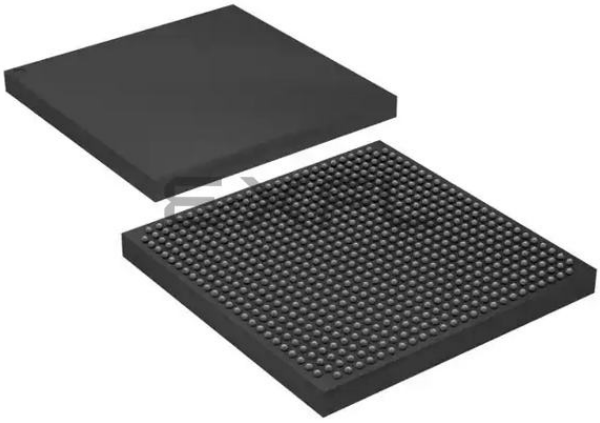

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

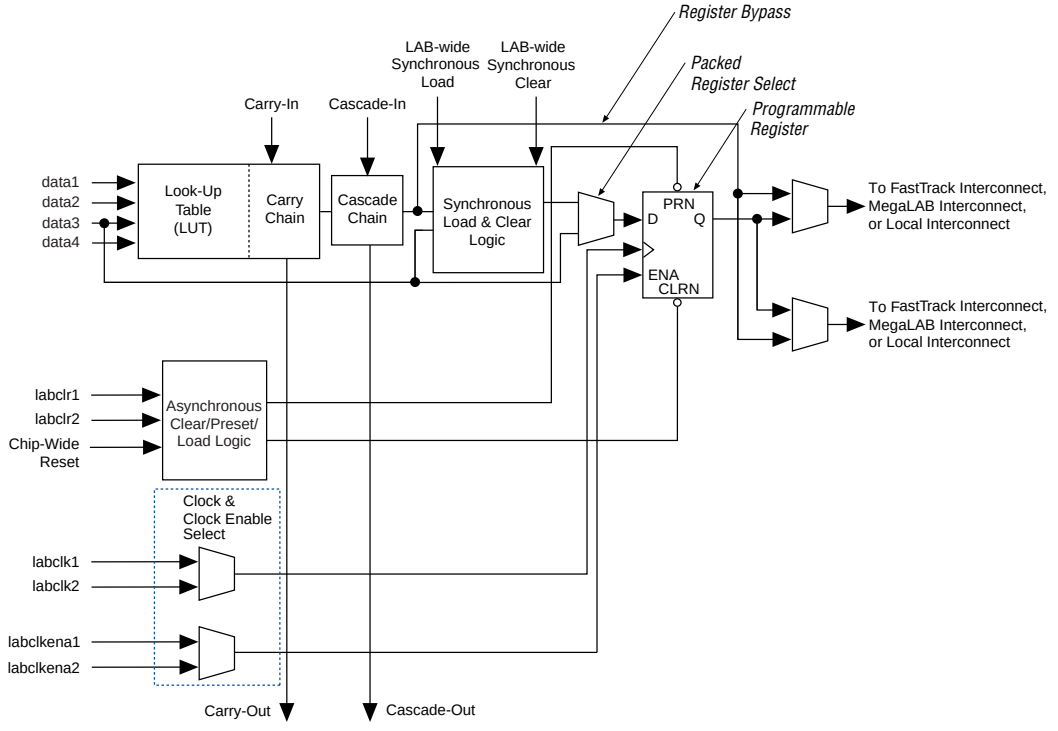
Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	3840
Number of Logic Elements/Cells	38400
Total RAM Bits	655360
Number of I/O	492
Number of Gates	3000000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	672-BBGA
Supplier Device Package	672-FBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep2a40f672c8

Figure 5. APEX II Logic Element



Each LE's programmable register can be configured for D, T, JK, or SR operation. The register's clock and clear control signals can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the register is bypassed and the output of the LUT drives the outputs of the LE.

The counter mode uses two three-input LUTs: one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading, and another AND gate provides synchronous clearing. If the cascade function is used by an LE in counter mode, the synchronous clear or load overrides any signal carried on the cascade chain. The synchronous clear overrides the synchronous load. LEs in arithmetic mode can drive out registered and unregistered versions of the LUT output.

Clear & Preset Logic Control

Logic for the register's clear and preset signals is controlled by LAB-wide signals. The LE directly supports an asynchronous clear function. The Quartus II Compiler can use a NOT-gate push-back technique to emulate an asynchronous preset. Moreover, the Quartus II Compiler can use a programmable NOT-gate push-back technique to emulate simultaneous preset and clear or asynchronous load. However, this technique uses three additional LEs per register. All emulation is performed automatically when the design is compiled. Registers that emulate simultaneous preset and load will enter an unknown state upon power-up or when the chip-wide reset is asserted.

In addition to the two clear and preset modes, APEX II devices provide a chip-wide reset pin (DEV_CLRn) that resets all registers in the device. Use of this pin is controlled through an option in the Quartus II software that is set before compilation. The chip-wide reset overrides all other control signals. Registers using an asynchronous preset are preset when the chip-wide reset is asserted; this effect results from the inversion technique used to implement the asynchronous preset.

FastTrack Interconnect

In the APEX II architecture, connections between LEs, ESBs, and I/O pins are provided by the FastTrack interconnect. The FastTrack interconnect is a series of continuous horizontal and vertical routing channels that traverse the device. This global routing structure provides predictable performance, even in complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.

