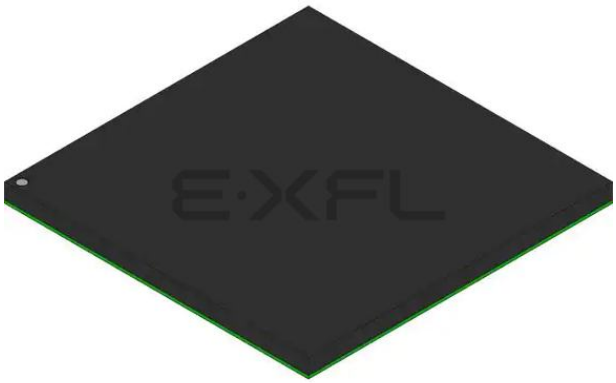


Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

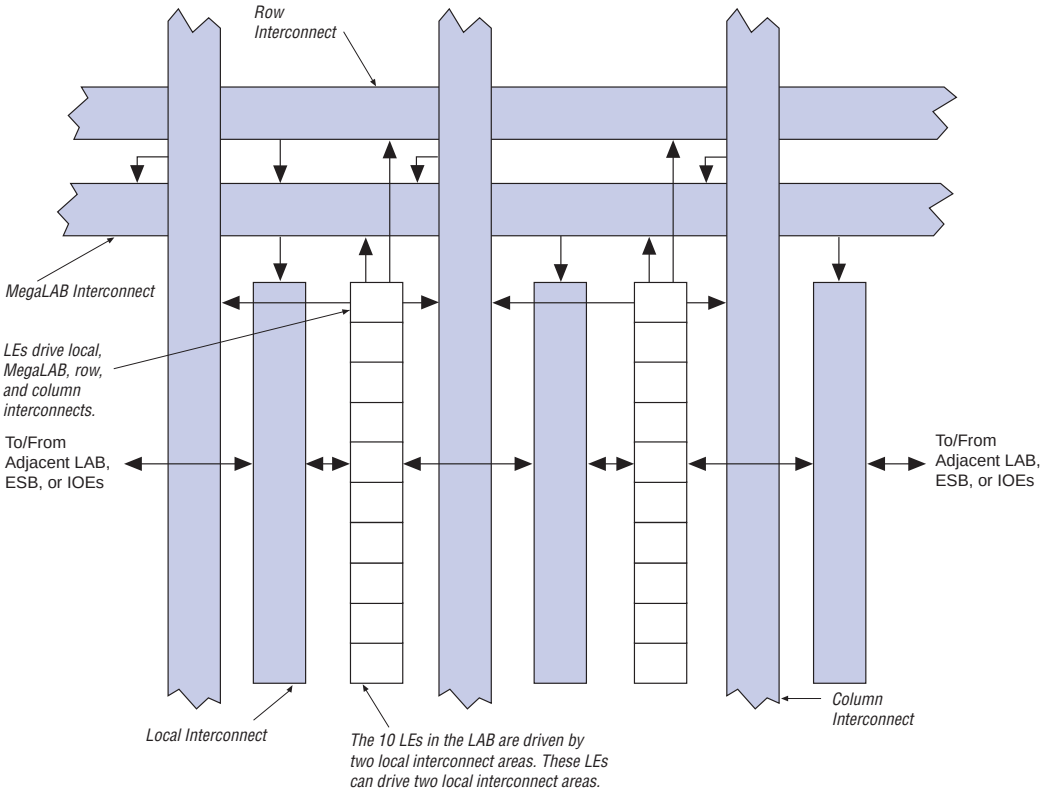
Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	492
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	672-BBGA, FCBGA
Supplier Device Package	672-FBGA (27x27)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=ep2a15f672i8

Figure 3 shows the APEX II LAB.

