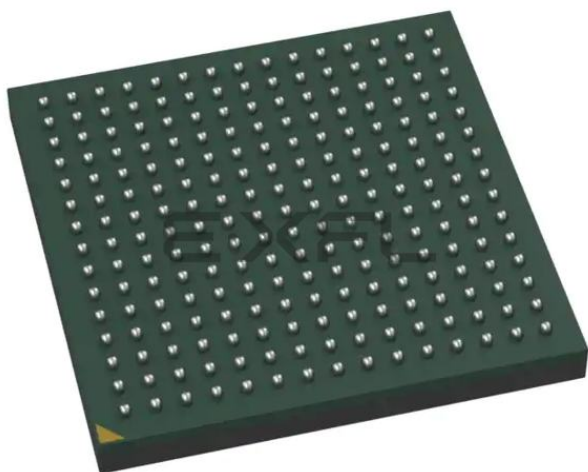




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Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details

Product Status	Obsolete
Core Processor	ARM920T
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	200MHz
Co-Processors/DSP	-
RAM Controllers	SDRAM
Graphics Acceleration	No
Display & Interface Controllers	LCD
Ethernet	-
SATA	-
USB	USB 1.x (1)
Voltage - I/O	1.8V, 3.0V
Operating Temperature	-30°C ~ 70°C (TA)
Security Features	-
Package / Case	225-LFBGA
Supplier Device Package	225-MAPBGA (13x13)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc9328mxldvp20r2

Table 2. i.MXL Signal Descriptions (Continued)

Signal Name	Function/Notes
Bootstrap	
BOOT [3:0]	System Boot Mode Select—The operational system boot mode of the i.MXL processor upon system reset is determined by the settings of these pins.
SDRAM Controller	
SDBA [4:0]	SDRAM non-interleave mode bank address multiplexed with address signals A [15:11]. These signals are logically equivalent to core address p_addr [25:21] in SDRAM cycles.
SDIBA [3:0]	SDRAM interleave addressing mode bank address multiplexed with address signals A [19:16]. These signals are logically equivalent to core address p_addr [12:9] in SDRAM cycles.
MA [11:10]	SDRAM address signals
MA [9:0]	SDRAM address signals which are multiplexed with address signals A [10:1]. MA [9:0] are selected on SDRAM cycles.
DQM [3:0]	SDRAM data enable
$\overline{\text{CSD0}}$	SDRAM Chip-select signal which is multiplexed with the $\overline{\text{CS2}}$ signal. These two signals are selectable by programming the system control register.
$\overline{\text{CSD1}}$	SDRAM Chip-select signal which is multiplexed with $\overline{\text{CS3}}$ signal. These two signals are selectable by programming the system control register. By default, $\overline{\text{CSD1}}$ is selected, so it can be used as boot chip-select by properly configuring BOOT [3:0] input pins.
$\overline{\text{RAS}}$	SDRAM Row Address Select signal
$\overline{\text{CAS}}$	SDRAM Column Address Select signal
$\overline{\text{SDWE}}$	SDRAM Write Enable signal
SDCKE0	SDRAM Clock Enable 0
SDCKE1	SDRAM Clock Enable 1
SDCLK	SDRAM Clock
$\overline{\text{RESET_SF}}$	Not Used
Clocks and Resets	
EXTAL16M	Crystal input (4 MHz to 16 MHz), or a 16 MHz oscillator input when the internal oscillator circuit is shut down.
XTAL16M	Crystal output
EXTAL32K	32 kHz crystal input
XTAL32K	32 kHz crystal output
CLKO	Clock Out signal selected from internal clock signals.
$\overline{\text{RESET_IN}}$	Master Reset—External active low Schmitt trigger input signal. When this signal goes active, all modules (except the reset module and the clock control module) are reset.
$\overline{\text{RESET_OUT}}$	Reset Out—Internal active low output signal from the Watchdog Timer module and is asserted from the following sources: Power-on reset, External reset ($\overline{\text{RESET_IN}}$), and Watchdog time-out.
POR	Power On Reset—Internal active high Schmitt trigger input signal. The POR signal is normally generated by an external RC circuit designed to detect a power-up event.

Table 3. MC9328MXLMC9328MXS Signal Multiplexing Scheme (Continued)

I/O Supply Voltage	225 BGA Ball	256 BGA Ball	Primary			Alternate		GPIO		AIN	BIN	AOUT	Default
			Signal	Dir	Pull-Up	Signal	Dir	Mux	Pull-Up				
NVDD1	N4	P3	$\overline{\text{EB0}}$	O									
NVDD1	M4	N3	D10	I/O	69K								
NVDD1	P4	P1	A3	O									
NVDD1	R3	N2	$\overline{\text{EB1}}$	O									
NVDD1	N5	P2	D9	I/O	69K								
NVDD1	R4	R1	$\overline{\text{EB2}}$	O									
NVDD1	P5	T2	A2	O									
NVDD1	M5	R2	$\overline{\text{EB3}}$	O									
NVDD1	N6	R5	D8	I/O	69K								
NVDD1	R5	T3	$\overline{\text{OE}}$	O									
NVDD1	P6	R3	A1	O									
NVDD1	L7	T4	$\overline{\text{CS5}}$	O				PA23	69K				PA23
NVDD1	R6	N4	D7	I/O	69K								
NVDD1	M7	R4	$\overline{\text{CS4}}$	O				PA22	69K				PA22
NVDD1	R7	N5	A0	O				PA21	69K				A0
NVDD1	N7	P4	$\overline{\text{CS3}}$	O		$\overline{\text{CSD1}}$							$\overline{\text{CSD1}}$
NVDD1	P7	P5	D6	I/O	69K								
NVDD1	K3	T5	$\overline{\text{CS2}}$	O		$\overline{\text{CSD0}}$							$\overline{\text{CSD0}}$
NVDD1	R8	M5	SDCLK	O									
NVDD1	M8	T6	$\overline{\text{CS1}}$	O									
NVDD1	N8	T7	$\overline{\text{CS0}}$	O									
NVDD1	P8	R6	D5	I/O	69K								
NVDD1	L9	P6	$\overline{\text{ECB}}$	I		ETMTRAC EPKT7		PA20	69K				$\overline{\text{ECB}}$
NVDD1	R9	N6	D4	I/O	69K								
NVDD1	R10	R7	$\overline{\text{LBA}}$	O		ETMTRAC EPKT6		PA19	69K				$\overline{\text{LBA}}$
NVDD1	R11	P8	D3	I/O	69K								
NVDD1	M9	R8	BCLK			ETMTRAC EPKT5		PA18	69K				BCLK
NVDD1	L8	P7	D2	I/O	69K								
NVDD1	N9	N7	PA17			ETMTRAC EPKT4		PA17	69K	SPI2_SS		$\overline{\text{DTACK}}$	PA17
NVDD1	K10	N8	D1	I/O	69K								
NVDD1	M10	M7	$\overline{\text{RW}}$										
NVDD1	P10	T8	MA11	O									
NVDD1	P9	M8	MA10	O									
NVDD1	N10	R9	D0	I/O	69K								
NVDD1	R12	P9	DQM3	O									

Table 3. MC9328MXLMC9328MXS Signal Multiplexing Scheme (Continued)

I/O Supply Voltage	225 BGA Ball	256 BGA Ball	Primary			Alternate		GPIO		AIN	BIN	AOUT	Default
			Signal	Dir	Pull-Up	Signal	Dir	Mux	Pull-Up				
NVDD1	N11	T9	DQM2	O									
NVDD1	P11	N9	DQM1	O									
NVDD1	N12	R10	DQM0	O									
NVDD1	P12	M9	$\overline{\text{RAS}}$	O									
NVDD1	R13	L8	$\overline{\text{CAS}}$	O									
NVDD1	R14	T10	$\overline{\text{SDWE}}$	O									
NVDD1	N13	R11	SDCKE0	O									
NVDD1	P13	P10	SDCKE1	O									
NVDD1	P15	N10	RESET_SF	O									
NVDD1	P14	T11	CLKO	O									
AVDD1	R15	T12	AVDD1	Static									
QVDD2	M13	R15	QVDD2	Static									
AVDD1	N15	P13	$\overline{\text{TRST}}$	I	69K								
AVDD1	N14	T13	TRISTATE ₁	I									
AVDD1	M15	T14	EXTAL16M	I									
AVDD1	L14	T15	XTAL16M	O									
AVDD1	L15	R16	EXTAL32K	I									
AVDD1	K15	P16	XTAL32K	O									
AVDD1	M14	M10	RESET_I _{N²}	I	69K								
AVDD1	K14	N11	RESET_O _{UT}	O									
AVDD1	L12	R12	POR ²	I									
AVDD1	K13	M11	BIG_ENDI _{AN³}	I									
AVDD1	M12	P11	BOOT3 ³	I									
AVDD1	K11	N12	BOOT2 ³	I									
AVDD1	J14	R13	BOOT1 ³	I									
AVDD1	J15	P12	BOOT0 ³	I									
NVDD2	J13	R14	$\overline{\text{TDO}}$ ⁴	O									
NVDD2	H15	N15	TMS	I	69K								
NVDD2	J12	L9	TCK	I	69K								
NVDD2	K12	N16	TDI	I	69K								
NVDD2	J11	P14	I2C_SCL	O				PA16	69K				PA16
NVDD2	H14	P15	I2C_SDA	I/O				PA15	69K				PA15

