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Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	360
Number of Logic Elements/Cells	2880
Total RAM Bits	40960
Number of I/O	147
Number of Gates	199000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1k50qi208-2

Email: info@E-XFL.COM

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General Description

Altera® ACEX 1K devices provide a die-efficient, low-cost architecture by combining look-up table (LUT) architecture with EABs. LUT-based logic provides optimized performance and efficiency for data-path, register intensive, mathematical, or digital signal processing (DSP) designs, while EABs implement RAM, ROM, dual-port RAM, or first-in first-out (FIFO) functions. These elements make ACEX 1K suitable for complex logic functions and memory functions such as digital signal processing, wide data-path manipulation, data transformation and microcontrollers, as required in high-performance communications applications. Based on reconfigurable CMOS SRAM elements, the ACEX 1K architecture incorporates all features necessary to implement common gate array megafunctions, along with a high pin count to enable an effective interface with system components. The advanced process and the low voltage requirement of the 2.5-V core allow ACEX 1K devices to meet the requirements of low-cost, high-volume applications ranging from DSL modems to low-cost switches.

The ability to reconfigure ACEX 1K devices enables complete testing prior to shipment and allows the designer to focus on simulation and design verification. ACEX 1K device reconfigurability eliminates inventory management for gate array designs and test vector generation for fault coverage.

Table 4 shows ACEX 1K device performance for some common designs. All performance results were obtained with Synopsys DesignWare or LPM functions. Special design techniques are not required to implement the applications; the designer simply infers or instantiates a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Application	Reso Us	urces ed		Performa	nce	
	LEs	EABs		Speed Grade		Units
			-1	-2	-3	
16-bit loadable counter	16	0	285	232	185	MHz
16-bit accumulator	16	0	285	232	185	MHz
16-to-1 multiplexer (1)	10	0	3.5	4.5	6.6	ns
16-bit multiplier with 3-stage pipeline(2)	592	0	156	131	93	MHz
256 × 16 RAM read cycle speed (2)	0	1	278	196	143	MHz
256 × 16 RAM write cycle speed (2)	0	1	185	143	111	MHz

Notes:

- This application uses combinatorial inputs and outputs.
- (2) This application uses registered inputs and outputs.

Table 5 shows ACEX 1K device performance for more complex designs. These designs are available as Altera MegaCore $^{\rm TM}$ functions.

Table 5. ACEX 1K Device Performance for Compl	ex Design	s			
Application	LEs		Perform	ance	
	Used		Speed Grade	!	Units
	·	-1	-2	-3	
16-bit, 8-tap parallel finite impulse response (FIR) filter	597	192	156	116	MSPS
8-bit, 512-point Fast Fourier transform (FFT)	1,854	23.4	28.7	38.9	μs
function		113	92	68	MHz
a16450 universal asynchronous receiver/transmitter (UART)	342	36	28	20.5	MHz

Each ACEX 1K device contains an embedded array and a logic array. The embedded array is used to implement a variety of memory functions or complex logic functions, such as digital signal processing (DSP), wide data-path manipulation, microcontroller applications, and data-transformation functions. The logic array performs the same function as the sea-of-gates in the gate array and is used to implement general logic such as counters, adders, state machines, and multiplexers. The combination of embedded and logic arrays provides the high performance and high density of embedded gate arrays, enabling designers to implement an entire system on a single device.

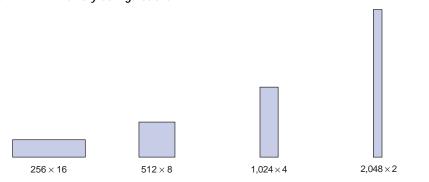
ACEX 1K devices are configured at system power-up with data stored in an Altera serial configuration device or provided by a system controller. Altera offers EPC16, EPC2, EPC1, and EPC1441 configuration devices, which configure ACEX 1K devices via a serial data stream. Configuration data can also be downloaded from system RAM or via the Altera MasterBlaster $^{\text{TM}}$, ByteBlasterMV $^{\text{TM}}$, or BitBlaster $^{\text{TM}}$ download cables. After an ACEX 1K device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Because reconfiguration requires less than 40 ms, real-time changes can be made during system operation.