Figure 3. APEX II LAB Structure

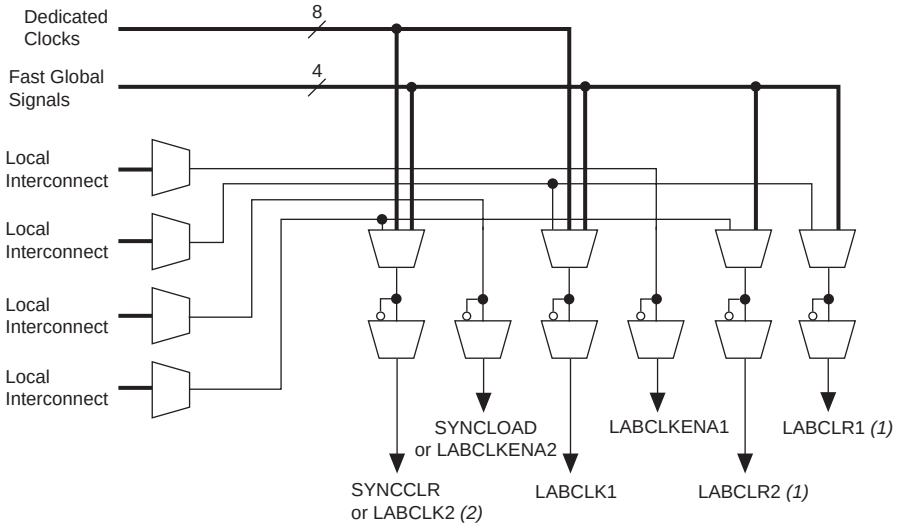


Each LAB contains dedicated logic for driving control signals to its LEs and ESBs. The control signals include clock, clock enable, asynchronous clear, asynchronous preset, asynchronous load, synchronous clear, and synchronous load signals. A maximum of six control signals can be used at a time. Although synchronous load and clear signals are generally used when implementing counters, they can also be used with other functions.

Each LAB can use two clocks and two clock enable signals. The LAB's clock and clock enable signals are linked (e.g., any LE in a particular LAB using CLK1 will also use CLKENA1). LEs with the same clock but different clock enable signals either use both clock signals in one LAB or are placed into separate LABs. If both the rising and falling edges of a clock are used in an LAB, both LAB-wide clock signals are used.

The LAB-wide control signals can be generated from the LAB local interconnect, global signals, and dedicated clock pins. The inherent low skew of the FastTrack interconnect enables it to be used for clock distribution. Figure 4 shows the LAB control signal generation circuit.

Figure 4. LAB Control Signal Generation



Notes to Figure 4:

(1) The LABCLR1 and LABCLR2 signals also control asynchronous load and asynchronous preset for LEs within the LAB.

(2) The SYNCCLR signal can be generated by the local interconnect or global signals.

Logic Element

The LE is the smallest unit of logic in the APEX II architecture. Each LE contains a four-input LUT, which is a function generator that can quickly implement any function of four variables. In addition, each LE contains a programmable register and carry and cascade chains. Each LE drives the local interconnect, MegaLAB interconnect, and FastTrack interconnect routing structures. See Figure 5.

Each LE has two outputs that drive the local, MegaLAB, or FastTrack interconnect routing structure. Each output can be driven independently by the LUT's or register's output. For example, the LUT can drive one output while the register drives the other output. This feature, called register packing, improves device utilization because the register and the LUT can be used for unrelated functions. The LE can also drive out registered and unregistered versions of the LUT output. The APEX II architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. A carry chain supports high-speed arithmetic functions such as counters and adders, while a cascade chain implements wide-input functions such as equality comparators with minimum delay. Carry and cascade chains connect LEs 1 through 10 in an LAB and all LABs in the same MegaLAB structure.

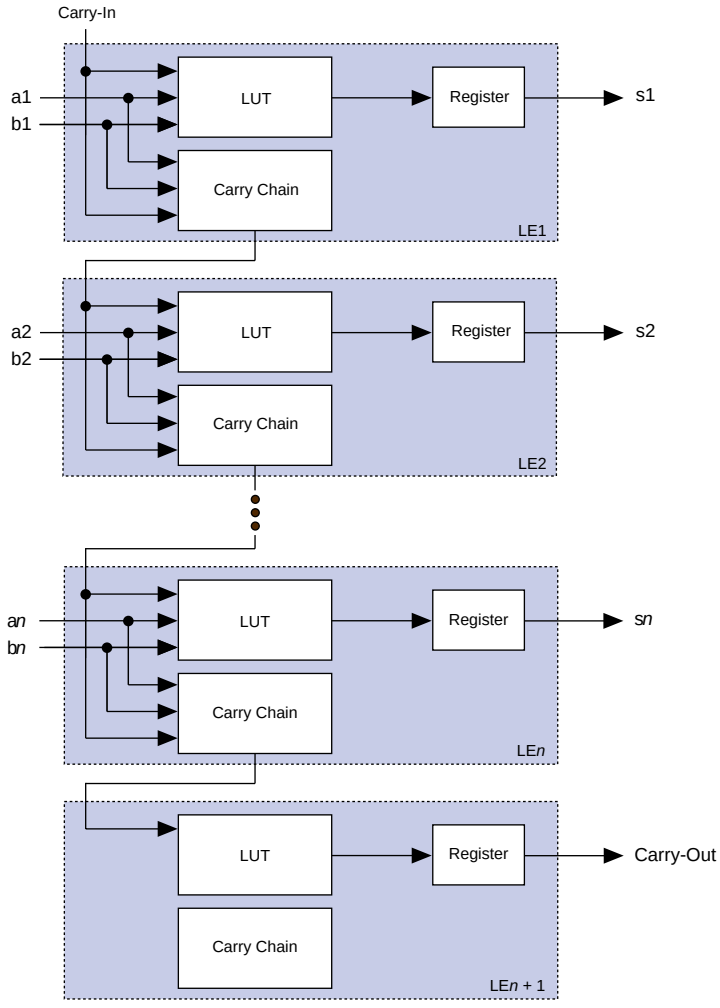
Carry Chain

The carry chain provides a fast carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the APEX II architecture to implement high-speed counters, adders, and comparators of arbitrary width. The Quartus II Compiler can create carry chain logic automatically during the design process, or the designer can create it manually during design entry. Parameterized functions such as DesignWare functions from Synopsys and library of parameterized modules (LPM) functions automatically take advantage of carry chains for the appropriate functions.

