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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	e200z0h
Core Size	32-Bit Single-Core
Speed	64MHz
Connectivity	CANbus, I²C, LINbus, SCI, SPI, UART/USART
Peripherals	DMA, LVD, POR, PWM, WDT
Number of I/O	149
Program Memory Size	1.5MB (1.5M x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	96K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 53x10/12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	176-LQFP
Supplier Device Package	176-LQFP (24x24)
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/spc560b64l7c6e0y

1 Introduction

1.1 Document overview

This document describes the features of the family and options available within the family members, and highlights important electrical and physical characteristics of the device.

1.2 Description

This family of 32-bit system-on-chip (SoC) microcontrollers is the latest achievement in integrated automotive application controllers. It belongs to an expanding family of automotive-focused products designed to address the next wave of body electronics applications within the vehicle.

The advanced and cost-efficient e200z0h host processor core of this automotive controller family complies with the Power Architecture technology and only implements the VLE (variable-length encoding) APU (Auxiliary Processor Unit), providing improved code density. It operates at speeds of up to 64 MHz and offers high performance processing optimized for low power consumption. It capitalizes on the available development infrastructure of current Power Architecture devices and is supported with software drivers, operating systems and configuration code to assist with users implementations.

Table 2. SPC560B54/6x family comparison(1)

Feature	SPC560B54		SPC560B60			SPC560B64			
CPU	e200z0h								
Execution speed(2)	Up to 64 MHz								
Code flash memory	768 KB		1 MB			1.5 MB			
Data flash memory	64 (4 × 16) KB								
SRAM	64 KB		80 KB			96 KB			
MPU	8-entry								
eDMA	16 ch								
10-bit ADC	Yes								
dedicated(3)	7 ch	15 ch	7 ch	15 ch	29 ch	7 ch	15 ch	29 ch	29 ch
shared with 12-bit ADC	19 ch								
12-bit ADC	Yes								
dedicated(4)	5 ch								
shared with 10-bit ADC	19 ch								
Total timer I/O(5) eMIOS	37 ch, 16-bit	64 ch, 16-bit	37 ch, 16-bit	64 ch, 16-bit	64 ch, 16-bit	37 ch, 16-bit	64 ch,16-bit	64 ch, 16-bit	64 ch, 16-bit
Counter / OPWM / ICOC(6)	10 ch								
O(I)PWM / OPWFMB / OPWMCB / ICOC(7)	7 ch								
O(I)PWM / ICOC(8)	7 ch	14 ch	7 ch	14 ch	14 ch	7 ch	14 ch	14 ch	14 ch

Table 3. SPC560B54/6x series block summary (continued)

Block	Function
Mode entry module (MC_ME)	Provides a mechanism for controlling the device operational mode and mode transition sequences in all functional states; also manages the power control unit, reset generation module and clock generation module, and holds the configuration, control and status registers accessible for applications
Non-maskable interrupt (NMI)	Handles external events that must produce an immediate response, such as power down detection
Periodic interrupt timer (PIT)	Produces periodic interrupts and triggers
Power control unit (MC_PCU)	Reduces the overall power consumption by disconnecting parts of the device from the power supply via a power switching device; device components are grouped into sections called “power domains” which are controlled by the PCU
Real-time counter (RTC)	A free running counter used for time keeping applications, the RTC can be configured to generate an interrupt at a predefined interval independent of the mode of operation (run mode or low-power mode)
Reset generation module (MC_RGM)	Centralizes reset sources and manages the device reset sequence of the device
Static random-access memory (SRAM)	Provides storage for program code, constants, and variables
System integration unit lite (SIUL)	Provides control over all the electrical pad controls and up 32 ports with 16 bits of bidirectional, general-purpose input and output signals and supports up to 32 external interrupts with trigger event configuration
System status and configuration module (SSCM)	Provides system configuration and status data (such as memory size and status, device mode and security status), device identification data, debug status port enable and selection, and bus and peripheral abort enable/disable
System timer module (STM)	Provides a set of output compare events to support AUTOSAR (Automotive Open System Architecture) and operating system tasks
Software watchdog timer (SWT)	Provides protection from runaway code
Wakeup unit (WKPU)	The wakeup unit supports up to 27 external sources that can generate interrupts or wakeup events, of which 1 can cause non-maskable interrupt requests or wakeup events.

