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Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	216
Number of Logic Elements/Cells	1728
Total RAM Bits	24576
Number of I/O	102
Number of Gates	119000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1k30tc144-3n

Email: info@E-XFL.COM

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General Description

Altera® ACEX 1K devices provide a die-efficient, low-cost architecture by combining look-up table (LUT) architecture with EABs. LUT-based logic provides optimized performance and efficiency for data-path, register intensive, mathematical, or digital signal processing (DSP) designs, while EABs implement RAM, ROM, dual-port RAM, or first-in first-out (FIFO) functions. These elements make ACEX 1K suitable for complex logic functions and memory functions such as digital signal processing, wide data-path manipulation, data transformation and microcontrollers, as required in high-performance communications applications. Based on reconfigurable CMOS SRAM elements, the ACEX 1K architecture incorporates all features necessary to implement common gate array megafunctions, along with a high pin count to enable an effective interface with system components. The advanced process and the low voltage requirement of the 2.5-V core allow ACEX 1K devices to meet the requirements of low-cost, high-volume applications ranging from DSL modems to low-cost switches.

The ability to reconfigure ACEX 1K devices enables complete testing prior to shipment and allows the designer to focus on simulation and design verification. ACEX 1K device reconfigurability eliminates inventory management for gate array designs and test vector generation for fault coverage.

Table 4 shows ACEX 1K device performance for some common designs. All performance results were obtained with Synopsys DesignWare or LPM functions. Special design techniques are not required to implement the applications; the designer simply infers or instantiates a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Application	Resources Used			Performance		
	LEs	EABs		Speed Grade		Units
			-1	-2	-3	
16-bit loadable counter	16	0	285	232	185	MHz
16-bit accumulator	16	0	285	232	185	MHz
16-to-1 multiplexer (1)	10	0	3.5	4.5	6.6	ns
16-bit multiplier with 3-stage pipeline(2)	592	0	156	131	93	MHz
256 × 16 RAM read cycle speed (2)	0	1	278	196	143	MHz
256 × 16 RAM write cycle speed (2)	0	1	185	143	111	MHz

Notes:

- This application uses combinatorial inputs and outputs.
- (2) This application uses registered inputs and outputs.

Table 5 shows ACEX 1K device performance for more complex designs. These designs are available as Altera MegaCore $^{\rm TM}$ functions.

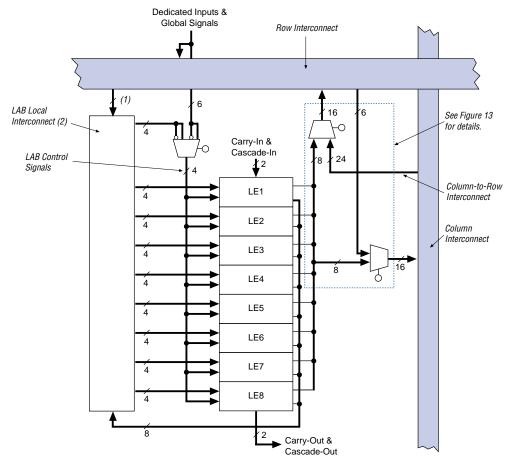
Table 5. ACEX 1K Device Performance for Compl	ex Design	s			
Application	LEs		Perform	ance	
	Used		Speed Grade	!	Units
	·	-1	-2	-3	
16-bit, 8-tap parallel finite impulse response (FIR) filter	597	192	156	116	MSPS
8-bit, 512-point Fast Fourier transform (FFT)	1,854	23.4	28.7	38.9	μs
function		113	92	68	MHz
a16450 universal asynchronous receiver/transmitter (UART)	342	36	28	20.5	MHz

Each ACEX 1K device contains an embedded array and a logic array. The embedded array is used to implement a variety of memory functions or complex logic functions, such as digital signal processing (DSP), wide data-path manipulation, microcontroller applications, and data-transformation functions. The logic array performs the same function as the sea-of-gates in the gate array and is used to implement general logic such as counters, adders, state machines, and multiplexers. The combination of embedded and logic arrays provides the high performance and high density of embedded gate arrays, enabling designers to implement an entire system on a single device.

