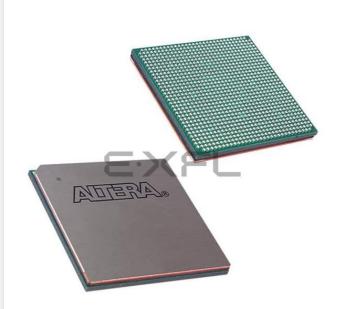
Intel - EP2A40F1020C7 Datasheet





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Understanding <u>Embedded - FPGAs (Field</u> <u>Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	3840
Number of Logic Elements/Cells	38400
Total RAM Bits	655360
Number of I/O	735
Number of Gates	3000000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	1020-BBGA
Supplier Device Package	1020-FBGA (33x33)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep2a40f1020c7

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

General Description

APEX II devices integrate high-speed differential I/O support using the True-LVDS interface. The dedicated serializer, deserializer, and CDS circuitry in the True-LVDS interface support the LVDS, LVPECL, HyperTransport, and PCML I/O standards. Flexible-LVDS pins located in regular user I/O banks offer additional differential support, increasing the total device bandwidth. This circuitry, together with enhanced IOEs and support for numerous I/O standards, allows APEX II devices to meet high-speed interface requirements.

APEX II devices also include other high-performance features such as bidirectional dual-port RAM, CAM, general-purpose PLLs, and numerous global clocks.

Configuration

The logic, circuitry, and interconnects in the APEX II architecture are configured with CMOS SRAM elements. APEX II devices are reconfigurable and are 100% tested prior to shipment. As a result, test vectors do not have to be generated for fault coverage. Instead, the designer can focus on simulation and design verification. In addition, the designer does not need to manage inventories of different ASIC designs; APEX II devices can be configured on the board for the specific functionality required.

APEX II devices are configured at system power-up with data either stored in an Altera configuration device or provided by a system controller. Altera offers in-system programmability (ISP)-capable configuration devices, which configure APEX II devices via a serial data stream. The enhanced configuration devices can configure any APEX II device in under 100 ms. Moreover, APEX II devices contain an optimized interface that permits microprocessors to configure APEX II devices serially or in parallel, synchronously or asynchronously. This interface also enables microprocessors to treat APEX II devices as memory and to configure the device by writing to a virtual memory location, simplifying reconfiguration.

APEX II devices also support a new byte-wide, synchronous configuration scheme at speeds of up to 66 MHz using EPC16 configuration devices or a microprocessor. This parallel configuration reduces configuration time by using eight data lines to send configuration data versus one data line in serial configuration.

APEX II devices support multi-voltage configuration; device configuration can be performed at 3.3 V and 2.5 V or 1.8 V.

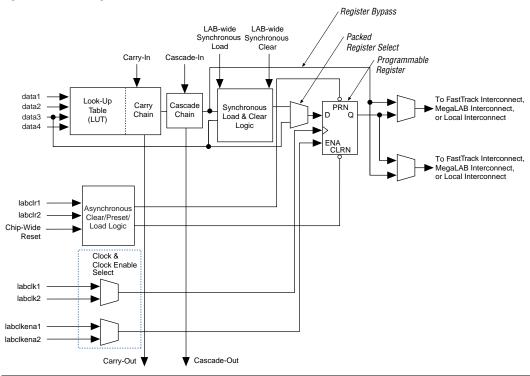


Figure 5. APEX II Logic Element

Each LE's programmable register can be configured for D, T, JK, or SR operation. The register's clock and clear control signals can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the register is bypassed and the output of the LUT drives the outputs of the LE.

LE Operating Modes

The APEX II LE can operate in one of the following three modes:

- Normal mode
- Arithmetic mode
- Counter mode

Each mode uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. LAB-wide signals provide clock, asynchronous clear, asynchronous preset, asynchronous load, synchronous clear, synchronous load, and clock enable control for the register. These LAB-wide signals are available in all LE modes.

The Quartus II software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions that specify which LE operating mode to use for optimal performance. Figure 8 shows the LE operating modes.

Normal Mode

The normal mode is suitable for general logic applications, combinatorial functions, or wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a four-input LUT. The Quartus II Compiler automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. LEs in normal mode support packed registers.