4.4.2.1 WAIT Read Cycle without DMA

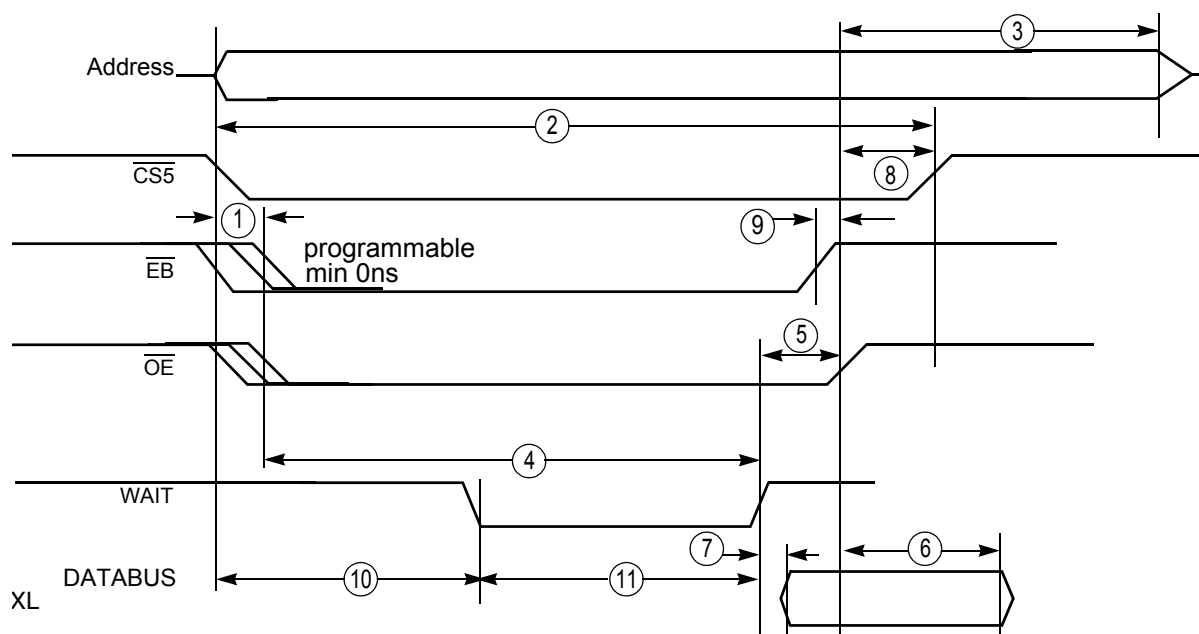


Figure 6. WAIT Read Cycle without DMA

Table 13. WAIT Read Cycle without DMA: WSC = 111111, DTACK_SEL=1, HCLK=96MHz

Number	Characteristic	3.0 ± 0.3 V		Unit
		Minimum	Maximum	
1	$\overline{OE}$ and $\overline{EB}$ assertion time	See note 2	–	ns
2	$\overline{CS5}$ pulse width	3T	–	ns
3	$\overline{OE}$ negated to address inactive	56.81	57.28	ns
4	Wait asserted after $\overline{OE}$ asserted	–	1020T	ns
5	Wait asserted to $\overline{OE}$ negated	2T+1.57	3T+7.33	ns
6	Data hold timing after $\overline{OE}$ negated	T-1.49	–	ns
7	Data ready after wait asserted	0	T	ns
8	OE negated to CS negated	1.5T-0.68	1.5T-0.06	ns
9	OE negated after EB negated	0.06	0.18	ns
10	Become low after CS5 asserted	0	1019T	ns
11	Wait pulse width	1T	1020T	ns

Note:

1. T is the system clock period. (For 96 MHz system clock, T=10.42 ns)
2. $\overline{OE}$ and $\overline{EB}$ assertion time is programmable by OEA bit in CS5L register. $\overline{EB}$ assertion in read cycle will occur only when EBC bit in CS5L register is clear.
3. Address becomes valid and $\overline{CS}$ asserts at the start of read access cycle.
4. The external wait input requirement is eliminated when CS5 is programmed to use internal wait state.

Table 16. WAIT Write Cycle DMA Enabled: WSC = 111111, DTACK_SEL=1, HCLK=96MHz

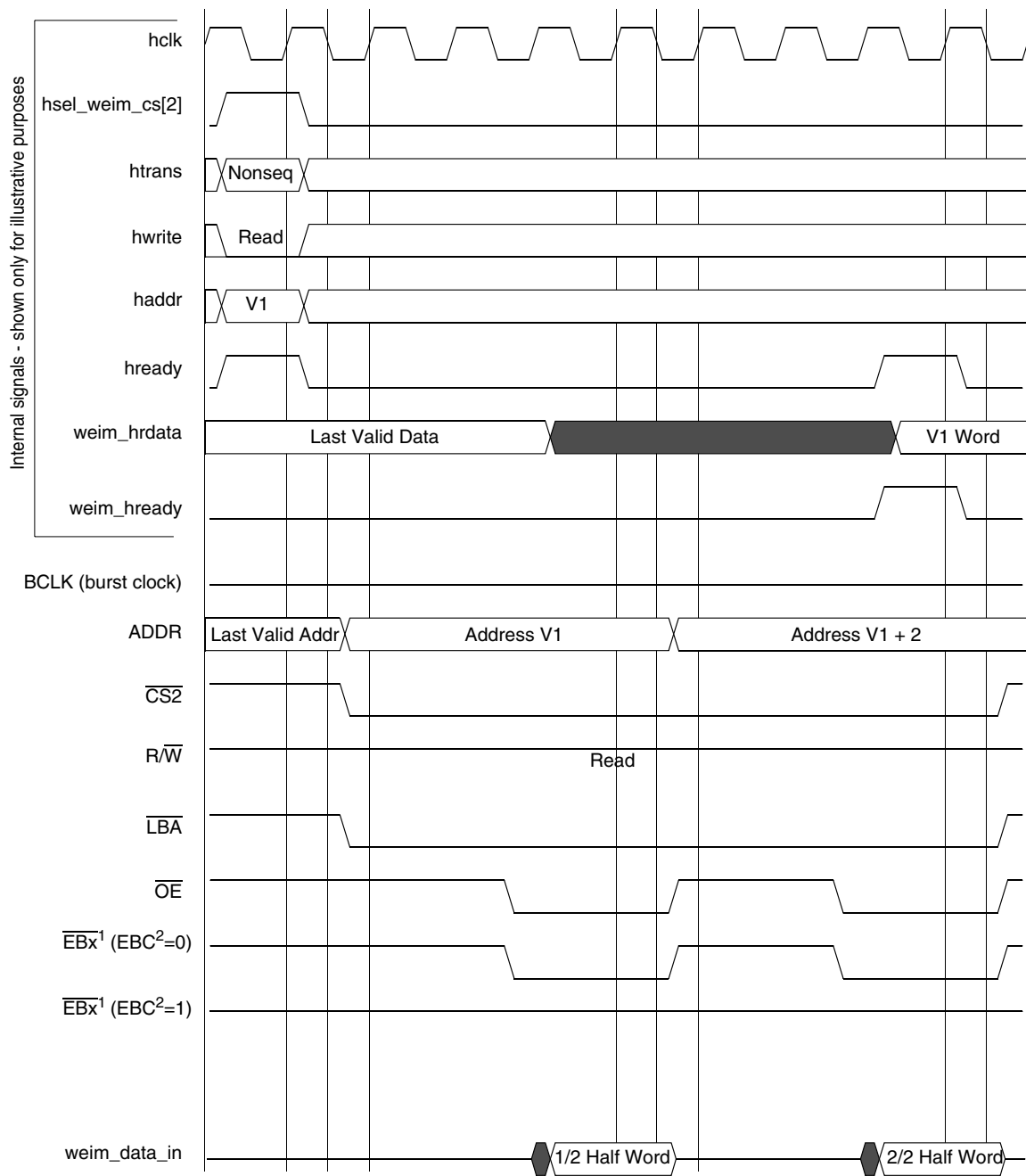
Number	Characteristic	3.0 ± 0.3 V		Unit
		Minimum	Maximum	
1	$\overline{\text{CS5}}$ assertion time	See note 2	–	ns
2	$\overline{\text{EB}}$ assertion time	See note 2	–	ns
3	$\overline{\text{CS5}}$ pulse width	3T	–	ns
4	$\overline{\text{RW}}$ negated before $\overline{\text{CS5}}$ is negated	2.5T-3.63	2.5T-1.16	ns
5	Address inactivated after $\overline{\text{CS}}$ negated	–	0.09	ns
6	Wait asserted after $\overline{\text{CS5}}$ asserted	–	1020T	ns
7	Wait asserted to $\overline{\text{RW}}$ negated	T+2.66	2T+7.96	ns
8	Data hold timing after $\overline{\text{RW}}$ negated	2T+0.03	–	ns
9	Data ready after $\overline{\text{CS5}}$ is asserted	–	T	ns
10	$\overline{\text{CS}}$ deactive to next $\overline{\text{CS}}$ active	T	–	ns
11	$\overline{\text{EB}}$ negate after $\overline{\text{CS}}$ negate	0.5T	0.5T+0.5	
12	Wait becomes low after $\overline{\text{CS5}}$ asserted	0	1019T	ns
13	Wait pulse width	1T	1020T	ns