ACEX 1K devices are configured at system power-up with data stored in an Altera serial configuration device or provided by a system controller. Altera offers EPC16, EPC2, EPC1, and EPC1441 configuration devices, which configure ACEX 1K devices via a serial data stream. Configuration data can also be downloaded from system RAM or via the Altera MasterBlaster $^{\text{TM}}$, ByteBlasterMV $^{\text{TM}}$, or BitBlaster $^{\text{TM}}$ download cables. After an ACEX 1K device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Because reconfiguration requires less than 40 ms, real-time changes can be made during system operation.

ACEX 1K devices contain an interface that permits microprocessors to configure ACEX 1K devices serially or in parallel, and synchronously or asynchronously. The interface also enables microprocessors to treat an ACEX 1K device as memory and configure it by writing to a virtual memory location, simplifying device reconfiguration.

Figure 7. ACEX 1K LAB



Notes:

- (1) EP1K10, EP1K30, and EP1K50 devices have 22 inputs to the LAB local interconnect channel from the row; EP1K100 devices have 26.
- (2) EP1K10, EP1K30, and EP1K50 devices have 30 LAB local interconnect channels; EP1K100 devices have 34.

Each LAB provides four control signals with programmable inversion that can be used in all eight LEs. Two of these signals can be used as clocks, the other two can be used for clear/preset control. The LAB clocks can be driven by the dedicated clock input pins, global signals, I/O signals, or internal signals via the LAB local interconnect. The LAB preset and clear control signals can be driven by the global signals, I/O signals, or internal signals via the LAB local interconnect. The global control signals are typically used for global clock, clear, or preset signals because they provide asynchronous control with very low skew across the device. If logic is required on a control signal, it can be generated in one or more LEs in any LAB and driven into the local interconnect of the target LAB. In addition, the global control signals can be generated from LE outputs.

Logic Element

The LE, the smallest unit of logic in the ACEX 1K architecture, has a compact size that provides efficient logic utilization. Each LE contains a 4-input LUT, which is a function generator that can quickly compute any function of four variables. In addition, each LE contains a programmable flipflop with a synchronous clock enable, a carry chain, and a cascade chain. Each LE drives both the local and the FastTrack Interconnect routing structure. Figure 8 shows the ACEX 1K LE.

Asynchronous Clear

The flipflop can be cleared by either LABCTRL1 or LABCTRL2. In this mode, the preset signal is tied to VCC to deactivate it.

Asynchronous Preset

An asynchronous preset is implemented as an asynchronous load, or with an asynchronous clear. If DATA3 is tied to VCC, asserting LABCTRL1 asynchronously loads a one into the register. Alternatively, the Altera software can provide preset control by using the clear and inverting the register's input and output. Inversion control is available for the inputs to both LEs and IOEs. Therefore, if a register is preset by only one of the two LABCTRL signals, the DATA3 input is not needed and can be used for one of the LE operating modes.

Asynchronous Preset & Clear

When implementing asynchronous clear and preset, LABCTRL1 controls the preset, and LABCTRL2 controls the clear. DATA3 is tied to VCC, so that asserting LABCTRL1 asynchronously loads a one into the register, effectively presetting the register. Asserting LABCTRL2 clears the register.

Asynchronous Load with Clear

When implementing an asynchronous load in conjunction with the clear, LABCTRL1 implements the asynchronous load of DATA3 by controlling the register preset and clear. LABCTRL2 implements the clear by controlling the register clear; LABCTRL2 does not have to feed the preset circuits.