The Quartus II Compiler creates carry chains longer than 10 LEs by linking LABs together automatically. For enhanced fitting, a long carry chain skips alternate LABs in a MegaLAB structure. A carry chain longer than one LAB skips either from an even-numbered LAB to the next even-numbered LAB, or from an odd-numbered LAB to the next odd-numbered LAB. For example, the last LE of the first LAB in the upper-left MegaLAB structure carries to the first LE of the third LAB in the MegaLAB structure.

Figure 6 shows how an *n*-bit full adder can be implemented in *n* + 1 LEs with the carry chain. One portion of the LUT generates the sum of two bits using the input signals and the carry-in signal; the sum is routed to the output of the LE. The register can be bypassed for simple adders or used for accumulator functions. Another portion of the LUT and the carry chain logic generates the carry-out signal, which is routed directly to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it is driven onto the local, MegaLAB, or FastTrack interconnect routing structures.

Figure 6. APEX II Carry Chain

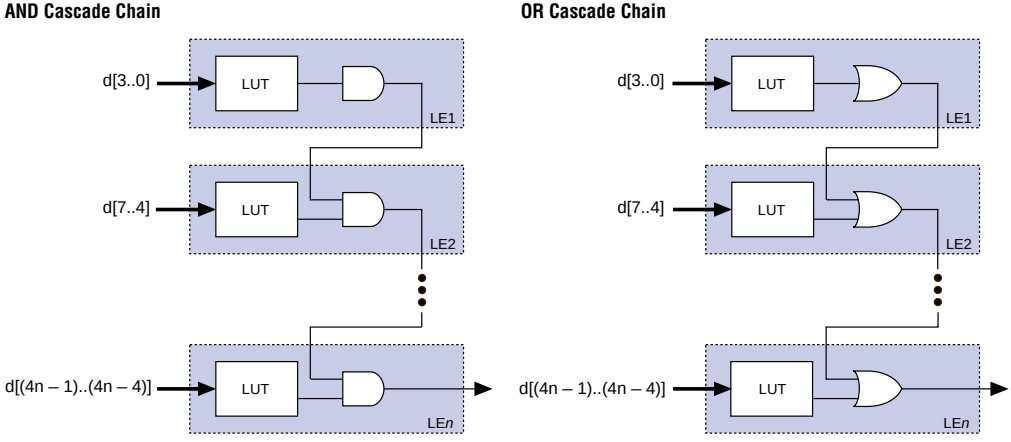


Cascade Chain

With the cascade chain, the APEX II architecture can implement functions with a very wide fan-in. Adjacent LUTs can compute portions of a function in parallel; the cascade chain serially connects the intermediate values. The cascade chain can use a logical AND or logical OR (via DeMorgan's inversion) to connect the outputs of adjacent LEs. Each additional LE provides four more inputs to the effective width of a function, with a short cascade delay. The Quartus II Compiler can create cascade chain logic automatically during the design process, or the designer can create it manually during design entry.

Cascade chains longer than 10 LEs are implemented automatically by linking LABs together. For enhanced fitting, a long cascade chain skips alternate LABs in a MegaLAB structure. A cascade chain longer than one LAB skips either from an even-numbered LAB to the next even-numbered LAB, or from an odd-numbered LAB to the next odd-numbered LAB. For example, the last LE of the first LAB in the upper-left MegaLAB structure carries to the first LE of the third LAB in the MegaLAB structure. Figure 7 shows how the cascade function can connect adjacent LEs to form functions with a wide fan-in.

Figure 7. APEX II Cascade Chain



Normal Mode

The normal mode is suitable for general logic applications, combinatorial functions, or wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a four-input LUT. The Quartus II Compiler automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. LEs in normal mode support packed registers.

Arithmetic Mode

The arithmetic mode is ideal for implementing adders, accumulators, and comparators. An LE in arithmetic mode uses two 3-input LUTs. One LUT computes a three-input function; the other generates a carry output. As shown in Figure 8, the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, when implementing an adder, this output is the sum of three signals: DATA1, DATA2, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain. LEs in arithmetic mode can drive out registered and unregistered versions of the LUT output.

The Quartus II software implements parameterized functions that use the arithmetic mode automatically where appropriate; the designer does not need to specify how the carry chain will be used.