ACEX 1K devices contain an interface that permits microprocessors to configure ACEX 1K devices serially or in parallel, and synchronously or asynchronously. The interface also enables microprocessors to treat an ACEX 1K device as memory and configure it by writing to a virtual memory location, simplifying device reconfiguration.

EABs can be used to implement synchronous RAM, which is easier to use than asynchronous RAM. A circuit using asynchronous RAM must generate the RAM write enable signal, while ensuring that its data and address signals meet setup and hold time specifications relative to the write enable signal. In contrast, the EAB's synchronous RAM generates its own write enable signal and is self-timed with respect to the input or write clock. A circuit using the EAB's self-timed RAM must only meet the setup and hold time specifications of the global clock.

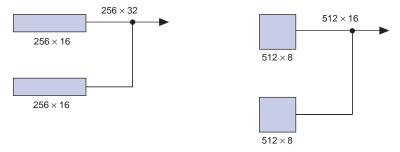
When used as RAM, each EAB can be configured in any of the following sizes: 256×16 ; 512×8 ; $1,024 \times 4$; or $2,048 \times 2$. Figure 5 shows the ACEX 1K EAB memory configurations.

Figure 5. ACEX 1K EAB Memory Configurations



Larger blocks of RAM are created by combining multiple EABs. For example, two 256×16 RAM blocks can be combined to form a 256×32 block, and two 512×8 RAM blocks can be combined to form a 512×16 block. Figure 6 shows examples of multiple EAB combination.

Figure 6. Examples of Combining ACEX 1K EABs



LE Operating Modes

The ACEX 1K LE can operate in the following four modes:

- Normal mode
- Arithmetic mode
- Up/down counter mode
- Clearable counter mode

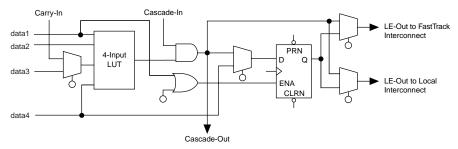
Each of these modes uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. Three inputs to the LE provide clock, clear, and preset control for the register. The Altera software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions that use a specific LE operating mode for optimal performance.

The architecture provides a synchronous clock enable to the register in all four modes. The Altera software can set DATA1 to enable the register synchronously, providing easy implementation of fully synchronous designs.

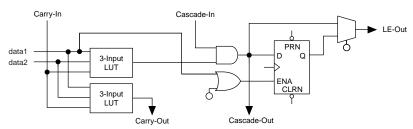
Figure 11 shows the ACEX 1K LE operating modes.

Figure 11. ACEX 1K LE Operating Modes

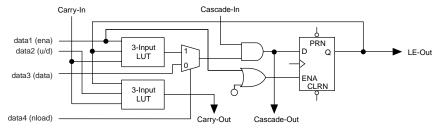
Normal Mode



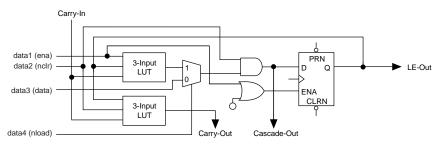
Arithmetic Mode



Up/Down Counter Mode



Clearable Counter Mode



FastTrack Interconnect Routing Structure

In the ACEX 1K architecture, connections between LEs, EABs, and device I/O pins are provided by the FastTrack Interconnect routing structure, which is a series of continuous horizontal and vertical routing channels that traverse the device. This global routing structure provides predictable performance, even in complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.

