAG port
  - ClockLock™ and ClockBoost™ options for reduced clock delay, clock skew, and clock multiplication
  - Built-in, low-skew clock distribution trees
  - 100% functional testing of all devices; test vectors or scan chains are not required
  - Pull-up on I/O pins before and during configuration
- Flexible interconnect
  - FastTrack® Interconnect continuous routing structure for fast, predictable interconnect delays
  - Dedicated carry chain that implements arithmetic functions such as fast adders, counters, and comparators (automatically used by software tools and megafunctions)
  - Dedicated cascade chain that implements high-speed, high-fan-in logic functions (automatically used by software tools and megafunctions)
  - Tri-state emulation that implements internal tri-state buses
  - Up to six global clock signals and four global clear signals
- Powerful I/O pins
  - Individual tri-state output enable control for each pin
  - Open-drain option on each I/O pin
  - Programmable output slew-rate control to reduce switching noise
  - Clamp to V_{CCIO} user-selectable on a pin-by-pin basis
  - Supports hot-socketing

General Description

Altera® ACEX 1K devices provide a die-efficient, low-cost architecture by combining look-up table (LUT) architecture with EABs. LUT-based logic provides optimized performance and efficiency for data-path, register intensive, mathematical, or digital signal processing (DSP) designs, while EABs implement RAM, ROM, dual-port RAM, or first-in first-out (FIFO) functions. These elements make ACEX 1K suitable for complex logic functions and memory functions such as digital signal processing, wide data-path manipulation, data transformation and microcontrollers, as required in high-performance communications applications. Based on reconfigurable CMOS SRAM elements, the ACEX 1K architecture incorporates all features necessary to implement common gate array megafunctions, along with a high pin count to enable an effective interface with system components. The advanced process and the low voltage requirement of the 2.5-V core allow ACEX 1K devices to meet the requirements of low-cost, high-volume applications ranging from DSL modems to low-cost switches.

The ability to reconfigure ACEX 1K devices enables complete testing prior to shipment and allows the designer to focus on simulation and design verification. ACEX 1K device reconfigurability eliminates inventory management for gate array designs and test vector generation for fault coverage.

Table 4 shows ACEX 1K device performance for some common designs. All performance results were obtained with Synopsys DesignWare or LPM functions. Special design techniques are not required to implement the applications; the designer simply infers or instantiates a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Table 4. ACEX 1K Device Performance

Application	Resources Used		Performance			
	LEs	EABs	Speed Grade			Units
			-1	-2	-3	
16-bit loadable counter	16	0	285	232	185	MHz
16-bit accumulator	16	0	285	232	185	MHz
16-to-1 multiplexer *(1)*	10	0	3.5	4.5	6.6	ns
16-bit multiplier with 3-stage pipeline*(2)*	592	0	156	131	93	MHz
256 × 16 RAM read cycle speed *(2)*	0	1	278	196	143	MHz
256 × 16 RAM write cycle speed *(2)*	0	1	185	143	111	MHz

Notes:

(1) This application uses combinatorial inputs and outputs.

(2) This application uses registered inputs and outputs.

The logic array consists of logic array blocks (LABs). Each LAB contains eight LEs and a local interconnect. An LE consists of a 4-input LUT, a programmable flipflop, and dedicated signal paths for carry and cascade functions. The eight LEs can be used to create medium-sized blocks of logic—such as 8-bit counters, address decoders, or state machines—or combined across LABs to create larger logic blocks. Each LAB represents about 96 usable logic gates.

Signal interconnections within ACEX 1K devices (as well as to and from device pins) are provided by the FastTrack Interconnect routing structure, which is a series of fast, continuous row and column channels that run the entire length and width of the device.

Each I/O pin is fed by an I/O element (IOE) located at the end of each row and column of the FastTrack Interconnect routing structure. Each IOE contains a bidirectional I/O buffer and a flipflop that can be used as either an output or input register to feed input, output, or bidirectional signals. When used with a dedicated clock pin, these registers provide exceptional performance. As inputs, they provide setup times as low as 1.1 ns and hold times of 0 ns. As outputs, these registers provide clock-to-output times as low as 2.5 ns. IOEs provide a variety of features, such as JTAG BST support, slew-rate control, tri-state buffers, and open-drain outputs.

Figure 1 shows a block diagram of the ACEX 1K device architecture. Each group of LEs is combined into an LAB; groups of LABs are arranged into rows and columns. Each row also contains a single EAB. The LABs and EABs are interconnected by the FastTrack Interconnect routing structure. IOEs are located at the end of each row and column of the FastTrack Interconnect routing structure.

If necessary, all EABs in a device can be cascaded to form a single RAM block. EABs can be cascaded to form RAM blocks of up to 2,048 words without impacting timing. Altera software automatically combines EABs to meet a designer's RAM specifications.