Note:

1. T is the system clock period. (For 96 MHz system clock, T=10.42 ns)
2. $\overline{\text{CS5}}$ assertion can be controlled by CSA bits. $\overline{\text{EB}}$ assertion also can be programmable by WEA bits in CS5L register.
3. Address becomes valid and $\overline{\text{RW}}$ asserts at the start of write access cycle.
4. The external wait input requirement is eliminated when $\overline{\text{CS5}}$ is programmed to use internal wait state.

4.4.3 EIM External Bus Timing

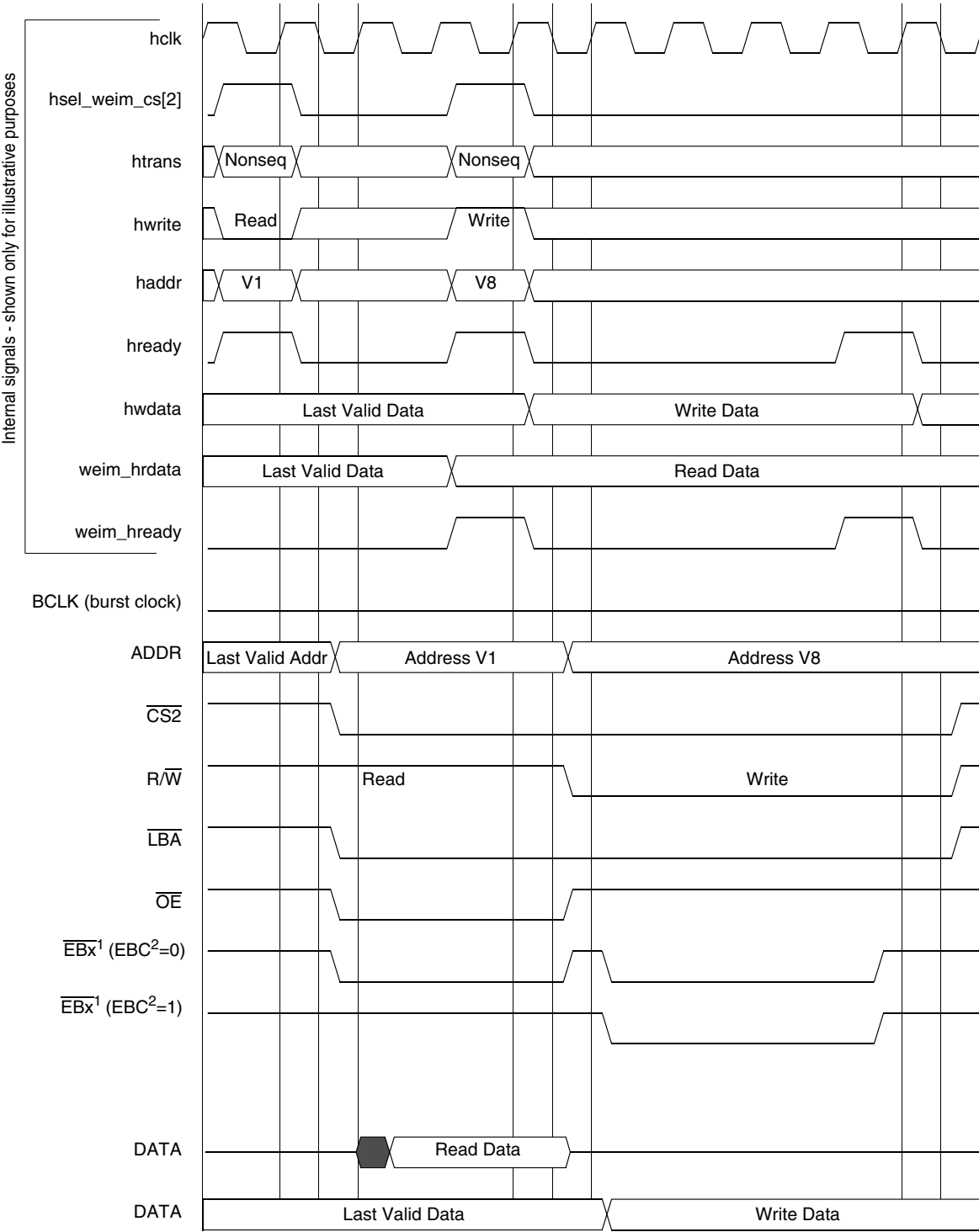
The External Interface Module (EIM) is the interface to devices external to the i.MXL, including generation of chip-selects for external peripherals and memory. The timing diagram for the EIM is shown in [Figure 5](#), and [Table 12](#) defines the parameters of signals.