Counter Mode

The counter mode offers clock enable, counter enable, synchronous up / down control, synchronous clear, and synchronous load options. The counter enable and synchronous up / down control signals are generated from the data inputs of the LAB local interconnect. The synchronous clear and synchronous load options are LAB-wide signals that affect all registers in the LAB. Consequently, if any of the LEs in an LAB use the counter mode, other LEs in that LAB must be used as part of the same counter or be used for a combinatorial function. The Quartus II software automatically places any registers that are not used by the counter into other LABs.

By combining multiple ESBs, the Quartus II software implements larger memory blocks automatically. For example, two 256 × 16 RAM blocks can be combined to form a 256 x 32 RAM block, and two 512 × 8 RAM blocks can be combined to form a 512 × 16 RAM block. Memory performance does not degrade for memory blocks up to 4,096 words deep. Each ESB can implement a 4,096-word-deep memory; the ESBs are used in parallel, eliminating the need for any external control logic that would increase delays. To create a high-speed memory block more than 4,096-words deep, the Quartus II software automatically combines ESBs with LE control logic.

Input/Output Clock Mode

The ESB implements input/output clock mode for both dual-port and bidirectional dual-port memory. An ESB using input/output clock mode can use up to two clocks. On each of the two ports, A or B, one clock controls all registers for inputs into the ESB: data input, WREN, read address, and write address. The other clock controls the ESB data output registers. Each ESB port, A or B, also supports independent read clock enable, write clock enable, and asynchronous clear signals. Input/output clock mode is commonly used for applications where the reads and writes occur at the same system frequency, but require different clock enable signals for the input and output registers. Figure 19 shows the ESB in input/output clock mode.

If the same data is written into multiple locations in the memory, a CAM block can be used in multiple-match or fast multiple-match modes. The ESB outputs the matched data's locations as an encoded or unencoded address. In multiple-match mode, it takes two clock cycles to write into a CAM block. For reading, there are 16 outputs from each ESB at each clock cycle. Therefore, it takes two clock cycles to represent the 32 words from a single ESB port. In this mode, encoded and unencoded outputs are available. To implement the encoded version, the Quartus II software adds a priority encoder with LEs. Fast multiple-match is identical to the multiple match mode, however, it only takes one clock cycle to read from a CAM block and generate valid outputs. To do this, the entire ESB is used to represent 16 outputs. In fast multiple-match mode, the ESB can implement a maximum CAM block size of 16 words.

A CAM block can be pre-loaded with data during configuration, or it can be written during system operation. In most cases, two clock cycles are required to write each word into CAM. When don't-care bits are used, a third clock cycle is required.

For more information on CAM, see *Application Note 119 (Implementing High-Speed Search Applications with APEX CAM)*.

Driving Signals to the ESB

ESBs provide flexible options for driving control signals. Different clocks can be used for the ESB inputs and outputs. Registers can be inserted independently on the data input, data output, read address, write address, WE, and RE signals. The global signals and the local interconnect can drive the WE and RE signals. The global signals, dedicated clock pins, and local interconnects can drive the ESB clock signals. Because the LEs drive the local interconnect, the LEs can control the WE and RE signals and the ESB clock, clock enable, and synchronous clear signals. Figure 24 shows the ESB control signal generation logic.

Table 10 describes the I/O standards supported by APEX II devices.

Table 10. APEX II Supported I/O Standards

I/O Standard	Type	Input Reference Voltage (V_{REF}) (V)	Output Supply Voltage (V_{CCIO}) (V)	Board Termination Voltage (V_{TT}) (V)
LVTTL	Single-ended	N/A	3.3	N/A
LVCMOS	Single-ended	N/A	3.3	N/A
2.5 V	Single-ended	N/A	2.5	N/A
1.8 V	Single-ended	N/A	1.8	N/A
1.5 V	Single-ended	N/A	1.5	N/A
3.3-V PCI	Single-ended	N/A	3.3	N/A
3.3-V PCI-X	Single-ended	N/A	3.3	N/A
LVDS	Differential	N/A	3.3	N/A
LVPECL	Differential	N/A	3.3	N/A
PCML	Differential	N/A	3.3	N/A
HyperTransport	Differential	N/A	2.5	N/A
Differential HSTL *(1)*	Differential	N/A	1.5	N/A
GTL+	Voltage referenced	1.0	N/A	1.5
HSTL class I and II	Voltage referenced	0.75	1.5	0.75
SSTL-2 class I and II	Voltage referenced	1.25	2.5	1.25
SSTL-3 class I and II	Voltage referenced	1.5	3.3	1.5
AGP (1× and 2×)	Voltage referenced	1.32	3.3	N/A
CTT	Voltage referenced	1.5	3.3	1.5

Note to Table 10:

(1) Differential HSTL is only supported on the eight dedicated global clock pins and four dedicated high-speed PLL clock pins.