The FastTrack Interconnect routing structure consists of row and column interconnect channels that span the entire device. Each row of LABs is served by a dedicated row interconnect. The row interconnect can drive I/O pins and feed other LABs in the row. The column interconnect routes signals between rows and can drive I/O pins.

Row channels drive into the LAB or EAB local interconnect. The row signal is buffered at every LAB or EAB to reduce the effect of fan-out on delay. A row channel can be driven by an LE or by one of three column channels. These four signals feed dual 4-to-1 multiplexers that connect to two specific row channels. These multiplexers, which are connected to each LE, allow column channels to drive row channels even when all eight LEs in a LAB drive the row interconnect.

Each column of LABs or EABs is served by a dedicated column interconnect. The column interconnect that serves the EABs has twice as many channels as other column interconnects. The column interconnect can then drive I/O pins or another row's interconnect to route the signals to other LABs or EABs in the device. A signal from the column interconnect, which can be either the output of a LE or an input from an I/O pin, must be routed to the row interconnect before it can enter a LAB or EAB. Each row channel that is driven by an IOE or EAB can drive one specific column channel.

Access to row and column channels can be switched between LEs in adjacent pairs of LABs. For example, a LE in one LAB can drive the row and column channels normally driven by a particular LE in the adjacent LAB in the same row, and vice versa. This flexibility enables routing resources to be used more efficiently. Figure 13 shows the ACEX 1K LAB.

Column-to-IOE Connections

When an IOE is used as an input, it can drive up to two separate column channels. When an IOE is used as an output, the signal is driven by a multiplexer that selects a signal from the column channels. Two IOEs connect to each side of the column channels. Each IOE can be driven by column channels via a multiplexer. The set of column channels is different for each IOE (see Figure 17).

Each IOE is driven by a m-to-1 multiplexer

Column Interconnect

Figure 17. ACEX 1K Column-to-IOE Connections Note (1)

Note:

The values for m and n are shown in Table 9.

Table 9 lists the ACEX 1K column-to-IOE interconnect resources.

Each IOE can drive two column channels.

Table 9. ACEX 1K Column-to-IOE Interconnect Resources						
Device	Channels per Column (n)	Column Channels per Pin (m)				
EP1K10	24	16				
EP1K30	24	16				
EP1K50	24	16				
EP1K100	24	16				

Tables 11 and 12 summarize the ClockLock and ClockBoost parameters for -1 and -2 speed-grade devices, respectively.

Table 11.	ClockLock & ClockBoost Parameters for -1	Speed-Grade De	vices			
Symbol	Parameter	Condition	Min	Тур	Max	Unit
t_R	Input rise time				5	ns
t_{F}	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f _{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		180	MHz
f _{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		90	MHz
f _{CLKDEV}	Input deviation from user specification in the Altera software (1)				25,000 <i>(</i> 2 <i>)</i>	PPM
t _{INCLKSTB}	Input clock stability (measured between adjacent clocks)				100	ps
t _{LOCK}	Time required for ClockLock or ClockBoost to acquire lock (3)				10	μs
t _{JITTER}	Jitter on ClockLock or ClockBoost-	t _{INCLKSTB} <100			250 (4)	ps
	generated clock (4)	t _{INCLKSTB} < 50			200 (4)	ps
t _{OUTDUTY}	Duty cycle for ClockLock or ClockBoost- generated clock		40	50	60	%

The VCCINT pins must always be connected to a 2.5-V power supply. With a 2.5-V $V_{\rm CCINT}$ level, input voltages are compatible with 2.5-V, 3.3-V, and 5.0-V inputs. The VCCIO pins can be connected to either a 2.5-V or 3.3-V power supply, depending on the output requirements. When the VCCIO pins are connected to a 2.5-V power supply, the output levels are compatible with 2.5-V systems. When the VCCIO pins are connected to a 3.3-V power supply, the output high is at 3.3 V and is therefore compatible with 3.3-V or 5.0-V systems. Devices operating with $V_{\rm CCIO}$ levels higher than 3.0 V achieve a faster timing delay of t_{OD2} instead of t_{OD1} .