Arithmetic Mode

The arithmetic mode is ideal for implementing adders, accumulators, and comparators. An LE in arithmetic mode uses two 3-input LUTs. One LUT computes a three-input function; the other generates a carry output. As shown in Figure 8, the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, when implementing an adder, this output is the sum of three signals: DATA1, DATA2, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain. LEs in arithmetic mode can drive out registered and unregistered versions of the LUT output.

The Quartus II software implements parameterized functions that use the arithmetic mode automatically where appropriate; the designer does not need to specify how the carry chain will be used.

Counter Mode

The counter mode offers clock enable, counter enable, synchronous up/down control, synchronous clear, and synchronous load options. The counter enable and synchronous up/down control signals are generated from the data inputs of the LAB local interconnect. The synchronous clear and synchronous load options are LAB-wide signals that affect all registers in the LAB. Consequently, if any of the LEs in an LAB use the counter mode, other LEs in that LAB must be used as part of the same counter or be used for a combinatorial function. The Quartus II software automatically places any registers that are not used by the counter into other LABs.

A column line can be directly driven by the LEs, IOEs, or ESBs in that column. Row IOEs can drive a column line on a device's left or right edge. The column line is used to route signals from one row to another. A column line can drive a row line; it can also drive the MegaLAB interconnect directly, allowing faster connections between rows.

Figure 10 shows how the FastTrack interconnect uses the local interconnect to drive LEs within MegaLAB structures.

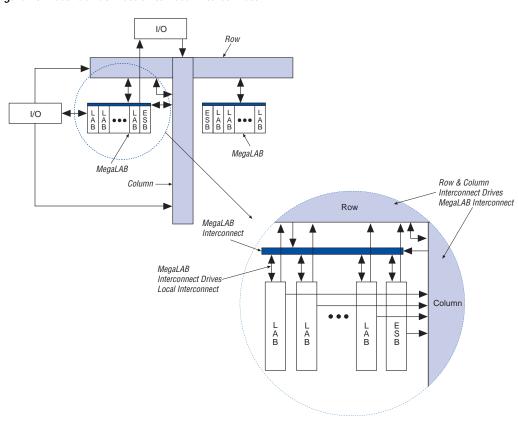


Figure 10. FastTrack Connection to Local Interconnect

Figure 11 shows the intersection of a row and column interconnect and how these forms of interconnects and LEs drive each other.

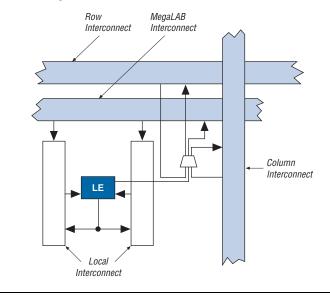


Figure 11. Driving the FastTrack Interconnect

APEX II devices feature FastRowTM lines for quickly routing input signals with high fan-out. Column I/O pins can drive the FastRow interconnect, which routes signals directly into the local interconnect without having to drive through the MegaLAB interconnect. FastRow lines traverse two MegaLAB structures. The FastRow interconnect drives the four MegaLABs in the top row and the four MegaLABs in the bottom row of the device. The FastRow interconnect drives all local interconnects in the appropriate MegaLABs. Column pins using the FastRow interconnect achieve a faster set-up time, because the signal does not need to use a MegaLab interconnect line to reach the destination LE. Figure 12 shows the FastRow interconnect.

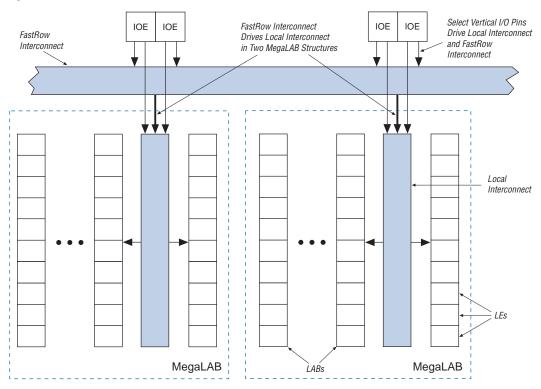




Table 5 summarizes how elements of the APEX II architecture drive each other.