EABs provide flexible options for driving and controlling clock signals. Different clocks and clock enables can be used for reading and writing to the EAB. Registers can be independently inserted on the data input, EAB output, write address, write enable signals, read address, and read enable signals. The global signals and the EAB local interconnect can drive write-enable, read-enable, and clock-enable signals. The global signals, dedicated clock pins, and EAB local interconnect can drive the EAB clock signals. Because the LEs drive the EAB local interconnect, the LEs can control write-enable, read-enable, clear, clock, and clock-enable signals.

An EAB is fed by a row interconnect and can drive out to row and column interconnects. Each EAB output can drive up to two row channels and up to two column channels; the unused row channel can be driven by other LEs. This feature increases the routing resources available for EAB outputs (see Figures 2 and 4). The column interconnect, which is adjacent to the EAB, has twice as many channels as other columns in the device.

Logic Array Block

An LAB consists of eight LEs, their associated carry and cascade chains, LAB control signals, and the LAB local interconnect. The LAB provides the coarse-grained structure to the ACEX 1K architecture, facilitating efficient routing with optimum device utilization and high performance. Figure 7 shows the ACEX 1K LAB.

Normal Mode

The normal mode is suitable for general logic applications and wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a 4-input LUT. The compiler automatically selects the carry-in or the `DATA3` signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. Either the register or the LUT can be used to drive both the local interconnect and the FastTrack Interconnect routing structure at the same time.

The LUT and the register in the LE can be used independently (register packing). To support register packing, the LE has two outputs; one drives the local interconnect, and the other drives the FastTrack Interconnect routing structure. The `DATA4` signal can drive the register directly, allowing the LUT to compute a function that is independent of the registered signal; a 3-input function can be computed in the LUT, and a fourth independent signal can be registered. Alternatively, a 4-input function can be generated, and one of the inputs to this function can be used to drive the register. The register in a packed LE can still use the clock enable, clear, and preset signals in the LE. In a packed LE, the register can drive the FastTrack Interconnect routing structure while the LUT drives the local interconnect, or vice versa.

Arithmetic Mode

The arithmetic mode offers two 3-input LUTs that are ideal for implementing adders, accumulators, and comparators. One LUT computes a 3-input function; the other generates a carry output. As shown in Figure 11, the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, in an adder, this output is the sum of three signals: `a`, `b`, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

Up/Down Counter Mode

The up/down counter mode offers counter enable, clock enable, synchronous up/down control, and data loading options. These control signals are generated by the data inputs from the LAB local interconnect, the carry-in signal, and output feedback from the programmable register. Two 3-input LUTs are used; one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading. Data can also be loaded asynchronously with the clear and preset register control signals without using the LUT resources.

Asynchronous Clear

The flipflop can be cleared by either `LABCTRL1` or `LABCTRL2`. In this mode, the preset signal is tied to `VCC` to deactivate it.

Asynchronous Preset

An asynchronous preset is implemented as an asynchronous load, or with an asynchronous clear. If `DATA3` is tied to `VCC`, asserting `LABCTRL1` asynchronously loads a one into the register. Alternatively, the Altera software can provide preset control by using the clear and inverting the register's input and output. Inversion control is available for the inputs to both LEs and IOEs. Therefore, if a register is preset by only one of the two `LABCTRL` signals, the `DATA3` input is not needed and can be used for one of the LE operating modes.

Asynchronous Preset & Clear

When implementing asynchronous clear and preset, `LABCTRL1` controls the preset, and `LABCTRL2` controls the clear. `DATA3` is tied to `VCC`, so that asserting `LABCTRL1` asynchronously loads a one into the register, effectively presetting the register. Asserting `LABCTRL2` clears the register.

Asynchronous Load with Clear

When implementing an asynchronous load in conjunction with the clear, `LABCTRL1` implements the asynchronous load of `DATA3` by controlling the register preset and clear. `LABCTRL2` implements the clear by controlling the register clear; `LABCTRL2` does not have to feed the preset circuits.

