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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	e200z0h
Core Size	32-Bit Single-Core
Speed	64MHz
Connectivity	CANbus, I²C, LINbus, SCI, SPI, UART/USART
Peripherals	DMA, LVD, POR, PWM, WDT
Number of I/O	77
Program Memory Size	768KB (768K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	64K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 53x10/12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP
Supplier Device Package	100-LQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/spc560b54l3b6e0x

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Table 2. SPC560B54/6x family comparison[1] (continued)

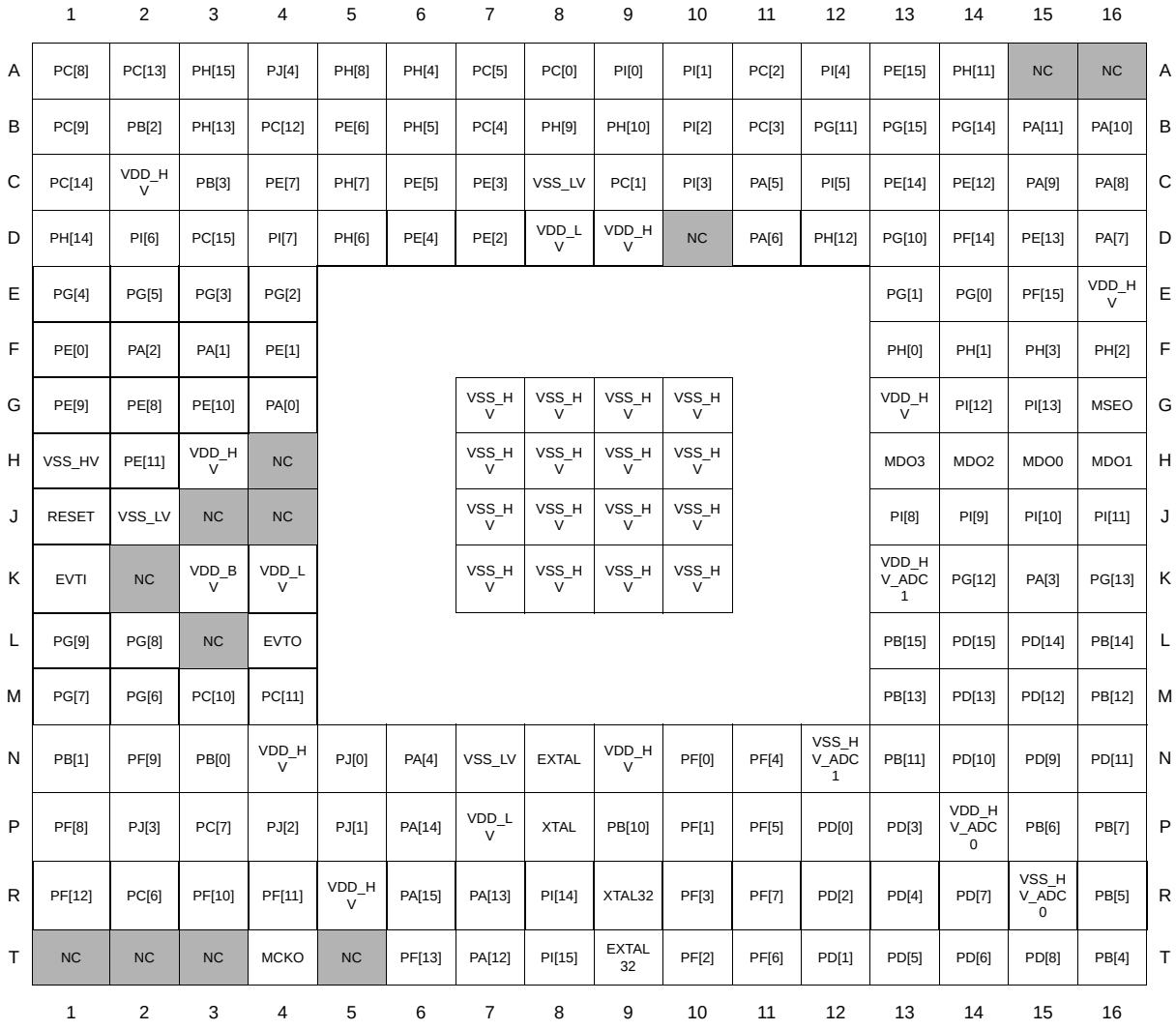
Feature	SPC560B54		SPC560B60			SPC560B64			
OPWM / ICOC[9]	13 ch	33 ch	13 ch	33 ch	33 ch	13 ch	33 ch	33 ch	33 ch
SCI (LINFlex)	4	8	4	8	10	4	8	10	10
SPI (DSPI)	3	5	3	5	6	3	5	6	6
CAN (FlexCAN)	6								
I2C	1								
32 KHz oscillator	Yes								
GPIO[10]	77	121	77	121	149	77	121	149	149
Debug	JTAG								N2+
Package	LQFP 100	LQFP 144	LQFP 100	LQFP 144	LQFP 176	LQFP 100	LQFP 144	LQFP 176	LBGA208[11]

1. Feature set dependent on selected peripheral multiplexing; table shows example.
2. Based on 125 °C ambient operating temperature.
3. Not shared with 12-bit ADC, but possibly shared with other alternate functions.
4. Not shared with 10-bit ADC, but possibly shared with other alternate functions.
5. See the eMIOS section of the chip reference manual for information on the channel configuration and functions.
6. Each channel supports a range of modes including Modulus counters, PWM generation, Input Capture, Output Compare.
7. Each channel supports a range of modes including PWM generation with dead time, Input Capture, Output Compare.
8. Each channel supports a range of modes including PWM generation, Input Capture, Output Compare, Period and Pulse width measurement.
9. Each channel supports a range of modes including PWM generation, Input Capture, and Output Compare.
10. Maximum I/O count based on multiplexing with peripherals.
11. LBGA208 available only as development package for Nexus2+.