Figure 12. APEX II FastRow Interconnect

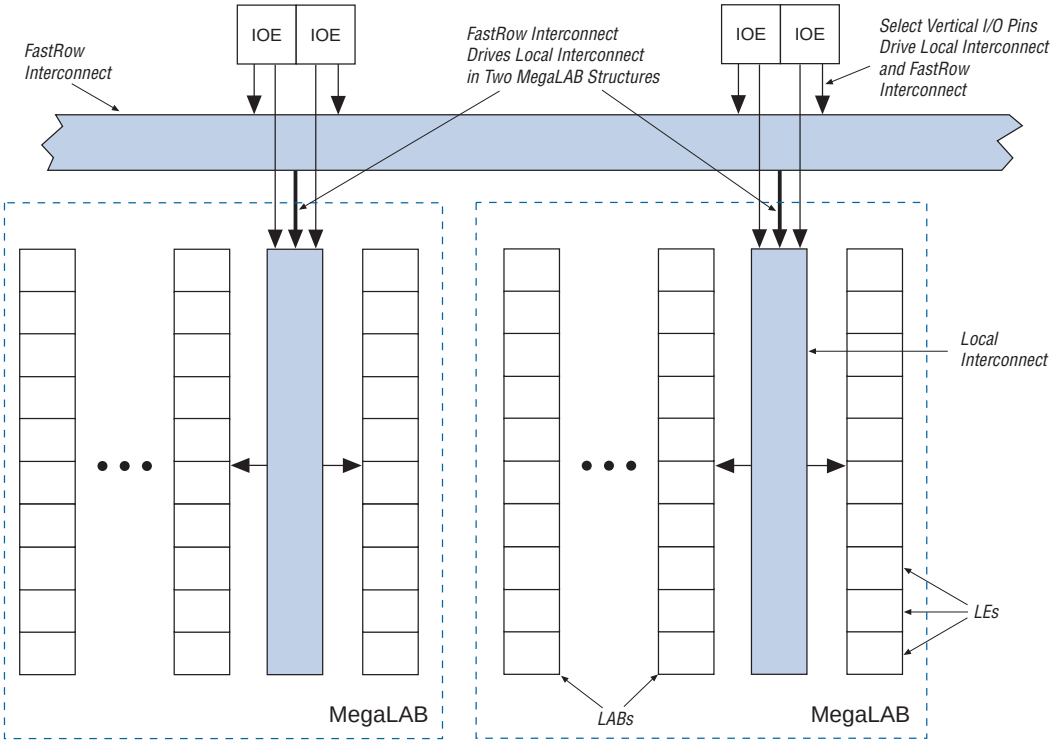


Table 5 summarizes how elements of the APEX II architecture drive each other.

Table 5. APEX II Routing Scheme

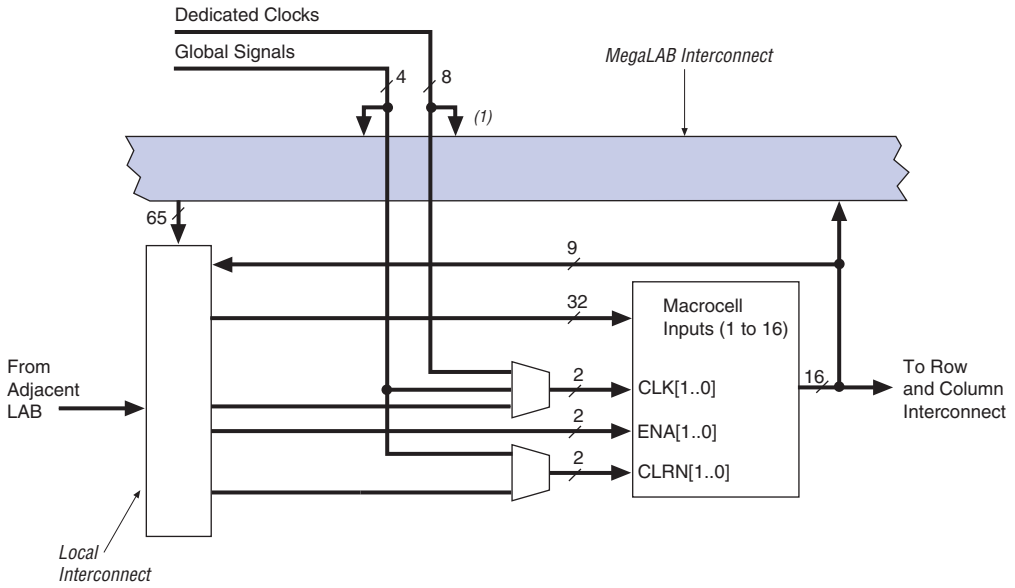
Source	Destination								
	Row I/O Pin	Column I/O Pin	LE	ESB	Local Interconnect	MegaLAB Interconnect	Row FastTrack Interconnect	Column FastTrack Interconnect	FastRow Interconnect
Row I/O pin					✓	✓	✓	✓	
Column I/O pin								✓	✓
LE					✓	✓	✓	✓	
ESB					✓	✓	✓	✓	
Local interconnect	✓	✓	✓	✓					
MegaLAB interconnect					✓				
Row FastTrack interconnect						✓		✓	
Column FastTrack interconnect						✓	✓		
FastRow interconnect					✓				

Product-Term Logic

The product-term portion of the MultiCore architecture is implemented with the ESB. The ESB can be configured to act as a block of macrocells on an ESB-by-ESB basis. 32 inputs from the adjacent local interconnect feed each ESB; therefore, the either MegaLAB or the adjacent LAB can drive the ESB. Also, nine ESB macrocells feed back into the ESB through the local interconnect for higher performance. Dedicated clock pins, global signals, and additional inputs from the local interconnect drive the ESB control signals.

In product-term mode, each ESB contains 16 macrocells. Each macrocell consists of two product terms and a programmable register. Figure 13 shows the ESB in product-term mode.