Table 5. API	EX II Ro	uting Sch	neme						
Source					De	stination			
	Row I/O Pin	Column I/O Pin	LE	ESB	Local Interconnect	MegaLAB Interconnect	Row FastTrack Interconnect	Column FastTrack Interconnect	FastRow Interconnect
Row I/O pin					\checkmark	~	\checkmark	\checkmark	
Column I/O pin								~	~
LE					\checkmark	~	\checkmark	\checkmark	
ESB					✓	\checkmark	\checkmark	\checkmark	
Local interconnect	~	~	~	~					
MegaLAB interconnect					~				
Row FastTrack interconnect						~		~	
Column FastTrack interconnect						~	~		
FastRow interconnect					~				

Product-Term Logic

The product-term portion of the MultiCore architecture is implemented with the ESB. The ESB can be configured to act as a block of macrocells on an ESB-by-ESB basis. 32 inputs from the adjacent local interconnect feed each ESB; therefore, the either MegaLAB or the adjacent LAB can drive the ESB. Also, nine ESB macrocells feed back into the ESB through the local interconnect for higher performance. Dedicated clock pins, global signals, and additional inputs from the local interconnect drive the ESB control signals.

In product-term mode, each ESB contains 16 macrocells. Each macrocell consists of two product terms and a programmable register. Figure 13 shows the ESB in product-term mode.

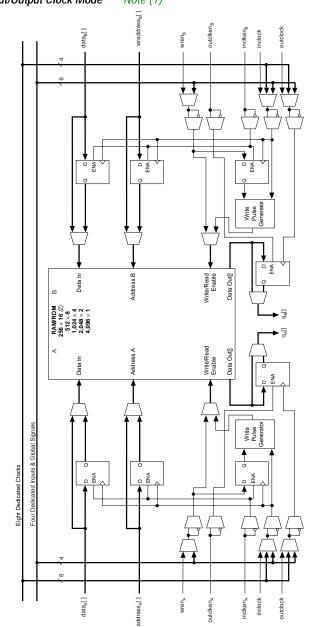


Figure 19. ESB in Input/Output Clock Mode Note (1)

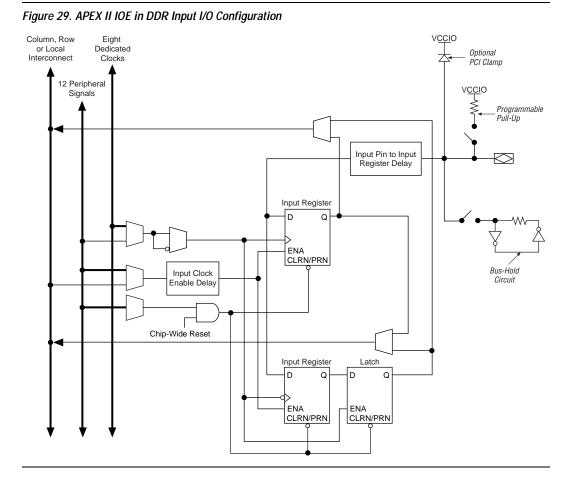
Notes to Figure 19:

- (1) All registers can be cleared asynchronously by ESB local interconnect signals, global signals, or the chip-wide reset.
- (2) This configuration is not supported for bidirectional dual-port configuration.

ble 7. Peripheral Control Bus Destina	ations
Peripheral Bus	I/O Control Signal
Output Enable 0 [OE0]	OE
Output Enable 1 [OE1]	OE
Output Enable 2 [OE2]	OE
Output Enable 3 [OE3]	OE
Output Enable 4 [OE4]	OE
Output Enable 5 [OE5]	OE
Clock Enable 0 [CE0]	CE, CLK
Clock Enable 1 [CE1]	CE, OE
Clock Enable 2 [CE2]	CE, CLK
Clock Enable 3 [CE3]	CE, OE
Clock Enable 4 [CE4]	CE, CLR
Clock Enable 5 [CE5]	CE, CLR

In normal bidirectional operation, the input register can be used for input data requiring fast setup times. The input register can have its own clock input and clock enable separate from the OE and output registers. The output register can be used for data requiring fast clock-to-output performance. The OE register can be used for fast clock-to-output enable timing. The OE and output register share the same clock source and the same clock enable source from local interconnect in the associated LAB, fast global signals, or row global signals. Figure 28 shows the IOE in bidirectional configuration.