Asynchronous Load with Preset

When implementing an asynchronous load in conjunction with preset, the Altera software provides preset control by using the clear and inverting the input and output of the register. Asserting LABCTRL2 presets the register, while asserting LABCTRL1 loads the register. The Altera software inverts the signal that drives DATA3 to account for the inversion of the register's output.

Asynchronous Load without Preset or Clear

When implementing an asynchronous load without preset or clear, LABCTRL1 implements the asynchronous load of DATA3 by controlling the register preset and clear.

For improved routing, the row interconnect consists of a combination of full-length and half-length channels. The full-length channels connect to all LABs in a row; the half-length channels connect to the LABs in half of the row. The EAB can be driven by the half-length channels in the left half of the row and by the full-length channels. The EAB drives out to the full-length channels. In addition to providing a predictable, row-wide interconnect, this architecture provides increased routing resources. Two neighboring LABs can be connected using a half-row channel, thereby saving the other half of the channel for the other half of the row.

Table 6 summarizes the FastTrack Interconnect routing structure resources available in each ACEX 1K device.

Table 6. ACEX 1	Table 6. ACEX 1K FastTrack Interconnect Resources							
Device	Rows	Channels per Row	Columns	Channels per Column				
EP1K10	3	144	24	24				
EP1K30	6	216	36	24				
EP1K50	10	216	36	24				
EP1K100	12	312	52	24				

In addition to general-purpose I/O pins, ACEX 1K devices have six dedicated input pins that provide low-skew signal distribution across the device. These six inputs can be used for global clock, clear, preset, and peripheral output-enable and clock-enable control signals. These signals are available as control signals for all LABs and IOEs in the device. The dedicated inputs can also be used as general-purpose data inputs because they can feed the local interconnect of each LAB in the device.

Figure 14 shows the interconnection of adjacent LABs and EABs, with row, column, and local interconnects, as well as the associated cascade and carry chains. Each LAB is labeled according to its location: a letter represents the row and a number represents the column. For example, LAB B3 is in row B, column 3.

SameFrame Pin-Outs

ACEX 1K devices support the SameFrame pin-out feature for FineLine BGA packages. The SameFrame pin-out feature is the arrangement of balls on FineLine BGA packages such that the lower-ball-count packages form a subset of the higher-ball-count packages. SameFrame pin-outs provide the flexibility to migrate not only from device to device within the same package, but also from one package to another. A given printed circuit board (PCB) layout can support multiple device density/package combinations. For example, a single board layout can support a range of devices from an EP1K10 device in a 256-pin FineLine BGA package to an EP1K100 device in a 484-pin FineLine BGA package.

The Altera software provides support to design PCBs with SameFrame pin-out devices. Devices can be defined for present and future use. The Altera software generates pin-outs describing how to lay out a board that takes advantage of this migration. Figure 18 shows an example of SameFrame pin-out.

Figure 18. SameFrame Pin-Out Example

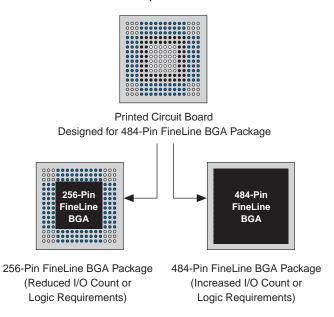


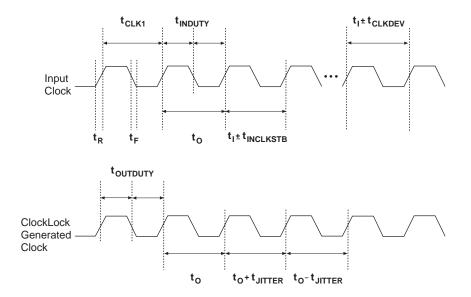
Table 10 shows the ACEX 1K device/package combinations that support SameFrame pin-outs for ACEX 1K devices. All FineLine BGA packages support SameFrame pin-outs, providing the flexibility to migrate not only from device to device within the same package, but also from one package to another. The I/O count will vary from device to device.