Figure 4. LQFP100 pin configuration

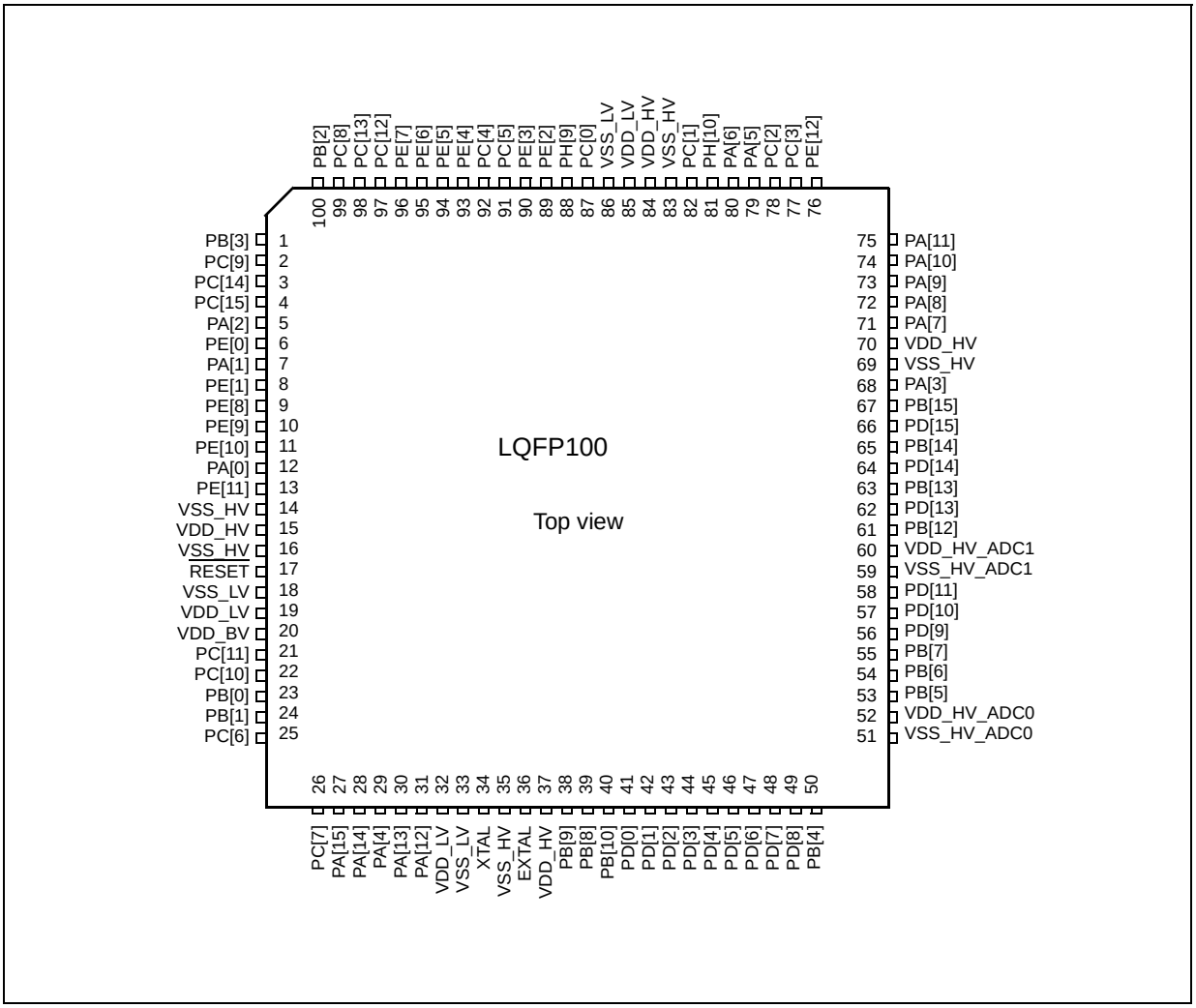


Figure 5 shows the SPC560B54/6x in the LBGA208 package.

Table 6. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								LQFP 100	LQFP 144	LQFP 176	LBGA 208[4]
Port C											
PC[0][10]	PCR[32]	AF0 AF1 AF2 AF3	GPIO[32] — TDI —	SIUL — JTAGC —	I/O — I —	M	Input, weak pull-up	87	126	154	A8
PC[1][10]	PCR[33]	AF0 AF1 AF2 AF3	GPIO[33] — TDO —	SIUL — JTAGC —	I/O — O —	F[11]	Tristate	82	121	149	C9
PC[2]	PCR[34]	AF0 AF1 AF2 AF3 —	GPIO[34] SCK_1 CAN4TX DEBUG[0] EIRQ[5]	SIUL DSPI_1 FlexCAN_4 SSCM SIUL	I/O I/O O O I	M	Tristate	78	117	145	A11
PC[3]	PCR[35]	AF0 AF1 AF2 AF3 — — —	GPIO[35] CS0_1 MA[0] DEBUG[1] EIRQ[6] CAN1RX CAN4RX	SIUL DSPI_1 ADC_0 SSCM SIUL FlexCAN_1 FlexCAN_4	I/O I/O O O I I I	S	Tristate	77	116	144	B11

Table 6. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								LQFP 100	LQFP 144	LQFP 176	LBGA 208[4]
PC[12]	PCR[44]	AF0 AF1 AF2 AF3 — —	GPIO[44] E0UC[12] — — EIRQ[19] SIN_2	SIUL eMIOS_0 — — SIUL DSPI_2	I/O I/O — — I I	M	Tristate	97	141	173	B4
PC[13]	PCR[45]	AF0 AF1 AF2 AF3	GPIO[45] E0UC[13] SOUT_2 —	SIUL eMIOS_0 DSPI_2 —	I/O I/O O —	S	Tristate	98	142	174	A2
PC[14]	PCR[46]	AF0 AF1 AF2 AF3 —	GPIO[46] E0UC[14] SCK_2 — EIRQ[8]	SIUL eMIOS_0 DSPI_2 — SIUL	I/O I/O I/O — I	S	Tristate	3	3	3	C1
PC[15]	PCR[47]	AF0 AF1 AF2 AF3 —	GPIO[47] E0UC[15] CS0_2 — EIRQ[20]	SIUL eMIOS_0 DSPI_2 — SIUL	I/O I/O I/O — I	M	Tristate	4	4	4	D3
Port D											

Table 6. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								LQFP 100	LQFP 144	LQFP 176	LBGA 208[4]
PE[14]	PCR[78]	AF0 AF1 AF2 AF3 —	GPIO[78] SCK_2 E1UC[21] — EIRQ[12]	SIUL DSPI_2 eMIOS_1 — SIUL	I/O I/O I/O — I	S	Tristate	—	112	136	C13
PE[15]	PCR[79]	AF0 AF1 AF2 AF3	GPIO[79] CS0_2 E1UC[22] —	SIUL DSPI_2 eMIOS_1 —	I/O I/O I/O —	M	Tristate	—	113	137	A13
Port F											
PF[0]	PCR[80]	AF0 AF1 AF2 AF3 —	GPIO[80] E0UC[10] CS3_1 — ADC0_S[8]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	—	55	63	N10
PF[1]	PCR[81]	AF0 AF1 AF2 AF3 —	GPIO[81] E0UC[11] CS4_1 — ADC0_S[9]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	—	56	64	P10
PF[2]	PCR[82]	AF0 AF1 AF2 AF3 —	GPIO[82] E0UC[12] CS0_2 — ADC0_S[10]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O I/O — I	J	Tristate	—	57	65	T10

