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### Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

### Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

#### Details

Product Status	Obsolete
Number of LABs/CLBs	360
Number of Logic Elements/Cells	2880
Total RAM Bits	40960
Number of I/O	147
Number of Gates	199000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/intel/ep1k50qc208-2">https://www.e-xfl.com/product-detail/intel/ep1k50qc208-2</a>

- Software design support and automatic place-and-route provided by Altera development systems for Windows-based PCs and Sun SPARCstation, and HP 9000 Series 700/800 workstations
- Flexible package options are available in 100 to 484 pins, including the innovative FineLine BGA™ packages (see [Tables 2 and 3](#))
- Additional design entry and simulation support provided by EDIF 2 0 0 and 3 0 0 netlist files, library of parameterized modules (LPM), DesignWare components, Verilog HDL, VHDL, and other interfaces to popular EDA tools from manufacturers such as Cadence, Exemplar Logic, Mentor Graphics, OrCAD, Synopsys, Synplcity, VeriBest, and Viewlogic

**Table 2. ACEX 1K Package Options & I/O Pin Count** *Notes (1), (2)*

Device	100-Pin TQFP	144-Pin TQFP	208-Pin PQFP	256-Pin FineLine BGA	484-Pin FineLine BGA
EP1K10	66	92	120	136	136 (3)
EP1K30		102	147	171	171 (3)
EP1K50		102	147	186	249
EP1K100			147	186	333

**Notes:**

- (1) ACEX 1K device package types include thin quad flat pack (TQFP), plastic quad flat pack (PQFP), and FineLine BGA packages.
- (2) Devices in the same package are pin-compatible, although some devices have more I/O pins than others. When planning device migration, use the I/O pins that are common to all devices.
- (3) This option is supported with a 256-pin FineLine BGA package. By using SameFrame™ pin migration, all FineLine BGA packages are pin-compatible. For example, a board can be designed to support 256-pin and 484-pin FineLine BGA packages.

**Table 3. ACEX 1K Package Sizes**

Device	100-Pin TQFP	144-Pin TQFP	208-Pin PQFP	256-Pin FineLine BGA	484-Pin FineLine BGA
Pitch (mm)	0.50	0.50	0.50	1.0	1.0
Area (mm <sup>2</sup> )	256	484	936	289	529
Length × width (mm × mm)	16 × 16	22 × 22	30.6 × 30.6	17 × 17	23 × 23

Table 5 shows ACEX 1K device performance for more complex designs. These designs are available as Altera MegaCore™ functions.

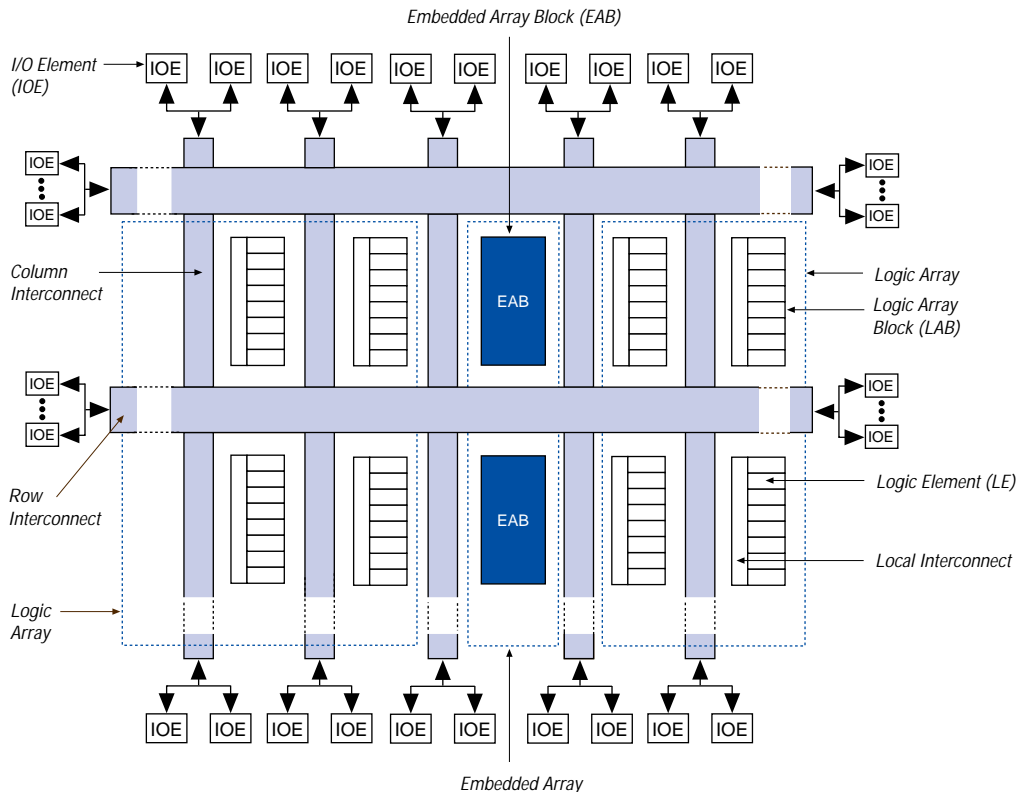
Table 5. ACEX 1K Device Performance for Complex Designs					
Application	LEs Used	Performance			
		Speed Grade			Units
		-1	-2	-3	
16-bit, 8-tap parallel finite impulse response (FIR) filter	597	192	156	116	MSPS
8-bit, 512-point Fast Fourier transform (FFT) function	1,854	23.4	28.7	38.9	μs
		113	92	68	MHz
a16450 universal asynchronous receiver/transmitter (UART)	342	36	28	20.5	MHz