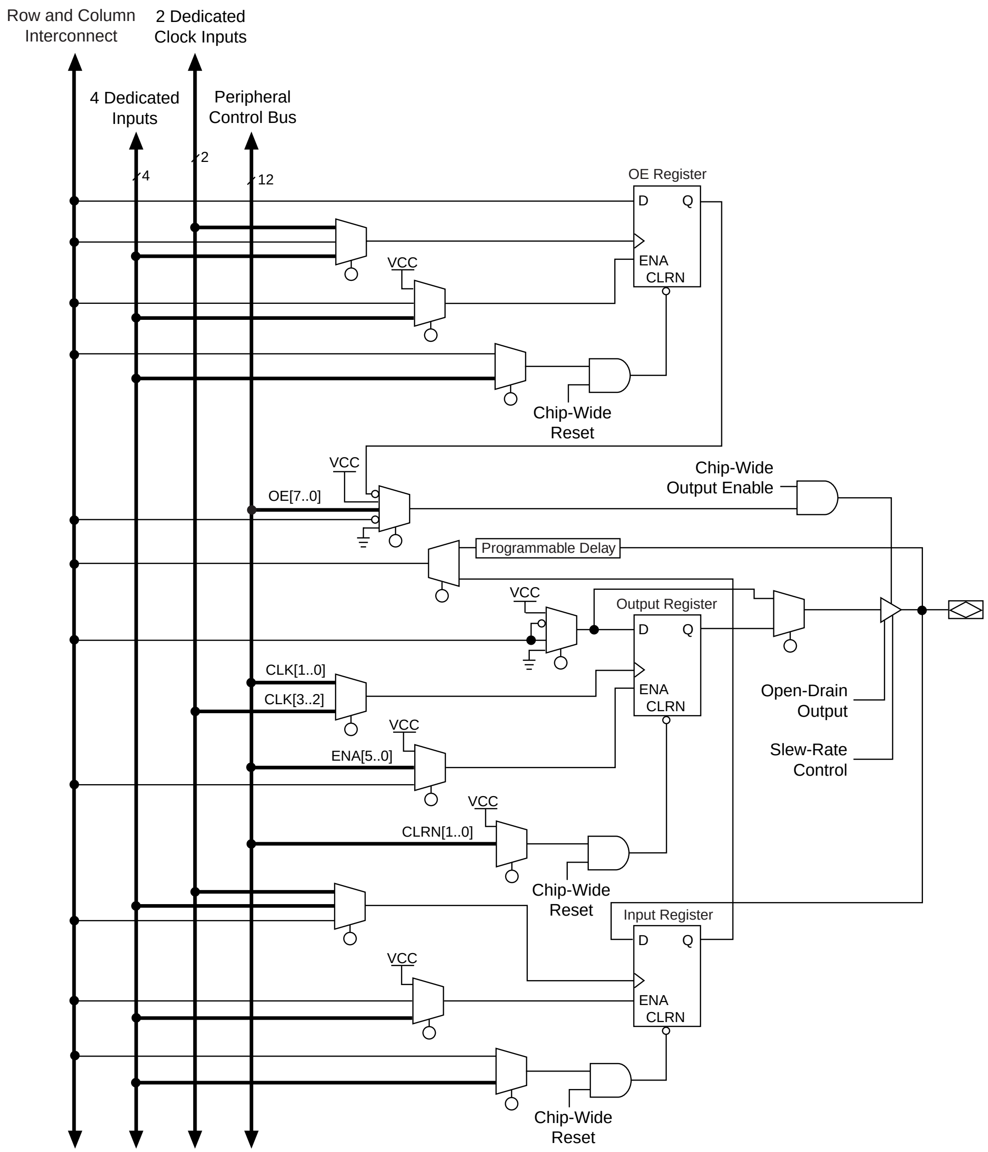
Asynchronous Load with Preset

When implementing an asynchronous load in conjunction with preset, the Altera software provides preset control by using the clear and inverting the input and output of the register. Asserting `LABCTRL2` presets the register, while asserting `LABCTRL1` loads the register. The Altera software inverts the signal that drives `DATA3` to account for the inversion of the register's output.

Asynchronous Load without Preset or Clear

When implementing an asynchronous load without preset or clear, `LABCTRL1` implements the asynchronous load of `DATA3` by controlling the register preset and clear.

Figure 15. ACEX 1K Bidirectional I/O Registers



When dedicated inputs drive non-inverted and inverted peripheral clears, clock enables, and output enables, two signals on the peripheral control bus will be used.

Table 7 lists the sources for each peripheral control signal and shows how the output enable, clock enable, clock, and clear signals share 12 peripheral control signals. Table 7 also shows the rows that can drive global signals.