Figure 13. Product-Term Logic in ESB



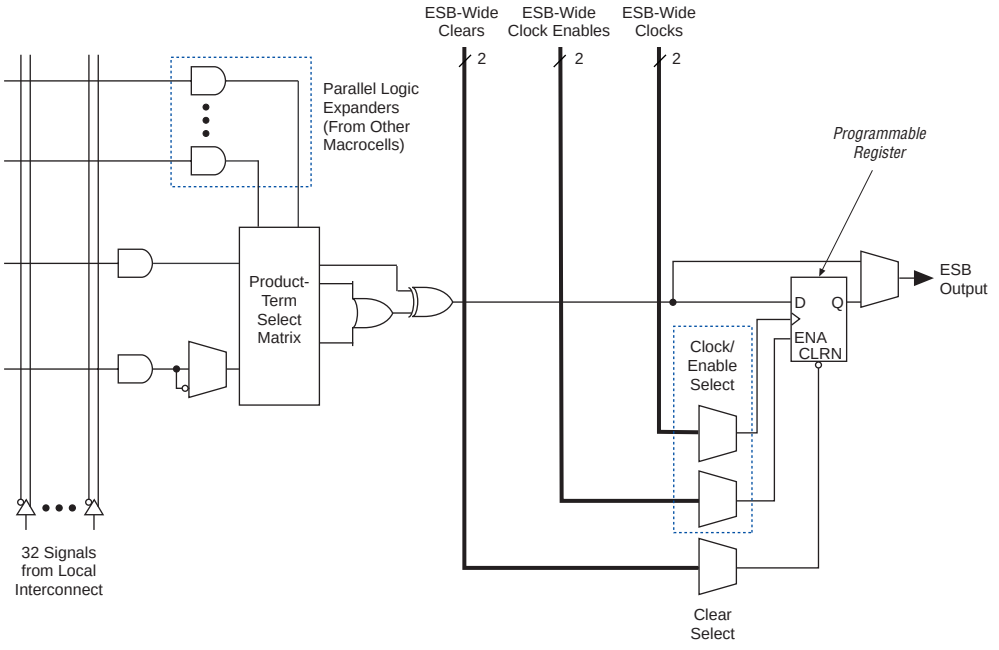
Note ot Figure 13:
(1) PLL outputs cannot drive data input ports.

Macrocells

APEX II macrocells can be configured individually for either sequential or combinatorial logic operation. The macrocell consists of three functional blocks: the logic array, the product-term select matrix, and the programmable register.

Combinatorial logic is implemented in the product terms. The product-term select matrix allocates these product terms for use as either primary logic inputs (to the OR and XOR gates) to implement combinatorial functions, or as parallel expanders to be used to increase the logic available to another macrocell. One product term can be inverted; the Quartus II software uses this feature to perform DeMorgan's inversion for more efficient implementation of wide OR functions. The Quartus II Compiler can use a NOT-gate push-back technique to emulate an asynchronous preset. Figure 14 shows the APEX II macrocell.

Figure 14. APEX II Macrocell

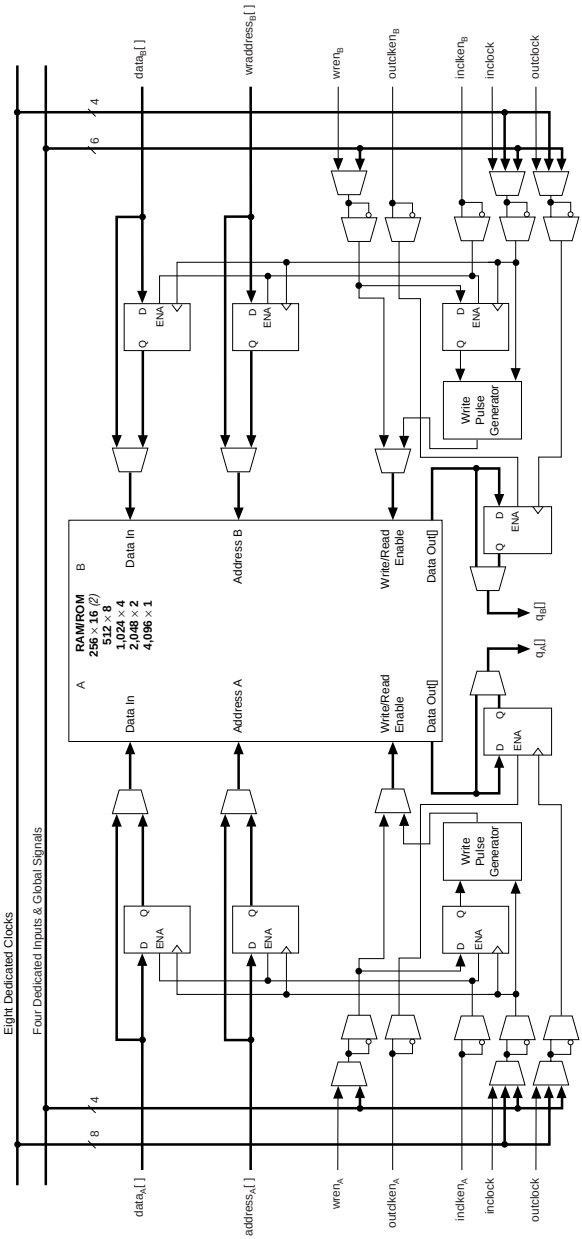


For registered functions, each macrocell register can be programmed individually to implement D, T, JK, or SR operation with programmable clock control. The register can be bypassed for combinatorial operation. During design entry, the designer specifies the desired register type; the Quartus II software then selects the most efficient register operation for each registered function to optimize resource utilization. The Quartus II software or other synthesis tools can also select the most efficient register operation automatically when synthesizing HDL designs.