Table 3 summarizes the functions of the blocks present on the SPC560B54/6x.

Table 3. SPC560B54/6x series block summary

Block	Function
Analog-to-digital converter (ADC)	Converts analog voltages to digital values
Boot assist module (BAM)	A block of read-only memory containing VLE code which is executed according to the boot mode of the device
Clock generation module (MC_CGM)	Provides logic and control required for the generation of system and peripheral clocks
Clock monitor unit (CMU)	Monitors clock source (internal and external) integrity
Cross triggering unit (CTU)	Enables synchronization of ADC conversions with a timer event from the eMIOS or from the PIT
Crossbar switch (XBAR)	Supports simultaneous connections between two master ports and three slave ports. The crossbar supports a 32-bit address bus width and a 64-bit data bus width.
Deserial serial peripheral interface (DSPI)	Provides a synchronous serial interface for communication with external devices
Enhanced direct memory access (eDMA)	Performs complex data transfers with minimal intervention from a host processor via "*n*" programmable channels
Enhanced modular input output system (eMIOS)	Provides the functionality to generate or measure events
Error correction status module (ECSM)	Provides a myriad of miscellaneous control functions for the device including program-visible information about configuration and revision levels, a reset status register, wakeup control for exiting sleep modes, and optional features such as information on memory errors reported by error-correcting codes
Flash memory	Provides non-volatile storage for program code, constants and variables
FlexCAN (controller area network)	Supports the standard CAN communications protocol
Frequency-modulated phase-locked loop (FMPLL)	Generates high-speed system clocks and supports programmable frequency modulation
Inter-integrated circuit (I^2C) bus	Two-wire bidirectional serial bus that provides a simple and efficient method of data exchange between devices
Internal multiplexer (IMUX) SIU subblock	Allows flexible mapping of peripheral interface on the different pins of the device
Interrupt controller (INTC)	Provides priority-based preemptive scheduling of interrupt requests
JTAG controller (JTAGC)	Provides the means to test chip functionality and connectivity while remaining transparent to system logic when not in test mode
LINFlex controller	Manages a high number of LIN (Local Interconnect Network protocol) messages efficiently with a minimum of CPU load
Memory protection unit (MPU)	Provides hardware access control for all memory references generated in a device



	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	
A	PC[8]	PC[13]	PH[15]	PJ[4]	PH[8]	PH[4]	PC[5]	PC[0]	PI[0]	PI[1]	PC[2]	PI[4]	PE[15]	PH[11]	NC	NC	A
B	PC[9]	PB[2]	PH[13]	PC[12]	PE[6]	PH[5]	PC[4]	PH[9]	PH[10]	PI[2]	PC[3]	PG[11]	PG[15]	PG[14]	PA[11]	PA[10]	B
C	PC[14]	VDD_HV	PB[3]	PE[7]	PH[7]	PE[5]	PE[3]	VSS_LV	PC[1]	PI[3]	PA[5]	PI[5]	PE[14]	PE[12]	PA[9]	PA[8]	C
D	PH[14]	PI[6]	PC[15]	PI[7]	PH[6]	PE[4]	PE[2]	VDD_LV	VDD_HV	NC	PA[6]	PH[12]	PG[10]	PF[14]	PE[13]	PA[7]	D
E	PG[4]	PG[5]	PG[3]	PG[2]									PG[1]	PG[0]	PF[15]	VDD_HV	E
F	PE[0]	PA[2]	PA[1]	PE[1]									PH[0]	PH[1]	PH[3]	PH[2]	F
G	PE[9]	PE[8]	PE[10]	PA[0]			VSS_HV	VSS_HV	VSS_HV	VSS_HV			VDD_HV	PI[12]	PI[13]	MSEO	G
H	VSS_HV	PE[11]	VDD_HV	NC			VSS_HV	VSS_HV	VSS_HV	VSS_HV			MDO3	MDO2	MDO0	MDO1	H
J	RESET	VSS_LV	NC	NC			VSS_HV	VSS_HV	VSS_HV	VSS_HV			PI[8]	PI[9]	PI[10]	PI[11]	J
K	EVTI	NC	VDD_BV	VDD_LV			VSS_HV	VSS_HV	VSS_HV	VSS_HV			VDD_HV_ADC1	PG[12]	PA[3]	PG[13]	K
L	PG[9]	PG[8]	NC	EVTO									PB[15]	PD[15]	PD[14]	PB[14]	L
M	PG[7]	PG[6]	PC[10]	PC[11]									PB[13]	PD[13]	PD[12]	PB[12]	M
N	PB[1]	PF[9]	PB[0]	VDD_HV	PJ[0]	PA[4]	VSS_LV	EXTAL	VDD_HV	PF[0]	PF[4]	VSS_HV_ADC1	PB[11]	PD[10]	PD[9]	PD[11]	N
P	PF[8]	PJ[3]	PC[7]	PJ[2]	PJ[1]	PA[14]	VDD_LV	XTAL	PB[10]	PF[1]	PF[5]	PD[0]	PD[3]	VDD_HV_ADC0	PB[6]	PB[7]	P
R	PF[12]	PC[6]	PF[10]	PF[11]	VDD_HV	PA[15]	PA[13]	PI[14]	XTAL32	PF[3]	PF[7]	PD[2]	PD[4]	PD[7]	VSS_HV_ADC0	PB[5]	R
T	NC	NC	NC	MCKO	NC	PF[13]	PA[12]	PI[15]	EXTAL32	PF[2]	PF[6]	PD[1]	PD[5]	PD[6]	PD[8]	PB[4]	T
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	

NOTE: The LBGA208 is available only as development package for Nexus 2+.