For more information on I/O standards supported by APEX II devices, see *Application Note 117 (Using Selectable I/O Standards in Altera Devices)*.

APEX II devices contain eight I/O banks, as shown in Figure 31. Two blocks within the right I/O banks contain circuitry to support high-speed True-LVDS, LVPECL, PCML, and HyperTransport inputs, and another two blocks within the left I/O banks support high-speed True-LVDS, LVPECL, PCML, and HyperTransport outputs. All other standards are supported by all I/O banks.

Each bank can support multiple standards with the same V_{CCIO} for input and output pins. Each bank can support one voltage-referenced I/O standard, but it can support multiple I/O standards with the same V_{CCIO} voltage level. For example, when V_{CCIO} is 3.3 V, a bank can support LVTTL, LVCMOS, 3.3-V PCI, and SSTL-3 for inputs and outputs. When the True-LVDS banks are not used for LVDS I/O pins, they support all of the other I/O standards except HSTL Class II output.

True-LVDS Interface

APEX II devices contain dedicated circuitry for supporting differential standards at speeds up to 1.0 Gbps. APEX II devices have dedicated differential buffers and circuitry to support LVDS, LVPECL, HyperTransport, and PCML I/O standards. Four dedicated high-speed PLLs (separate from the general-purpose PLLs) multiply reference clocks and drive high-speed differential serializer/deserializer channels. In addition, CDS circuitry at each receiver channel corrects any fixed clock-to-data skew. All APEX II devices support 36 input channels, 36 output channels, two dedicated receiver PLLs, and two dedicated transmitter PLLs.

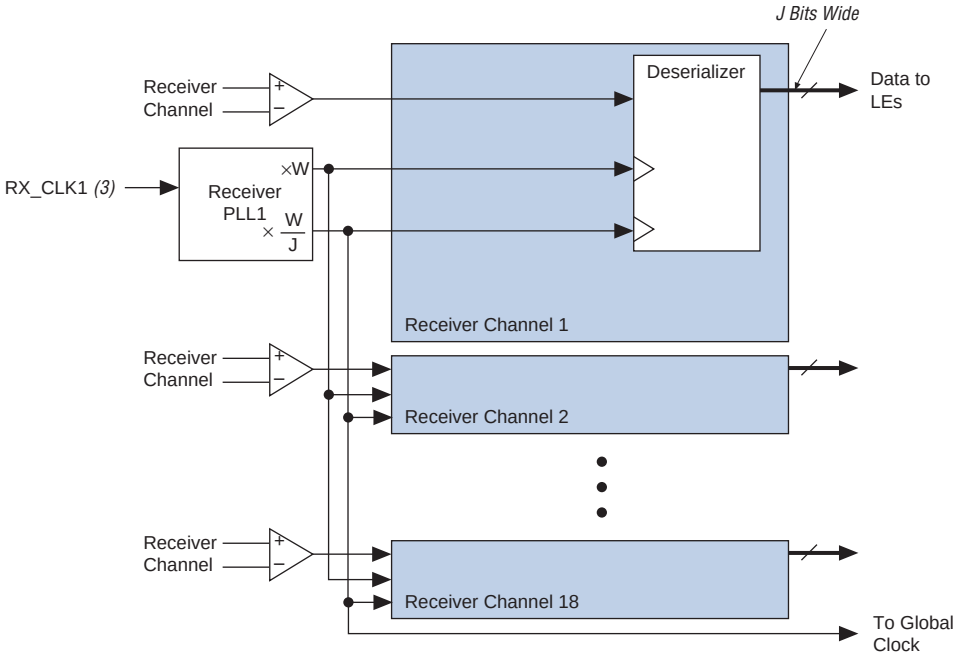
The True-LVDS circuitry supports the following standards and applications:

- RapidIO
- POS-PHY Level 4
- Utopia IV
- HyperTransport

APEX II devices support source-synchronous interfacing with LVDS, LVPECL,PCML, or HyperTransport signaling at up to 1 Gbps. Serial channels are transmitted and received along with a low-speed clock. The receiving device then multiplies the clock by a factor of 1, 2, or 4 to 10. The serialization/deserialization rate can be any number from 1, 2, or 4 to 10 and does not have to equal the clock multiplication value.

For example, an 840-Mbps LVDS channel can be received along with an 84-MHz clock. The 84-MHz clock is multiplied by 10 to drive the serial shift register, but the register can be clocked out in parallel at 8- or 10-bits wide at 84 or 105 MHz. See Figures 32 and 33.

Figure 32. True-LVDS Receiver Diagram *Notes (1), (2)*



Notes to Figure 32:

(1) Two sets of 18 receiver channels are located in each APEX II device. Each set of 18 channels has one receiver PLL.

(2) *W* = 1, 2, 4 to 10
J = 1, 2, 4 to 10
W does not have to equal *J*. When *J* = 1 or 2, the deserializer is bypassed. When *J* = 2, DDR I/O registers are used.