Each programmable register can be clocked by one of two ESB-wide clocks. The ESB-wide clocks can be generated from device dedicated clock pins, global signals, or local interconnect. Each clock also has an associated clock enable, generated from the local interconnect. The clock and clock enable signals are related for a particular ESB; any macrocell using a clock also uses the associated clock enable.

If both the rising and falling edges of a clock are used in an ESB, both ESB-wide clock signals are used.

Figure 19. ESB in Input/Output Clock Mode *Note (1)*

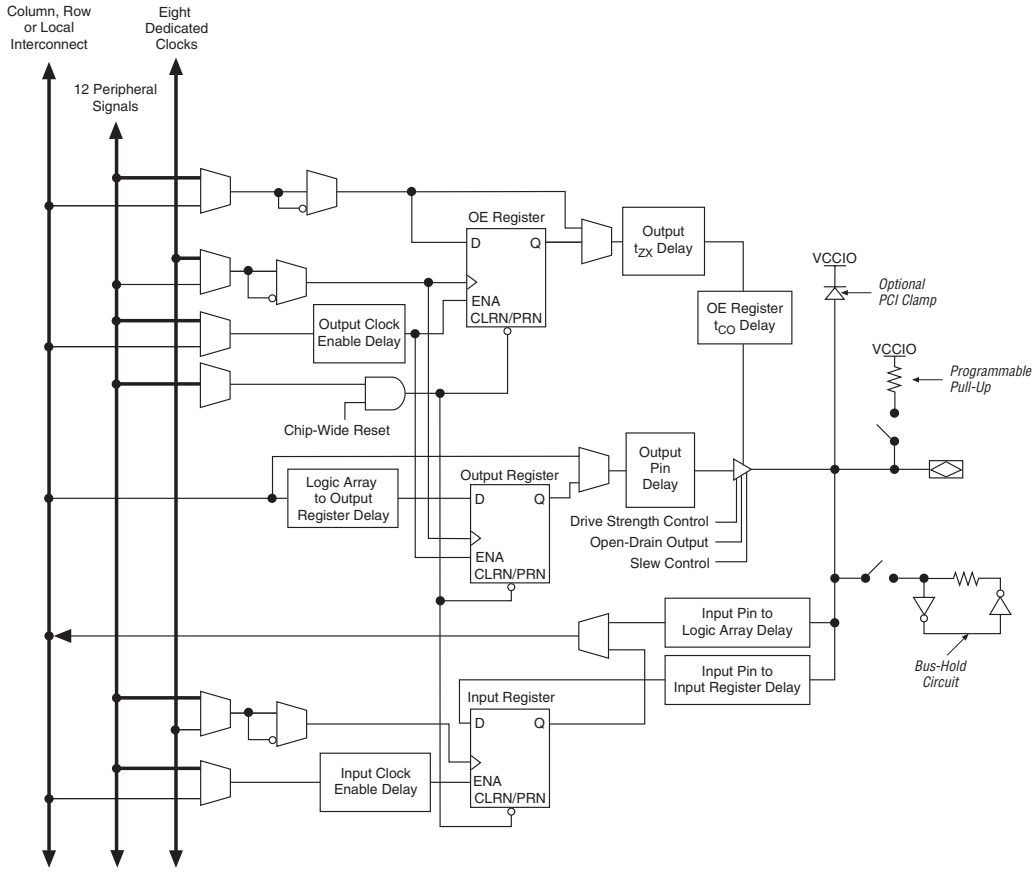


Notes to Figure 19:

(1) All registers can be cleared asynchronously by ESB local interconnect signals, global signals, or the chip-wide reset.

(2) This configuration is not supported for bidirectional dual-port configuration.

Figure 28. APEX II IOE in Bidirectional I/O Configuration



The APEX II IOE includes programmable delays that can be activated to ensure zero hold times, minimum clock-to-output times, input IOE register-to-logic array register transfers, or logic array-to-output IOE register transfers.

A path in which a pin directly drives a register may require the delay to ensure zero hold time, whereas a path in which a pin drives a register through combinatorial logic may not require the delay. Programmable delays exist for decreasing input pin to logic array and IOE input register delays. The Quartus II Compiler can program these delays automatically to minimize setup time while providing a zero hold time. Delays are also programmable for increasing the register to pin delays for output and / or output enable registers. A programmable delay exists for increasing the t_{ZX} delay to the output pin, which is required for ZBT interfaces. Table 8 shows the programmable delays for APEX II devices.

Table 8. APEX II Programmable Delay Chain

Programmable Delays	Quartus II Logic Option
Input pin to logic array delay *(1)*	Decrease input delay to internal cells
Input pin to input register delay	Decrease input delay to input register
Output propagation delay	Increase delay to output pin
Output enable register t_{CO} delay	Increase delay to output enable pin
Output t_{ZX} delay	Increase t_{ZX} delay to output pin
Output clock enable delay	Increase output clock enable delay
Input clock enable delay	Increase input clock enable delay
Logic array to output register delay	Decrease input delay to output register

Note to Table 8:
(1) This delay has four settings: off and three levels of delay.