NC = Not connected

Figure 5. LBGA208 configuration

3.2 Pad configuration during reset phases

All pads have a fixed configuration under reset.

During the power-up phase, all pads are forced to tristate.

After power-up phase, all pads are tristate with the following exceptions:

- PA[9] (FAB) is pull-down. Without external strong pull-up the device starts fetching from flash.
- PA[8], PC[0] and PH[9:10] are in input weak pull-up when out of reset.
- RESET pad is driven low by the device till 40 FIRC clock cycles after phase2 completion. Minimum phase3 duration is 40 FIRC cycles.
- Nexus output pads (MDO[n], MCKO, EVTO, MSEO) are forced to output.

Table 4. Voltage supply pin descriptions (continued)

Port pin	Function	Pin number			
		LQFP100	LQFP144	LQFP176	LBGA208
VDD_BV	Internal regulator supply voltage	20	24	32	K3
VSS_HV_ADC0	Reference ground and analog ground for the A/D converter 0 (10-bit)	51	73	89	R15
VDD_HV_ADC0	Reference voltage and analog supply for the A/D converter 0 (10-bit)	52	74	90	P14
VSS_HV_ADC1	Reference ground and analog ground for the A/D converter 1 (12-bit)	59	81	98	N12
VDD_HV_ADC1	Reference voltage and analog supply for the A/D converter 1 (12-bit)	60	82	99	K13

1. A decoupling capacitor must be placed between each of the three VDD_LV/VSS_LV supply pairs to ensure stable voltage (see the recommended operating conditions in the device datasheet).

3.5 Pad types

In the device the following types of pads are available for system pins and functional port pins:

S = Slow[d]

M = Medium[d] [e]

F = Fast[d] [e]

I = Input only with analog feature[d]

J = Input/Output ('S' pad) with analog feature

X = Oscillator

3.6 System pins

The system pins are listed in *Table 5*.

d. See the I/O pad electrical characteristics in the chip datasheet for details.

e. All medium and fast pads are in slow configuration by default at reset and can be configured as fast or medium. The only exception is PC[1] which is in medium configuration by default (see the PCR.SRC description in the chip reference manual, Pad Configuration Registers (PCR0–PCR148)).

Table 6. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								LQFP 100	LQFP 144	LQFP 176	LBGA 208[4]
PA[4]	PCR[4]	AF0 AF1 AF2 AF3 — —	GPIO[4] E0UC[4] — CS0_1 LIN5RX WKPU[9][5]	SIUL eMIOS_0 — DSPI_1 LINFlex_5 WKPU	I/O I/O — I/O I I	S	Tristate	29	43	51	N6
PA[5]	PCR[5]	AF0 AF1 AF2 AF3	GPIO[5] E0UC[5] LIN4TX —	SIUL eMIOS_0 LINFlex_4 —	I/O I/O O —	M	Tristate	79	118	146	C11
PA[6]	PCR[6]	AF0 AF1 AF2 AF3 — —	GPIO[6] E0UC[6] — CS1_1 EIRQ[1] LIN4RX	SIUL eMIOS_0 — DSPI_1 SIUL LINFlex_4	I/O I/O — O I I	S	Tristate	80	119	147	D11
PA[7]	PCR[7]	AF0 AF1 AF2 AF3 — —	GPIO[7] E0UC[7] LIN3TX — EIRQ[2] ADC1_S[1]	SIUL eMIOS_0 LINFlex_3 — SIUL ADC_1	I/O I/O O — I I	J	Tristate	71	104	128	D16

Table 6. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								LQFP 100	LQFP 144	LQFP 176	LBGA 208[4]
PC[12]	PCR[44]	AF0 AF1 AF2 AF3 — —	GPIO[44] E0UC[12] — — EIRQ[19] SIN_2	SIUL eMIOS_0 — — SIUL DSPI_2	I/O I/O — — I I	M	Tristate	97	141	173	B4
PC[13]	PCR[45]	AF0 AF1 AF2 AF3	GPIO[45] E0UC[13] SOUT_2 —	SIUL eMIOS_0 DSPI_2 —	I/O I/O O —	S	Tristate	98	142	174	A2
PC[14]	PCR[46]	AF0 AF1 AF2 AF3 —	GPIO[46] E0UC[14] SCK_2 — EIRQ[8]	SIUL eMIOS_0 DSPI_2 — SIUL	I/O I/O I/O — I	S	Tristate	3	3	3	C1
PC[15]	PCR[47]	AF0 AF1 AF2 AF3 —	GPIO[47] E0UC[15] CS0_2 — EIRQ[20]	SIUL eMIOS_0 DSPI_2 — SIUL	I/O I/O I/O — I	M	Tristate	4	4	4	D3
Port D											