(3) These clock pins drive receiver PLLs only. They do not drive directly to the logic array. However, the receiver PLL can drive the logic array via a global clock line.

Single-Bit Mode

Single-bit CDS corrects a fixed clock-to-data skew of up to ±50% of the data bit period, which allows receiver input skew margin (RSKM) to increase by 50% of the data period. To use single-bit CDS, the deserialization factor, *J*, must be equal to the multiplication factor, *W*. The combination of allowable *W* / *J* factors and the associated CDS training patterns automatically determine byte alignment (see Table 11).

Table 11. Single-Bit CDS Training Patterns

W/J Factor	Single-Bit CDS Pattern
10	0000011111
9	000001111
8	00001111
7	0000111
6	000111
5	00011
4	0011

Multi-Bit Mode

Multi-bit CDS corrects any fixed clock-to-data skew. This feature enables flexible board topologies, such as an N:1 topology (see Figure 34), a switch topology, or a matrix topology. Multi-bit CDS corrects for the skews inherent with these topologies, making them possible to use.

IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

All APEX II devices provide JTAG BST circuitry that complies with the IEEE Std. 1149.1-1990 specification. JTAG boundary-scan testing can be performed before or after configuration, but not during configuration. APEX II devices can also use the JTAG port for configuration with the Quartus II software or with hardware using either Jam™ Standard Test and Programming Language (STAPL) Files (**.jam**) or Jam Byte-Code Files (**.jbc**). Finally, APEX II devices use the JTAG port to monitor the logic operation of the device with the SignalTap embedded logic analyzer. APEX II devices support the JTAG instructions shown in Table 16.

Table 16. APEX II JTAG Instructions

JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins. Also used by the SignalTap embedded logic analyzer.
EXTEST *(1)*	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through selected devices to adjacent devices during normal device operation.
USERCODE	Selects the 32-bit USERCODE register and places it between the TDI and TDO pins, allowing the USERCODE to be serially shifted out of TDO.
IDCODE	Selects the IDCODE register and places it between TDI and TDO, allowing the IDCODE to be serially shifted out of TDO.
HIGHZ *(1)*	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through selected devices to adjacent devices during normal device operation, while tri-stating all of the I/O pins.
CLAMP *(1)*	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through selected devices to adjacent devices during normal device operation while holding I/O pins to a state defined by the data in the boundary-scan register.
ICR instructions	Used when configuring an APEX II device via the JTAG port with a MasterBlaster™ or ByteBlasterMV™ download cable, or when using a Jam File or Jam Byte-Code File via an embedded processor.
SignalTap instructions	Monitors internal device operation with the SignalTap embedded logic analyzer.

Note to Table 16:

(1) Bus hold and weak pull-up features override the high-impedance state of HIGHZ, CLAMP, and EXTEST.

The APEX II device instruction register length is 10 bits. The APEX II device USERCODE register length is 32 bits. Tables 17 and 18 show the boundary-scan register length and device IDCODE information for APEX II devices.

Tables 20 through 41 provide information on absolute maximum ratings, recommended operating conditions, and DC operating conditions for 1.5-V APEX II devices.

Table 20. APEX II Device Absolute Maximum Ratings *Notes (1), (2)*

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCINT}	Supply voltage	With respect to ground *(3)*	–0.5	2.4	V
V_{CCIO}			–0.5	4.6	V
V_I	DC input voltage		–0.5	4.6	V
I_{OUT}	DC output current, per pin		–25	25	mA
T_{STG}	Storage temperature	No bias	–65	150	° C
T_{AMB}	Ambient temperature	Under bias	–65	135	° C
T_J	Junction temperature	BGA packages under bias		135	° C

Table 21. APEX II Device Recommended Operating Conditions

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCINT}	Supply voltage for internal logic and input buffers	*(4)*	1.425	1.575	V
V_{CCIO}	Supply voltage for output buffers, 3.3-V operation	*(4)*, *(5)*	3.00 (3.135)	3.60 (3.465)	V
	Supply voltage for output buffers, 2.5-V operation	*(4)*	2.375	2.625	V
	Supply voltage for output buffers, 1.8-V operation	*(4)*	1.71	1.89	V
	Supply voltage for output buffers, 1.5-V operation	*(4)*	1.4	1.6	V
V_I	Input voltage	*(3)*, *(6)*	–0.5	4.1	V
V_O	Output voltage		0	V_{CCIO}	V
T_J	Operating junction temperature	For commercial use	0	85	° C
		For industrial use	–40	100	° C
t_R	Input rise time			40	ns
t_F	Input fall time			40	ns