Table 7. Peripheral Bus Sources for ACEX Devices

Peripheral Control Signal	EP1K10	EP1K30	EP1K50	EP1K100
`OE0`	Row A	Row A	Row A	Row A
`OE1`	Row A	Row B	Row B	Row C
`OE2`	Row B	Row C	Row D	Row E
`OE3`	Row B	Row D	Row F	Row L
`OE4`	Row C	Row E	Row H	Row I
`OE5`	Row C	Row F	Row J	Row K
`CLKENA0/CLK0/GLOBAL0`	Row A	Row A	Row A	Row F
`CLKENA1/OE6/GLOBAL1`	Row A	Row B	Row C	Row D
`CLKENA2/CLR0`	Row B	Row C	Row E	Row B
`CLKENA3/OE7/GLOBAL2`	Row B	Row D	Row G	Row H
`CLKENA4/CLR1`	Row C	Row E	Row I	Row J
`CLKENA5/CLK1/GLOBAL3`	Row C	Row F	Row J	Row G

Signals on the peripheral control bus can also drive the four global signals, referred to as `GLOBAL0` through `GLOBAL3`. An internally generated signal can drive a global signal, providing the same low-skew, low-delay characteristics as a signal driven by an input pin. An LE drives the global signal by driving a row line that drives the peripheral bus which then drives the global signal. This feature is ideal for internally generated clear or clock signals with high fan-out. However, internally driven global signals offer no advantage over the general-purpose interconnect for routing data signals.

The chip-wide output enable pin is an active-high pin that can be used to tri-state all pins on the device. This option can be set in the Altera software. The built-in I/O pin pull-up resistors (which are active during configuration) are active when the chip-wide output enable pin is asserted. The registers in the IOE can also be reset by the chip-wide reset pin.

For more information, search for “SameFrame” in MAX+PLUS II Help.

Table 10. ACEX 1K SameFrame Pin-Out Support

Device	256-Pin FineLine BGA	484-Pin FineLine BGA
EP1K10	✓	*(1)*
EP1K30	✓	*(1)*
EP1K50	✓	✓
EP1K100	✓	✓

Note:

(1) This option is supported with a 256-pin FineLine BGA package and SameFrame migration.

ClockLock & ClockBoost Features

To support high-speed designs, -1 and -2 speed grade ACEX 1K devices offer ClockLock and ClockBoost circuitry containing a phase-locked loop (PLL) that is used to increase design speed and reduce resource usage. The ClockLock circuitry uses a synchronizing PLL that reduces the clock delay and skew within a device. This reduction minimizes clock-to-output and setup times while maintaining zero hold times. The ClockBoost circuitry, which provides a clock multiplier, allows the designer to enhance device area efficiency by sharing resources within the device. The ClockBoost feature allows the designer to distribute a low-speed clock and multiply that clock on-device. Combined, the ClockLock and ClockBoost features provide significant improvements in system performance and bandwidth.

The ClockLock and ClockBoost features in ACEX 1K devices are enabled through the Altera software. External devices are not required to use these features. The output of the ClockLock and ClockBoost circuits is not available at any of the device pins.

The ClockLock and ClockBoost circuitry lock onto the rising edge of the incoming clock. The circuit output can drive the clock inputs of registers only; the generated clock cannot be gated or inverted.

The dedicated clock pin (`GCLK1`) supplies the clock to the ClockLock and ClockBoost circuitry. When the dedicated clock pin is driving the ClockLock or ClockBoost circuitry, it cannot drive elsewhere in the device.

Tables 11 and 12 summarize the ClockLock and ClockBoost parameters for -1 and -2 speed-grade devices, respectively.

Table 11. ClockLock & ClockBoost Parameters for -1 Speed-Grade Devices

Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_R	Input rise time				5	ns
t_F	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		180	MHz
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		90	MHz
f_{CLKDEV}	Input deviation from user specification in the Altera software *(1)*				25,000 *(2)*	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock *(3)*				10	µs
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock *(4)*	$t_{INCLKSTB}$ <100			250 *(4)*	ps
		$t_{INCLKSTB}$ < 50			200 *(4)*	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Table 12. ClockLock & ClockBoost Parameters for -2 Speed-Grade Devices

Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_R	Input rise time				5	ns
t_F	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		80	MHz
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		40	MHz
f_{CLKDEV}	Input deviation from user specification in the software *(1)*				25,000	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock *(3)*				10	µs
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock *(4)*	$t_{INCLKSTB} < 100$			250 *(4)*	ps
		$t_{INCLKSTB} < 50$			200 *(4)*	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Notes to tables:

(1) To implement the ClockLock and ClockBoost circuitry with the Altera software, designers must specify the input frequency. The Altera software tunes the PLL in the ClockLock and ClockBoost circuitry to this frequency. The f_{CLKDEV} parameter specifies how much the incoming clock can differ from the specified frequency during device operation. Simulation does not reflect this parameter.

(2) Twenty-five thousand parts per million (PPM) equates to 2.5% of input clock period.

(3) During device configuration, the ClockLock and ClockBoost circuitry is configured before the rest of the device. If the incoming clock is supplied during configuration, the ClockLock and ClockBoost circuitry locks during configuration because the t_{LOCK} value is less than the time required for configuration.

(4) The t_{JITTER} specification is measured under long-term observation. The maximum value for t_{JITTER} is 200 ps if $t_{INCLKSTB}$ is lower than 50 ps.