Table 13 summarizes ACEX 1K MultiVolt I/O support.

Table 13. ACEX 1	K MultiVo	It I/O Supp	oort			
V _{CCIO} (V) Input Signal (V) Output Signal (V)						(V)
	2.5	3.3	5.0	2.5	3.3	5.0
2.5	✓	✓ (1)	√ (1)	✓		
3.3	✓	✓	√ (1)	√ (2)	✓	✓

Notes:

- (1) The PCI clamping diode must be disabled on an input which is driven with a voltage higher than $V_{\rm CCIO}$.
- (2) When $V_{\rm CCIO}$ = 3.3 V, an ACEX 1K device can drive a 2.5-V device that has 3.3-V tolerant inputs.

Open-drain output pins on ACEX 1K devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a higher V_{IH} than LVTTL. When the open-drain pin is active, it will drive low. When the pin is inactive, the resistor will pull up the trace to 5.0 V, thereby meeting the CMOS V_{OH} requirement. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The I_{OL} current specification should be considered when selecting a pull-up resistor.

Power Sequencing & Hot-Socketing

Because ACEX 1K devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The $V_{\rm CCIO}$ and $V_{\rm CCINT}$ power planes can be powered in any order.

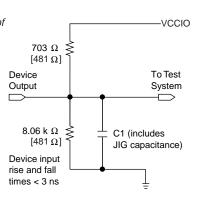
Signals can be driven into ACEX 1K devices before and during power up without damaging the device. Additionally, ACEX 1K devices do not drive out during power up. Once operating conditions are reached, ACEX 1K devices operate as specified by the user.

Generic Testing

Each ACEX 1K device is functionally tested. Complete testing of each configurable static random access memory (SRAM) bit and all logic functionality ensures 100% yield. AC test measurements for ACEX 1K devices are made under conditions equivalent to those shown in Figure 21. Multiple test patterns can be used to configure devices during all stages of the production flow.

Figure 21. ACEX 1K AC Test Conditions

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-groundcurrent transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers in brackets are for 2.5-V devices or outputs. Numbers without brackets are for 3.3-V devices or outputs.



Operating Conditions

Tables 18 through 21 provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for 2.5-V ACEX 1K devices.

Table 1	8. ACEX 1K Device Absolute I	Maximum Ratings Note (1)			
Symbol	Parameter	Conditions	Min	Max	Unit
V _{CCINT}	Supply voltage	With respect to ground (2)	-0.5	3.6	V
V _{CCIO}			-0.5	4.6	V
V _I	DC input voltage		-2.0	5.75	V
I _{OUT}	DC output current, per pin		-25	25	mA
T _{STG}	Storage temperature	No bias	-65	150	° C
T _{AMB}	Ambient temperature	Under bias	-65	135	° C
TJ	Junction temperature	PQFP, TQFP, and BGA packages, under bias		135	° C

Figure 22 shows the required relationship between V_{CCIO} and V_{CCINT} to satisfy 3.3-V PCI compliance.

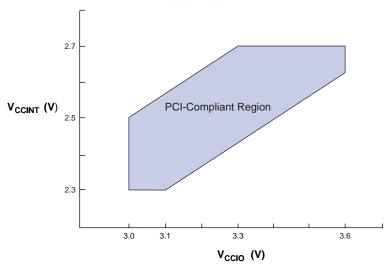


Figure 22. Relationship between V_{CCIO} & V_{CCINT} for 3.3-V PCI Compliance

Figure 23 shows the typical output drive characteristics of ACEX 1K devices with 3.3-V and 2.5-V $V_{\rm CCIO}$. The output driver is compliant to the 3.3-V *PCI Local Bus Specification, Revision 2.2* (when VCCIO pins are connected to 3.3 V). ACEX 1K devices with a -1 speed grade also comply with the drive strength requirements of the *PCI Local Bus Specification, Revision 2.2* (when VCCINT pins are powered with a minimum supply of 2.375 V, and VCCIO pins are connected to 3.3 V). Therefore, these devices can be used in open 5.0-V PCI systems.