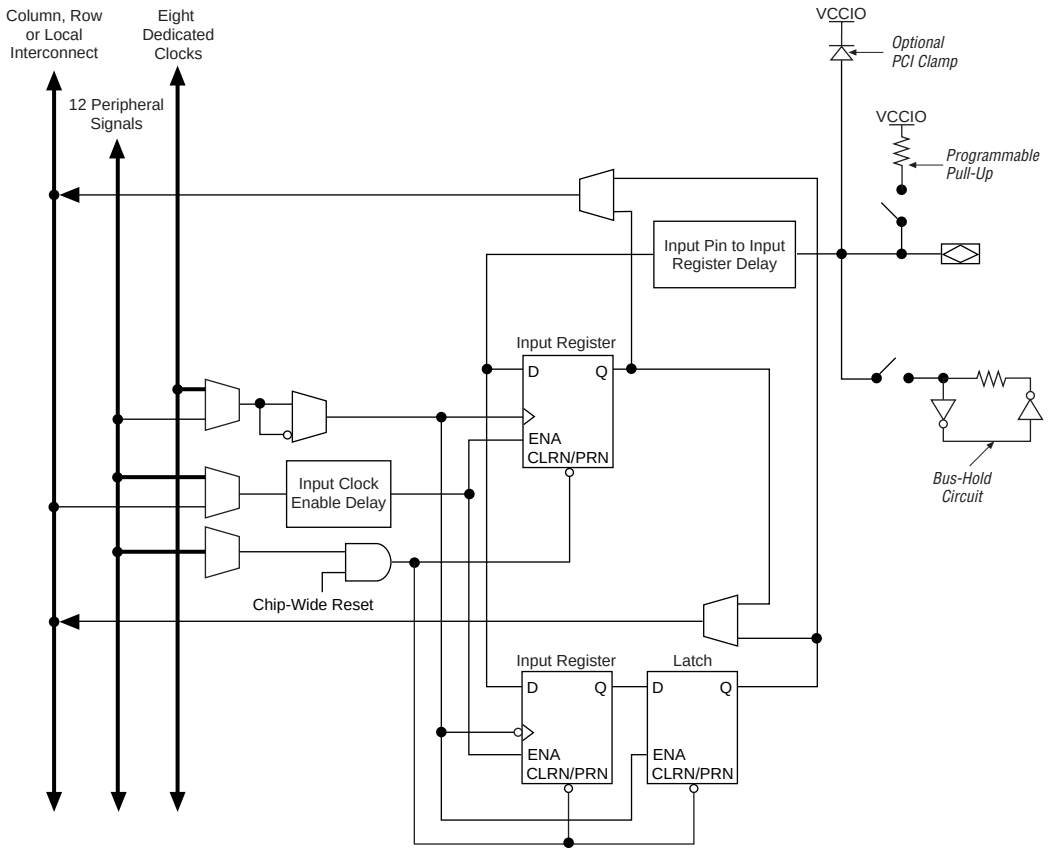
The IOE registers in APEX II devices share the same source for clear or preset. The designer can program preset and clear for each individual IOE. The registers can be programmed to power up high or low after configuration is complete. If programmed to power up low, an asynchronous clear can control the registers. If programmed to power up high, an asynchronous preset can control the registers. This feature prevents the inadvertent activation of another device's active-low input upon power-up. If one register in an IOE uses a preset or clear signal then all registers in the IOE must use that preset or clear signal.

Double Data Rate I/O

APEX II devices have six-register IOEs which support DDR interfacing by clocking data on both positive and negative clock edges. The IOEs in APEX II devices support DDR inputs, DDR outputs, and bidirectional DDR modes.

When using the IOE for DDR inputs, the two input registers are used to clock double rate input data on alternating edges. An input latch is also used within the IOE for DDR input acquisition. The latch holds the data that is present during the clock high times. This allows both bits of data to be synchronous to the same clock edge (either rising or falling). Figure 29 shows an IOE configured for DDR input.

Figure 29. APEX II IOE in DDR Input I/O Configuration



When using the IOE for DDR outputs, the two output registers are configured to clock two data paths from LEs on rising clock edges. These register outputs are multiplexed by the clock to drive the output pin at a ×2 rate. One output register clocks the first bit out on the clock high time, while the other output register clocks the second bit out on the clock low time. Figure 30 shows the IOE configured for DDR output.

Bus Hold

Each APEX II device I/O pin provides an optional bus-hold feature. When this feature is enabled for an I/O pin, the bus-hold circuitry weakly holds the signal at its last driven state. By holding the last driven state of the pin until the next input signal is present, the bus-hold feature eliminates the need to add external pull-up or pull-down resistors to hold a signal level when the bus is tri-stated. The bus-hold circuitry also pulls undriven pins away from the input threshold voltage where noise can cause unintended high-frequency switching. This feature can be selected individually for each I/O pin. The bus-hold output will drive no higher than V_{CCIO} to prevent overdriving signals. If the bus-hold feature is enabled, the programmable pull-up option cannot be used. The bus-hold feature should also be disabled if open-drain outputs are used with the GTL+ I/O standard.

The bus-hold circuitry weakly pulls the signal level to the last driven state through a resistor with a nominal resistance (R_{BH}) of approximately 7 kΩ. Table 41 on page 74 gives specific sustaining current that will be driven through this resistor and overdrive current that will identify the next driven input level. This information is provided for each V_{CCIO} voltage level.

The bus-hold circuitry is active only after configuration. When going into user mode, the bus-hold circuit captures the value on the pin present at the end of configuration.

Programmable Pull-Up Resistor

Each APEX II device I/O pin provides an optional programmable pull-up resistor during user mode. When this feature is enabled for an I/O pin, the pull-up resistor (typically 25 kΩ) weakly holds the output to the V_{CCIO} level of the bank that the output pin resides in.

Dedicated Fast I/O Pins

APEX II devices incorporate dedicated bidirectional pins for signals with high internal fanout, such as PCI control signals. These pins are called dedicated fast I/O pins (FAST1, FAST2, FAST3, and FAST4) and can drive the four global fast lines throughout the device, ideal for fast clock, clock enable, preset, clear, or high fanout logic signal distribution. The dedicated fast I/O pins have one output register and one OE register, but they do not have input registers. The dedicated fast lines can also be driven by a LE local interconnect to generate internal global signals.