For designs that require both a multiplied and non-multiplied clock, the clock trace on the board can be connected to the GCLK1 pin. In the Altera software, the GCLK1 pin can feed both the ClockLock and ClockBoost circuitry in the ACEX 1K device. However, when both circuits are used, the other clock pin cannot be used.

ClockLock & ClockBoost Timing Parameters

For the ClockLock and ClockBoost circuitry to function properly, the incoming clock must meet certain requirements. If these specifications are not met, the circuitry may not lock onto the incoming clock, which generates an erroneous clock within the device. The clock generated by the ClockLock and ClockBoost circuitry must also meet certain specifications. If the incoming clock meets these requirements during configuration, the ClockLock and ClockBoost circuitry will lock onto the clock during configuration. The circuit will be ready for use immediately after configuration. Figure 19 shows the incoming and generated clock specifications.

Figure 19. Specifications for the Incoming & Generated Clocks Note (1)



Note:

(1) The $\mathbf{t_I}$ parameter refers to the nominal input clock period; the $\mathbf{t_O}$ parameter refers to the nominal output clock period.

Figure 20. ACEX 1K JTAG Waveforms

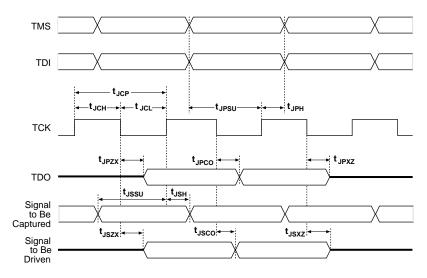


Table 17 shows the timing parameters and values for ACEX 1K devices.

Symbol	Parameter	Min	Max	Unit
t _{JCP}	TCK clock period	100		ns
t _{JCH}	TCK clock high time	50		ns
t _{JCL}	TCK clock low time	50		ns
t _{JPSU}	JTAG port setup time	20		ns
t _{JPH}	JTAG port hold time	45		ns
t _{JPCO}	JTAG port clock to output		25	ns
t _{JPZX}	JTAG port high impedance to valid output		25	ns
t _{JPXZ}	JTAG port valid output to high impedance		25	ns
t _{JSSU}	Capture register setup time	20		ns
t _{JSH}	Capture register hold time	45		ns
t _{JSCO}	Update register clock to output		35	ns
t _{JSZX}	Update register high impedance to valid output		35	ns
t _{JSXZ}	Update register valid output to high impedance		35	ns

Figure 22 shows the required relationship between V_{CCIO} and V_{CCINT} to satisfy 3.3-V PCI compliance.

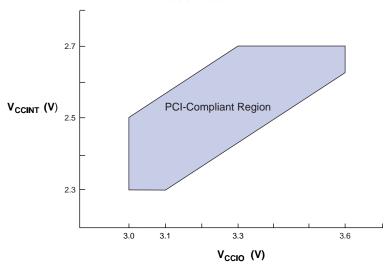
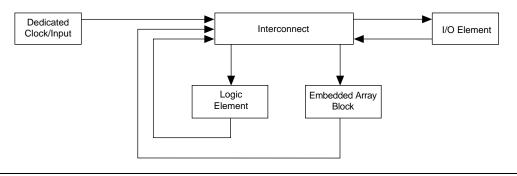


Figure 22. Relationship between V_{CCIO} & V_{CCINT} for 3.3-V PCI Compliance

Figure 23 shows the typical output drive characteristics of ACEX 1K devices with 3.3-V and 2.5-V $V_{\rm CCIO}$. The output driver is compliant to the 3.3-V *PCI Local Bus Specification, Revision 2.2* (when VCCIO pins are connected to 3.3 V). ACEX 1K devices with a -1 speed grade also comply with the drive strength requirements of the *PCI Local Bus Specification, Revision 2.2* (when VCCINT pins are powered with a minimum supply of 2.375 V, and VCCIO pins are connected to 3.3 V). Therefore, these devices can be used in open 5.0-V PCI systems.