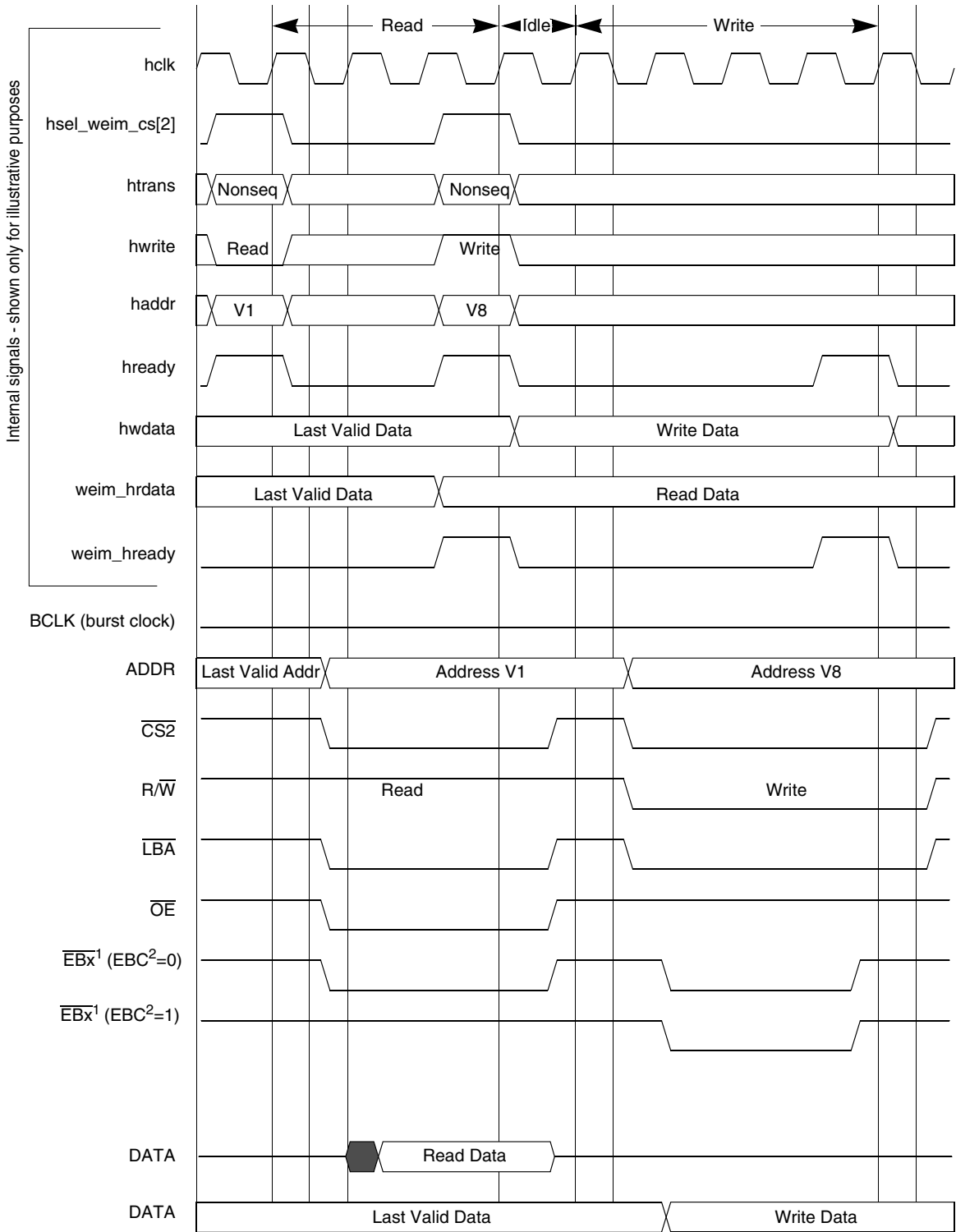
Note 1: x = 0, 1, 2 or 3

Note 2: EBC = Enable Byte Control bit (bit 11) on the Chip Select Control Register

Figure 16. WSC = 3, OEA = 4, A.WORD/E.HALF

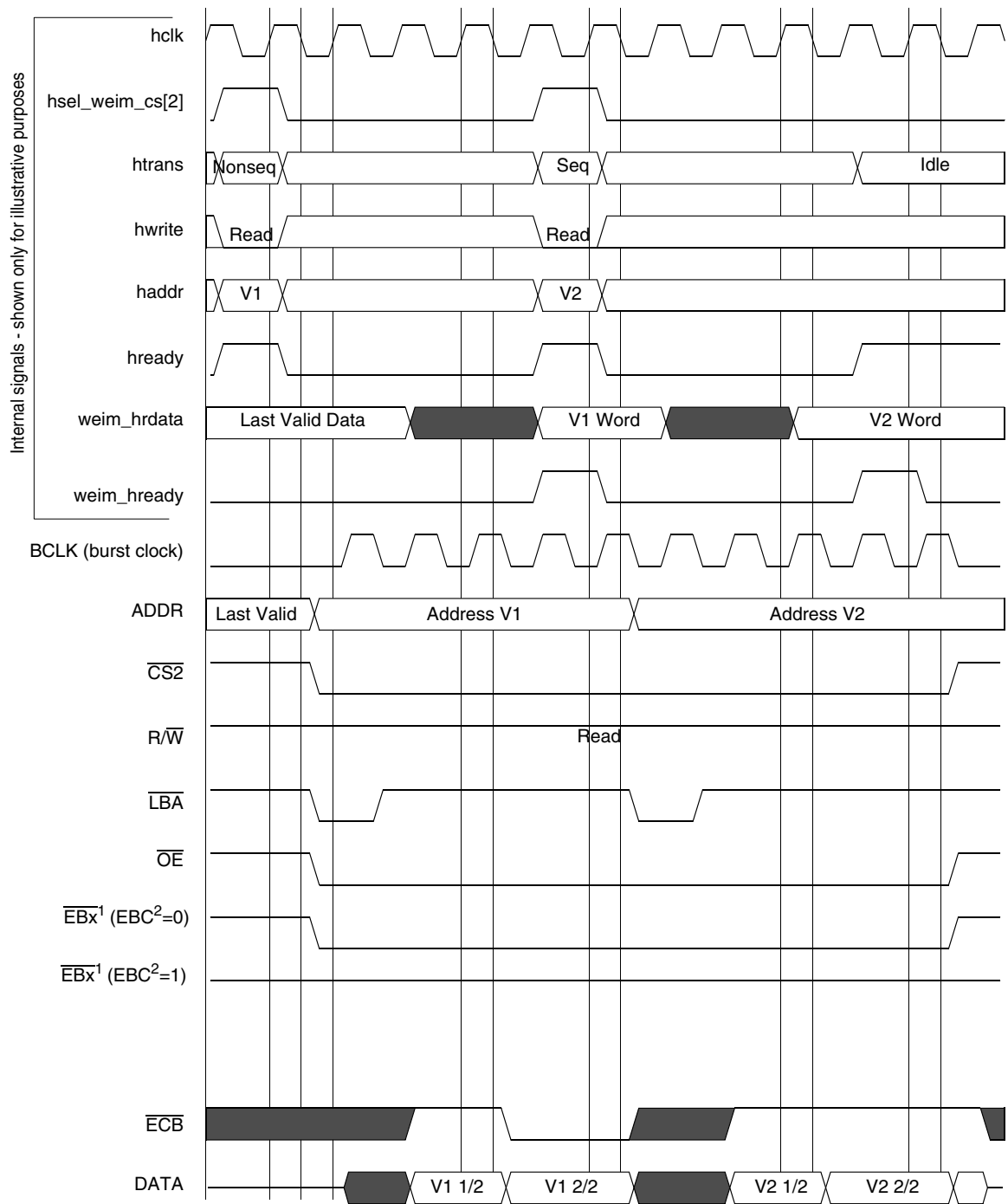


Note 1: x = 0, 1, 2 or 3
 Note 2: EBC = Enable Byte Control bit (bit 11) on the Chip Select Control Register
Figure 22. WSC = 2, WWS = 2, WEA = 1, WEN = 2, A.HALF/E.HALF



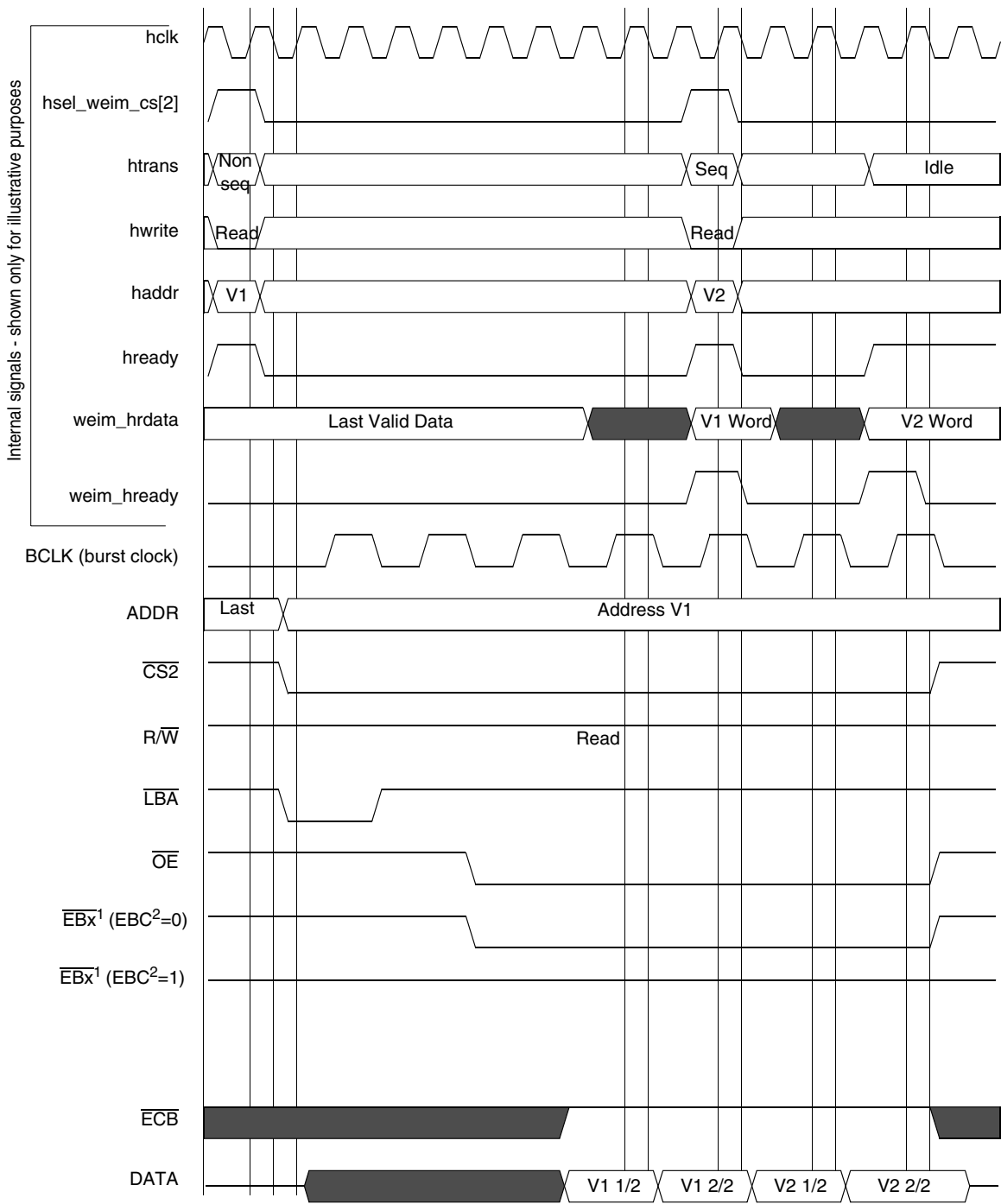
Note 1: x = 0, 1, 2 or 3
 Note 2: EBC = Enable Byte Control bit (bit 11) on the Chip Select Control Register