Each bank can support multiple standards with the same V_{CCIO} for input and output pins. Each bank can support one voltage-referenced I/O standard, but it can support multiple I/O standards with the same V_{CCIO} voltage level. For example, when V_{CCIO} is 3.3 V, a bank can support LVTTL, LVCMOS, 3.3-V PCI, and SSTL-3 for inputs and outputs. When the True-LVDS banks are not used for LVDS I/O pins, they support all of the other I/O standards except HSTL Class II output.

True-LVDS Interface

APEX II devices contain dedicated circuitry for supporting differential standards at speeds up to 1.0 Gbps. APEX II devices have dedicated differential buffers and circuitry to support LVDS, LVPECL, HyperTransport, and PCML I/O standards. Four dedicated high-speed PLLs (separate from the general-purpose PLLs) multiply reference clocks and drive high-speed differential serializer/deserializer channels. In addition, CDS circuitry at each receiver channel corrects any fixed clock-to-data skew. All APEX II devices support 36 input channels, 36 output channels, two dedicated receiver PLLs, and two dedicated transmitter PLLs.

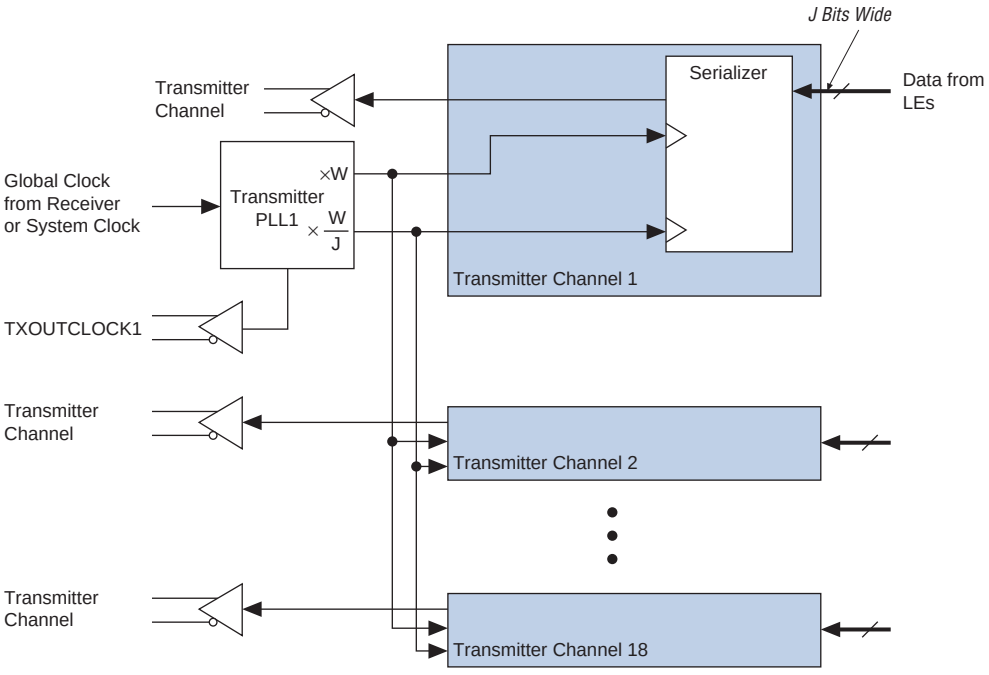
The True-LVDS circuitry supports the following standards and applications:

- RapidIO
- POS-PHY Level 4
- Utopia IV
- HyperTransport

APEX II devices support source-synchronous interfacing with LVDS, LVPECL,PCML, or HyperTransport signaling at up to 1 Gbps. Serial channels are transmitted and received along with a low-speed clock. The receiving device then multiplies the clock by a factor of 1, 2, or 4 to 10. The serialization/deserialization rate can be any number from 1, 2, or 4 to 10 and does not have to equal the clock multiplication value.

For example, an 840-Mbps LVDS channel can be received along with an 84-MHz clock. The 84-MHz clock is multiplied by 10 to drive the serial shift register, but the register can be clocked out in parallel at 8- or 10-bits wide at 84 or 105 MHz. See Figures 32 and 33.

Figure 33. True-LVDS Transmitter Diagram *Notes (1), (2)*



Notes to Figure 33:

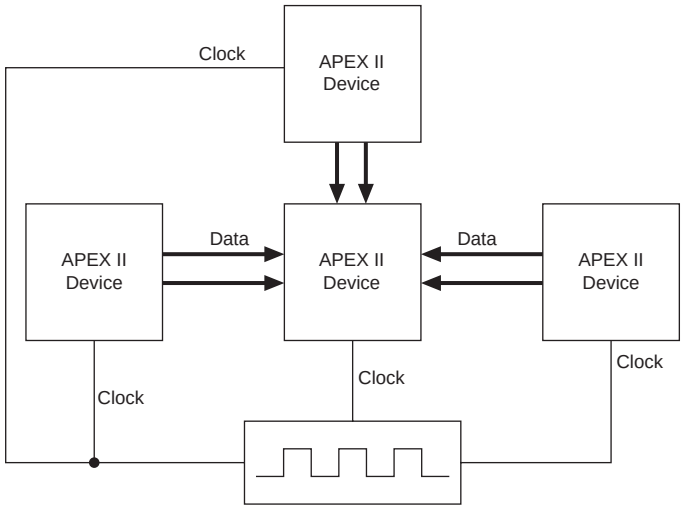
(1) Two sets of 18 transmitter channels are located in each APEX II device. Each set of 18 channels has one transmitter PLL.

(2) *W* = 1, 2, 4 to 10
J = 1, 2, 4 to 10
W does not have to equal *J*. When *J* = 1 or 2, the deserializer is bypassed. When *J* = 2, DDR I/O registers are used.