I/O Configuration

This section discusses the PCI pull-up clamping diode option, slew-rate control, open-drain output option, and MultiVolt I/O interface for ACEX 1K devices. The PCI pull-up clamping diode, slew-rate control, and open-drain output options are controlled pin-by-pin via Altera software logic options. The MultiVolt I/O interface is controlled by connecting V_{CCIO} to a different voltage than V_{CCINT}. Its effect can be simulated in the Altera software via the **Global Project Device Options** dialog box (Assign menu).

The `VCCINT` pins must always be connected to a 2.5-V power supply. With a 2.5-V V_{CCINT} level, input voltages are compatible with 2.5-V, 3.3-V, and 5.0-V inputs. The `VCCIO` pins can be connected to either a 2.5-V or 3.3-V power supply, depending on the output requirements. When the `VCCIO` pins are connected to a 2.5-V power supply, the output levels are compatible with 2.5-V systems. When the `VCCIO` pins are connected to a 3.3-V power supply, the output high is at 3.3 V and is therefore compatible with 3.3-V or 5.0-V systems. Devices operating with V_{CCIO} levels higher than 3.0 V achieve a faster timing delay of t_{OD2} instead of t_{OD1}.

Table 13 summarizes ACEX 1K MultiVolt I/O support.

Table 13. ACEX 1K MultiVolt I/O Support

V_{CCIO} (V)	Input Signal (V)			Output Signal (V)		
	2.5	**3.3**	**5.0**	**2.5**	**3.3**	**5.0**
2.5	✓	✓ *(1)*	✓ *(1)*	✓		
3.3	✓	✓	✓ *(1)*	✓ *(2)*	✓	✓

Notes:

(1) The PCI clamping diode must be disabled on an input which is driven with a voltage higher than V_{CCIO}.

(2) When V_{CCIO} = 3.3 V, an ACEX 1K device can drive a 2.5-V device that has 3.3-V tolerant inputs.

Open-drain output pins on ACEX 1K devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a higher V_{IH} than LVTTL. When the open-drain pin is active, it will drive low. When the pin is inactive, the resistor will pull up the trace to 5.0 V, thereby meeting the CMOS V_{OH} requirement. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The I_{OL} current specification should be considered when selecting a pull-up resistor.

Power Sequencing & Hot-Socketing

Because ACEX 1K devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The V_{CCIO} and V_{CCINT} power planes can be powered in any order.

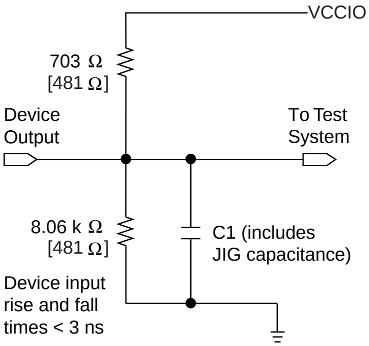
Signals can be driven into ACEX 1K devices before and during power up without damaging the device. Additionally, ACEX 1K devices do not drive out during power up. Once operating conditions are reached, ACEX 1K devices operate as specified by the user.

Generic Testing

Each ACEX 1K device is functionally tested. Complete testing of each configurable static random access memory (SRAM) bit and all logic functionality ensures 100% yield. AC test measurements for ACEX 1K devices are made under conditions equivalent to those shown in Figure 21. Multiple test patterns can be used to configure devices during all stages of the production flow.

Figure 21. ACEX 1K AC Test Conditions

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers in brackets are for 2.5-V devices or outputs. Numbers without brackets are for 3.3-V devices or outputs.



Operating Conditions

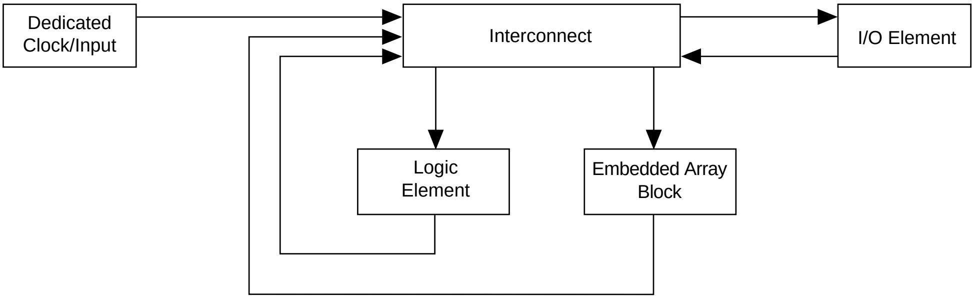
Tables 18 through 21 provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for 2.5-V ACEX 1K devices.