Figure 23. WSC = 2, WWS = 1, WEA = 1, WEN = 2, EDC = 1, A.HALF/E.HALF



Note 1: x = 0, 1, 2 or 3
 Note 2: EBC = Enable Byte Control bit (bit 11) on the Chip Select Control Register

Figure 30. WSC = 2, SYNC = 1, DOL = [1/0], A.WORD/E.HALF



Note 1: x = 0, 1, 2 or 3
 Note 2: EBC = Enable Byte Control bit (bit 11) on the Chip Select Control Register

Figure 32. WSC = 7, OEA = 8, SYNC = 1, DOL = 1, BCD = 1, BCS = 1, A.WORD/E.HALF

4.4.4 Non-TFT Panel Timing

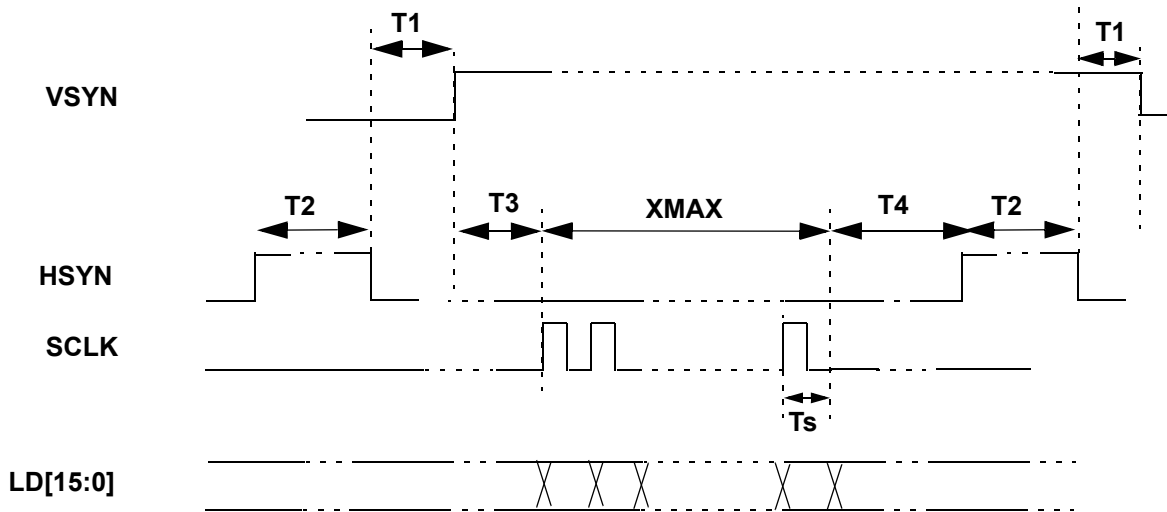


Figure 33. Non-TFT Panel Timing

Table 17. Non TFT Panel Timing Diagram

Symbol	Parameter	Allowed Register Minimum Value ^{1, 2}	Actual Value	Unit
T1	HSYN to VSYNC delay ³	0	HWAIT2+2	Tpix ⁴
T2	HSYN pulse width	0	HWIDTH+1	Tpix
T3	VSYNC to SCLK	–	$0 \leq T3 \leq Ts^5$	–
T4	SCLK to HSYN	0	HWAIT1+1	Tpix

¹ Maximum frequency of LCDC_CLK is 48 MHz, which is controlled by Peripheral Clock Divider Register.

² Maximum frequency of SCLK is HCLK / 5, otherwise LD output will be wrong.

³ VSYNC, HSYN and SCLK can be programmed as active high or active low. In the above timing diagram, all these 3 signals are active high.

⁴ Tpix is the pixel clock period which equals LCDC_CLK period * (PCD + 1).

⁵ Ts is the shift clock period. Ts = Tpix * (panel data bus width).

4.5 SPI Timing Diagrams

To use the internal transmit (TX) and receive (RX) data FIFOs when the SPI 1 module is configured as a master, two control signals are used for data transfer rate control: the $\overline{SS}$ signal (output) and the $\overline{SPI_RDY}$ signal (input). The SPI1 Sample Period Control Register (PERIODREG1) and the SPI2 Sample Period Control Register (PERIODREG2) can also be programmed to a fixed data transfer rate for either SPI 1 or SPI 2. When the SPI 1 module is configured as a slave, the user can configure the SPI1 Control Register (CONTROLREG1) to match the external SPI master's timing. In this configuration, $\overline{SS}$ becomes an input signal, and is used to latch data into or load data out to the internal data shift registers, as well as to increment the data FIFO. Figure 34 through Figure 38 show the timing relationship of the master SPI using different triggering mechanisms.