Clock-Data Synchronization

In addition to dedicated serial-to-parallel converters, APEX II True-LVDS circuitry contains CDS circuitry in every receiver channel. The CDS feature can be turned on or off independently for each receiver channel. There are two modes for the CDS circuitry: single-bit mode, which corrects a fixed clock-to-data skew of up to ±50% of the data bit period, and multi-bit mode, which corrects any fixed clock-to-data skew.

Figure 34. Multi-Bit CDS Supports N:1 Topology



When using multi-bit CDS, the *J* and *W* factors do not need to be the same value. The byte boundary cannot be distinguished with multi-bit CDS patterns (see Table 12). Therefore, the byte must be aligned using internal logic. Table 12 shows the possible training patterns for multi-bit CDS. Either pattern can be used.

Table 12. Multi-Bit CDS Patterns

W Factor	J Factor	Multi-Bit CDS Pattern
1, 2, 4 to 10	4 to 10	3 × J-bits of `010101` pattern
1, 2, 4 to 10	4 to 10	3 × J-bits of `101010` pattern

Pre-Programmed CDS

When the fixed clock-to-data skew is known, CDS can be pre-programmed into the device during configuration. If CDS is pre-programmed into the device, the training patterns do not need to be transmitted to the receiver channels. The resolution of each pre-programmed setting is 25% of the data period, to compensate for skew up to ±50% of the data period.

For more information, see the following documents:

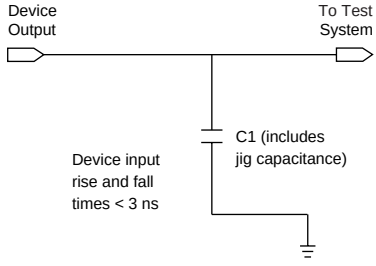
- *Application Note 39 (IEEE Std. 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices)*
- *Jam Programming & Test Language Specification*

Generic Testing

Each APEX II device is functionally tested. Complete testing of each configurable static random access memory (SRAM) bit and all logic functionality ensures 100% yield. AC test measurements for APEX II devices are made under conditions equivalent to those shown in Figure 37. Multiple test patterns can be used to configure devices during all stages of the production flow. AC test criteria include:

- Power supply transients can affect AC measurements.
- Simultaneous transitions of multiple outputs should be avoided for accurate measurement.
- Threshold tests must not be performed under AC conditions.
- Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result.

Figure 37. APEX II AC Test Conditions



Operating Conditions

APEX II devices are offered in both commercial and industrial grades. However, industrial-grade devices may have limited speed-grade availability.

Table 22. APEX II Device DC Operating Conditions *Note (7)*

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
I_I	Input pin leakage current	$V_I = V_{CCIO}$ to 0 V *(8)*	–10		10	µA
I_{OZ}	Tri-stated I/O pin leakage current	$V_O = V_{CCIO}$ to 0 V *(8)*	–10		10	µA
I_{CC0}	V_{CC} supply current (standby) (All ESBs in power-down mode)	V_I = ground, no load, no toggling inputs, -7 speed grade		10		mA
		V_I = ground, no load, no toggling inputs, -8, -9 speed grades		5		mA
R_{CONF}	Value of I/O pin pull-up resistor before and during configuration	V_{CCIO} = 3.0 V *(9)*	20		50	kΩ
		V_{CCIO} = 2.375 V *(9)*	30		80	kΩ
		V_{CCIO} = 1.71 V *(9)*	60		150	kΩ

Table 23. LVTTL Specifications

Symbol	Parameter	Conditions	Minimum	Maximum	Units
V_{CCIO}	Output supply voltage		3.0	3.6	V
V_{IH}	High-level input voltage		1.7	4.1	V
V_{IL}	Low-level input voltage		–0.5	0.8	V
I_I	Input pin leakage current	V_{IN} = 0 V or V_{CCIO}	–5	5	µA
V_{OH}	High-level output voltage	I_{OH} = –4 to –24 mA *(10)*	2.4		V
V_{OL}	Low-level output voltage	I_{OL} = 4 to 24 mA *(10)*		0.45	V

Table 24. LVCMOS Specifications

Symbol	Parameter	Conditions	Minimum	Maximum	Units
V_{CCIO}	Output supply voltage		3.0	3.6	V
V_{IH}	High-level input voltage		1.7	4.1	V
V_{IL}	Low-level input voltage		–0.5	0.7	V
I_I	Input pin leakage current	V_{IN} = 0 V or V_{CCIO}	–10	10	µA
V_{OH}	High-level output voltage	V_{CCIO} = 3.0, I_{OH} = –0.1 mA	V_{CCIO} – 0.2		V
V_{OL}	Low-level output voltage	V_{CCIO} = 3.0, I_{OL} = 0.1 mA		0.2	V

Table 28. 3.3-V PCI Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
V_{CCIO}	Output supply voltage		3.0	3.3	3.6	V
V_{IH}	High-level input voltage		0.5 × V_{CCIO}		V_{CCIO} + 0.5	V
V_{IL}	Low-level input voltage		–0.5		0.3 × V_{CCIO}	V
I_I	Input pin leakage current	$0 < V_{IN} < V_{CCIO}$	–10		10	µA
V_{OH}	High-level output voltage	I_{OUT} = –500 µA	0.9 × V_{CCIO}			V
V_{OL}	Low-level output voltage	I_{OUT} = 1,500 µA			0.1 × V_{CCIO}	V