Table 6. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								LQFP 100	LQFP 144	LQFP 176	LBGA 208[4]
PD[4]	PCR[52]	AF0 AF1 AF2 AF3 — —	GPIO[52] — — — ADC0_P[8] ADC1_P[8]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	45	67	81	R13
PD[5]	PCR[53]	AF0 AF1 AF2 AF3 — —	GPIO[53] — — — ADC0_P[9] ADC1_P[9]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	46	68	82	T13
PD[6]	PCR[54]	AF0 AF1 AF2 AF3 — —	GPIO[54] — — — ADC0_P[10] ADC1_P[10]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	47	69	83	T14
PD[7]	PCR[55]	AF0 AF1 AF2 AF3 — —	GPIO[55] — — — ADC0_P[11] ADC1_P[11]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	48	70	84	R14

Table 6. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								LQFP 100	LQFP 144	LQFP 176	LBGA 208[4]
PI[5]	PCR[133]	AF0 AF1 AF2 AF3	GPIO[133] E1UC[29] SCK_4 —	SIUL eMIOS_1 DSPI_4 —	I/O I/O I/O —	S	Tristate	—	—	142	C12
PI[6]	PCR[134]	AF0 AF1 AF2 AF3	GPIO[134] E1UC[30] CS0_4 —	SIUL eMIOS_1 DSPI_4 —	I/O I/O I/O —	S	Tristate	—	—	11	D2
PI[7]	PCR[135]	AF0 AF1 AF2 AF3	GPIO[135] E1UC[31] CS1_4 —	SIUL eMIOS_1 DSPI_4 —	I/O I/O O —	S	Tristate	—	—	12	D3
PI[8]	PCR[136]	AF0 AF1 AF2 AF3 —	GPIO[136] — — — ADC0_S[16]	SIUL — — — ADC_0	I/O — — — I	J	Tristate	—	—	108	J13
PI[9]	PCR[137]	AF0 AF1 AF2 AF3 —	GPIO[137] — — — ADC0_S[17]	SIUL — — — ADC_0	I/O — — — I	J	Tristate	—	—	109	J14

3.8 Nexus 2+ pins

In the LBGA208 package, eight additional debug pins are available (see *Table 7*).

Table 7. Nexus 2+ pin descriptions

Port pin	Function	I/O direction	Pad type	Function after reset	Pin number		
					LQFP 100	LQFP 144	LBGA 208(1)
MCKO	Message clock out	O	F	—	—	—	T4
MDO0	Message data out 0	O	M	—	—	—	H15
MDO1	Message data out 1	O	M	—	—	—	H16
MDO2	Message data out 2	O	M	—	—	—	H14
MDO3	Message data out 3	O	M	—	—	—	H13
EVTI	Event in	I	M	Pull-up	—	—	K1
EVTO	Event out	O	M	—	—	—	L4
MSEO	Message start/end out	O	M	—	—	—	G16

1. LBGA208 available only as development package for Nexus2+.

The internal voltage regulator requires external capacitance (C_{REGn}) to be connected to the device in order to provide a stable low voltage digital supply to the device. Capacitances should be placed on the board as near as possible to the associated pins. Care should also be taken to limit the serial inductance of the board to less than 5 nH.

Each decoupling capacitor must be placed between each of the three V_{DD_LV}/V_{SS_LV} supply pairs to ensure stable voltage (see *Section 4.4: Recommended operating conditions*).

Table 26. Voltage regulator electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
C_{REGn}	SR	—	Internal voltage regulator external capacitance	—	200	—	500	nF
R_{REG}	SR	—	Stability capacitor equivalent serial resistance	Range: 10 kHz to 20 MHz	—	—	0.2	W
C_{DEC1}	SR	—	Decoupling capacitance[2] ballast	V_{DD_BV}/V_{SS_LV} pair: V_{DD_BV} = 4.5 V to 5.5 V	100[3]	470[4]	—	nF
				V_{DD_BV}/V_{SS_LV} pair: V_{DD_BV} = 3 V to 3.6 V	400		—	
C_{DEC2}	SR	—	Decoupling capacitance regulator supply	V_{DD}/V_{SS} pair	10	100	—	nF
V_{MREG}	CC	T	Main regulator output voltage	Before exiting from reset	—	1.32	—	V
		P		After trimming	1.16	1.28	—	
I_{MREG}	SR	—	Main regulator current provided to V_{DD_LV} domain	—	—	—	150	mA
$I_{MREGINT}$	CC	D	Main regulator module current consumption	I_{MREG} = 200 mA	—	—	2	mA
				I_{MREG} = 0 mA	—	—	1	
V_{LPREG}	CC	P	Low-power regulator output voltage	After trimming	1.16	1.28	—	V
I_{LPREG}	SR	—	Low-power regulator current provided to V_{DD_LV} domain	—	—	—	15	mA
$I_{LPREGINT}$	CC	D	Low-power regulator module current consumption	I_{LPREG} = 15 mA; T_A = 55 °C	—	—	600	µA
		—		I_{LPREG} = 0 mA; T_A = 55 °C	—	5	—	
V_{ULPREG}	CC	P	Ultra low power regulator output voltage	After trimming	1.16	1.28	—	V
I_{ULPREG}	SR	—	Ultra low power regulator current provided to V_{DD_LV} domain	—	—	—	5	mA
$I_{ULPREGINT}$	CC	D	Ultra low power regulator module current consumption	I_{ULPREG} = 5 mA; T_A = 55 °C	—	—	100	µA
				I_{ULPREG} = 0 mA; T_A = 55 °C	—	2	—	
I_{DD_BV}	CC	D	In-rush average current on V_{DD_BV} during power-up[5]	—	—	—	300[6]	mA

1. V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified.
2. This capacitance value is driven by the constraints of the external voltage regulator supplying the V_{DD_BV} voltage. A typical value is in the range of 470 nF.
3. This value is acceptable to guarantee operation from 4.5 V to 5.5 V.
4. External regulator and capacitance circuitry must be capable of providing I_{DD_BV} while maintaining supply V_{DD_BV} in operating range.
5. In-rush average current is seen only for short time during power-up and on standby exit (maximum 20 µs, depending on external capacitances to be loaded).
6. The duration of the in-rush current depends on the capacitance placed on LV pins. BV decoupling capacitors must be sized accordingly. Refer to I_{MREG} value for minimum amount of current to be provided in cc.