Table 18. ACEX 1K Device Absolute Maximum Ratings Note (1)

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CCINT}	Supply voltage	With respect to ground (2)	–0.5	3.6	V
V_{CCIO}			–0.5	4.6	V
V_I	DC input voltage		–2.0	5.75	V
I_{OUT}	DC output current, per pin		–25	25	mA
T_{STG}	Storage temperature	No bias	–65	150	° C
T_{AMB}	Ambient temperature	Under bias	–65	135	° C
T_J	Junction temperature	PQFP, TQFP, and BGA packages, under bias		135	° C

Figure 24 shows the overall timing model, which maps the possible paths to and from the various elements of the ACEX 1K device.

Figure 24. ACEX 1K Device Timing Model



Figures 25 through 28 show the delays that correspond to various paths and functions within the LE, IOE, EAB, and bidirectional timing models.

Figure 25. ACEX 1K Device LE Timing Model

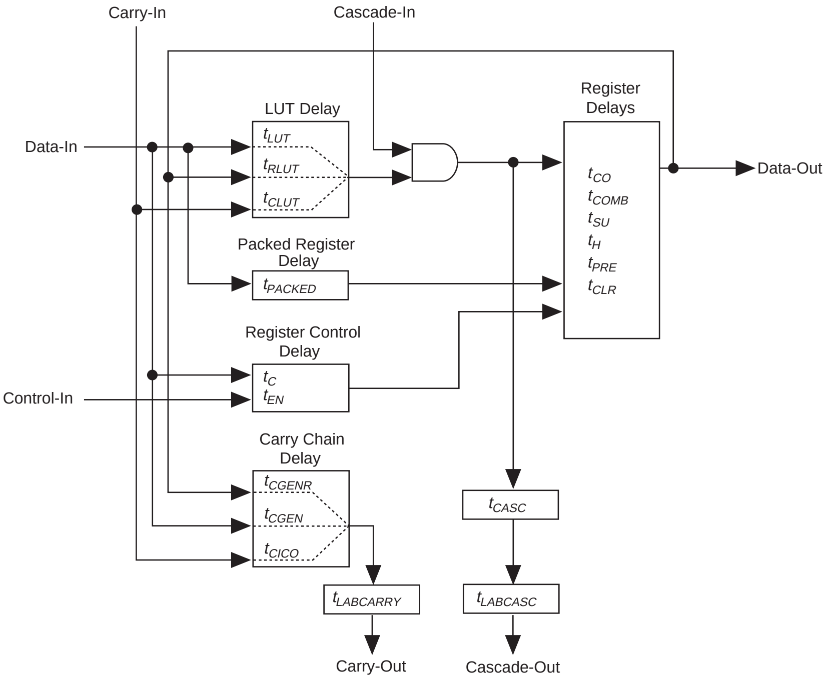


Table 24. EAB Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
$t_{EABDATA1}$	Data or address delay to EAB for combinatorial input	
$t_{EABDATA2}$	Data or address delay to EAB for registered input	
t_{EABWE1}	Write enable delay to EAB for combinatorial input	
t_{EABWE2}	Write enable delay to EAB for registered input	
t_{EABRE1}	Read enable delay to EAB for combinatorial input	
t_{EABRE2}	Read enable delay to EAB for registered input	
t_{EABCLK}	EAB register clock delay	
t_{EABCO}	EAB register clock-to-output delay	
$t_{EABBYPASS}$	Bypass register delay	
t_{EABSU}	EAB register setup time before clock	
t_{EABH}	EAB register hold time after clock	
t_{EABCLR}	EAB register asynchronous clear time to output delay	
t_{AA}	Address access delay (including the read enable to output delay)	
t_{WP}	Write pulse width	
t_{RP}	Read pulse width	
t_{WDSU}	Data setup time before falling edge of write pulse	*(5)*
t_{WDH}	Data hold time after falling edge of write pulse	*(5)*
t_{WASU}	Address setup time before rising edge of write pulse	*(5)*
t_{WAH}	Address hold time after falling edge of write pulse	*(5)*
t_{RASU}	Address setup time before rising edge of read pulse	
t_{RAH}	Address hold time after falling edge of read pulse	
t_{WO}	Write enable to data output valid delay	
t_{DD}	Data-in to data-out valid delay	
t_{EABOUT}	Data-out delay	
t_{EABCH}	Clock high time	
t_{EABCL}	Clock low time	