Table 22. LE	Timing Microparameters (Part 2 of 2) Note (1)	
Symbol	Parameter	Conditions
t _{CASC}	Cascade-in to cascade-out delay	
t_{C}	LE register control signal delay	
t_{CO}	LE register clock-to-output delay	
t _{COMB}	Combinatorial delay	
t _{SU}	LE register setup time for data and enable signals before clock; LE register recovery time after asynchronous clear, preset, or load	
t_H	LE register hold time for data and enable signals after clock	
t _{PRE}	LE register preset delay	
t _{CLR}	LE register clear delay	
t _{CH}	Minimum clock high time from clock pin	
t_{CL}	Minimum clock low time from clock pin	

Table 23. 10	E Timing Microparameters Note (1)	
Symbol	Parameter	Conditions
t_{IOD}	IOE data delay	
t_{IOC}	IOE register control signal delay	
t _{IOCO}	IOE register clock-to-output delay	
t _{IOCOMB}	IOE combinatorial delay	
t _{IOSU}	IOE register setup time for data and enable signals before clock; IOE register recovery time after asynchronous clear	
t _{IOH}	IOE register hold time for data and enable signals after clock	
t _{IOCLR}	IOE register clear time	
t _{OD1}	Output buffer and pad delay, slow slew rate = off, V _{CCIO} = 3.3 V	C1 = 35 pF (2)
t _{OD2}	Output buffer and pad delay, slow slew rate = off, V _{CCIO} = 2.5 V	C1 = 35 pF (3)
t _{OD3}	Output buffer and pad delay, slow slew rate = on	C1 = 35 pF (4)
t_{XZ}	IOE output buffer disable delay	
t_{ZX1}	IOE output buffer enable delay, slow slew rate = off, V _{CCIO} = 3.3 V	C1 = 35 pF (2)
t_{ZX2}	IOE output buffer enable delay, slow slew rate = off, V _{CCIO} = 2.5 V	C1 = 35 pF (3)
t _{ZX3}	IOE output buffer enable delay, slow slew rate = on	C1 = 35 pF (4)
t _{INREG}	IOE input pad and buffer to IOE register delay	
t _{IOFD}	IOE register feedback delay	
t _{INCOMB}	IOE input pad and buffer to FastTrack Interconnect delay	

Table 26. Inte	erconnect Timing Microparameters Note (1)	
Symbol	Parameter	Conditions
t _{DIN2IOE}	Delay from dedicated input pin to IOE control input	(7)
t _{DIN2LE}	Delay from dedicated input pin to LE or EAB control input	(7)
t _{DIN2DATA}	Delay from dedicated input or clock to LE or EAB data	(7)
t _{DCLK2IOE}	Delay from dedicated clock pin to IOE clock	(7)
t _{DCLK2LE}	Delay from dedicated clock pin to LE or EAB clock	(7)
t _{SAMELAB}	Routing delay for an LE driving another LE in the same LAB	(7)
t _{SAMEROW}	Routing delay for a row IOE, LE, or EAB driving a row IOE, LE, or EAB in the same row	(7)
t _{SAME} COLUMN	Routing delay for an LE driving an IOE in the same column	(7)
t _{DIFFROW}	Routing delay for a column IOE, LE, or EAB driving an LE or EAB in a different row	(7)
t _{TWOROWS}	Routing delay for a row IOE or EAB driving an LE or EAB in a different row	(7)
t _{LEPERIPH}	Routing delay for an LE driving a control signal of an IOE via the peripheral control bus	(7)
t _{LABCARRY}	Routing delay for the carry-out signal of an LE driving the carry-in signal of a different LE in a different LAB	
t _{LABCASC}	Routing delay for the cascade-out signal of an LE driving the cascade-in signal of a different LE in a different LAB	