When using the IOE for DDR inputs, the two input registers are used to clock double rate input data on alternating edges. An input latch is also used within the IOE for DDR input acquisition. The latch holds the data that is present during the clock high times. This allows both bits of data to be synchronous to the same clock edge (either rising or falling). Figure 29 shows an IOE configured for DDR input.



When using the IOE for DDR outputs, the two output registers are configured to clock two data paths from LEs on rising clock edges. These register outputs are multiplexed by the clock to drive the output pin at a \times 2 rate. One output register clocks the first bit out on the clock high time, while the other output register clocks the second bit out on the clock low time. Figure 30 shows the IOE configured for DDR output.

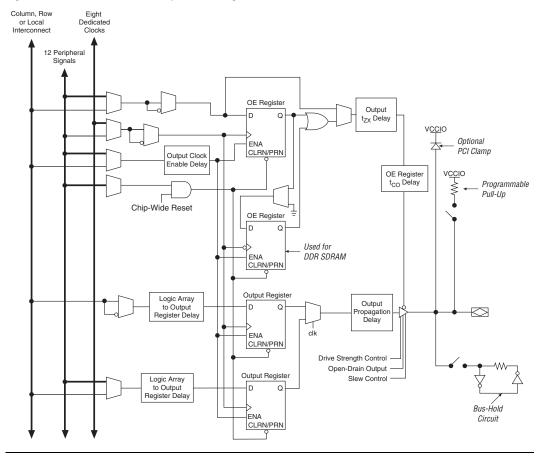


Figure 30. APEX II IOE in DDR Output I/O Configuration

The APEX II IOE operates in bidirectional DDR mode by combining the DDR input and DDR output configurations.

APEX II I/O pins transfer data on a DDR bidirectional bus to support DDR SDRAM at 167 MHz (334 Mbps). The negative-edge-clocked OE register is used to hold the OE signal inactive until the falling edge of the clock. This is done to meet DDR SDRAM timing requirements. QDR SRAMs are also supported with DDR I/O pins on separate read and write ports.

Zero Bus Turnaround SRAM Interface Support

In addition to DDR SDRAM support, APEX II device I/O pins also support interfacing with ZBT SRAM devices at up to 200 MHz. ZBT SRAM blocks are designed to eliminate dead bus cycles when turning a bidirectional bus around between reads and writes, or writes and reads. ZBT allows for 100% bus utilization because ZBT SRAM can be read or written on every clock cycle.

To avoid bus contention, the output clock-to-low-impedance time (t_{ZX}) delay ensures that the t_{ZX} is greater than the clock-to-high-impedance time (t_{XZ}). Phase delay control of clocks to the OE/output and input registers using two general-purpose PLLs enable the APEX II device to meet ZBT t_{CO} and t_{SU} times.

Programmable Drive Strength

The output buffer for each APEX II device I/O pin has a programmable drive strength control for certain I/O standards. The LVTTL standard has several levels of drive strength that the user can control. SSTL-3 class I and II, SSTL-2 class I and II, HSTL class I and II, 3.3-V GTL+, PCI, and PCI-X support a minimum setting. The minimum setting is the lowest drive strength that guarantees the I_{OH}/I_{OL} of the standard. Using minimum settings provides signal slew rate control to reduce system noise and signal overshoot. Table 9 shows the possible settings for the I/O standards with drive strength control.

Bus Hold

Each APEX II device I/O pin provides an optional bus-hold feature. When this feature is enabled for an I/O pin, the bus-hold circuitry weakly holds the signal at its last driven state. By holding the last driven state of the pin until the next input signal is present, the bus-hold feature eliminates the need to add external pull-up or pull-down resistors to hold a signal level when the bus is tri-stated. The bus-hold circuitry also pulls undriven pins away from the input threshold voltage where noise can cause unintended high-frequency switching. This feature can be selected individually for each I/O pin. The bus-hold output will drive no higher than $V_{\rm CCIO}$ to prevent overdriving signals. If the bus-hold feature is enabled, the programmable pull-up option cannot be used. The bus-hold feature should also be disabled if open-drain outputs are used with the GTL+I/O standard.