Each ACEX 1K device contains an embedded array and a logic array. The embedded array is used to implement a variety of memory functions or complex logic functions, such as digital signal processing (DSP), wide data-path manipulation, microcontroller applications, and data-transformation functions. The logic array performs the same function as the sea-of-gates in the gate array and is used to implement general logic such as counters, adders, state machines, and multiplexers. The combination of embedded and logic arrays provides the high performance and high density of embedded gate arrays, enabling designers to implement an entire system on a single device.

ACEX 1K devices are configured at system power-up with data stored in an Altera serial configuration device or provided by a system controller. Altera offers EPC16, EPC2, EPC1, and EPC1441 configuration devices, which configure ACEX 1K devices via a serial data stream. Configuration data can also be downloaded from system RAM or via the Altera MasterBlaster™, ByteBlasterMV™, or BitBlaster™ download cables. After an ACEX 1K device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Because reconfiguration requires less than 40 ms, real-time changes can be made during system operation.

ACEX 1K devices contain an interface that permits microprocessors to configure ACEX 1K devices serially or in parallel, and synchronously or asynchronously. The interface also enables microprocessors to treat an ACEX 1K device as memory and configure it by writing to a virtual memory location, simplifying device reconfiguration.

Figure 1. ACEX 1K Device Block Diagram



ACEX 1K devices provide six dedicated inputs that drive the flipflops' control inputs and ensure the efficient distribution of high-speed, low-skew (less than 1.0 ns) control signals. These signals use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect routing structure. Four of the dedicated inputs drive four global signals. These four global signals can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device.

### Normal Mode

The normal mode is suitable for general logic applications and wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a 4-input LUT. The compiler automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. Either the register or the LUT can be used to drive both the local interconnect and the FastTrack Interconnect routing structure at the same time.

The LUT and the register in the LE can be used independently (register packing). To support register packing, the LE has two outputs; one drives the local interconnect, and the other drives the FastTrack Interconnect routing structure. The DATA4 signal can drive the register directly, allowing the LUT to compute a function that is independent of the registered signal; a 3-input function can be computed in the LUT, and a fourth independent signal can be registered. Alternatively, a 4-input function can be generated, and one of the inputs to this function can be used to drive the register. The register in a packed LE can still use the clock enable, clear, and preset signals in the LE. In a packed LE, the register can drive the FastTrack Interconnect routing structure while the LUT drives the local interconnect, or vice versa.

### Arithmetic Mode

The arithmetic mode offers two 3-input LUTs that are ideal for implementing adders, accumulators, and comparators. One LUT computes a 3-input function; the other generates a carry output. As shown in [Figure 11](#), the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, in an adder, this output is the sum of three signals: a, b, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

### Up/Down Counter Mode

The up/down counter mode offers counter enable, clock enable, synchronous up/down control, and data loading options. These control signals are generated by the data inputs from the LAB local interconnect, the carry-in signal, and output feedback from the programmable register. Two 3-input LUTs are used; one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading. Data can also be loaded asynchronously with the clear and preset register control signals without using the LUT resources.

### Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but it supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Two 3-input LUTs are used; one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is ANDed with a synchronous clear signal.