Figure 24 shows the overall timing model, which maps the possible paths to and from the various elements of the ACEX 1K device.

Figure 24. ACEX 1K Device Timing Model



Figures 25 through 28 show the delays that correspond to various paths and functions within the LE, IOE, EAB, and bidirectional timing models.

Figure 25. ACEX 1K Device LE Timing Model

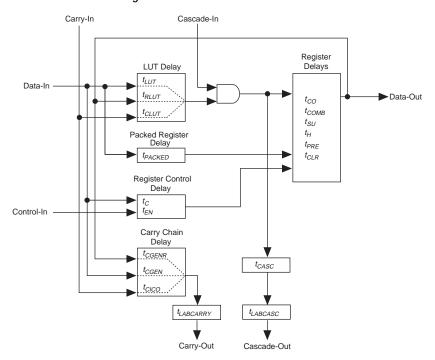


Table 22. LE	Timing Microparameters (Part 2 of 2) Note (1)	
Symbol	Parameter	Conditions
t _{CASC}	Cascade-in to cascade-out delay	
t_{C}	LE register control signal delay	
t_{CO}	LE register clock-to-output delay	
t _{COMB}	Combinatorial delay	
t _{SU}	LE register setup time for data and enable signals before clock; LE register recovery time after asynchronous clear, preset, or load	
t_H	LE register hold time for data and enable signals after clock	
t _{PRE}	LE register preset delay	
t _{CLR}	LE register clear delay	
t _{CH}	Minimum clock high time from clock pin	
t_{CL}	Minimum clock low time from clock pin	

Table 23. 10	E Timing Microparameters Note (1)	
Symbol	Parameter	Conditions
t_{IOD}	IOE data delay	
t_{IOC}	IOE register control signal delay	
t _{IOCO}	IOE register clock-to-output delay	
t _{IOCOMB}	IOE combinatorial delay	
t _{IOSU}	IOE register setup time for data and enable signals before clock; IOE register recovery time after asynchronous clear	
t _{IOH}	IOE register hold time for data and enable signals after clock	
t _{IOCLR}	IOE register clear time	
t _{OD1}	Output buffer and pad delay, slow slew rate = off, V _{CCIO} = 3.3 V	C1 = 35 pF (2)
t _{OD2}	Output buffer and pad delay, slow slew rate = off, V _{CCIO} = 2.5 V	C1 = 35 pF (3)
t _{OD3}	Output buffer and pad delay, slow slew rate = on	C1 = 35 pF (4)
t_{XZ}	IOE output buffer disable delay	
t_{ZX1}	IOE output buffer enable delay, slow slew rate = off, V _{CCIO} = 3.3 V	C1 = 35 pF (2)
t_{ZX2}	IOE output buffer enable delay, slow slew rate = off, V _{CCIO} = 2.5 V	C1 = 35 pF (3)
t _{ZX3}	IOE output buffer enable delay, slow slew rate = on	C1 = 35 pF (4)
t _{INREG}	IOE input pad and buffer to IOE register delay	
t _{IOFD}	IOE register feedback delay	
t _{INCOMB}	IOE input pad and buffer to FastTrack Interconnect delay	