4.8.2 Low voltage detector electrical characteristics

The device implements a power-on reset (POR) module to ensure correct power-up initialization, as well as five low voltage detectors (LVDs) to monitor the V_{DD} and the V_{DD_LV} voltage while device is supplied:

- POR monitors V_{DD} during the power-up phase to ensure device is maintained in a safe reset state (refer to RGM Destructive Event Status (RGM_DES) Register flag F_POR in device reference manual)
- LVDHV3 monitors V_{DD} to ensure device reset below minimum functional supply (refer to RGM Destructive Event Status (RGM_DES) Register flag F_LVD27 in device reference manual)
- LVDHV3B monitors V_{DD_BV} to ensure device reset below minimum functional supply (refer to RGM Destructive Event Status (RGM_DES) Register flag F_LVD27_VREG in device reference manual)
- LVDHV5 monitors V_{DD} when application uses device in the 5.0 V ± 10% range (refer to RGM Functional Event Status (RGM_FES) Register flag F_LVD45 in device reference manual)
- LVDLVCOR monitors power domain No. 1 (refer to RGM Destructive Event Status (RGM_DES) Register flag F_LVD12_PD1 in device reference manual)
- LVDLVBKP monitors power domain No. 0 (refer to RGM Destructive Event Status (RGM_DES) Register flag F_LVD12_PD0 in device reference manual)

Note: When enabled, power domain No. 2 is monitored through LVDLVBKP.

Therefore it is recommended that the user apply EMC software optimization and prequalification tests in relation with the EMC level requested for the application.

- Software recommendations – The software flowchart must include the management of runaway conditions such as:
  - Corrupted program counter
  - Unexpected reset
  - Critical data corruption (control registers...)
- Prequalification trials – Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the reset pin or the oscillator pins for 1 second.

  To complete these trials, ESD stress can be applied directly on the device. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note *Software Techniques For Improving Microcontroller EMC Performance* (AN1015)).

4.11.2 Electromagnetic interference (EMI)

The product is monitored in terms of emission based on a typical application. This emission test conforms to the IEC61967-1 standard, which specifies the general conditions for EMI measurements.

Table 34. EMI radiated emission measurement(1)(2)

Symbol		C	Parameter	Conditions		Value			Unit
						Min	Typ	Max	
—	SR	—	Scan range	—		0.150		1000	MHz
f_{CPU}	SR	—	Operating frequency	—		—	64	—	MHz
V_{DD_LV}	SR	—	LV operating voltages	—		—	1.28	—	V
S_{EMI}	CC	T	Peak level	V_{DD} = 5 V, T_A = 25 °C, LQFP144 package Test conforming to IEC 61967-2, f_{OSC} = 8 MHz/f_{CPU} = 64 MHz	No PLL frequency modulation	—	—	18	dBµV
					± 2% PLL frequency modulation	—	—	14	dBµV

1. EMI testing and I/O port waveforms per IEC 61967-1, -2, -4.
2. For information on conducted emission and susceptibility measurement (norm IEC 61967-4), please contact your local marketing representative.

4.11.3 Absolute maximum ratings (electrical sensitivity)

Based on two different tests (ESD and LU) using specific measurement methods, the product is stressed in order to determine its performance in terms of electrical sensitivity.