### *Internal Tri-State Emulation*

Internal tri-state emulation provides internal tri-states without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

### *Clear & Preset Logic Control*

Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

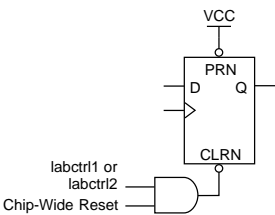
The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

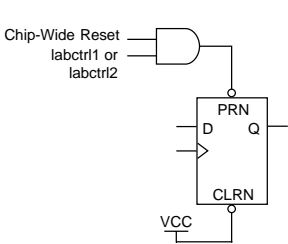
In addition to the six clear and preset modes, ACEX 1K devices provide a chip-wide reset pin that can reset all registers in the device. Use of this feature is set during design entry. In any of the clear and preset modes, the chip-wide reset overrides all other signals. Registers with asynchronous presets may be preset when the chip-wide reset is asserted. Inversion can be used to implement the asynchronous preset. Figure 12 shows examples of how to setup the preset and clear inputs for the desired functionality.

Figure 12. ACEX 1K LE Clear & Preset Modes

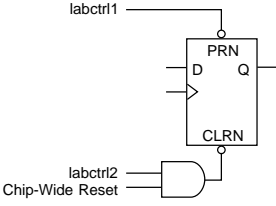
Asynchronous Clear



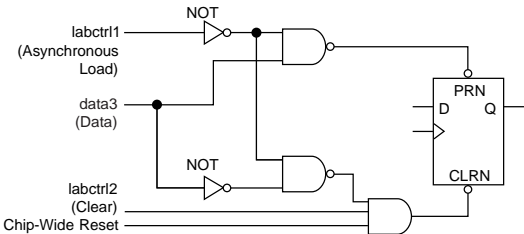
Asynchronous Preset



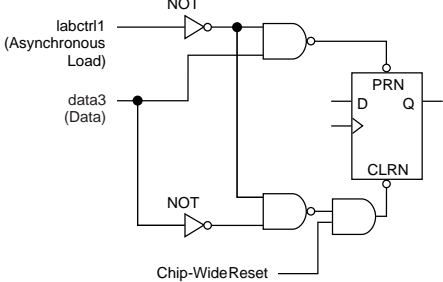
Asynchronous Preset & Clear



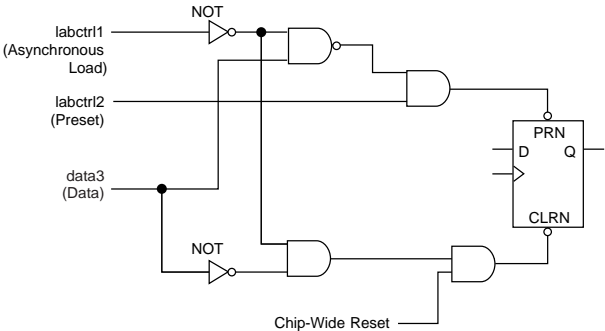
Asynchronous Load with Clear



Asynchronous Load without Clear or Preset



Asynchronous Load with Preset



## FastTrack Interconnect Routing Structure

In the ACEX 1K architecture, connections between LEs, EABs, and device I/O pins are provided by the FastTrack Interconnect routing structure, which is a series of continuous horizontal and vertical routing channels that traverse the device. This global routing structure provides predictable performance, even in complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.

The FastTrack Interconnect routing structure consists of row and column interconnect channels that span the entire device. Each row of LABs is served by a dedicated row interconnect. The row interconnect can drive I/O pins and feed other LABs in the row. The column interconnect routes signals between rows and can drive I/O pins.

Row channels drive into the LAB or EAB local interconnect. The row signal is buffered at every LAB or EAB to reduce the effect of fan-out on delay. A row channel can be driven by an LE or by one of three column channels. These four signals feed dual 4-to-1 multiplexers that connect to two specific row channels. These multiplexers, which are connected to each LE, allow column channels to drive row channels even when all eight LEs in a LAB drive the row interconnect.

Each column of LABs or EABs is served by a dedicated column interconnect. The column interconnect that serves the EABs has twice as many channels as other column interconnects. The column interconnect can then drive I/O pins or another row's interconnect to route the signals to other LABs or EABs in the device. A signal from the column interconnect, which can be either the output of a LE or an input from an I/O pin, must be routed to the row interconnect before it can enter a LAB or EAB. Each row channel that is driven by an IOE or EAB can drive one specific column channel.

Access to row and column channels can be switched between LEs in adjacent pairs of LABs. For example, a LE in one LAB can drive the row and column channels normally driven by a particular LE in the adjacent LAB in the same row, and vice versa. This flexibility enables routing resources to be used more efficiently. [Figure 13](#) shows the ACEX 1K LAB.