Symbol	Parameter	Conditions
t _{EABDATA1}	Data or address delay to EAB for combinatorial input	
t _{EABDATA2}	Data or address delay to EAB for registered input	
t _{EABWE1}	Write enable delay to EAB for combinatorial input	
t _{EABWE2}	Write enable delay to EAB for registered input	
t _{EABRE1}	Read enable delay to EAB for combinatorial input	
t _{EABRE2}	Read enable delay to EAB for registered input	
t _{EABCLK}	EAB register clock delay	
t _{EABCO}	EAB register clock-to-output delay	
t _{EABBYPASS}	Bypass register delay	
t _{EABSU}	EAB register setup time before clock	
t _{EABH}	EAB register hold time after clock	
t _{EABCLR}	EAB register asynchronous clear time to output delay	
t_{AA}	Address access delay (including the read enable to output delay)	
t_{WP}	Write pulse width	
t_{RP}	Read pulse width	
t _{WDSU}	Data setup time before falling edge of write pulse	(5)
t _{WDH}	Data hold time after falling edge of write pulse	(5)
t _{WASU}	Address setup time before rising edge of write pulse	(5)
t _{WAH}	Address hold time after falling edge of write pulse	(5)
t _{RASU}	Address setup time before rising edge of read pulse	
t _{RAH}	Address hold time after falling edge of read pulse	
t_{WO}	Write enable to data output valid delay	
t_{DD}	Data-in to data-out valid delay	
t _{EABOUT}	Data-out delay	
t _{EABCH}	Clock high time	
t _{EABCL}	Clock low time	

Table 26. Inte	erconnect Timing Microparameters Note (1)	
Symbol	Parameter	Conditions
t _{DIN2IOE}	Delay from dedicated input pin to IOE control input	(7)
t _{DIN2LE}	Delay from dedicated input pin to LE or EAB control input	(7)
t _{DIN2DATA}	Delay from dedicated input or clock to LE or EAB data	(7)
t _{DCLK2IOE}	Delay from dedicated clock pin to IOE clock	(7)
t _{DCLK2LE}	Delay from dedicated clock pin to LE or EAB clock	(7)
t _{SAMELAB}	Routing delay for an LE driving another LE in the same LAB	(7)
t _{SAMEROW}	Routing delay for a row IOE, LE, or EAB driving a row IOE, LE, or EAB in the same row	(7)
t _{SAME} COLUMN	Routing delay for an LE driving an IOE in the same column	(7)
t _{DIFFROW}	Routing delay for a column IOE, LE, or EAB driving an LE or EAB in a different row	(7)
t _{TWOROWS}	Routing delay for a row IOE or EAB driving an LE or EAB in a different row	(7)
t _{LEPERIPH}	Routing delay for an LE driving a control signal of an IOE via the peripheral control bus	(7)
t _{LABCARRY}	Routing delay for the carry-out signal of an LE driving the carry-in signal of a different LE in a different LAB	
t _{LABCASC}	Routing delay for the cascade-out signal of an LE driving the cascade-in signal of a different LE in a different LAB	

Notes to tables:

- Microparameters are timing delays contributed by individual architectural elements. These parameters cannot be measured explicitly.
- Operating conditions: $V_{CCIO} = 3.3 \text{ V} \pm 10\%$ for commercial or industrial and extended use in ACEX 1K devices
- Operating conditions: $V_{CCIO} = 2.5 \text{ V} \pm 5\%$ for commercial or industrial and extended use in ACEX 1K devices. Operating conditions: $V_{CCIO} = 2.5 \text{ V} \text{ or } 3.3 \text{ V}$. (3)
- (4)
- Because the RAM in the EAB is self-timed, this parameter can be ignored when the WE signal is registered.
- EAB macroparameters are internal parameters that can simplify predicting the behavior of an EAB at its boundary; these parameters are calculated by summing selected microparameters.
- These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.

Symbol	Speed Grade							
	-1		-2		-3			
	Min	Max	Min	Max	Min	Max		
t _{DIN2IOE}		2.3		2.7		3.6	ns	
t _{DIN2LE}		0.8		1.1		1.4	ns	
t _{DIN2DATA}		1.1		1.4		1.8	ns	
t _{DCLK2IOE}		2.3		2.7		3.6	ns	
t _{DCLK2LE}		0.8		1.1		1.4	ns	
t _{SAMELAB}		0.1		0.1		0.2	ns	
t _{SAMEROW}		1.8		2.1		2.9	ns	
t _{SAME} COLUMN		0.3		0.4		0.7	ns	
t _{DIFFROW}		2.1		2.5		3.6	ns	
t _{TWOROWS}		3.9		4.6		6.5	ns	
t _{LEPERIPH}		3.3		3.7		4.8	ns	
t _{LABCARRY}		0.3		0.4		0.5	ns	
t _{LABCASC}		0.9		1.0		1.4	ns	