Table 29. PCI-X Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
V_{CCIO}	Output supply voltage		3.0		3.6	V
V_{IH}	High-level input voltage		0.5 × V_{CCIO}		V_{CCIO} + 0.5	V
V_{IL}	Low-level input voltage		–0.5		0.35 × V_{CCIO}	V
V_{IPU}	Input pull-up voltage		0.7 × V_{CCIO}			V
I_{IL}	Input leakage current	$0 < V_{IN} < V_{CCIO}$	–10		10	µA
V_{OH}	High-level output voltage	I_{OUT} = –500 µA	0.9 × V_{CCIO}			V
V_{OL}	Low-level output voltage	I_{OUT} = 1,500 µA			0.1 × V_{CCIO}	V
L_{PIN}	Pin inductance				15	nH

Table 30. GTL+ I/O Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
V_{TT}	Termination voltage		1.35	1.5	1.65	V
V_{REF}	Reference voltage		0.88	1.0	1.12	V
V_{IH}	High-level input voltage		V_{REF} + 0.1			V
V_{IL}	Low-level input voltage				V_{REF} – 0.1	V
V_{OL}	Low-level output voltage	I_{OL} = 36 mA *(10)*			0.65	V

Figure 41. f_{MAX} Timing Model

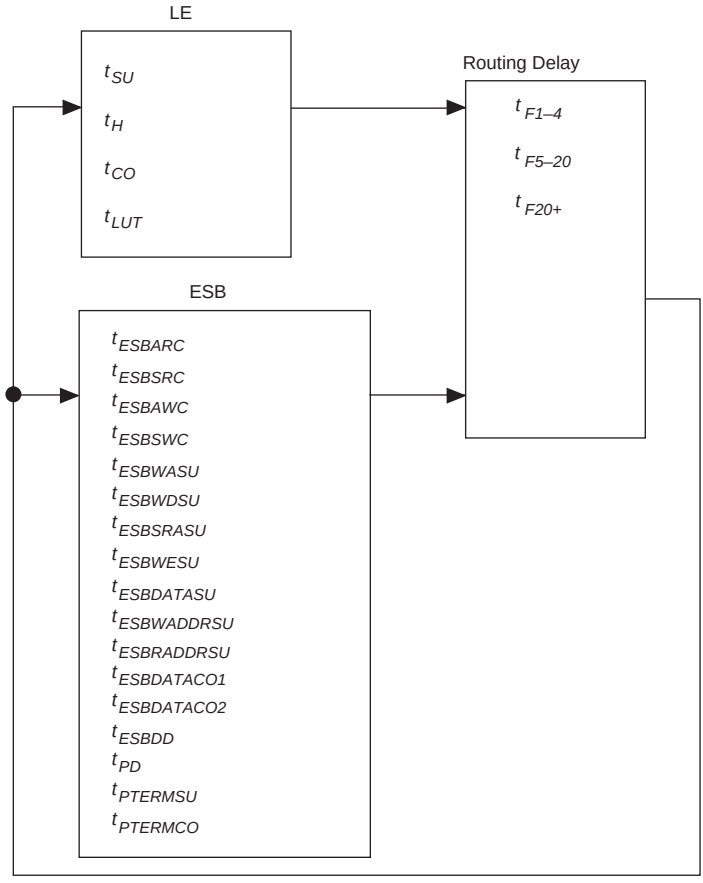


Table 63. EP2A40 Minimum Pulse Width Timing Parameters

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{CH}	0.89		1.33		1.88		ns
t_{CL}	0.89		1.33		1.88		ns
t_{CLRP}	0.12		0.14		0.16		ns
t_{PREP}	0.12		0.14		0.16		ns
t_{ESBCH}	0.89		1.33		1.88		ns
t_{ESBCL}	0.89		1.33		1.88		ns
t_{ESBWP}	1.05		1.20		1.38		ns
t_{ESBRP}	0.78		0.90		1.03		ns

Table 64. EP2A70 f_{MAX} LE Timing Parameters

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{SU}	0.30		0.34		0.39		ns
t_{H}	0.30		0.34		0.39		ns
t_{CO}		0.22		0.25		0.29	ns
t_{LUT}		0.66		0.76		0.87	ns

Table 77. APEX II Selectable I/O Standards Output Adder Delays

Symbol	-7 Speed Grade		-8 Speed Grade		-9 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
LVCMOS		0.00		0.00		0.00	ns
LVTTL		0.00		0.00		0.00	ns
1.5 V		3.32		3.82		4.20	ns
1.8 V		2.65		3.05		3.36	ns
2.5 V		1.20		1.38		1.52	ns
3.3-V PCI		– 0.68		– 0.78		– 0.85	ns
3.3-V PCI-X		– 0.68		– 0.78		– 0.85	ns
GTL+		– 0.45		– 0.52		– 0.57	ns
SSTL-3 Class I		– 0.52		– 0.60		– 0.66	ns
SSTL-3 Class II		– 0.52		– 0.60		– 0.66	ns
SSTL-2 Class I		– 0.68		– 0.78		– 0.86	ns
SSTL-2 Class II		– 0.81		– 0.93		– 1.02	ns
HSTL Class I		– 0.08		– 0.09		– 0.10	ns
HSTL Class II		– 0.23		– 0.27		– 0.30	ns
LVDS		– 1.41		– 1.62		– 1.79	ns
LVPECL		– 1.38		– 1.58		– 1.74	ns
PCML		– 1.30		– 1.50		– 1.65	ns
CTT		0.00		0.00		0.00	ns
3.3-V AGP 1×		0.00		0.00		0.00	ns
3.3-V AGP 2×		0.00		0.00		0.00	ns
HyperTransport		– 1.22		– 1.41		– 1.55	ns
Differential HSTL		– 1.41		– 1.62		– 1.79	ns

Power Consumption

Detailed power consumption information for APEX II devices will be released via a future interactive power estimator on the Altera web site.

Device Pin-Outs

See the Altera web site (**http://www.altera.com**) or the ***Altera Digital Library*** for pin-out information.