Table 6. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								LQFP 100	LQFP 144	LQFP 176	LBGA 208[4]
PF[3]	PCR[83]	AF0 AF1 AF2 AF3 —	GPIO[83] E0UC[13] CS1_2 — ADC0_S[11]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	—	58	66	R10
PF[4]	PCR[84]	AF0 AF1 AF2 AF3 —	GPIO[84] E0UC[14] CS2_2 — ADC0_S[12]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	—	59	67	N11
PF[5]	PCR[85]	AF0 AF1 AF2 AF3 —	GPIO[85] E0UC[22] CS3_2 — ADC0_S[13]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	—	60	68	P11
PF[6]	PCR[86]	AF0 AF1 AF2 AF3 —	GPIO[86] E0UC[23] CS1_1 — ADC0_S[14]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	—	61	69	T11
PF[7]	PCR[87]	AF0 AF1 AF2 AF3 —	GPIO[87] — CS2_1 — ADC0_S[15]	SIUL — DSPI_1 — ADC_0	I/O — O — I	J	Tristate	—	62	70	R11

Table 6. Functional port pin descriptions (continued)

Port pin	PCR	Alternate function[1]	Function	Peripheral	I/O direction[2]	Pad type	RESET configuration[3]	Pin number			
								LQFP 100	LQFP 144	LQFP 176	LBGA 208[4]
PJ[2]	PCR[146]	AF0 AF1 AF2 AF3 —	GPIO[146] CS0_5 — — ADC0_S[26]	SIUL DSPI_5 — — ADC_0	I/O I/O — — I	J	Tristate	—	—	72	P4
PJ[3]	PCR[147]	AF0 AF1 AF2 AF3 —	GPIO[147] CS1_5 — — ADC0_S[27]	SIUL DSPI_5 — — ADC_0	I/O O — — I	J	Tristate	—	—	71	P2
PJ[4]	PCR[148]	AF0 AF1 AF2 AF3	GPIO[148] SCK_5 E1UC[18] —	SIUL DSPI_5 eMIOS_1 —	I/O I/O I/O —	M	Tristate	—	—	5	A4

1. Alternate functions are chosen by setting the values of the PCR.PA bitfields inside the SIUL module. PCR.PA = 00 → AF0; PCR.PA = 01 → AF1; PCR.PA = 10 → AF2; PCR.PA = 11 → AF2. This is intended to select the output functions; to use one of the input functions, the PCR.IBE bit must be written to '1', regardless of the values selected in the PCR.PA bitfields. For this reason, the value corresponding to an input only function is reported as "—".
2. Multiple inputs are routed to all respective modules internally. The input of some modules must be configured by setting the values of the PSMIO.PADSELx bitfields inside the SIUL module.
3. The RESET configuration applies during and after reset.
4. LBGA208 available only as development package for Nexus2+
5. All WKPU pins also support external interrupt capability. See the WKPU chapter for further details.
6. NMI has higher priority than alternate function. When NMI is selected, the PCR.AF field is ignored.
7. "Not applicable" because these functions are available only while the device is booting. Refer to the BAM information for details.
8. Value of PCR.IBE bit must be 0.
9. This wakeup input cannot be used to exit STANDBY mode.

4 Electrical characteristics

This section contains electrical characteristics of the device as well as temperature and power considerations.

This product contains devices to protect the inputs against damage due to high static voltages. However, it is advisable to take precautions to avoid application of any voltage higher than the specified maximum rated voltages.

To enhance reliability, unused inputs can be driven to an appropriate logic voltage level (V_{DD} or V_{SS}). This could be done by the internal pull-up and pull-down, which is provided by the product for most general purpose pins.

The parameters listed in the following tables represent the characteristics of the device and its demands on the system.

In the tables where the device logic provides signals with their respective timing characteristics, the symbol “CC” for Controller Characteristics is included in the Symbol column.

In the tables where the external system must provide signals with their respective timing characteristics to the device, the symbol “SR” for System Requirement is included in the Symbol column.

4.1 Parameter classification

The electrical parameters shown in this supplement are guaranteed by various methods. To give the customer a better understanding, the classifications listed in *Table 8* are used and the parameters are tagged accordingly in the tables where appropriate.

Table 8. Parameter classifications

Classification tag	Tag description
P	Those parameters are guaranteed during production testing on each individual device.
C	Those parameters are achieved by the design characterization by measuring a statistically relevant sample size across process variations.
T	Those parameters are achieved by design characterization on a small sample size from typical devices under typical conditions unless otherwise noted. All values shown in the typical column are within this category.
D	Those parameters are derived mainly from simulations.

Note: The classification is shown in the column labeled “C” in the parameter tables where appropriate.

4.2 NVUSRO register

Bit values in the Non-Volatile User Options (NVUSRO) Register control portions of the device configuration, namely electrical parameters such as high voltage supply and oscillator margin, as well as digital functionality (watchdog enable/disable after reset).