The bus-hold circuitry weakly pulls the signal level to the last driven state through a resistor with a nominal resistance (R_{BH}) of approximately 7 k Ω . Table 41 on page 74 gives specific sustaining current that will be driven through this resistor and overdrive current that will identify the next driven input level. This information is provided for each V_{CCIO} voltage level.

The bus-hold circuitry is active only after configuration. When going into user mode, the bus-hold circuit captures the value on the pin present at the end of configuration.

Programmable Pull-Up Resistor

Each APEX II device I/O pin provides an optional programmable pull-up resistor during user mode. When this feature is enabled for an I/O pin, the pull-up resistor (typically 25 k Ω) weakly holds the output to the V_{CCIO} level of the bank that the output pin resides in.

Dedicated Fast I/O Pins

APEX II devices incorporate dedicated bidirectional pins for signals with high internal fanout, such as PCI control signals. These pins are called dedicated fast I/O pins (FAST1, FAST2, FAST3, and FAST4) and can drive the four global fast lines throughout the device, ideal for fast clock, clock enable, preset, clear, or high fanout logic signal distribution. The dedicated fast I/O pins have one output register and one OE register, but they do not have input registers. The dedicated fast lines can also be driven by a LE local interconnect to generate internal global signals.

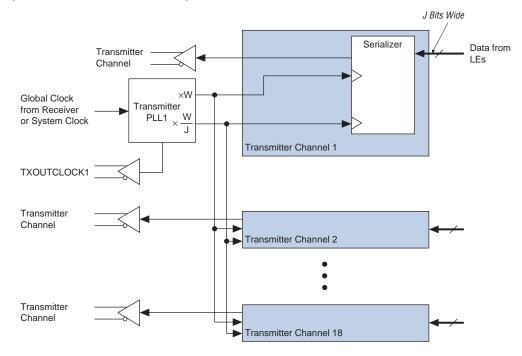


Figure 33. True-LVDS Transmitter Diagram Notes (1), (2)

Notes to Figure 33:

- (1) Two sets of 18 transmitter channels are located in each APEX II device. Each set of 18 channels has one transmitter PLL.
- (2) W = 1, 2, 4 to 10J = 1, 2, 4 to 10

W does not have to equal J. When J = 1 or 2, the deserializer is bypassed. When J = 2, DDR I/O registers are used.

Clock-Data Synchronization

In addition to dedicated serial-to-parallel converters, APEX II True-LVDS circuitry contains CDS circuitry in every receiver channel. The CDS feature can be turned on or off independently for each receiver channel. There are two modes for the CDS circuitry: single-bit mode, which corrects a fixed clock-to-data skew of up to $\pm 50\%$ of the data bit period, and multi-bit mode, which corrects any fixed clock-to-data skew.

The APEX II VCCINT pins must always be connected to a 1.5-V power supply. With a 1.5-V V_{CCINT} level, input pins are 1.5-V, 1.8-V, 2.5-V and 3.3-V tolerant. The VCCIO pins can be connected to either a 1.5-V, 1.8-V, 2.5-V or 3.3-V power supply, depending on the output requirements. The output levels are compatible with systems of the same voltage as the power supply (i.e., when VCCIO pins are connected to a 1.5-V power supply, the output levels are compatible with 1.5-V systems). When VCCIO pins are connected to a 3.3-V power supply, the output high is 3.3 V and is compatible with 3.3-V or 5.0-V systems.

Table 14 summarizes APEX II MultiVolt I/O support.

Table 14.	Table 14. APEX II MultiVolt I/O Support Note (1)										
V _{CCIO} (V)	0 (V) Input Signal Output Signal										
	1.5 V	1.8 V	2.5 V	3.3 V	5.0 V	1.5 V	1.8 V	2.5 V	3.3 V	5.0 V	
1.5	\checkmark	~	\checkmark	~		~					
1.8	 (2) 	 Image: A set of the set of the	\checkmark	~		 (3) 	~				
2.5	 (2) 	🗸 (2)	\checkmark	~		(4)	(4)	~			
3.3	 (2) 	 (2) 	\checkmark	\checkmark	 (5) 	 (6) 	 (6) 	 (6) 	\checkmark	 	

Notes to Table 14:

The PCI clamping diode must be disabled to drive an input with voltages higher than V_{CCIO} , except for with a 5.0-V (1)input.