Table 25. EAB Timing Macroparameters *Notes (1), (6)*

Symbol	Parameter	Conditions
t_{EABAA}	EAB address access delay	
$t_{EABRCCOMB}$	EAB asynchronous read cycle time	
$t_{EABRCREG}$	EAB synchronous read cycle time	
t_{EABWP}	EAB write pulse width	
$t_{EABWCCOMB}$	EAB asynchronous write cycle time	
$t_{EABWCREG}$	EAB synchronous write cycle time	
t_{EABDD}	EAB data-in to data-out valid delay	
$t_{EABDATACO}$	EAB clock-to-output delay when using output registers	
$t_{EABDATASU}$	EAB data/address setup time before clock when using input register	
$t_{EABDATAH}$	EAB data/address hold time after clock when using input register	
$t_{EABWESU}$	EAB `WE` setup time before clock when using input register	
t_{EABWEH}	EAB `WE` hold time after clock when using input register	
$t_{EABWDSU}$	EAB data setup time before falling edge of write pulse when not using input registers	
t_{EABWDH}	EAB data hold time after falling edge of write pulse when not using input registers	
$t_{EABWASU}$	EAB address setup time before rising edge of write pulse when not using input registers	
t_{EABWAH}	EAB address hold time after falling edge of write pulse when not using input registers	
t_{EABWO}	EAB write enable to data output valid delay	

Table 26. Interconnect Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
$t_{DIN2IOE}$	Delay from dedicated input pin to IOE control input	*(7)*
t_{DIN2LE}	Delay from dedicated input pin to LE or EAB control input	*(7)*
$t_{DIN2DATA}$	Delay from dedicated input or clock to LE or EAB data	*(7)*
$t_{DCLK2IOE}$	Delay from dedicated clock pin to IOE clock	*(7)*
$t_{DCLK2LE}$	Delay from dedicated clock pin to LE or EAB clock	*(7)*
$t_{SAMELAB}$	Routing delay for an LE driving another LE in the same LAB	*(7)*
$t_{SAMEROW}$	Routing delay for a row IOE, LE, or EAB driving a row IOE, LE, or EAB in the same row	*(7)*
$t_{SAMECOLUMN}$	Routing delay for an LE driving an IOE in the same column	*(7)*
$t_{DIFFROW}$	Routing delay for a column IOE, LE, or EAB driving an LE or EAB in a different row	*(7)*
$t_{TWOROWS}$	Routing delay for a row IOE or EAB driving an LE or EAB in a different row	*(7)*
$t_{LEPERIPH}$	Routing delay for an LE driving a control signal of an IOE via the peripheral control bus	*(7)*
$t_{LABCARRY}$	Routing delay for the carry-out signal of an LE driving the carry-in signal of a different LE in a different LAB	
$t_{LABCASC}$	Routing delay for the cascade-out signal of an LE driving the cascade-in signal of a different LE in a different LAB	

Notes to tables:

(1) Microparameters are timing delays contributed by individual architectural elements. These parameters cannot be measured explicitly.

(2) Operating conditions: V_{CCIO} = 3.3 V ± 10% for commercial or industrial and extended use in ACEX 1K devices

(3) Operating conditions: V_{CCIO} = 2.5 V ± 5% for commercial or industrial and extended use in ACEX 1K devices.

(4) Operating conditions: V_{CCIO} = 2.5 V or 3.3 V.

(5) Because the RAM in the EAB is self-timed, this parameter can be ignored when the `WE` signal is registered.

(6) EAB macroparameters are internal parameters that can simplify predicting the behavior of an EAB at its boundary; these parameters are calculated by summing selected microparameters.

(7) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.

Table 37. EP1K30 Device LE Timing Microparameters (Part 2 of 2) Note (1)							
Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
t_{COMB}		0.4		0.4		0.6	ns
t_{SU}	0.4		0.6		0.6		ns
t_{H}	0.7		1.0		1.3		ns
t_{PRE}		0.8		0.9		1.2	ns
t_{CLR}		0.8		0.9		1.2	ns
t_{CH}	2.0		2.5		2.5		ns
t_{CL}	2.0		2.5		2.5		ns

Table 38. EP1K30 Device IOE Timing Microparameters Note (1)							
Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
t_{IOD}		2.4		2.8		3.8	ns
t_{IOC}		0.3		0.4		0.5	ns
t_{IOCO}		1.0		1.1		1.6	ns
t_{IOCOMB}		0.0		0.0		0.0	ns
t_{IOSU}	1.2		1.4		1.9		ns
t_{IOH}	0.3		0.4		0.5		ns
t_{IOCLR}		1.0		1.1		1.6	ns
t_{OD1}		1.9		2.3		3.0	ns
t_{OD2}		1.4		1.8		2.5	ns
t_{OD3}		4.4		5.2		7.0	ns
t_{XZ}		2.7		3.1		4.3	ns
t_{ZX1}		2.7		3.1		4.3	ns
t_{ZX2}		2.2		2.6		3.8	ns
t_{ZX3}		5.2		6.0		8.3	ns
t_{INREG}		3.4		4.1		5.5	ns
t_{IOFD}		0.8		1.3		2.4	ns
t_{INCOMB}		0.8		1.3		2.4	ns