Table 18. SLOW configuration output buffer electrical characteristics (continued)

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
V_{OL}	CC	P	Output low level SLOW configuration	Push Pull	I_{OL} = 2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	—	—	$0.1V_{DD}$	V
		C			I_{OL} = 2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[2]	—	—	$0.1V_{DD}$	
		C			I_{OL} = 1 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	—	—	0.5	

1. V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified.
2. The configuration PAD3V5 = 1 when V_{DD} = 5 V is only a transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

Table 19. MEDIUM configuration output buffer electrical characteristics

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
V_{OH}	CC	C	Output high level MEDIUM configuration	Push Pull	I_{OH} = –3.8 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	$0.8V_{DD}$	—	—	V
		P			I_{OH} = –2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	$0.8V_{DD}$	—	—	
		C			I_{OH} = –1 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[2]	$0.8V_{DD}$	—	—	
		C			I_{OH} = –1 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	V_{DD} – 0.8	—	—	
		C			I_{OH} = –100 µA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	$0.8V_{DD}$	—	—	
V_{OL}	CC	C	Output low level MEDIUM configuration	Push Pull	I_{OL} = 3.8 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	$0.2V_{DD}$	V
		P			I_{OL} = 2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	—	—	$0.1V_{DD}$	
		C			I_{OL} = 1 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[2]	—	—	$0.1V_{DD}$	
		C			I_{OL} = 1 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	—	—	0.5	
		C			I_{OL} = 100 µA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	$0.1V_{DD}$	

Table 28. Power consumption on VDD_BV and VDD_HV

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
I_{DDMAX}[2]	CC	D	RUN mode maximum average current	—		—	115	140[3]	mA
I_{DDRUN}[4]	CC	T	RUN mode typical average current[5]	f_{CPU} = 8 MHz		—	12	—	mA
		T		f_{CPU} = 16 MHz		—	27	—	
		T		f_{CPU} = 32 MHz		—	43	—	
		P		f_{CPU} = 48 MHz		—	56	100	
		P		f_{CPU} = 64 MHz		—	70	125	
I_{DDHALT}	CC	C	HALT mode current[6]	Slow internal RC oscillator (128 kHz) running	T_A = 25 °C	—	10	18	mA
		P			T_A = 125 °C	—	17	28	
I_{DDSTOP}	CC	P	STOP mode current[7]	Slow internal RC oscillator (128 kHz) running	T_A = 25 °C	—	350	900[8]	µA
		D			T_A = 55 °C	—	750	—	
		D			T_A = 85 °C	—	2	7	mA
		D			T_A = 105 °C	—	4	10	
		P			T_A = 125 °C	—	7	14	
$I_{DDSTDBY2}$	CC	P	STANDBY2 mode current[9]	Slow internal RC oscillator (128 kHz) running	T_A = 25 °C	—	30	100	µA
		D			T_A = 55 °C	—	75	—	
		D			T_A = 85 °C	—	180	700	
		D			T_A = 105 °C	—	315	1000	
		P			T_A = 125 °C	—	560	1700	
$I_{DDSTDBY1}$	CC	T	STANDBY1 mode current[10]	Slow internal RC oscillator (128 kHz) running	T_A = 25 °C	—	20	60	µA
		D			T_A = 55 °C	—	45	—	
		D			T_A = 85 °C	—	100	350	
		D			T_A = 105 °C	—	165	500	
		D			T_A = 125 °C	—	280	900	

1. V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified.
2. I_{DDMAX} is drawn only from the VDD_BV pin. Running consumption does not include I/Os toggling which is highly dependent on the application. The given value is thought to be a worst case value with all peripherals running, and code fetched from code flash while modify operation ongoing on data flash. Notice that this value can be significantly reduced by application: switch off not used peripherals (default), reduce peripheral frequency through internal prescaler, fetch from RAM most used functions, use low power mode when possible.
3. Higher current may be sunk by device during power-up and standby exit. Please refer to in-rush average current in *Table 26*.
4. I_{DDRUN} is drawn only from the VDD_BV pin. RUN current measured with typical application with accesses on both Flash and RAM.
5. Only for the “P” classification: Data and Code Flash in Normal Power. Code fetched from RAM: Serial IPs CAN and LIN in loop back mode, DSPI as Master, PLL as system clock (4 x Multiplier) peripherals on (eMIOS/CTU/ADC) and running at max frequency, periodic SW/WDG timer reset enabled.

Therefore it is recommended that the user apply EMC software optimization and prequalification tests in relation with the EMC level requested for the application.

- Software recommendations – The software flowchart must include the management of runaway conditions such as:
  - Corrupted program counter
  - Unexpected reset
  - Critical data corruption (control registers...)
- Prequalification trials – Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the reset pin or the oscillator pins for 1 second.

  To complete these trials, ESD stress can be applied directly on the device. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note *Software Techniques For Improving Microcontroller EMC Performance* (AN1015)).

4.11.2 Electromagnetic interference (EMI)

The product is monitored in terms of emission based on a typical application. This emission test conforms to the IEC61967-1 standard, which specifies the general conditions for EMI measurements.

Table 34. EMI radiated emission measurement(1)(2)

Symbol		C	Parameter	Conditions		Value			Unit
						Min	Typ	Max	
—	SR	—	Scan range	—		0.150		1000	MHz
f_{CPU}	SR	—	Operating frequency	—		—	64	—	MHz
V_{DD_LV}	SR	—	LV operating voltages	—		—	1.28	—	V
S_{EMI}	CC	T	Peak level	V_{DD} = 5 V, T_A = 25 °C, LQFP144 package Test conforming to IEC 61967-2, f_{OSC} = 8 MHz/f_{CPU} = 64 MHz	No PLL frequency modulation	—	—	18	dBµV
					± 2% PLL frequency modulation	—	—	14	dBµV

1. EMI testing and I/O port waveforms per IEC 61967-1, -2, -4.
2. For information on conducted emission and susceptibility measurement (norm IEC 61967-4), please contact your local marketing representative.

4.11.3 Absolute maximum ratings (electrical sensitivity)

Based on two different tests (ESD and LU) using specific measurement methods, the product is stressed in order to determine its performance in terms of electrical sensitivity.