Table 35. EP1K10	External Til	ming Param	eters No	te (1)			
Symbol			Speed	Grade			Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
t _{DRR}		7.5		9.5		12.5	ns
t _{INSU} (2), (3)	2.4		2.7		3.6		ns
t _{INH} (2), (3)	0.0		0.0		0.0		ns
t _{оитсо} (2), (3)	2.0	6.6	2.0	7.8	2.0	9.6	ns
t _{INSU} (4), (3)	1.4		1.7		-		ns
t _{INH} (4), (3)	0.5	5.1	0.5	6.4	_	_	ns
t _{оитсо} (4), (3)	0.0		0.0		-		ns
t _{PCISU} (3)	3.0		4.2		6.4		ns
t _{PCIH} (3)	0.0		0.0		_		ns
t _{PCICO} (3)	2.0	6.0	2.0	7.5	2.0	10.2	ns

Symbol	Speed Grade							
	-1		-	-2		3		
	Min	Max	Min	Max	Min	Max		
t_{IOD}		1.7		2.0		2.6	ns	
t _{IOC}		0.0		0.0		0.0	ns	
t _{IOCO}		1.4		1.6		2.1	ns	
t _{IOCOMB}		0.5		0.7		0.9	ns	
t _{IOSU}	0.8		1.0		1.3		ns	
t _{IOH}	0.7		0.9		1.2		ns	
t _{IOCLR}		0.5		0.7		0.9	ns	
t _{OD1}		3.0		4.2		5.6	ns	
t _{OD2}		3.0		4.2		5.6	ns	
t _{OD3}		4.0		5.5		7.3	ns	
t_{XZ}		3.5		4.6		6.1	ns	
t_{ZX1}		3.5		4.6		6.1	ns	
t_{ZX2}		3.5		4.6		6.1	ns	
t_{ZX3}		4.5		5.9		7.8	ns	
t _{INREG}		2.0		2.6		3.5	ns	
t _{IOFD}		0.5		0.8		1.2	ns	
t _{INCOMB}		0.5		0.8		1.2	ns	

Symbol	Speed Grade							
	-1		-2		-3			
	Min	Max	Min	Max	Min	Max		
t _{EABDATA1}		1.5		2.0		2.6	ns	
t _{EABDATA1}		0.0		0.0		0.0	ns	
t _{EABWE1}		1.5		2.0		2.6	ns	
t _{EABWE2}		0.3		0.4		0.5	ns	
t _{EABRE1}		0.3		0.4		0.5	ns	
t _{EABRE2}		0.0		0.0		0.0	ns	
t _{EABCLK}		0.0		0.0		0.0	ns	
t _{EABCO}		0.3		0.4		0.5	ns	
t _{EABBYPASS}		0.1		0.1		0.2	ns	
t _{EABSU}	0.8		1.0		1.4		ns	
t _{EABH}	0.1		0.1		0.2		ns	
t _{EABCLR}	0.3		0.4		0.5		ns	
t_{AA}		4.0		5.1		6.6	ns	
t_{WP}	2.7		3.5		4.7		ns	
t _{RP}	1.0		1.3		1.7		ns	
t _{WDSU}	1.0		1.3		1.7		ns	
t_{WDH}	0.2		0.2		0.3		ns	
t _{WASU}	1.6		2.1		2.8		ns	
t _{WAH}	1.6		2.1		2.8		ns	
t _{RASU}	3.0		3.9		5.2		ns	
t _{RAH}	0.1		0.1		0.2		ns	
t_{WO}		1.5		2.0		2.6	ns	
t_{DD}		1.5		2.0		2.6	ns	
t _{EABOUT}		0.2		0.3		0.3	ns	
t _{EABCH}	1.5		2.0		2.5		ns	
t _{EABCL}	2.7		3.5		4.7		ns	