When dedicated inputs drive non-inverted and inverted peripheral clears, clock enables, and output enables, two signals on the peripheral control bus will be used.

**Table 7** lists the sources for each peripheral control signal and shows how the output enable, clock enable, clock, and clear signals share 12 peripheral control signals. **Table 7** also shows the rows that can drive global signals.

*Table 7. Peripheral Bus Sources for ACEX Devices*

Peripheral Control Signal	EP1K10	EP1K30	EP1K50	EP1K100
OE0	Row A	Row A	Row A	Row A
OE1	Row A	Row B	Row B	Row C
OE2	Row B	Row C	Row D	Row E
OE3	Row B	Row D	Row F	Row L
OE4	Row C	Row E	Row H	Row I
OE5	Row C	Row F	Row J	Row K
CLKENA0/CLK0/GLOBAL0	Row A	Row A	Row A	Row F
CLKENA1/OE6/GLOBAL1	Row A	Row B	Row C	Row D
CLKENA2/CLR0	Row B	Row C	Row E	Row B
CLKENA3/OE7/GLOBAL2	Row B	Row D	Row G	Row H
CLKENA4/CLR1	Row C	Row E	Row I	Row J
CLKENA5/CLK1/GLOBAL3	Row C	Row F	Row J	Row G

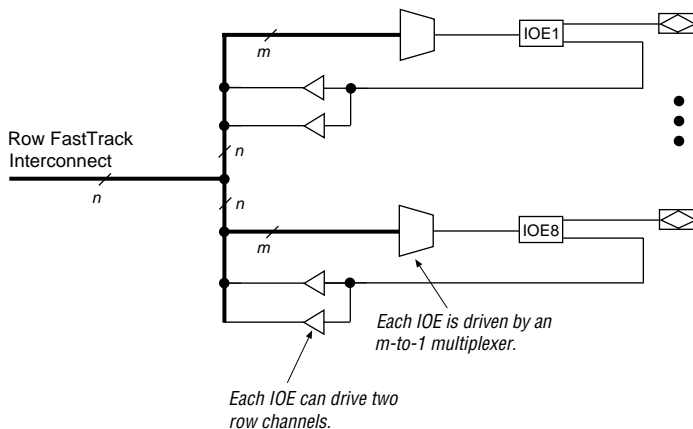
Signals on the peripheral control bus can also drive the four global signals, referred to as GLOBAL0 through GLOBAL3. An internally generated signal can drive a global signal, providing the same low-skew, low-delay characteristics as a signal driven by an input pin. An LE drives the global signal by driving a row line that drives the peripheral bus which then drives the global signal. This feature is ideal for internally generated clear or clock signals with high fan-out. However, internally driven global signals offer no advantage over the general-purpose interconnect for routing data signals.

The chip-wide output enable pin is an active-high pin that can be used to tri-state all pins on the device. This option can be set in the Altera software. The built-in I/O pin pull-up resistors (which are active during configuration) are active when the chip-wide output enable pin is asserted. The registers in the IOE can also be reset by the chip-wide reset pin.

### Row-to-IOE Connections

When an IOE is used as an input signal, it can drive two separate row channels. The signal is accessible by all LEs within that row. When an IOE is used as an output, the signal is driven by a multiplexer that selects a signal from the row channels. Up to eight IOEs connect to each side of each row channel (see Figure 16).

Figure 16. ACEX 1K Row-to-IOE Connections *Note (1)*



**Note:**

(1) The values for  $m$  and  $n$  are shown in Table 8.

Table 8 lists the ACEX 1K row-to-IOE interconnect resources.

Table 8. ACEX 1K Row-to-IOE Interconnect Resources		
Device	Channels per Row ( $n$ )	Row Channels per Pin ( $m$ )
EP1K10	144	18
EP1K30	216	27
EP1K50	216	27
EP1K100	312	39

Tables 11 and 12 summarize the ClockLock and ClockBoost parameters for -1 and -2 speed-grade devices, respectively.