These input levels are only allowed if the input standard is set to any V_{REF} standard (i.e., SSTL-3, SSTL-2, HSTL, (2)GTL+, and AGP 2×). The V_{REF} standard inputs are powered by V_{CCINT}. LVTTL, PCI, PCI-X, and AGP 1× standard inputs are powered by V_{CCIO} . As a result, input levels below the V_{CCIO} setting cannot drive these standards. When $V_{CCIO} = 1.8$ V, an APEX II device can drive a 1.5-V device with 1.8-V tolerant inputs.

(3)

When $V_{CCIO} = 2.5$ V, an APEX II device can drive a 1.5-V or 1.8-V device with 2.5-V tolerant inputs. (4)

(5) APEX II devices can be 5.0-V tolerant with the use of an external series resistor and enabling the PCI clamping diode.

When V_{CCIO} = 3.3 V, an APEX II device can drive a 1.5-V, 1.8-V, or 2.5-V device with 3.3-V tolerant inputs. (6)

> Open-drain output pins with a pull-up resistor to the 5.0-V supply and a series register to the I/O pin can drive 5.0-V CMOS input pins that require a V_{IH} of 3.5 V. When the pin is inactive, the trace will be pulled up to 5.0 V by the resistor. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The IOL current specification should be considered when selecting a pull-up resistor.

Because APEX II devices can be used in a mixed-voltage environment, Power they have been designed specifically for any possible power-up sequence. Sequencing & Therefore, the V_{CCIO} and V_{CCINT} power supplies may be powered in any Hot Socketing order.

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
V _{CCIO}	Output supply voltage		1.4	1.5	1.6	V
V _{REF}	Input reference voltage		0.68	0.75	0.9	V
V _{TT}	Termination voltage		0.7	0.75	0.8	V
V _{IH} (DC)	DC high-level input voltage		V _{REF} + 0.1			V
V _{IL} (DC)	DC low-level input voltage		-0.3		V _{REF} – 0.1	V
V _{IH} (AC)	AC high-level input voltage		V _{REF} + 0.2			V
V _{IL} (AC)	AC low-level input voltage				V _{REF} – 0.2	V
V _{OH}	High-level output voltage	I _{OH} = 8 mA (10)	V _{CCIO} – 0.4			V
V _{OL}	Low-level output voltage	$I_{OL} = -8 \text{ mA} (10)$			0.4	V

Table 38. 1.5-V HSTL Class II Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
V _{CCIO}	Output supply voltage		1.4	1.5	1.6	V
V _{REF}	Input reference voltage		0.68	0.75	0.9	V
V _{TT}	Termination voltage		0.7	0.75	0.8	V
V _{IH} (DC)	DC high-level input voltage		V _{REF} + 0.1			V
V _{IL} (DC)	DC low-level input voltage		-0.3		V _{REF} – 0.1	V
V _{IH} (AC)	AC high-level input voltage		V _{REF} + 0.2			V
V _{IL} (AC)	AC low-level input voltage				V _{REF} - 0.2	V
V _{OH}	High-level output voltage	I _{OH} = 16 mA <i>(10)</i>	$V_{CCIO} - 0.4$			V
V _{OL}	Low-level output voltage	I _{OL} = -16 mA (10)			0.4	V

Table 39. 1.5-V Differential HSTL Specifications									
Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units			
V _{CCIO}	I/O supply voltage		1.4	1.5	1.6	V			
V _{DIF} (DC)	DC input differential voltage		0.2			V			
V _{CM} (DC)	DC common mode input voltage		0.68		0.9	V			
V _{DIF} (AC)	AC differential input voltage		0.4			V			

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Table 54. EP2A15 f _{MAX} Routing Delays									
Symbol	-7 Spee	d Grade	-8 Spee	d Grade	-9 Spee	Unit			
	Min	Max	Min	Max	Min	Мах			
t _{F1-4}	0.19		0.21		0.25		ns		
t _{F5-20}	0.64		0.73		0.84		ns		
t _{F20+}	1.18		1.35		1.56		ns		