Table 42. Fast internal RC oscillator (16 MHz) electrical characteristics (continued)

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
$I_{FIRCSTOP}$	CC	T	Fast internal RC oscillator high frequency and system clock current in stop mode	T_A = 25 °C	sysclk = off	—	500	—	µA
					sysclk = 2 MHz	—	600	—	
					sysclk = 4 MHz	—	700	—	
					sysclk = 8 MHz	—	900	—	
					sysclk = 16 MHz	—	1250	—	
t_{FIRCSU}	CC	C	Fast internal RC oscillator start-up time	V_{DD} = 5.0 V ± 10%		—	1.1	2.0	µs
$\Delta_{FIRCPRE}$	CC	C	Fast internal RC oscillator precision after software trimming of f_{FIRC}	T_A = 25 °C		–1	—	1	%
$\Delta_{FIRCTRIM}$	CC	C	Fast internal RC oscillator trimming step	T_A = 25 °C		—	1.6		%
$\Delta_{FIRCVAR}$	CC	C	Fast internal RC oscillator variation over temperature and supply with respect to f_{FIRC} at T_A = 25 °C in high-frequency configuration	—		–5	—	5	%

1. V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified.
2. This does not include consumption linked to clock tree toggling and peripherals consumption when RC oscillator is ON.

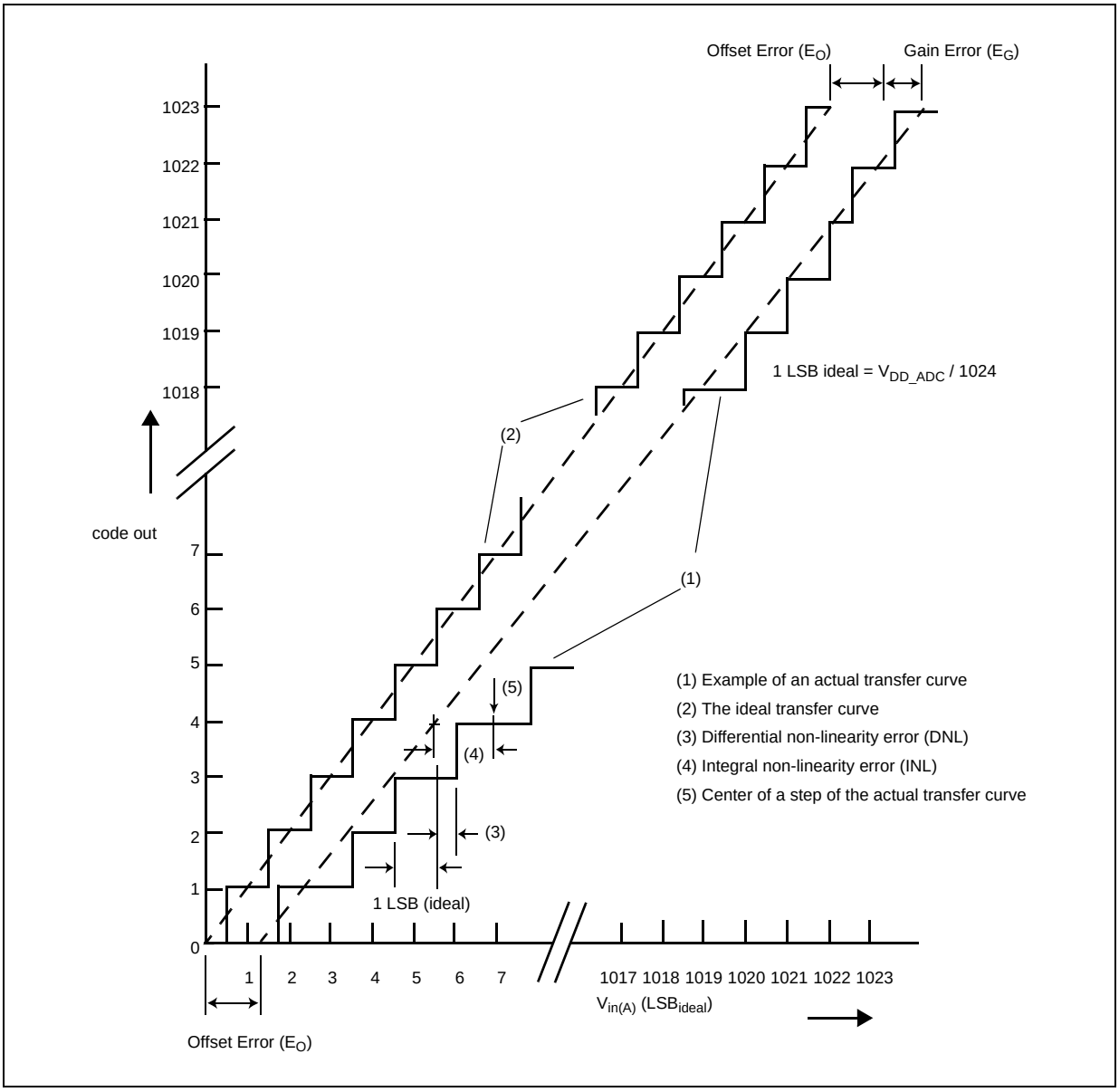
4.16 Slow internal RC oscillator (128 kHz) electrical characteristics

The device provides a 128 kHz low power internal RC oscillator. This can be used as the reference clock for the RTC module.