Figure 15. Slow external crystal oscillator (32 kHz) timing diagram

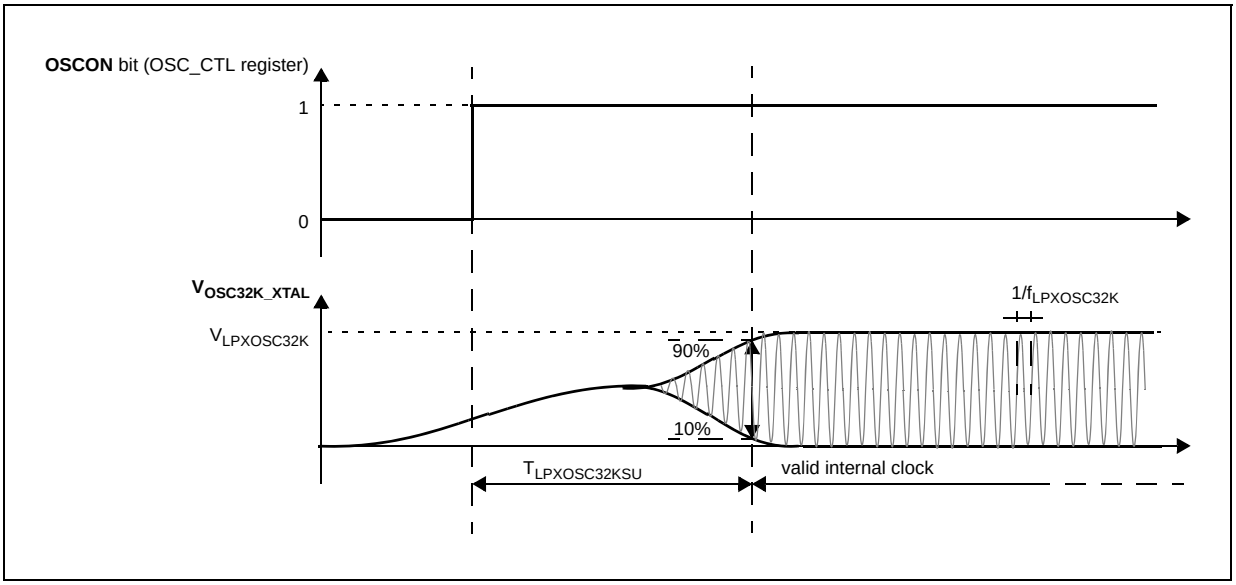


Table 40. Slow external crystal oscillator (32 kHz) electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
f_{SXOSC}	SR	—	Slow external crystal oscillator frequency	—	32	32.768	40	kHz
V_{SXOSC}	CC	T	Oscillation amplitude	—	—	2.1	—	V
$I_{SXOSCBIAS}$	CC	T	Oscillation bias current	—		2.5		µA
I_{SXOSC}	CC	T	Slow external crystal oscillator consumption	—	—	—	8	µA
$t_{SXOSCSU}$	CC	T	Slow external crystal oscillator start-up time	—	—	—	2[2]	s

1. V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified. Values are specified for no neighbor GPIO pin activity. If oscillator is enabled (OSC32K_XTAL and OSC32K_EXTAL pins), neighboring pins should not toggle.
2. Start-up time has been measured with EPSON TOYOCOM MC306 crystal. Variation may be seen with other crystal.

4.14 FMPLL electrical characteristics

The device provides a frequency modulated phase locked loop (FMPLL) module to generate a fast system clock from the main oscillator driver.

1. A first and quick charge transfer from the internal capacitance C_{P1} and C_{P2} to the sampling capacitance C_S occurs (C_S is supposed initially completely discharged): considering a worst case (since the time constant in reality would be faster) in which C_{P2} is reported in parallel to C_{P1} (call $C_P = C_{P1} + C_{P2}$), the two capacitances C_P and C_S are in series, and the time constant is

Equation 5

$$\tau_1 = (R_{SW} + R_{AD}) \bullet \frac{C_P \bullet C_S}{C_P + C_S}$$

Equation 5 can again be simplified considering only C_S as an additional worst condition. In reality, the transient is faster, but the A/D converter circuitry has been designed to be robust also in the very worst case: the sampling time t_S is always much longer than the internal time constant:

Equation 6

$$\tau_1 < (R_{SW} + R_{AD}) \bullet C_S \ll t_s$$

The charge of C_{P1} and C_{P2} is redistributed also on C_S, determining a new value of the voltage V_{A1} on the capacitance according to *Equation 7*:

Equation 7

$$V_{A1} \bullet (C_S + C_{P1} + C_{P2}) = V_A \bullet (C_{P1} + C_{P2})$$

2. A second charge transfer involves also C_F (that is typically bigger than the on-chip capacitance) through the resistance R_L: again considering the worst case in which C_{P2} and C_S were in parallel to C_{P1} (since the time constant in reality would be faster), the time constant is:

Equation 8

$$\tau_2 < R_L \bullet (C_S + C_{P1} + C_{P2})$$

In this case, the time constant depends on the external circuit: in particular imposing that the transient is completed well before the end of sampling time t_s, a constraints on R_L sizing is obtained:

Equation 9 ADC_0 (10-bit)

$$8.5 \bullet \tau_2 = 8.5 \bullet R_L \bullet (C_S + C_{P1} + C_{P2}) < t_s$$

Equation 10 ADC_1 (12-bit)

$$10 \bullet \tau_2 = 10 \bullet R_L \bullet (C_S + C_{P1} + C_{P2}) < t_s$$

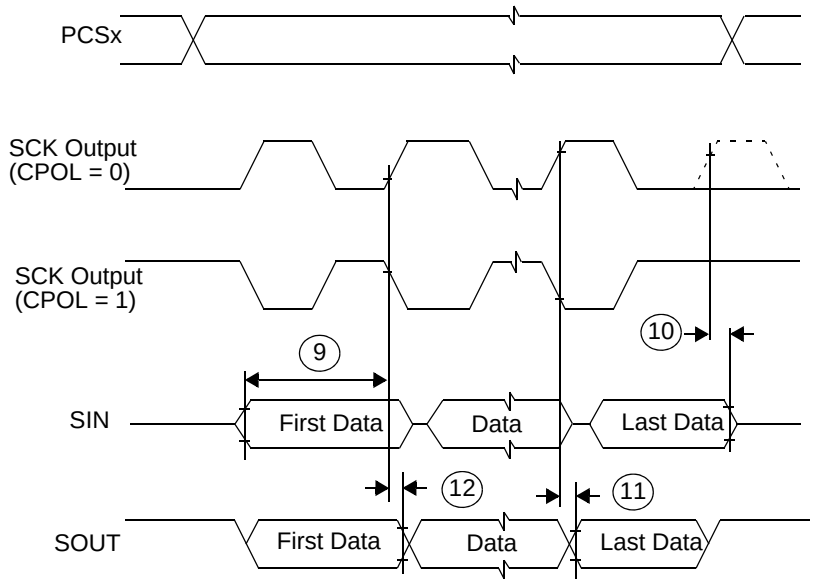
Of course, R_L shall be sized also according to the current limitation constraints, in combination with R_S (source impedance) and R_F (filter resistance). Being C_F definitively bigger than C_{P1}, C_{P2} and C_S, then the final voltage V_{A2} (at the end of the charge transfer transient) will be much higher than V_{A1}. *Equation 11* must be respected (charge balance assuming now C_S already charged at V_{A1}):