Symbol	Speed Grade						
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
t _{INSUBIDIR} (3)	1.7		2.5		3.3		ns
t _{INHBIDIR} (3)	0.0		0.0		0.0		ns
t _{INSUBIDIR} (4)	2.0		2.8		-		ns
t _{INHBIDIR} (4)	0.0		0.0		-		ns
toutcobidir (3)	2.0	5.2	2.0	6.9	2.0	9.1	ns
t _{XZBIDIR} (3)		5.6		7.5		10.1	ns
t _{ZXBIDIR} (3)		5.6		7.5		10.1	ns
toutcobidir (4)	0.5	3.0	0.5	4.6	-	-	ns
t _{XZBIDIR} (4)		4.6		6.5		-	ns
t _{ZXBIDIR} (4)		4.6		6.5		_	ns

Notes to tables:

- (1) All timing parameters are described in Tables 22 through 29 in this data sheet.
- (2) These parameters are specified by characterization.
- (3) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
- (4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Power Consumption

The supply power (P) for ACEX 1K devices can be calculated with the following equation:

$$P = P_{INT} + P_{IO} = (I_{CCSTANDBY} + I_{CCACTIVE}) \times V_{CC} + P_{IO}$$

The $I_{CCACTIVE}$ value depends on the switching frequency and the application logic. This value is calculated based on the amount of current that each LE typically consumes. The P_{IO} value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.



Compared to the rest of the device, the embedded array consumes a negligible amount of power. Therefore, the embedded array can be ignored when calculating supply current.

During initialization, which occurs immediately after configuration, the device resets registers, enables I/O pins, and begins to operate as a logic device. Before and during configuration, all I/O pins (except dedicated inputs, clock, or configuration pins) are pulled high by a weak pull-up resistor. Together, the configuration and initialization processes are called *command mode*; normal device operation is called *user mode*.

SRAM configuration elements allow ACEX 1K devices to be reconfigured in-circuit by loading new configuration data into the device. Real-time reconfiguration is performed by forcing the device into command mode with a device pin, loading different configuration data, re-initializing the device, and resuming user-mode operation. The entire reconfiguration process requires less than 40 ms and can be used to reconfigure an entire system dynamically. In-field upgrades can be performed by distributing new configuration files.

Configuration Schemes

The configuration data for an ACEX 1K device can be loaded with one of five configuration schemes (see Table 59), chosen on the basis of the target application. An EPC16, EPC2, EPC1, or EPC1441 configuration device, intelligent controller, or the JTAG port can be used to control the configuration of a ACEX 1K device, allowing automatic configuration on system power-up.

Multiple ACEX 1K devices can be configured in any of the five configuration schemes by connecting the configuration enable (nCE) and configuration enable output (nCEO) pins on each device. Additional APEX 20K, APEX 20KE, FLEX 10K, FLEX 10KA, FLEX 10KE, ACEX 1K, and FLEX 6000 devices can be configured in the same serial chain.

Table 59. Data Sources for ACEX 1K Configuration				
Configuration Scheme	Data Source			
Configuration device	EPC16, EPC2, EPC1, or EPC1441 configuration device			
Passive serial (PS)	BitBlaster or ByteBlasterMV download cables, or serial data source			
Passive parallel asynchronous (PPA)	Parallel data source			
Passive parallel synchronous (PPS)	Parallel data source			
JTAG	BitBlaster or ByteBlasterMV download cables, or microprocessor with a Jam STAPL File or JBC File			

Device Pin-Outs

See the Altera web site (http://www.altera.com) or the *Altera Documentation Library* for pin-out information.



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