Table 43. Slow internal RC oscillator (128 kHz) electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
f_{SIRC}	CC	P	Slow internal RC oscillator low frequency	T_A = 25 °C, trimmed	—	128	—	kHz
	SR	—		—	100	—	150	
I_{SIRC}[2]	CC	C	Slow internal RC oscillator low frequency current	T_A = 25 °C, trimmed	—	—	5	µA
t_{SIRCSU}	CC	P	Slow internal RC oscillator start-up time	T_A = 25 °C, V_{DD} = 5.0 V ± 10%	—	8	12	µs

Figure 16. ADC_0 characteristic and error definitions



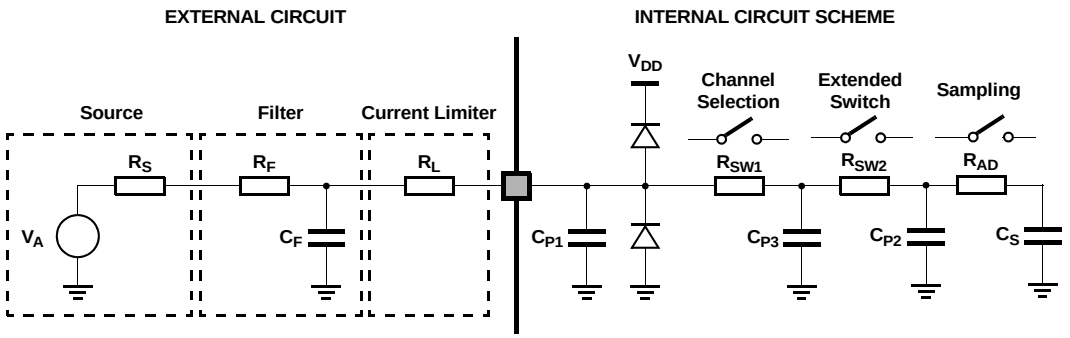
4.17.2 Input impedance and ADC accuracy

In the following analysis, the input circuit corresponding to the precise channels is considered.

To preserve the accuracy of the A/D converter, it is necessary that analog input pins have low AC impedance. Placing a capacitor with good high frequency characteristics at the input pin of the device can be effective: the capacitor should be as large as possible, ideally infinite. This capacitor contributes to attenuating the noise present on the input pin; furthermore, it sources charge during the sampling phase, when the analog signal source is a high-impedance source.

A real filter can typically be obtained by using a series resistance with a capacitor on the input pin (simple RC filter). The RC filtering may be limited according to the value of source

Figure 18. Input equivalent circuit (extended channels)



R_S Source Impedance

R_F Filter Resistance

C_F Filter Capacitance

R_L Current Limiter Resistance

R_{SW} Channel Selection Switch Impedance (two contributions R_{SW1} and R_{SW2})

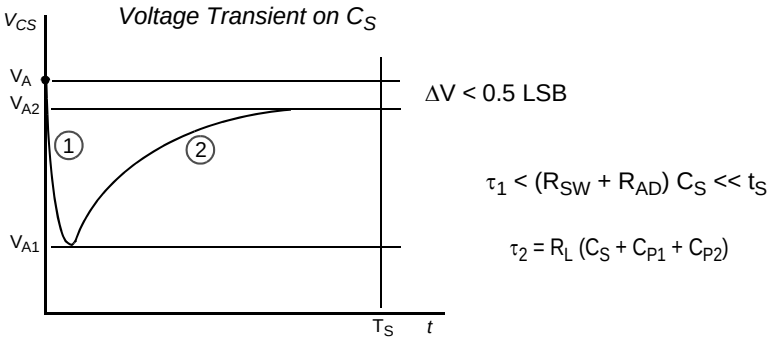
R_{AD}Sampling Switch Impedance

C_P Pin Capacitance (three contributions, C_{P1}, C_{P2} and C_{P3})

C_S Sampling Capacitance

A second aspect involving the capacitance network shall be considered. Assuming the three capacitances C_F, C_{P1} and C_{P2} are initially charged at the source voltage V_A (refer to the equivalent circuit reported in *Figure 17*): A charge sharing phenomenon is installed when the sampling phase is started (A/D switch close).

Figure 19. Transient behavior during sampling phase



In particular two different transient periods can be distinguished:

1. A first and quick charge transfer from the internal capacitance C_{P1} and C_{P2} to the sampling capacitance C_S occurs (C_S is supposed initially completely discharged): considering a worst case (since the time constant in reality would be faster) in which C_{P2} is reported in parallel to C_{P1} (call $C_P = C_{P1} + C_{P2}$), the two capacitances C_P and C_S are in series, and the time constant is

Equation 5

$$\tau_1 = (R_{SW} + R_{AD}) \bullet \frac{C_P \bullet C_S}{C_P + C_S}$$

Equation 5 can again be simplified considering only C_S as an additional worst condition. In reality, the transient is faster, but the A/D converter circuitry has been designed to be robust also in the very worst case: the sampling time t_S is always much longer than the internal time constant:

Equation 6

$$\tau_1 < (R_{SW} + R_{AD}) \bullet C_S \ll t_s$$

The charge of C_{P1} and C_{P2} is redistributed also on C_S, determining a new value of the voltage V_{A1} on the capacitance according to *Equation 7*:

Equation 7

$$V_{A1} \bullet (C_S + C_{P1} + C_{P2}) = V_A \bullet (C_{P1} + C_{P2})$$

2. A second charge transfer involves also C_F (that is typically bigger than the on-chip capacitance) through the resistance R_L: again considering the worst case in which C_{P2} and C_S were in parallel to C_{P1} (since the time constant in reality would be faster), the time constant is:

Equation 8

$$\tau_2 < R_L \bullet (C_S + C_{P1} + C_{P2})$$

In this case, the time constant depends on the external circuit: in particular imposing that the transient is completed well before the end of sampling time t_s, a constraints on R_L sizing is obtained:

Equation 9 ADC_0 (10-bit)

$$8.5 \bullet \tau_2 = 8.5 \bullet R_L \bullet (C_S + C_{P1} + C_{P2}) < t_s$$

Equation 10 ADC_1 (12-bit)

$$10 \bullet \tau_2 = 10 \bullet R_L \bullet (C_S + C_{P1} + C_{P2}) < t_s$$

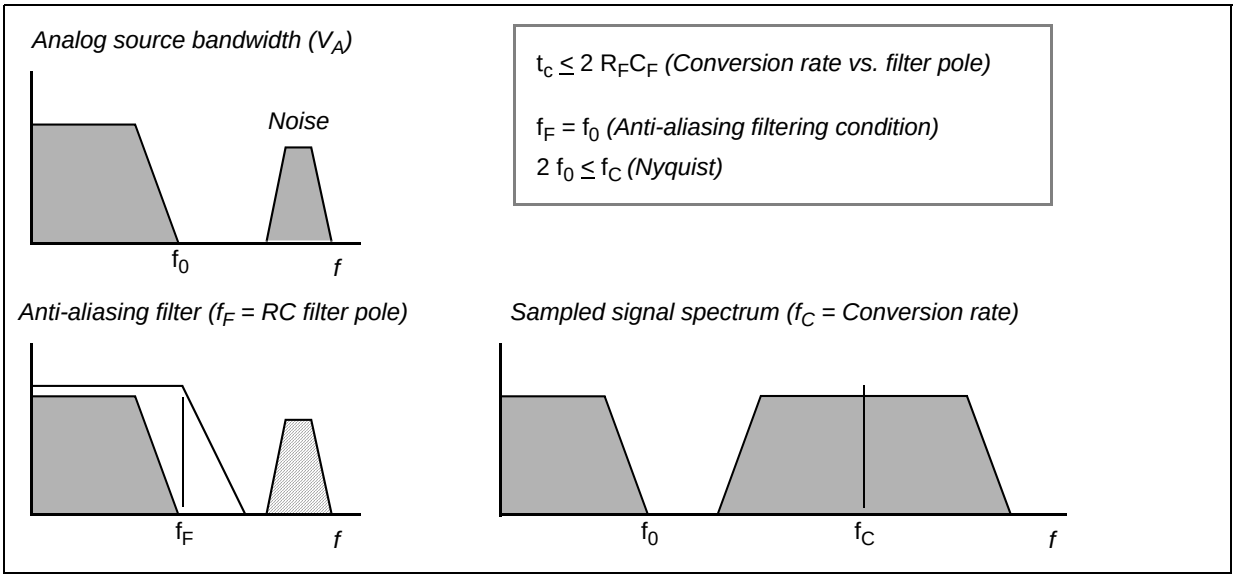
Of course, R_L shall be sized also according to the current limitation constraints, in combination with R_S (source impedance) and R_F (filter resistance). Being C_F definitively bigger than C_{P1}, C_{P2} and C_S, then the final voltage V_{A2} (at the end of the charge transfer transient) will be much higher than V_{A1}. *Equation 11* must be respected (charge balance assuming now C_S already charged at V_{A1}):

Equation 11

$$V_{A2} \bullet (C_S + C_{P1} + C_{P2} + C_F) = V_A \bullet C_F + V_{A1} \bullet (C_{P1} + C_{P2} + C_S)$$

The two transients above are not influenced by the voltage source that, due to the presence of the $R_F C_F$ filter, is not able to provide the extra charge to compensate the voltage drop on C_S with respect to the ideal source V_A; the time constant $R_F C_F$ of the filter is very high with respect to the sampling time (t_s). The filter is typically designed to act as antialiasing.

Figure 20. Spectral representation of input signal



Calling f_0 the bandwidth of the source signal (and as a consequence the cut-off frequency of the antialiasing filter, f_F), according to the Nyquist theorem the conversion rate f_C must be at least $2f_0$; it means that the constant time of the filter is greater than or at least equal to twice the conversion period (t_c). Again the conversion period t_c is longer than the sampling time t_s, which is just a portion of it, even when fixed channel continuous conversion mode is selected (fastest conversion rate at a specific channel): in conclusion it is evident that the time constant of the filter $R_F C_F$ is definitively much higher than the sampling time t_s, so the charge level on C_S cannot be modified by the analog signal source during the time in which the sampling switch is closed.

The considerations above lead to impose new constraints on the external circuit, to reduce the accuracy error due to the voltage drop on C_S; from the two charge balance equations above, it is simple to derive *Equation 12* between the ideal and real sampled voltage on C_S:

Equation 12

$$\frac{V_{A2}}{V_A} = \frac{C_{P1} + C_{P2} + C_F}{C_{P1} + C_{P2} + C_F + C_S}$$

From this formula, in the worst case (when V_A is maximum, that is for instance 5 V), assuming to accept a maximum error of half a count, a constraint is evident on C_F value:

Equation 13 ADC_0 (10-bit)

$$C_F > 2048 \bullet C_S$$

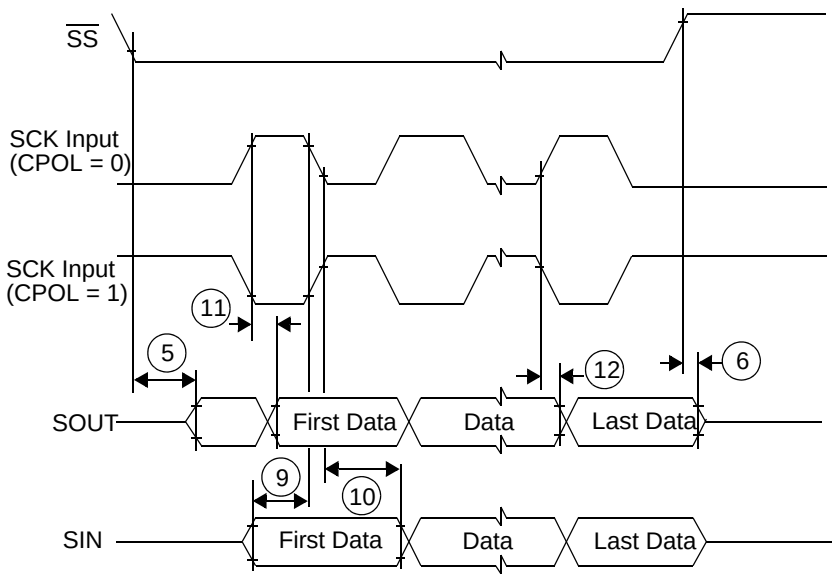
4.18 On-chip peripherals

4.18.1 Current consumption

Table 47. On-chip peripherals current consumption(1)