Notes to tables:

- Microparameters are timing delays contributed by individual architectural elements. These parameters cannot be measured explicitly.
- Operating conditions: $V_{CCIO} = 3.3 \text{ V} \pm 10\%$ for commercial or industrial and extended use in ACEX 1K devices
- Operating conditions: $V_{CCIO} = 2.5 \text{ V} \pm 5\%$ for commercial or industrial and extended use in ACEX 1K devices. Operating conditions: $V_{CCIO} = 2.5 \text{ V} \text{ or } 3.3 \text{ V}$. (3)
- (4)
- Because the RAM in the EAB is self-timed, this parameter can be ignored when the WE signal is registered.
- EAB macroparameters are internal parameters that can simplify predicting the behavior of an EAB at its boundary; these parameters are calculated by summing selected microparameters.
- These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.

Tables 30 through 36 show EP1K10 device internal and external timing parameters.

Table 30. EP1K10 Device LE Timing Microparameters Note (1)								
Symbol		Speed Grade						
	-	1	-2		-3			
	Min	Max	Min	Max	Min	Max		
t_{LUT}		0.7		0.8		1.1	ns	
t_{CLUT}		0.5		0.6		0.8	ns	
t _{RLUT}		0.6		0.7		1.0	ns	
t _{PACKED}		0.4		0.4		0.5	ns	
t_{EN}		0.9		1.0		1.3	ns	
t_{CICO}		0.1		0.1		0.2	ns	
t _{CGEN}		0.4		0.5		0.7	ns	
t _{CGENR}		0.1		0.1		0.2	ns	
t _{CASC}		0.7		0.9		1.1	ns	
t_{C}		1.1		1.3		1.7	ns	
$t_{\rm CO}$		0.5		0.7		0.9	ns	
t _{COMB}		0.4		0.5		0.7	ns	
t _{SU}	0.7		0.8		1.0		ns	
t _H	0.9		1.0		1.1		ns	
t _{PRE}		0.8		1.0		1.4	ns	
t _{CLR}		0.9		1.0		1.4	ns	
t _{CH}	2.0		2.5		2.5		ns	
t_{CL}	2.0		2.5		2.5		ns	

Symbol			Speed	Grade			Unit	
	-	1	-	-2		3		
	Min	Max	Min	Max	Min	Max		
t_{IOD}		2.6		3.1		4.0	ns	
t _{IOC}		0.3		0.4		0.5	ns	
t _{IOCO}		0.9		1.0		1.4	ns	
t _{IOCOMB}		0.0		0.0		0.0	ns	
t _{iosu}	1.3		1.5		2.0		ns	
t _{IOH}	0.9		1.0		1.4		ns	
t _{IOCLR}		1.1		1.3		1.7	ns	
t _{OD1}		3.1		3.7		4.1	ns	
t _{OD2}		2.6		3.3		3.9	ns	
t _{OD3}		5.8		6.9		8.3	ns	
t_{XZ}		3.8		4.5		5.9	ns	
t_{ZX1}		3.8		4.5		5.9	ns	
t_{ZX2}		3.3		4.1		5.7	ns	
t_{ZX3}		6.5		7.7		10.1	ns	
t _{INREG}		3.7		4.3		5.7	ns	
t _{IOFD}		0.9		1.0		1.4	ns	
t _{INCOMB}		1.9		2.3		3.0	ns	