*Table 11. ClockLock & ClockBoost Parameters for -1 Speed-Grade Devices*

Symbol	Parameter	Condition	Min	Typ	Max	Unit
$t_R$	Input rise time				5	ns
$t_F$	Input fall time				5	ns
$t_{INDUTY}$	Input duty cycle		40		60	%
$f_{CLK1}$	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		180	MHz
$f_{CLK2}$	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		90	MHz
$f_{CLKDEV}$	Input deviation from user specification in the Altera software (1)				25,000 (2)	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
$t_{LOCK}$	Time required for ClockLock or ClockBoost to acquire lock (3)				10	μs
$t_{JITTER}$	Jitter on ClockLock or ClockBoost-generated clock (4)	$t_{INCLKSTB} < 100$			250 (4)	ps
		$t_{INCLKSTB} < 50$			200 (4)	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

## PCI Pull-Up Clamping Diode Option

ACEX 1K devices have a pull-up clamping diode on every I/O, dedicated input, and dedicated clock pin. PCI clamping diodes clamp the signal to the  $V_{CCIO}$  value and are required for 3.3-V PCI compliance. Clamping diodes can also be used to limit overshoot in other systems.

Clamping diodes are controlled on a pin-by-pin basis. When  $V_{CCIO}$  is 3.3 V, a pin that has the clamping diode option turned on can be driven by a 2.5-V or 3.3-V signal, but not a 5.0-V signal. When  $V_{CCIO}$  is 2.5 V, a pin that has the clamping diode option turned on can be driven by a 2.5-V signal, but not a 3.3-V or 5.0-V signal. Additionally, a clamping diode can be activated for a subset of pins, which allows a device to bridge between a 3.3-V PCI bus and a 5.0-V device.

## Slew-Rate Control

The output buffer in each IOE has an adjustable output slew rate that can be configured for low-noise or high-speed performance. A slower slew rate reduces system noise and adds a maximum delay of 4.3 ns. The fast slew rate should be used for speed-critical outputs in systems that are adequately protected against noise. Designers can specify the slew rate pin-by-pin or assign a default slew rate to all pins on a device-wide basis. The slow slew rate setting affects only the falling edge of the output.

## Open-Drain Output Option

ACEX 1K devices provide an optional open-drain output (electrically equivalent to open-collector output) for each I/O pin. This open-drain output enables the device to provide system-level control signals (e.g., interrupt and write enable signals) that can be asserted by any of several devices. It can also provide an additional wired-OR plane.

## MultiVolt I/O Interface

The ACEX 1K device architecture supports the MultiVolt I/O interface feature, which allows ACEX 1K devices in all packages to interface with systems of differing supply voltages. These devices have one set of  $V_{CC}$  pins for internal operation and input buffers ( $V_{CCINT}$ ), and another set for I/O output drivers ( $V_{CCIO}$ ).

The  $V_{CCINT}$  pins must always be connected to a 2.5-V power supply. With a 2.5-V  $V_{CCINT}$  level, input voltages are compatible with 2.5-V, 3.3-V, and 5.0-V inputs. The  $V_{CCIO}$  pins can be connected to either a 2.5-V or 3.3-V power supply, depending on the output requirements. When the  $V_{CCIO}$  pins are connected to a 2.5-V power supply, the output levels are compatible with 2.5-V systems. When the  $V_{CCIO}$  pins are connected to a 3.3-V power supply, the output high is at 3.3 V and is therefore compatible with 3.3-V or 5.0-V systems. Devices operating with  $V_{CCIO}$  levels higher than 3.0 V achieve a faster timing delay of  $t_{OD2}$  instead of  $t_{OD1}$ .

**Table 13** summarizes ACEX 1K MultiVolt I/O support.

<i>Table 13. ACEX 1K MultiVolt I/O Support</i>						
$V_{CCIO}$ (V)	Input Signal (V)			Output Signal (V)		
	2.5	3.3	5.0	2.5	3.3	5.0
2.5	✓	✓ (1)	✓ (1)	✓		
3.3	✓	✓	✓ (1)	✓ (2)	✓	✓

**Notes:**

- (1) The PCI clamping diode must be disabled on an input which is driven with a voltage higher than  $V_{CCIO}$ .
- (2) When  $V_{CCIO} = 3.3$  V, an ACEX 1K device can drive a 2.5-V device that has 3.3-V tolerant inputs.

Open-drain output pins on ACEX 1K devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a higher  $V_{IH}$  than LVTTL. When the open-drain pin is active, it will drive low. When the pin is inactive, the resistor will pull up the trace to 5.0 V, thereby meeting the CMOS  $V_{OH}$  requirement. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The  $I_{OL}$  current specification should be considered when selecting a pull-up resistor.

## Power Sequencing & Hot-Socketing

Because ACEX 1K devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The  $V_{CCIO}$  and  $V_{CCINT}$  power planes can be powered in any order.