Table 46. ADC_1 conversion characteristics (12-bit ADC_1) (continued)

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
I_{INJ}	SR	—	Input current Injection	Current injection on one ADC_1 input, different from the converted one	V_{DD} = 3.3 V ± 10%	–5	—	5	mA
					V_{DD} = 5.0 V ± 10%	–5	—	5	
| INLP |	CC	T	Absolute integral nonlinearity – Precise channels	No overload		—	1	3	LSB
| INLX |	CC	T	Absolute integral nonlinearity – Extended channels	No overload		—	1.5	5	LSB
| DNL |	CC	T	Absolute differential nonlinearity	No overload		—	0.5	1	LSB
| E_O |	CC	T	Absolute offset error	—		—	2	—	LSB
| E_G |	CC	T	Absolute gain error	—		—	2	—	LSB
TUEP[7]	CC	P	Total unadjusted error for precise channels, input only pins	Without current injection		–6	—	6	LSB
		T		With current injection		–8	—	8	
TUEX[7]	CC	T	Total unadjusted error for extended channel	Without current injection		–10	—	10	LSB
		T		With current injection		–12	—	12	

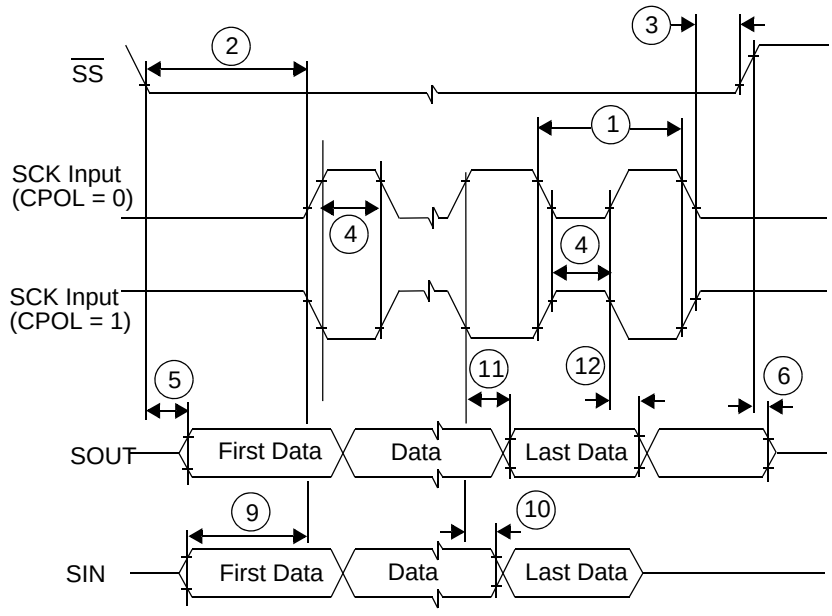
1. V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified
2. Analog and digital V_{SS} **must** be common (to be tied together externally).
3. V_{AINx} may exceed V_{SS_ADC1} and V_{DD_ADC1} limits, remaining on absolute maximum ratings, but the results of the conversion will be clamped respectively to 0x000 or 0xFFF.
4. During the sampling time the input capacitance C_S can be charged/discharged by the external source. The internal resistance of the analog source must allow the capacitance to reach its final voltage level within t_{ADC1_S}. After the end of the sampling time t_{ADC1_S}, changes of the analog input voltage have no effect on the conversion result. Values for the sampling clock t_{ADC1_S} depend on programming.
5. This parameter does not include the sampling time t_{ADC1_S}, but only the time for determining the digital result and the time to load the result's register with the conversion result.
6. Duty cycle is ensured by using system clock without prescaling. When ADCLKSEL = 0, the duty cycle is ensured by internal divider by 2.
7. Total Unadjusted Error: The maximum error that occurs without adjusting Offset and Gain errors. This error is a combination of Offset, Gain and Integral Linearity errors.

Figure 27. DSPI modified transfer format timing — master, CPHA = 1



Note: Numbers shown reference *Table 47*.