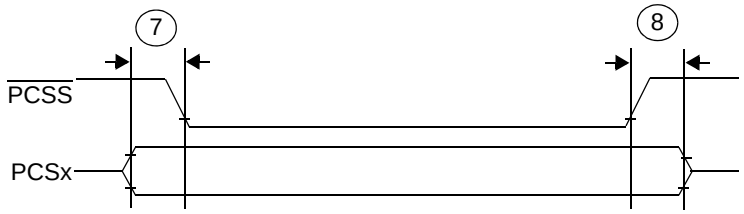
Symbol		C	Parameter	Conditions		Typical value(2)	Unit
$I_{DD\_BV(CAN)}$	CC	T	CAN (FlexCAN) supply current on V_{DD_BV}	Bitrate: 500 Kbyte/s	Total (static + dynamic) consumption: – FlexCAN in loop-back mode – XTAL at 8 MHz used as CAN engine clock source – Message sending period is 580 µs	8 * f_{periph} + 85	µA
				Bitrate: 125 Kbyte/s		8 * f_{periph} + 27	
$I_{DD\_BV(eMIOS)}$	CC	T	eMIOS supply current on V_{DD_BV}	Static consumption: – eMIOS channel OFF – Global prescaler enabled		29 * f_{periph}	µA
				Dynamic consumption: – It does not change varying the frequency (0.003 mA)		3	
$I_{DD\_BV(SCI)}$	CC	T	SCI (LINFlex) supply current on V_{DD_BV}	Total (static + dynamic) consumption: – LIN mode – Baudrate: 20 Kbyte/s		5 * f_{periph} + 31	µA
$I_{DD\_BV(SPI)}$	CC	T	SPI (DSPI) supply current on V_{DD_BV}	Ballast static consumption (only clocked)		1	µA
				Ballast dynamic consumption (continuous communication): – Baudrate: 2 Mbit/s – Transmission every 8 µs – Frame: 16 bits		16 * f_{periph}	
I_{DD_BV} (ADC_0/ADC_1)	CC	T	ADC_0/ADC_1 supply current on V_{DD_BV}	V_{DD} = 5.5 V	Ballast static consumption (no conversion)(3)	41 * f_{periph}	µA
					Ballast dynamic consumption (continuous conversion)(3)	46 * f_{periph}	
$I_{DD\_HV\_ADC0}$	CC	T	ADC_0 supply current on $V_{DD\_HV\_ADC0}$	V_{DD} = 5.5 V	Analog static consumption (no conversion)	200	µA
					Analog dynamic consumption (continuous conversion)	3	mA
$I_{DD\_HV\_ADC1}$	CC	T	ADC_1 supply current on $V_{DD\_HV\_ADC1}$	V_{DD} = 5.5 V	Analog static consumption (no conversion)	300 * f_{periph}	µA
					Analog dynamic consumption (continuous conversion)	4	mA

Figure 29. DSPI modified transfer format timing — slave, CPHA = 1



Note: Numbers shown reference *Table 47*.

Figure 30. DSPI PCS strobe ($\overline{\text{PCSS}}$) timing



Note: Numbers shown reference *Table 47*.

4.18.3 Nexus characteristics

Table 49. Nexus characteristics

No.	Symbol		C	Parameter	Value			Unit
					Min	Typ	Max	
1	t_{TCYC}	CC	D	TCK cycle time	64	—	—	ns
2	t_{MCYC}	CC	D	MCKO cycle time	32	—	—	ns
3	t_{MDOV}	CC	D	MCKO low to MDO data valid	—	—	8	ns
4	t_{MSEOV}	CC	D	MCKO low to MSEO_b data valid	—	—	8	ns

Table 56. Revision history (continued)

Date	Revision	Changes
13-Sep-2010	4 (cont.)	*Table 26* – Updated all values – Removed $I_{VREGREF}$ and $I_{VREDLVD12}$ rows – Added the footnote “The duration of the in-rush current depends on the capacitance placed on LV pins. BV decaps must be sized accordingly. Refer to IMREG value for minimum amount of current to be provided in cc.” to the I_{DD_BV} specification. *Table 27* – Updated V_{PORH} min/max value – Updated $V_{LVDLVCORL}$ min value Updated *Table 28* *Table 29* – $T_{dwprogram}$: added initial max value – Inserted T_{eslat} row *Table 30*: removed the “To be confirmed” footnote In the “Crystal oscillator and resonator connection scheme” figure, removed R_P. *Table 40* – Removed g_{mSXOSC} row – $I_{SXOSCBIAS}$: added min/typ/max value *Table 41*: – Added f_{VCO} row – Added Δt_{STJIT} row *Table 42* – $I_{FIRCPWD}$: removed row for T_A = 55 °C – Updated T_{FIRCSU} row *Table 45*: Added two rows: $I_{ADC0pwd}$ and $I_{ADC0run}$ *Table 46* – Added two rows: $I_{ADC1pwd}$ and $I_{ADC1run}$ – Updated values of f_{ADC_1} and t_{ADC1_PU} – Updated t_{ADC1_C} row Updated *Table 47* Updated *Table 48* Added *Table 55*
29-Oct- 2010	5	Removed “Preliminary—Subject to Change Without Notice” marking. This data sheet contains specifications based on characterization data. Updated *Table 55* Added *Table 56* Updated *Figure 37*