Table 34. SSTL-3 Class II Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
V_{CCIO}	Output supply voltage		3.0	3.3	3.6	V
V_{TT}	Termination voltage		V_{REF} – 0.05	V_{REF}	V_{REF} + 0.05	V
V_{REF}	Reference voltage		1.3	1.5	1.7	V
V_{IH}	High-level input voltage		V_{REF} + 0.2		V_{CCIO} + 0.3	V
V_{IL}	Low-level input voltage		–0.3		V_{REF} – 0.2	V
V_{OH}	High-level output voltage	I_{OH} = –16 mA *(10)*	V_{TT} + 0.8			V
V_{OL}	Low-level output voltage	I_{OL} = 16 mA *(10)*			V_{TT} – 0.8	V

Table 35. 3.3-V AGP 2× Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
V_{CCIO}	Output supply voltage		3.15	3.3	3.45	V
V_{REF}	Reference voltage		$0.39 \times V_{CCIO}$		$0.41 \times V_{CCIO}$	V
V_{IH}	High-level input voltage *(11)*		$0.5 \times V_{CCIO}$		V_{CCIO} + 0.5	V
V_{IL}	Low-level input voltage *(11)*				$0.3 \times V_{CCIO}$	V
V_{OH}	High-level output voltage	I_{OUT} = –20 µA	$0.9 \times V_{CCIO}$		3.6	V
V_{OL}	Low-level output voltage	I_{OUT} = 20 µA			$0.1 \times V_{CCIO}$	V
I_I	Input pin leakage current	$0 < V_{IN} < V_{CCIO}$	–10		10	µA

Table 36. 3.3-V AGP 1× Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
V_{CCIO}	Output supply voltage		3.15	3.3	3.45	V
V_{IH}	High-level input voltage *(11)*		$0.5 \times V_{CCIO}$		V_{CCIO} + 0.5	V
V_{IL}	Low-level input voltage *(11)*				$0.3 \times V_{CCIO}$	V
V_{OH}	High-level output voltage	I_{OUT} = –20 µA	$0.9 \times V_{CCIO}$		3.6	V
V_{OL}	Low-level output voltage	I_{OUT} = 20 µA			$0.1 \times V_{CCIO}$	V
I_I	Input pin leakage current	$0 < V_{IN} < V_{CCIO}$	–10		10	µA

Tables 52 through 67 show the APEX II device f_{MAX} and functional timing parameters.

Table 52. EP2A15 f_{MAX} LE Timing Parameters

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{SU}	0.25		0.29		0.33		ns
t_{H}	0.25		0.29		0.33		ns
t_{CO}		0.18		0.20		0.23	ns
t_{LUT}		0.53		0.61		0.70	ns

Table 53. EP2A15 f_{MAX} ESB Timing Parameters

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{ESBARC}		1.28		1.47		1.69	ns
t_{ESBSRC}		2.49		2.86		3.29	ns
t_{ESBAWC}		2.20		2.53		2.91	ns
t_{ESBSWC}		3.02		3.47		3.99	ns
$t_{ESBWASU}$	– 0.55		– 0.64		– 0.73		ns
t_{ESBWAH}	0.15		0.18		0.20		ns
$t_{ESBWDSU}$	0.37		0.43		0.49		ns
t_{ESBWDH}	0.16		0.18		0.21		ns
$t_{ESBRASU}$	0.84		0.96		1.11		ns
t_{ESBRAH}	0.00		0.00		0.00		ns
$t_{ESBWESU}$	0.14		0.16		0.19		ns
$t_{ESBDATASU}$	– 0.02		– 0.03		– 0.03		ns
$t_{ESBWADDRSU}$	– 0.40		– 0.46		– 0.53		ns
$t_{ESBRADDRSU}$	– 0.38		– 0.44		– 0.51		ns
$t_{ESBDATACO1}$		1.30		1.50		1.72	ns
$t_{ESBDATACO2}$		1.84		2.12		2.44	ns
t_{ESBDD}		2.42		2.78		3.19	ns
t_{PD}		1.69		1.94		2.23	ns
$t_{PTERMSU}$	1.10		1.26		1.45		ns
$t_{PTERMCO}$		0.82		0.94		1.08	ns

Table 61. EP2A40 f_{MAX} ESB Timing Parameters

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{ESBARC}		2.28		2.62		3.01	ns
t_{ESBSRC}		2.23		2.56		2.95	ns
t_{ESBAWC}		3.13		3.60		4.13	ns
t_{ESBSWC}		2.76		3.18		3.65	ns
$t_{ESBWASU}$	1.19		1.37		1.57		ns
t_{ESBWAH}	0.00		0.00		0.00		ns
$t_{ESBWDSU}$	1.44		1.66		1.91		ns
t_{ESBWDH}	0.00		0.00		0.00		ns
$t_{ESBRASU}$	1.88		2.17		2.49		ns
t_{ESBRAH}	0.00		0.00		0.00		ns
$t_{ESBWESU}$	1.60		1.85		2.12		ns
$t_{ESBDATASU}$	0.74		0.85		0.98		ns
$t_{ESBWADDRSU}$	0.82		0.94		1.08		ns
$t_{ESBRADDRSU}$	0.73		0.84		.97		ns
$t_{ESBDATACO1}$		1.09		1.25		1.44	ns
$t_{ESBDATACO2}$		1.73		1.99		2.29	ns
t_{ESBDD}		3.26		3.75		4.32	ns
t_{PD}		1.55		1.78		2.05	ns
$t_{PTERMSU}$	0.99		1.13		1.30		ns
$t_{PTERMCO}$		0.79		0.90		1.04	ns