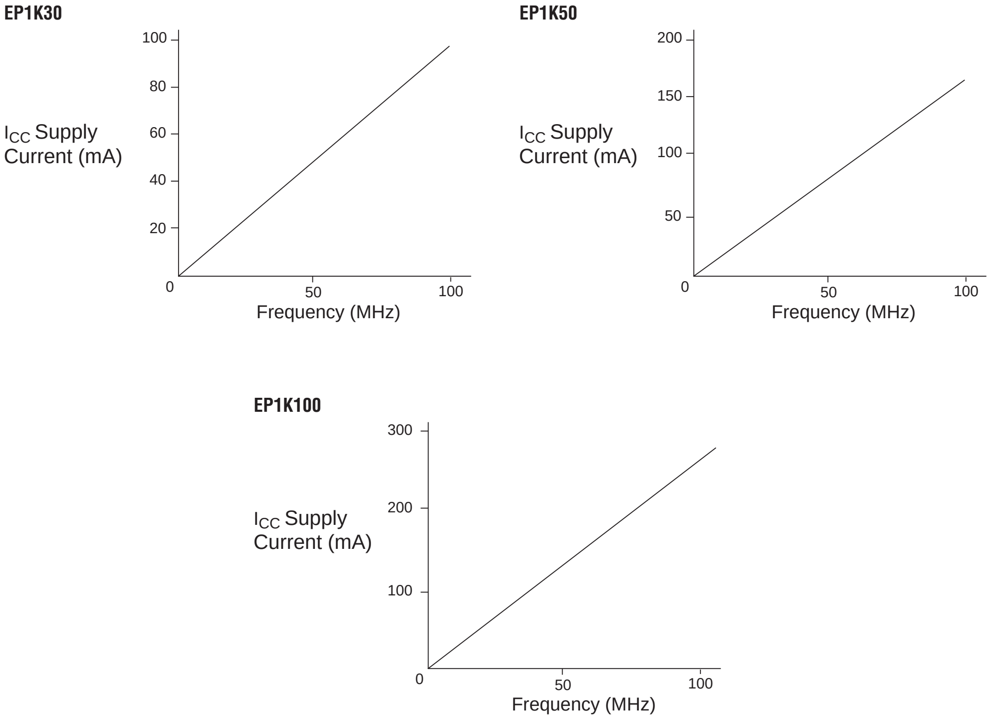
Table 44. EP1K50 Device LE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
t_{CO}		0.6		0.6		0.7	ns
t_{COMB}		0.3		0.4		0.5	ns
t_{SU}	0.5		0.6		0.7		ns
t_{H}	0.5		0.6		0.8		ns
t_{PRE}		0.4		0.5		0.7	ns
t_{CLR}		0.8		1.0		1.2	ns
t_{CH}	2.0		2.5		3.0		ns
t_{CL}	2.0		2.5		3.0		ns

Table 45. EP1K50 Device IOE Timing Microparameters *Note (1)*

Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.3		1.3		1.9	ns
t_{IOC}		0.3		0.4		0.4	ns
t_{IOCO}		1.7		2.1		2.6	ns
t_{IOCOMB}		0.5		0.6		0.8	ns
t_{IOSU}	0.8		1.0		1.3		ns
t_{IOH}	0.4		0.5		0.6		ns
t_{IOCLR}		0.2		0.2		0.4	ns
t_{OD1}		1.2		1.2		1.9	ns
t_{OD2}		0.7		0.8		1.7	ns
t_{OD3}		2.7		3.0		4.3	ns
t_{XZ}		4.7		5.7		7.5	ns
t_{ZX1}		4.7		5.7		7.5	ns
t_{ZX2}		4.2		5.3		7.3	ns
t_{ZX3}		6.2		7.5		9.9	ns
t_{INREG}		3.5		4.2		5.6	ns
t_{IOFD}		1.1		1.3		1.8	ns
t_{INCOMB}		1.1		1.3		1.8	ns

Figure 31. ACEX 1K $I_{CCACTIVE}$ vs. Operating Frequency



Configuration & Operation

The ACEX 1K architecture supports several configuration schemes. This section summarizes the device operating modes and available device configuration schemes.

Operating Modes

The ACEX 1K architecture uses SRAM configuration elements that require configuration data to be loaded every time the circuit powers up. The process of physically loading the SRAM data into the device is called *configuration*. Before configuration, as V_{CC} rises, the device initiates a Power-On Reset (POR). This POR event clears the device and prepares it for configuration. The ACEX 1K POR time does not exceed 50 µs.

When configuring with a configuration device, refer to the relevant configuration device data sheet for POR timing information.