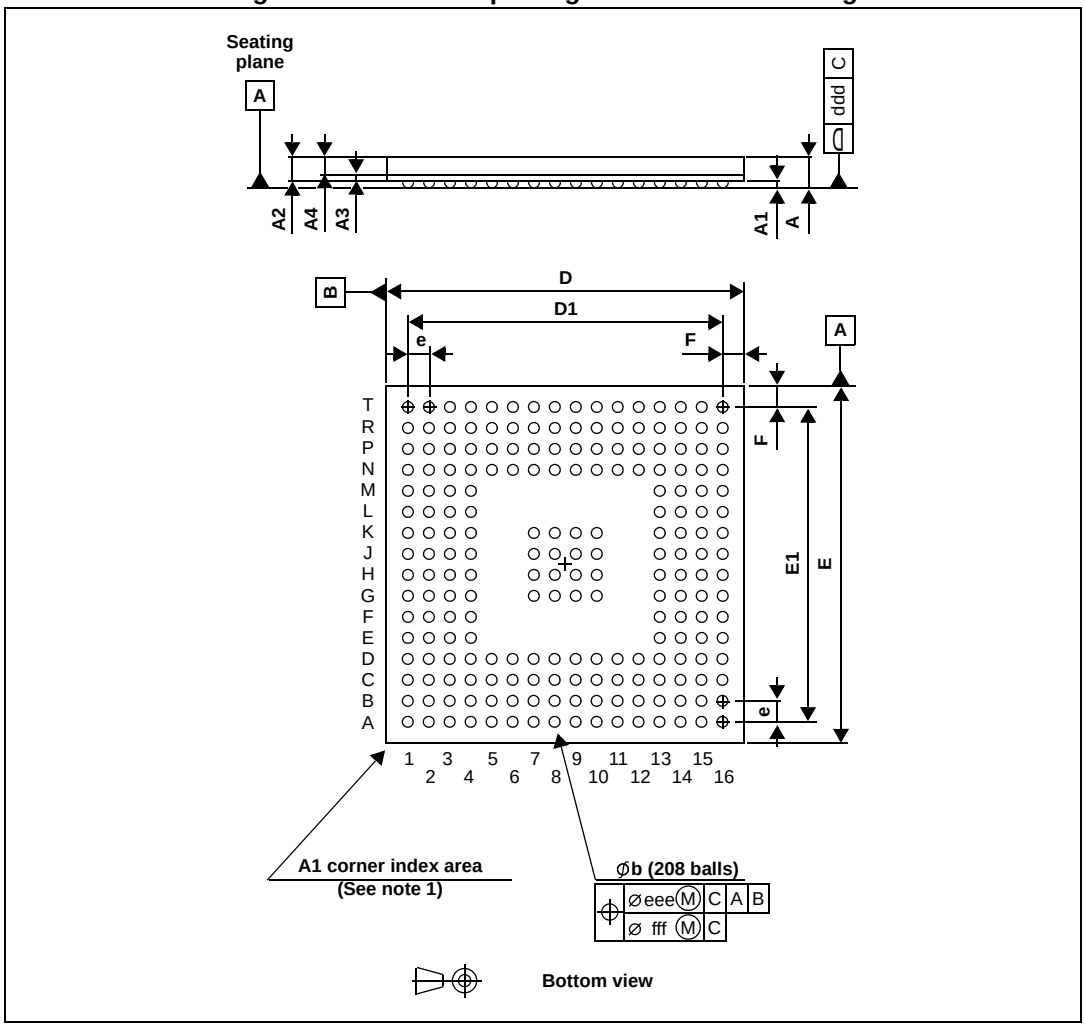
Figure 28. DSPI modified transfer format timing — slave, CPHA = 0



Note: Numbers shown reference *Table 47*.

5.2.4 LBGA208

Figure 36. LBGA208 package mechanical drawing



1. The terminal A1 corner must be identified on the top surface by using a corner chamfer, ink or metalized markings, or other feature of package body or integral heatslug.
A distinguishing feature is allowable on the bottom surface of the package to identify the terminal A1 corner. Exact shape of each corner is optional.

Table 54. LBGA208 mechanical data

Symbol	mm			inches(1)			Notes
	Min	Typ	Max	Min	Typ	Max	
A	—	—	1.70	—	—	0.0669	(2)
A1	0.30	—	—	0.0118	—	—	—
A2	—	1.085	—	—	0.0427	—	—
A3	—	0.30	—	—	0.0118	—	—
A4	—	—	0.80	—	—	0.0315	—
b	0.50	0.60	0.70	0.0197	0.0236	0.0276	(3)

Table 54. LBGA208 mechanical data (continued)

Symbol	mm			inches[1]			Notes
	Min	Typ	Max	Min	Typ	Max	
D	16.80	17.00	17.20	0.6614	0.6693	0.6772	—
D1	—	15.00	—	—	0.5906	—	—
E	16.80	17.00	17.20	0.6614	0.6693	0.6772	—
E1	—	15.00	—	—	0.5906	—	—
e	—	1.00	—	—	0.0394	—	—
F	—	1.00	—	—	0.0394	—	—
ddd	—	—	0.20	—	—	0.0079	
eee	—	—	0.25	—	—	0.0098	(4)
fff	—	—	0.10	—	—	0.0039	(5)

1. Values in inches are converted from mm and rounded to 4 decimal digits.
2. LBGA stands for **L**ow profile **B**all **G**rid **A**rray.
   – Low profile: The total profile height (Dim A) is measured from the seating plane to the top of the component
   – The maximum total package height is calculated by the following methodology:
   A2 (Typ) + A1 (Typ) + $\sqrt{(A1^2 + A3^2 + A4^2}$ tolerance values)
   – Low profile: 1.20 mm < A ≤ 1.70 mm
3. The typical ball diameter before mounting is 0.60mm.
4. The tolerance of position that controls the location of the pattern of balls with respect to datums A and B.
   For each ball there is a cylindrical tolerance zone eee perpendicular to datum C and located on true position with respect to datums A and B as defined by e. The axis perpendicular to datum C of each ball must lie within this tolerance zone.
5. The tolerance of position that controls the location of the balls within the matrix with respect to each other.
   For each ball there is a cylindrical tolerance zone fff perpendicular to datum C and located on true position as defined by e.
   The axis perpendicular to datum C of each ball must lie within this tolerance zone.
   Each tolerance zone fff in the array is contained entirely in the respective zone eee above.
   The axis of each ball must lie simultaneously in both tolerance zones.