Table 62. EP2A40 f_{MAX} Routing Delays

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{F1-4}	0.17		0.19		0.22		ns
t_{F5-20}	1.12		1.28		1.48		ns
t_{F20+}	1.49		1.72		1.98		ns

Table 65. EP2A70 f_{MAX} ESB Timing Parameters

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{ESBARC}		3.12		3.58		4.12	ns
t_{ESBSRC}		3.11		3.58		4.11	ns
t_{ESBAWC}		4.41		5.07		5.83	ns
t_{ESBSWC}		3.82		4.39		5.05	ns
$t_{ESBWASU}$	1.73		1.99		2.28		ns
t_{ESBWAH}	0.00		0.00		0.00		ns
$t_{ESBWDSU}$	1.87		2.15		2.47		ns
t_{ESBWDH}	0.00		0.00		0.00		ns
$t_{ESBRASU}$	2.76		3.17		3.65		ns
t_{ESBRAH}	0.00		0.00		0.00		ns
$t_{ESBWESU}$	1.98		2.27		2.61		ns
$t_{ESBDATASU}$	1.06		1.22		1.40		ns
$t_{ESBWADDRSU}$	1.17		1.34		1.54		ns
$t_{ESBRADDRSU}$	1.02		1.17		1.35		ns
$t_{ESBDATACO1}$		1.52		1.75		2.01	ns
$t_{ESBDATACO2}$		2.35		2.71		3.11	ns
t_{ESBDD}		4.43		5.10		5.87	ns
t_{PD}		2.17		2.49		2.87	ns
$t_{PTERMSU}$	1.40		1.62		1.86		ns
$t_{PTERMCO}$		1.08		1.24		1.42	ns

Table 66. EP2A70 f_{MAX} Routing Delays

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{F1-4}	0.15		0.18		0.20		ns
t_{F5-20}	1.21		1.39		1.60		ns
t_{F20+}	1.87		2.15		2.55		ns

Table 73. EP2A40 External Timing Parameters for Column I/O Pins

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.00		2.16		2.33		ns
t_{INH}	0.00		0.00		0.00		ns
t_{OUTCO}	2.00	4.96	2.00	5.29	2.00	5.64	ns
t_{XZ}		7.04		7.59		8.19	ns
t_{ZX}		7.04		7.59		8.19	ns
$t_{INSUPLL}$	1.20		1.31		1.43		ns
t_{INHPLL}	0.00		0.00		0.00		ns
$t_{OUTCOPLL}$	0.50	2.66	0.50	2.87	0.50	3.09	ns
t_{XZPLL}		4.74		5.17		5.64	ns
t_{ZXPLL}		4.74		5.17		5.64	ns

Table 74. EP2A70 External Timing Parameters for Row I/O Pins

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.48		2.68		2.90		ns
t_{INH}	0.00		0.00		0.00		ns
t_{OUTCO}	2.00	4.76	2.00	5.12	2.00	5.51	ns
t_{XZ}		5.68		6.19		6.76	ns
t_{ZX}		5.68		6.19		6.76	ns
$t_{INSUPLL}$	1.19		1.30		1.43		ns
t_{INHPLL}	0.00		0.00		0.00		ns
$t_{OUTCOPLL}$	0.50	2.52	0.50	2.74	0.50	2.98	ns
t_{XZPLL}		3.44		3.82		4.23	ns
t_{ZXPLL}		3.44		3.82		4.23	ns

Revision History

The information contained in the *APEX II Programmable Logic Device Family Data Sheet* version 3.0 supersedes information published in previous versions. The following changes were made to the *APEX II Programmable Logic Device Family Data Sheet* version 3.0:

- Changed the value from 624 to 400 Mbps throughout the document.
- Deleted the pin count (612) for the EP2A25 device in the 1,020-pin FineLine BGA package (see Table 3).
- Added Table 13.
- Changed the maximum value of 3.6 to 2.4 in Table 20.
- Updated Tables 60 through 67 and Tables 72 through 75.
- Updated Figures 25, 28, and 30.
- Added *Note (1)* to Figure 13.
- Added Figure 43.

101 Innovation Drive
San Jose, CA 95134
(408) 544-7000
http://www.altera.com
Applications Hotline:
(800) 800-EPLD
Customer Marketing:
(408) 544-7104
Literature Services:
lit_req@altera.com