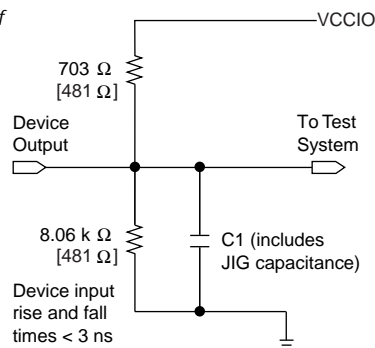
Signals can be driven into ACEX 1K devices before and during power up without damaging the device. Additionally, ACEX 1K devices do not drive out during power up. Once operating conditions are reached, ACEX 1K devices operate as specified by the user.

## Generic Testing

Each ACEX 1K device is functionally tested. Complete testing of each configurable static random access memory (SRAM) bit and all logic functionality ensures 100% yield. AC test measurements for ACEX 1K devices are made under conditions equivalent to those shown in [Figure 21](#). Multiple test patterns can be used to configure devices during all stages of the production flow.

**Figure 21. ACEX 1K AC Test Conditions**

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers in brackets are for 2.5-V devices or outputs. Numbers without brackets are for 3.3-V devices or outputs.



## Operating Conditions

[Tables 18](#) through [21](#) provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for 2.5-V ACEX 1K devices.

**Table 18. ACEX 1K Device Absolute Maximum Ratings** *Note (1)*

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{CCINT}$	Supply voltage	With respect to ground (2)	–0.5	3.6	V
$V_{CCIO}$			–0.5	4.6	V
$V_I$			–2.0	5.75	V
$I_{OUT}$	DC output current, per pin		–25	25	mA
$T_{STG}$	Storage temperature	No bias	–65	150	°C
$T_{AMB}$	Ambient temperature	Under bias	–65	135	°C
$T_J$	Junction temperature	PQFP, TQFP, and BGA packages, under bias		135	°C

Tables 27 through 29 describe the ACEX 1K external timing parameters and their symbols.

Table 27. External Reference Timing Parameters *Note (1)*

Symbol	Parameter	Conditions
$t_{\text{DRR}}$	Register-to-register delay via four LEs, three row interconnects, and four local interconnects	(2)

Table 28. External Timing Parameters

Symbol	Parameter	Conditions
$t_{\text{INSU}}$	Setup time with global clock at IOE register	(3)
$t_{\text{INH}}$	Hold time with global clock at IOE register	(3)
$t_{\text{OUTCO}}$	Clock-to-output delay with global clock at IOE register	(3)
$t_{\text{PCISU}}$	Setup time with global clock for registers used in PCI designs	(3), (4)
$t_{\text{PCIH}}$	Hold time with global clock for registers used in PCI designs	(3), (4)
$t_{\text{PCICO}}$	Clock-to-output delay with global clock for registers used in PCI designs	(3), (4)

Table 29. External Bidirectional Timing Parameters *Note (3)*

Symbol	Parameter	Conditions
$t_{\text{INSUBIDIR}}$	Setup time for bidirectional pins with global clock at same-row or same-column LE register	
$t_{\text{INHBIDIR}}$	Hold time for bidirectional pins with global clock at same-row or same-column LE register	
$t_{\text{OUTCOBIDIR}}$	Clock-to-output delay for bidirectional pins with global clock at IOE register	CI = 35 pF
$t_{\text{XZBIDIR}}$	Synchronous IOE output buffer disable delay	CI = 35 pF
$t_{\text{ZXBIDIR}}$	Synchronous IOE output buffer enable delay, slow slew rate = off	CI = 35 pF

**Notes to tables:**

- (1) External reference timing parameters are factory-tested, worst-case values specified by Altera. A representative subset of signal paths is tested to approximate typical device applications.
- (2) Contact Altera Applications for test circuit specifications and test conditions.
- (3) These timing parameters are sample-tested only.
- (4) This parameter is measured with the measurement and test conditions, including load, specified in the *PCI Local Bus Specification, Revision 2.2*.

Table 37. EP1K30 Device LE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{COMB}$		0.4		0.4		0.6	ns
$t_{SU}$	0.4		0.6		0.6		ns
$t_H$	0.7		1.0		1.3		ns
$t_{PRE}$		0.8		0.9		1.2	ns
$t_{CLR}$		0.8		0.9		1.2	ns
$t_{CH}$	2.0		2.5		2.5		ns
$t_{CL}$	2.0		2.5		2.5		ns