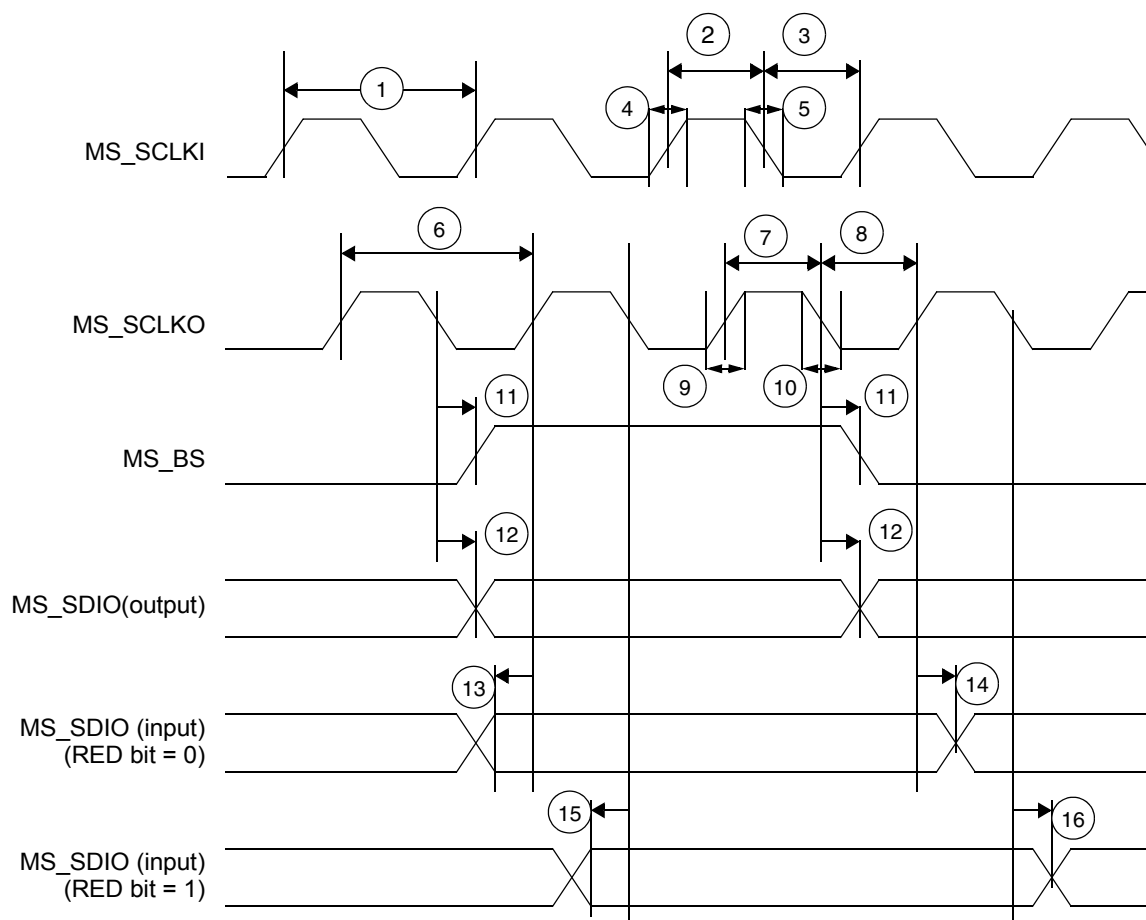


Figure 50. MSHC Signal Timing Diagram

Table 25. MSHC Signal Timing Parameter Table

Ref No.	Parameter	3.0 ± 0.3 V		Unit
		Minimum	Maximum	
1	MS_SCLKI frequency	–	25	MHz
2	MS_SCLKI high pulse width	20	–	ns
3	MS_SCLKI low pulse width	20	–	ns
4	MS_SCLKI rise time	–	3	ns
5	MS_SCLKI fall time	–	3	ns
6	MS_SCLKO frequency ¹	–	25	MHz
7	MS_SCLKO high pulse width ¹	20	–	ns
8	MS_SCLKO low pulse width ¹	15	–	ns
9	MS_SCLKO rise time ¹	–	5	ns
10	MS_SCLKO fall time ¹	–	5	ns
11	MS_BS delay time ¹	–	3	ns

Table 27. SDRAM Read Timing Parameter Table

Ref No.	Parameter	1.8 ± 0.1 V		3.0 ± 0.3 V		Unit
		Minimum	Maximum	Minimum	Maximum	
1	SDRAM clock high-level width	2.67	–	4	–	ns
2	SDRAM clock low-level width	6	–	4	–	ns
3	SDRAM clock cycle time	11.4	–	10	–	ns
3S	CS, RAS, CAS, WE, DQM setup time	3.42	–	3	–	ns
3H	CS, RAS, CAS, WE, DQM hold time	2.28	–	2	–	ns
4S	Address setup time	3.42	–	3	–	ns
4H	Address hold time	2.28	–	2	–	ns
5	SDRAM access time (CL = 3)	–	6.84	–	6	ns
5	SDRAM access time (CL = 2)	–	6.84	–	6	ns
5	SDRAM access time (CL = 1)	–	22	–	22	ns
6	Data out hold time	2.85	–	2.5	–	ns
7	Data out high-impedance time (CL = 3)	–	6.84	–	6	ns
7	Data out high-impedance time (CL = 2)	–	6.84	–	6	ns
7	Data out high-impedance time (CL = 1)	–	22	–	22	ns
8	Active to read/write command period (RC = 1)	t_{RCD}^1	–	t_{RCD1}	–	ns

¹ t_{RCD} = SDRAM clock cycle time. This settings can be found in the *MC9328MXL reference manual*.

Table 34. SSI (Port B Alternate Function) Timing Parameter Table

Ref No.	Parameter	1.8 ± 0.1 V		3.0 ± 0.3 V		Unit
		Minimum	Maximum	Minimum	Maximum	
Internal Clock Operation ¹ (Port B Alternate Function ²)						
1	STCK/SRCK clock period ¹	95	–	83.3	–	ns
2	STCK high to STFS (bl) high ³	1.7	4.8	1.5	4.2	ns
3	SRCK high to SRFS (bl) high ³	-0.1	1.0	-0.1	1.0	ns
4	STCK high to STFS (bl) low ³	3.08	5.24	2.7	4.6	ns
5	SRCK high to SRFS (bl) low ³	1.25	2.28	1.1	2.0	ns
6	STCK high to STFS (wl) high ³	1.71	4.79	1.5	4.2	ns
7	SRCK high to SRFS (wl) high ³	-0.1	1.0	-0.1	1.0	ns
8	STCK high to STFS (wl) low ³	3.08	5.24	2.7	4.6	ns
9	SRCK high to SRFS (wl) low ³	1.25	2.28	1.1	2.0	ns
10	STCK high to STXD valid from high impedance	14.93	16.19	13.1	14.2	ns
11a	STCK high to STXD high	1.25	3.42	1.1	3.0	ns
11b	STCK high to STXD low	2.51	3.99	2.2	3.5	ns
12	STCK high to STXD high impedance	12.43	14.59	10.9	12.8	ns
13	SRXD setup time before SRCK low	20	–	17.5	–	ns
14	SRXD hold time after SRCK low	0	–	0	–	ns
External Clock Operation (Port B Alternate Function ²)						
15	STCK/SRCK clock period ¹	92.8	–	81.4	–	ns
16	STCK/SRCK clock high period	27.1	–	40.7	–	ns
17	STCK/SRCK clock low period	61.1	–	40.7	–	ns
18	STCK high to STFS (bl) high ³	–	92.8	0	81.4	ns
19	SRCK high to SRFS (bl) high ³	–	92.8	0	81.4	ns
20	STCK high to STFS (bl) low ³	–	92.8	0	81.4	ns
21	SRCK high to SRFS (bl) low ³	–	92.8	0	81.4	ns
22	STCK high to STFS (wl) high ³	–	92.8	0	81.4	ns
23	SRCK high to SRFS (wl) high ³	–	92.8	0	81.4	ns
24	STCK high to STFS (wl) low ³	–	92.8	0	81.4	ns
25	SRCK high to SRFS (wl) low ³	–	92.8	0	81.4	ns
26	STCK high to STXD valid from high impedance	18.9	29.07	16.6	25.5	ns
27a	STCK high to STXD high	9.23	20.75	8.1	18.2	ns
27b	STCK high to STXD low	10.60	21.32	9.3	18.7	ns

The limitation on pixel clock rise time / fall time are not specified. It should be calculated from the hold time and setup time, according to:

Rising-edge latch data

$$\begin{aligned} \text{max rise time allowed} &= (\text{positive duty cycle} - \text{hold time}) \\ \text{max fall time allowed} &= (\text{negative duty cycle} - \text{setup time}) \end{aligned}$$

In most of case, duty cycle is 50 / 50, therefore

$$\begin{aligned} \text{max rise time} &= (\text{period} / 2 - \text{hold time}) \\ \text{max fall time} &= (\text{period} / 2 - \text{setup time}) \end{aligned}$$

For example: Given pixel clock period = 10ns, duty cycle = 50 / 50, hold time = 1ns, setup time = 1ns.