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Мах	Min	Мах	Min	Мах	
t _{CH}	1.00		1.15		1.32		ns
t _{CL}	1.00		1.15		1.32		ns
t _{CLRP}	0.13		0.15		0.17		ns
t _{PREP}	0.13		0.15		0.17		ns
t _{ESBCH}	1.00		1.15		1.32		ns
t _{ESBCL}	1.00		1.15		1.32		ns
t _{ESBWP}	1.12		1.28		1.48		ns
t _{ESBRP}	0.88		1.02		1.17		ns

Table 56.	EP2A25 f _{MAX} LE	Timing	Parameters
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Symbol	-7 Spee	d Grade	-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Мах	
t _{SU}	0.25		0.29		0.33		ns
t _H	0.25		0.29		0.33		ns
t _{CO}		0.18		0.20		0.23	ns
t _{LUT}		0.53		0.61		0.70	ns

Table 67. EP2A70 Minimum Pulse Width Timing Parameters									
Symbol	-7 Speed Grade		-8 Spee	-8 Speed Grade		-9 Speed Grade			
	Min	Мах	Min	Мах	Min	Мах			
t _{CH}	1.19		1.78		2.53		ns		
t _{CL}	1.19		1.78		2.53		ns		
t _{CLRP}	0.16		0.19		0.21		ns		
t _{PREP}	0.16		0.19		0.21		ns		
t _{ESBCH}	1.19		1.78		2.53		ns		
t _{ESBCL}	1.19		1.78		2.53		ns		
t _{ESBWP}	1.35		1.56		1.79		ns		
t _{ESBRP}	1.13		1.30		1.50		ns		

Tables 68 through 77 show the IOE external timing parameter values for APEX II devices.

Table 68. EP2A	15 External Ti	ming Parame	eters for Row	I/O Pins			
Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Мах	1
t _{INSU}	2.06		2.25		2.46		ns
t _{INH}	0.00		0.00		0.00		ns
t _{оитсо}	2.00	4.05	2.00	4.45	2.00	4.90	ns
t _{xz}		4.98		5.59		6.26	ns
t _{ZX}		4.98		5.59		6.26	ns
t _{INSUPLL}	1.15		1.28		1.42		ns
t _{INHPLL}	0.00		0.00		0.00		ns
t _{OUTCOPLL}	0.50	2.60	0.50	2.87	0.50	3.16	ns
t _{XZPLL}		3.53		4.00		4.52	ns
t _{ZXPLL}		3.53		4.00		4.52	ns

Table 73. EP2A40 External Timing Parameters for Column I/O Pins							
Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{INSU}	2.00		2.16		2.33		ns
t _{INH}	0.00		0.00		0.00		ns
t _{оитсо}	2.00	4.96	2.00	5.29	2.00	5.64	ns
t _{XZ}		7.04		7.59		8.19	ns
t _{ZX}		7.04		7.59		8.19	ns
	1.20		1.31		1.43		ns
	0.00		0.00		0.00		ns
t _{OUTCOPLL}	0.50	2.66	0.50	2.87	0.50	3.09	ns
t _{XZPLL}		4.74		5.17		5.64	ns
tZXPLL		4.74		5.17		5.64	ns

Table 74. EP2A	70 External Ti	ming Parame	eters for Row	I/O Pins			
Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Мах	Min	Max	Min	Max	1
t _{INSU}	2.48		2.68		2.90		ns
t _{INH}	0.00		0.00		0.00		ns
t _{outco}	2.00	4.76	2.00	5.12	2.00	5.51	ns
t _{XZ}		5.68		6.19		6.76	ns
t _{ZX}		5.68		6.19		6.76	ns
t _{INSUPLL}	1.19		1.30		1.43		ns
t _{INHPLL}	0.00		0.00		0.00		ns
t _{OUTCOPLL}	0.50	2.52	0.50	2.74	0.50	2.98	ns
t _{XZPLL}		3.44		3.82		4.23	ns
tZXPLL		3.44		3.82		4.23	ns