Table 38. EP1K30 Device IOE Timing Microparameters *Note (1)*

Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{IOD}$		2.4		2.8		3.8	ns
$t_{IOC}$		0.3		0.4		0.5	ns
$t_{IOCO}$		1.0		1.1		1.6	ns
$t_{IOCOMB}$		0.0		0.0		0.0	ns
$t_{IOSU}$	1.2		1.4		1.9		ns
$t_{IOH}$	0.3		0.4		0.5		ns
$t_{IOCLR}$		1.0		1.1		1.6	ns
$t_{OD1}$		1.9		2.3		3.0	ns
$t_{OD2}$		1.4		1.8		2.5	ns
$t_{OD3}$		4.4		5.2		7.0	ns
$t_{XZ}$		2.7		3.1		4.3	ns
$t_{ZX1}$		2.7		3.1		4.3	ns
$t_{ZX2}$		2.2		2.6		3.8	ns
$t_{ZX3}$		5.2		6.0		8.3	ns
$t_{INREG}$		3.4		4.1		5.5	ns
$t_{IOFD}$		0.8		1.3		2.4	ns
$t_{INCOMB}$		0.8		1.3		2.4	ns



Table 39. EP1K30 Device EAB Internal Microparameters *Note (1)*

Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.7		2.0		2.3	ns
$t_{EABDATA1}$		0.6		0.7		0.8	ns
$t_{EABWE1}$		1.1		1.3		1.4	ns
$t_{EABWE2}$		0.4		0.4		0.5	ns
$t_{EABRE1}$		0.8		0.9		1.0	ns
$t_{EABRE2}$		0.4		0.4		0.5	ns
$t_{EABCLK}$		0.0		0.0		0.0	ns
$t_{EABCO}$		0.3		0.3		0.4	ns
$t_{EABYPASS}$		0.5		0.6		0.7	ns
$t_{EABSU}$	0.9		1.0		1.2		ns
$t_{EABH}$	0.4		0.4		0.5		ns
$t_{EABCLR}$	0.3		0.3		0.3		ns
$t_{AA}$		3.2		3.8		4.4	ns
$t_{WP}$	2.5		2.9		3.3		ns
$t_{RP}$	0.9		1.1		1.2		ns
$t_{WDSU}$	0.9		1.0		1.1		ns
$t_{WDH}$	0.1		0.1		0.1		ns
$t_{WASU}$	1.7		2.0		2.3		ns
$t_{WAH}$	1.8		2.1		2.4		ns
$t_{RASU}$	3.1		3.7		4.2		ns
$t_{RAH}$	0.2		0.2		0.2		ns
$t_{WO}$		2.5		2.9		3.3	ns
$t_{DD}$		2.5		2.9		3.3	ns
$t_{EABOUT}$		0.5		0.6		0.7	ns
$t_{EABCH}$	1.5		2.0		2.3		ns
$t_{EABCL}$	2.5		2.9		3.3		ns

Table 46. EP1K50 Device EAB Internal Microparameters *Note (1)*

Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.7		2.4		3.2	ns
$t_{EABDATA2}$		0.4		0.6		0.8	ns
$t_{EABWE1}$		1.0		1.4		1.9	ns
$t_{EABWE2}$		0.0		0.0		0.0	ns
$t_{EABRE1}$		0.0		0.0		0.0	
$t_{EABRE2}$		0.4		0.6		0.8	
$t_{EABCLK}$		0.0		0.0		0.0	ns
$t_{EABCO}$		0.8		1.1		1.5	ns
$t_{EABYPASS}$		0.0		0.0		0.0	ns
$t_{EABSU}$	0.7		1.0		1.3		ns
$t_{EABH}$	0.4		0.6		0.8		ns
$t_{EABCLR}$	0.8		1.1		1.5		
$t_{AA}$		2.0		2.8		3.8	ns
$t_{WP}$	2.0		2.8		3.8		ns
$t_{RP}$	1.0		1.4		1.9		
$t_{WDSU}$	0.5		0.7		0.9		ns
$t_{WDH}$	0.1		0.1		0.2		ns
$t_{WASU}$	1.0		1.4		1.9		ns
$t_{WAH}$	1.5		2.1		2.9		ns
$t_{RASU}$	1.5		2.1		2.8		
$t_{RAH}$	0.1		0.1		0.2		
$t_{WO}$		2.1		2.9		4.0	ns
$t_{DD}$		2.1		2.9		4.0	ns
$t_{EABOUT}$		0.0		0.0		0.0	ns
$t_{EABCH}$	1.5		2.0		2.5		ns
$t_{EABCL}$	1.5		2.0		2.5		ns

Tables 51 through 57 show EP1K100 device internal and external timing parameters.