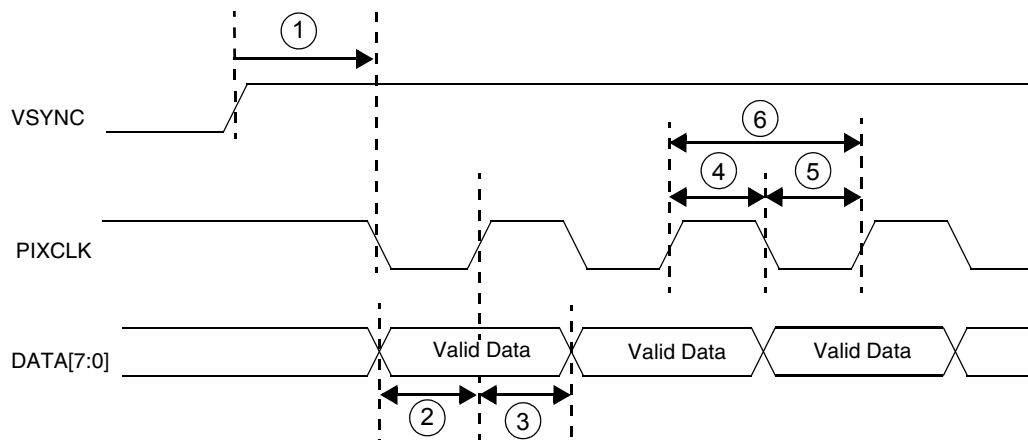
$$\begin{aligned} \text{positive duty cycle} &= 10 / 2 = 5\text{ns} \\ \Rightarrow \text{max rise time allowed} &= 5 - 1 = 4\text{ns} \\ \text{negative duty cycle} &= 10 / 2 = 5\text{ns} \\ \Rightarrow \text{max fall time allowed} &= 5 - 1 = 4\text{ns} \end{aligned}$$

Falling-edge latch data

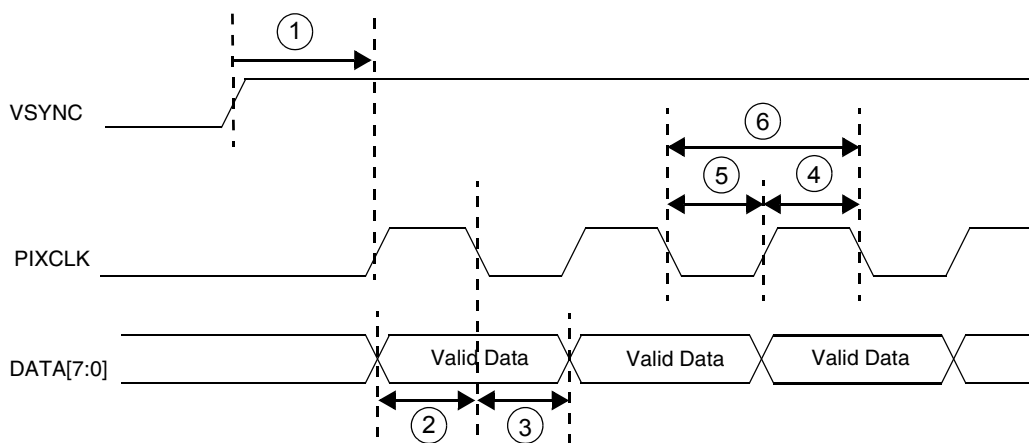
$$\begin{aligned} \text{max fall time allowed} &= (\text{negative duty cycle} - \text{hold time}) \\ \text{max rise time allowed} &= (\text{positive duty cycle} - \text{setup time}) \end{aligned}$$

4.14.2 Non-Gated Clock Mode

Figure 65 shows the timing diagram when the CMOS sensor output data is configured for negative edge and the CSI is programmed to received data on the positive edge. Figure 66 shows the timing diagram when the CMOS sensor output data is configured for positive edge and the CSI is programmed to received data in negative edge. The parameters for the timing diagrams are listed in Table 36.



**Figure 65. Sensor Output Data on Pixel Clock Falling Edge
CSI Latches Data on Pixel Clock Rising Edge**



**Figure 66. Sensor Output Data on Pixel Clock Rising Edge
CSI Latches Data on Pixel Clock Falling Edge**

Table 36. Non-Gated Clock Mode Parameters

Ref No.	Parameter	Min	Max	Unit
1	csi_vsync to csi_pixclk	180	—	ns
2	csi_d setup time	1	—	ns
3	csi_d hold time	1	—	ns
4	csi_pixclk high time	10.42	—	ns
5	csi_pixclk low time	10.42	—	ns
6	csi_pixclk frequency	0	48	MHz

The limitation on pixel clock rise time / fall time are not specified. It should be calculated from the hold time and setup time, according to:

$$\begin{aligned} \text{max rise time allowed} &= (\text{positive duty cycle} - \text{hold time}) \\ \text{max fall time allowed} &= (\text{negative duty cycle} - \text{setup time}) \end{aligned}$$

In most of case, duty cycle is 50 / 50, therefore:

$$\begin{aligned} \text{max rise time} &= (\text{period} / 2 - \text{hold time}) \\ \text{max fall time} &= (\text{period} / 2 - \text{setup time}) \end{aligned}$$

For example: Given pixel clock period = 10ns, duty cycle = 50 / 50, hold time = 1ns, setup time = 1ns.

$$\begin{aligned} \text{positive duty cycle} &= 10 / 2 = 5\text{ns} \\ \Rightarrow \text{max rise time allowed} &= 5 - 1 = 4\text{ns} \\ \text{negative duty cycle} &= 10 / 2 = 5\text{ns} \\ \Rightarrow \text{max fall time allowed} &= 5 - 1 = 4\text{ns} \end{aligned}$$

Falling-edge latch data

$$\begin{aligned} \text{max fall time allowed} &= (\text{negative duty cycle} - \text{hold time}) \\ \text{max rise time allowed} &= (\text{positive duty cycle} - \text{setup time}) \end{aligned}$$

5 Pin-Out and Package Information

Table 37 illustrates the package pin assignments for the 256-pin MAPBGA package. For a complete listing of signals, see the Signal Multiplexing Table 3 on page 9.

Table 37. i.MXL 256 MAPBGA Pin Assignments

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	
A	NVSS	SD_DAT3	SD_CLK	NVSS	USBD_AFE	NVDD4	NVSS	UART1_RTS	UART1_RXD	NVDD3	N.C.	N.C.	QVDD4	N.C.	N.C.	N.C.	A
B	A24	SD_DAT1	SD_CMD	PB16	USBD_ROE	USBD_VP	SSI_RXCLK	SSI_TXCLK	SPI1_SCLK	N.C.	N.C.	N.C.	QVSS	N.C.	N.C.	N.C.	B
C	A23	D31	SD_DAT0	PB15	USBD_RCV	UART2_CTS	UART2_RXD	SSI_RXFS	UART1_TXD	N.C.	N.C.	N.C.	N.C.	N.C.	N.C.	N.C.	C
D	A22	D30	D29	PB14	USBD_SUSPND	USBD_VPO	USBD_VMO	SSI_RXDAT	SPI1_SPL_RDY	N.C.	N.C.	N.C.	N.C.	N.C.	N.C.	N.C.	D
E	A20	A21	D28	D26	SD_DAT2	USBD_VM	UART2_RTS	SSI_TXDAT	SPI1_SS	N.C.	N.C.	N.C.	N.C.	N.C.	N.C.	N.C.	E
F	A18	D27	D25	A19	A16	PB18	UART2_TXD	SSI_TXFS	SPI1_MISO	N.C.	N.C.	REV	N.C.	N.C.	LSCLK	SPL_SPR	F
G	A15	A17	D24	D23	D21	PB17	PB19	UART1_CTS	SPI1_MOSI	N.C.	CLS	CONTRAST	ACD/OE	LP/HSYNC	FLM/VSYN	LD1	G
H	A13	D22	A14	D20	NVDD1	NVDD1	NVSS	QVSS	QVDD1	PS	LD0	LD2	LD4	LD5	LD9	LD3	H
J	A12	A11	D18	D19	NVDD1	NVDD1	NVSS	NVDD1	NVSS	NVSS	LD6	LD7	LD8	LD11	QVDD3	QVSS	J
K	A10	D16	A9	D17	NVDD1	NVSS	NVSS	NVDD1	NVDD2	NVDD2	LD10	LD12	LD13	LD14	TMR2OUT	LD15	K
L	A8	A7	D13	D15	D14	NVDD1	NVSS	CAS	TCK	TIN	PWMO	CSI_MCLK	CSI_D0	CSI_D1	CSI_D2	CSI_D3	L
M	A5	D12	D11	A6	SDCLK	NVSS	RW	MA10	RAS	RESET_IN	BIG_ENDIAN	CSI_D4	CSI_HSYNC	CSI_VSYNC	CSI_D6	CSI_D5	M
N	A4	EB1	D10	D7	A0	D4	PA17	D1	DQM1	RESET_SF1	RESET_OUT	BOOT2	CSI_PIXCLK	CSI_D7	TMS	TDI	N
P	A3	D9	EB0	CS3	D6	ECB	D2	D3	DQM3	SDCKE1	BOOT3	BOOT0	TRST	I2C_SCL	I2C_SDA	XTAL32K	P
R	EB2	EB3	A1	CS4	D8	D5	LB ¹	BCLK ²	D0	DQM0	SDCKE0	POR	BOOT1	TD0	QVDD2	EXTAL32K	R
T	NVSS	A2	OE	CS5	CS2	CS1	CS0	MA11	DQM2	SDWE	CLKO	AVDD1	TRISTATE	EXTAL16M	XTAL16M	QVSS	T
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	

¹ This signal is not used and should be floated in an actual application.