Symbol	Speed Grade							
	-1		-2		-3			
	Min	Max	Min	Max	Min	Max		
t _{DIN2IOE}		2.3		2.7		3.6	ns	
t _{DIN2LE}		0.8		1.1		1.4	ns	
t _{DIN2DATA}		1.1		1.4		1.8	ns	
t _{DCLK2IOE}		2.3		2.7		3.6	ns	
t _{DCLK2LE}		0.8		1.1		1.4	ns	
t _{SAMELAB}		0.1		0.1		0.2	ns	
t _{SAMEROW}		1.8		2.1		2.9	ns	
t _{SAME} COLUMN		0.3		0.4		0.7	ns	
t _{DIFFROW}		2.1		2.5		3.6	ns	
t _{TWOROWS}		3.9		4.6		6.5	ns	
t _{LEPERIPH}		3.3		3.7		4.8	ns	
t _{LABCARRY}		0.3		0.4		0.5	ns	
t _{LABCASC}		0.9		1.0		1.4	ns	

Table 35. EP1K10 External Timing Parameters Note (1)									
Symbol		Unit							
	-1		-2		-3				
	Min	Max	Min	Max	Min	Max			
t _{DRR}		7.5		9.5		12.5	ns		
t _{INSU} (2), (3)	2.4		2.7		3.6		ns		
t _{INH} (2), (3)	0.0		0.0		0.0		ns		
t _{оитсо} (2), (3)	2.0	6.6	2.0	7.8	2.0	9.6	ns		
t _{INSU} (4), (3)	1.4		1.7		_		ns		
t _{INH} (4), (3)	0.5	5.1	0.5	6.4	_	_	ns		
t _{оитсо} (4), (3)	0.0		0.0		_		ns		
t _{PCISU} (3)	3.0		4.2		6.4		ns		
t _{PCIH} (3)	0.0		0.0		-		ns		
t _{PCICO} (3)	2.0	6.0	2.0	7.5	2.0	10.2	ns		

Symbol	Speed Grade								
	-1		-2		-3				
	Min	Max	Min	Max	Min	Max			
t _{INSUBIDIR} (2)	2.2		2.3		3.2		ns		
t _{INHBIDIR} (2)	0.0		0.0		0.0		ns		
t _{OUTCOBIDIR} (2)	2.0	6.6	2.0	7.8	2.0	9.6	ns		
t _{XZBIDIR} (2)		8.8		11.2		14.0	ns		
t _{ZXBIDIR} (2)		8.8		11.2		14.0	ns		
t _{INSUBIDIR} (4)	3.1		3.3		_	-			
t _{INHBIDIR} (4)	0.0		0.0				•		
toutcobidir (4)	0.5	5.1	0.5	6.4	-	_	ns		
t _{XZBIDIR} (4)		7.3		9.2		_	ns		
t _{ZXBIDIR} (4)		7.3		9.2		_	ns		

Notes to tables:

- (1) All timing parameters are described in Tables 22 through 29 in this data sheet.
- (2) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
- (3) These parameters are specified by characterization.
- (4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Tables 37 through 43 show EP1K30 device internal and external timing parameters.

Symbol	Speed Grade							
	-1		-2		-3			
	Min	Max	Min	Max	Min	Max		
t_{LUT}		0.7		0.8		1.1	ns	
t _{CLUT}		0.5		0.6		0.8	ns	
t _{RLUT}		0.6		0.7		1.0	ns	
t _{PACKED}		0.3		0.4		0.5	ns	
t_{EN}		0.6		0.8		1.0	ns	
t _{CICO}		0.1		0.1		0.2	ns	
t _{CGEN}		0.4		0.5		0.7	ns	
t _{CGENR}		0.1		0.1		0.2	ns	
t _{CASC}		0.6		0.8		1.0	ns	
t _C		0.0		0.0		0.0	ns	
t _{co}		0.3		0.4		0.5	ns	