Table 51. EP1K100 Device LE Timing Microparameters *Note (1)*

Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{LUT}$		0.7		1.0		1.5	ns
$t_{CLUT}$		0.5		0.7		0.9	ns
$t_{RLUT}$		0.6		0.8		1.1	ns
$t_{PACKED}$		0.3		0.4		0.5	ns
$t_{EN}$		0.2		0.3		0.3	ns
$t_{CICO}$		0.1		0.1		0.2	ns
$t_{CGEN}$		0.4		0.5		0.7	ns
$t_{CGENR}$		0.1		0.1		0.2	ns
$t_{CASC}$		0.6		0.9		1.2	ns
$t_C$		0.8		1.0		1.4	ns
$t_{CO}$		0.6		0.8		1.1	ns
$t_{COMB}$		0.4		0.5		0.7	ns
$t_{SU}$	0.4		0.6		0.7		ns
$t_H$	0.5		0.7		0.9		ns
$t_{PRE}$		0.8		1.0		1.4	ns
$t_{CLR}$		0.8		1.0		1.4	ns
$t_{CH}$	1.5		2.0		2.5		ns
$t_{CL}$	1.5		2.0		2.5		ns

Table 53. EP1K100 Device EAB Internal Microparameters *Note (1)*

Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.5		2.0		2.6	ns
$t_{EABDATA1}$		0.0		0.0		0.0	ns
$t_{EABWE1}$		1.5		2.0		2.6	ns
$t_{EABWE2}$		0.3		0.4		0.5	ns
$t_{EABRE1}$		0.3		0.4		0.5	ns
$t_{EABRE2}$		0.0		0.0		0.0	ns
$t_{EABCLK}$		0.0		0.0		0.0	ns
$t_{EABCO}$		0.3		0.4		0.5	ns
$t_{EABYPASS}$		0.1		0.1		0.2	ns
$t_{EABSU}$	0.8		1.0		1.4		ns
$t_{EABH}$	0.1		0.1		0.2		ns
$t_{EABCLR}$	0.3		0.4		0.5		ns
$t_{AA}$		4.0		5.1		6.6	ns
$t_{WP}$	2.7		3.5		4.7		ns
$t_{RP}$	1.0		1.3		1.7		ns
$t_{WDSU}$	1.0		1.3		1.7		ns
$t_{WDH}$	0.2		0.2		0.3		ns
$t_{WASU}$	1.6		2.1		2.8		ns
$t_{WAH}$	1.6		2.1		2.8		ns
$t_{RASU}$	3.0		3.9		5.2		ns
$t_{RAH}$	0.1		0.1		0.2		ns
$t_{WO}$		1.5		2.0		2.6	ns
$t_{DD}$		1.5		2.0		2.6	ns
$t_{EABOUT}$		0.2		0.3		0.3	ns
$t_{EABCH}$	1.5		2.0		2.5		ns
$t_{EABCL}$	2.7		3.5		4.7		ns

Table 55. EP1K100 Device Interconnect Timing Microparameters *Note (1)*

Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		3.1		3.6		4.4	ns
$t_{DIN2LE}$		0.3		0.4		0.5	ns
$t_{DIN2DATA}$		1.6		1.8		2.0	ns
$t_{DCLK2IOE}$		0.8		1.1		1.4	ns
$t_{DCLK2LE}$		0.3		0.4		0.5	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		1.5		2.5		3.4	ns
$t_{SAMECOLUMN}$		0.4		1.0		1.6	ns
$t_{DIFFROW}$		1.9		3.5		5.0	ns
$t_{TWOROWS}$		3.4		6.0		8.4	ns
$t_{LEPERIPH}$		4.3		5.4		6.5	ns
$t_{LABCARRY}$		0.5		0.7		0.9	ns
$t_{LABCASC}$		0.8		1.0		1.4	ns

Table 56. EP1K100 External Timing Parameters *Notes (1), (2)*

Symbol	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
t <sub>DDR</sub>		9.0		12.0		16.0	ns
t <sub>INSU</sub> (3)	2.0		2.5		3.3		ns
t <sub>INH</sub> (3)	0.0		0.0		0.0		ns
t <sub>OUTCO</sub> (3)	2.0	5.2	2.0	6.9	2.0	9.1	ns
t <sub>INSU</sub> (4)	2.0		2.2		–		ns
t <sub>INH</sub> (4)	0.0		0.0		–		ns
t <sub>OUTCO</sub> (4)	0.5	3.0	0.5	4.6	–	–	ns
t <sub>PCISU</sub>	3.0		6.2		–		ns
t <sub>PCIH</sub>	0.0		0.0		–		ns
t <sub>PCICO</sub>	2.0	6.0	2.0	6.9	–	–	ns