² burst clock

5.1 MAPBGA 256 Package Dimensions

Figure 67 illustrates the 256 MAPBGA 14 mm × 14 mm × 1.30 mm package, with an 0.8 mm pad pitch. The device designator for the MAPBGA package is VH.

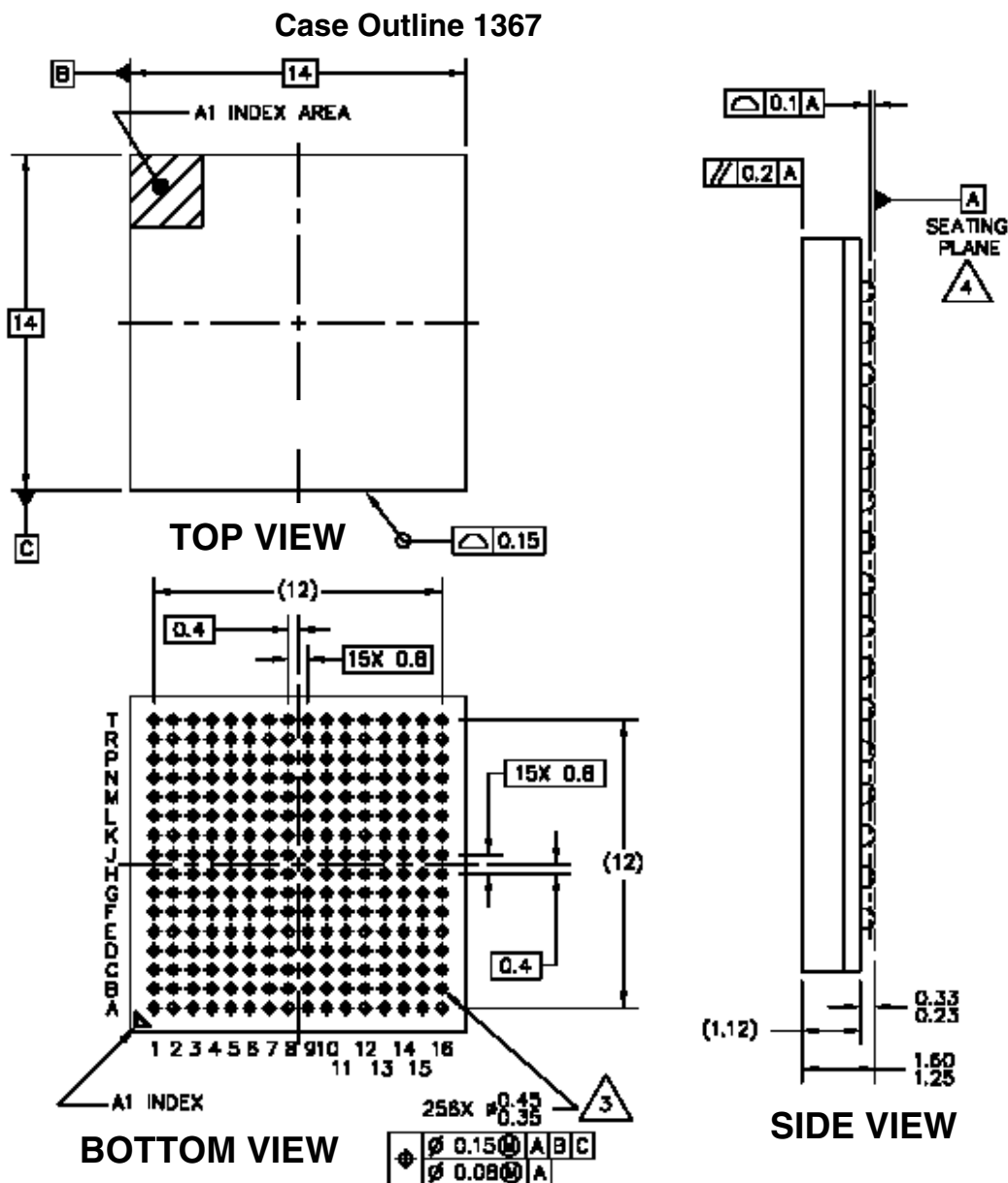


Figure 67. i.MXL 256 MAPBGA Mechanical Drawing

5.2 MAPBGA 225 Package Dimensions

Figure 68 illustrates the 225 MAPBGA 13 mm × 13 mm package.

Case Outline 1304B

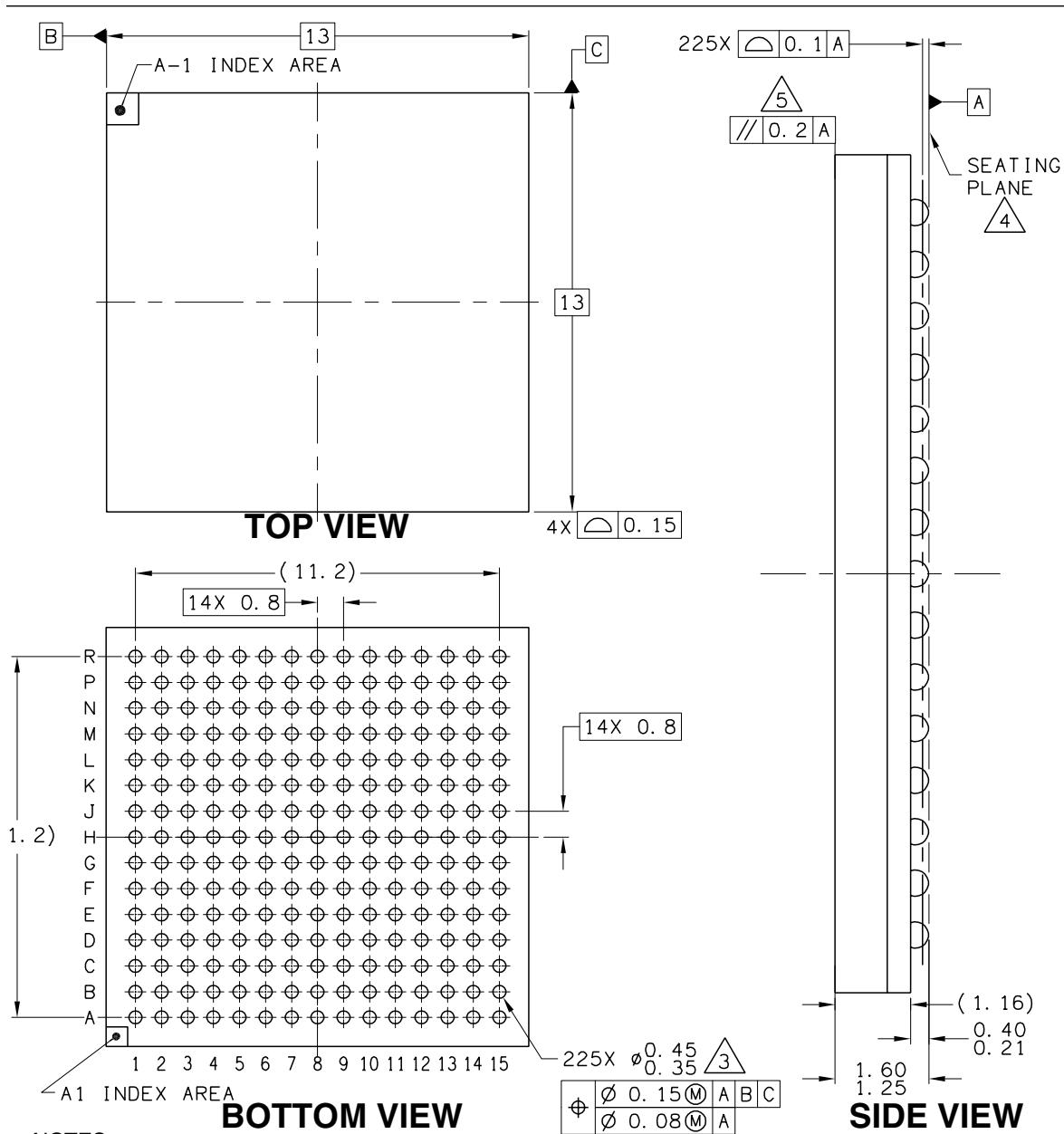


Figure 68. i.MXL 225 MAPBGA Mechanical Drawing

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