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Symbol	Speed Grade							
	-1		-2		-3			
	Min	Max	Min	Max	Min	Max		
t _{EABAA}		6.4		7.6		8.8	ns	
t _{EABRCOMB}	6.4		7.6		8.8		ns	
t _{EABRCREG}	4.4		5.1		6.0		ns	
t _{EABWP}	2.5		2.9		3.3		ns	
t _{EABWCOMB}	6.0		7.0		8.0		ns	
t _{EABWCREG}	6.8		7.8		9.0		ns	
t _{EABDD}		5.7		6.7		7.7	ns	
t _{EABDATA} CO		0.8		0.9		1.1	ns	
t _{EABDATASU}	1.5		1.7		2.0		ns	
t _{EABDATAH}	0.0		0.0		0.0		ns	
t _{EABWESU}	1.3		1.4		1.7		ns	
t _{EABWEH}	0.0		0.0		0.0		ns	
t _{EABWDSU}	1.5		1.7		2.0		ns	
t _{EABWDH}	0.0		0.0		0.0		ns	
t _{EABWASU}	3.0		3.6		4.3		ns	
t _{EABWAH}	0.5		0.5		0.4		ns	
t _{EABWO}		5.1		6.0		6.8	ns	

Symbol	Speed Grade							
	-1		-2		-3			
	Min	Max	Min	Max	Min	Max		
t _{EABDATA1}		1.5		2.0		2.6	ns	
t _{EABDATA1}		0.0		0.0		0.0	ns	
t _{EABWE1}		1.5		2.0		2.6	ns	
t _{EABWE2}		0.3		0.4		0.5	ns	
t _{EABRE1}		0.3		0.4		0.5	ns	
t _{EABRE2}		0.0		0.0		0.0	ns	
t _{EABCLK}		0.0		0.0		0.0	ns	
t _{EABCO}		0.3		0.4		0.5	ns	
t _{EABBYPASS}		0.1		0.1		0.2	ns	
t _{EABSU}	0.8		1.0		1.4		ns	
t _{EABH}	0.1		0.1		0.2		ns	
t _{EABCLR}	0.3		0.4		0.5		ns	
t_{AA}		4.0		5.1		6.6	ns	
t_{WP}	2.7		3.5		4.7		ns	
t _{RP}	1.0		1.3		1.7		ns	
t _{WDSU}	1.0		1.3		1.7		ns	
t_{WDH}	0.2		0.2		0.3		ns	
t _{WASU}	1.6		2.1		2.8		ns	
t _{WAH}	1.6		2.1		2.8		ns	
t _{RASU}	3.0		3.9		5.2		ns	
t _{RAH}	0.1		0.1		0.2		ns	
t_{WO}		1.5		2.0		2.6	ns	
t_{DD}		1.5		2.0		2.6	ns	
t _{EABOUT}		0.2		0.3		0.3	ns	
t _{EABCH}	1.5		2.0		2.5		ns	
t _{EABCL}	2.7		3.5		4.7		ns	

The I_{CCACTIVE} value can be calculated with the following equation:

$$I_{CCACTIVE} = K \times f_{MAX} \times N \times tog_{LC} (\mu A)$$

Where:

f_{MAX} = Maximum operating frequency in MHz
 N = Total number of LEs used in the device

tog_{LC} = Average percent of LEs toggling at each clock

(typically 12.5%)

K = Constant

Table 58 provides the constant (K) values for ACEX 1K devices.

Table 58. ACEX 1K Constant Values						
Device	K Value					
EP1K10	4.5					
EP1K30	4.5					
EP1K50	4.5					
EP1K100	4.5					

This supply power calculation provides an I_{CC} estimate based on typical conditions with no output load. The actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

To better reflect actual designs, the power model (and the constant K in the power calculation equations) for continuous interconnect ACEX 1K devices assumes that LEs drive FastTrack Interconnect channels. In contrast, the power model of segmented FPGAs assumes that all LEs drive only one short interconnect segment. This assumption may lead to inaccurate results when compared to measured power consumption for actual designs in segmented FPGAs.

Figure 31 shows the relationship between the current and operating frequency of ACEX 1K devices. For information on other ACEX 1K devices, contact Altera Applications at (800) 800-EPLD.