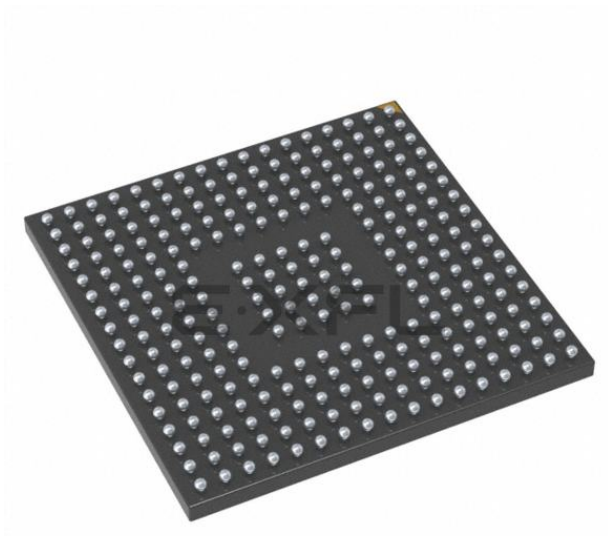




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Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Active
Core Processor	ARM® Cortex®-M7
Core Size	32-Bit Single-Core
Speed	216MHz
Connectivity	CANbus, EBI/EMI, Ethernet, I ² C, IrDA, LINbus, SAI, SD, SPDIF-Rx, SPI, UART/USART, USB OTG
Peripherals	Brown-out Detect/Reset, DMA, I ² S, POR, PWM, WDT
Number of I/O	140
Program Memory Size	512KB (512K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	320K x 8
Voltage - Supply (Vcc/Vdd)	1.7V ~ 3.6V
Data Converters	A/D 24x12b; D/A 2x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	201-UFBGA
Supplier Device Package	176+25UFBGA (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/stm32f745iek7

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1 Description

The STM32F745xx and STM32F746xx devices are based on the high-performance ARM® Cortex®-M7 32-bit RISC core operating at up to 216 MHz frequency. The Cortex®-M7 core features a single floating point unit (SFPU) precision which supports all ARM® single-precision data-processing instructions and data types. It also implements a full set of DSP instructions and a memory protection unit (MPU) which enhances the application security.

The STM32F745xx and STM32F746xx devices incorporate high-speed embedded memories with a Flash memory up to 1 Mbyte, 320 Kbytes of SRAM (including 64 Kbytes of Data TCM RAM for critical real-time data), 16 Kbytes of instruction TCM RAM (for critical real-time routines), 4 Kbytes of backup SRAM available in the lowest power modes, and an extensive range of enhanced I/Os and peripherals connected to two APB buses, two AHB buses, a 32-bit multi-AHB bus matrix and a multi layer AXI interconnect supporting internal and external memories access.

All the devices offer three 12-bit ADCs, two DACs, a low-power RTC, thirteen general-purpose 16-bit timers including two PWM timers for motor control and one low-power timer available in Stop mode, two general-purpose 32-bit timers, a true random number generator (RNG). They also feature standard and advanced communication interfaces.

- Up to four I²Cs
- Six SPIs, three I²Ss in duplex mode. To achieve the audio class accuracy, the I²S peripherals can be clocked via a dedicated internal audio PLL or via an external clock to allow synchronization.
- Four USARTs plus four UARTs
- An USB OTG full-speed and a USB OTG high-speed with full-speed capability (with the ULPI),
- Two CANs
- Two SAI serial audio interfaces
- An SDMMC host interface
- Ethernet and camera interfaces
- LCD-TFT display controller
- Chrom-ART Accelerator™
- SPDIFRX interface
- HDMI-CEC

Advanced peripherals include an SDMMC interface, a flexible memory control (FMC) interface, a Quad-SPI Flash memory interface, a camera interface for CMOS sensors. Refer to [Table 2: STM32F745xx and STM32F746xx features and peripheral counts](#) for the list of peripherals available on each part number.

The STM32F745xx and STM32F746xx devices operate in the –40 to +105 °C temperature range from a 1.7 to 3.6 V power supply. A dedicated supply input for USB (OTG_FS and OTG_HS) is available on all the packages except LQFP100 for a greater power supply choice.

The supply voltage can drop to 1.7 V with the use of an external power supply supervisor (refer to [Section 2.17.2: Internal reset OFF](#)). A comprehensive set of power-saving mode allows the design of low-power applications.

The STM32F745xx and STM32F746xx devices offer devices in 8 packages ranging from 100 pins to 216 pins. The set of included peripherals changes with the device chosen.

2.12 Nested vectored interrupt controller (NVIC)

The devices embed a nested vectored interrupt controller able to manage 16 priority levels, and handle up to 97 maskable interrupt channels plus the 16 interrupt lines of the Cortex[®]-M7 with FPU core.

- Closely coupled NVIC gives low-latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Allows early processing of interrupts
- Processing of late arriving, higher-priority interrupts
- Support tail chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimum interrupt latency.

2.13 External interrupt/event controller (EXTI)

The external interrupt/event controller consists of 24 edge-detector lines used to generate interrupt/event requests. Each line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the Internal APB2 clock period. Up to 168 GPIOs can be connected to the 16 external interrupt lines.

2.14 Clocks and startup

On reset the 16 MHz internal HSI RC oscillator is selected as the default CPU clock. The 16 MHz internal RC oscillator is factory-trimmed to offer 1% accuracy. The application can then select as system clock either the RC oscillator or an external 4-26 MHz clock source. This clock can be monitored for failure. If a failure is detected, the system automatically switches back to the internal RC oscillator and a software interrupt is generated (if enabled). This clock source is input to a PLL thus allowing to increase the frequency up to 216 MHz. Similarly, full interrupt management of the PLL clock entry is available when necessary (for example if an indirectly used external oscillator fails).

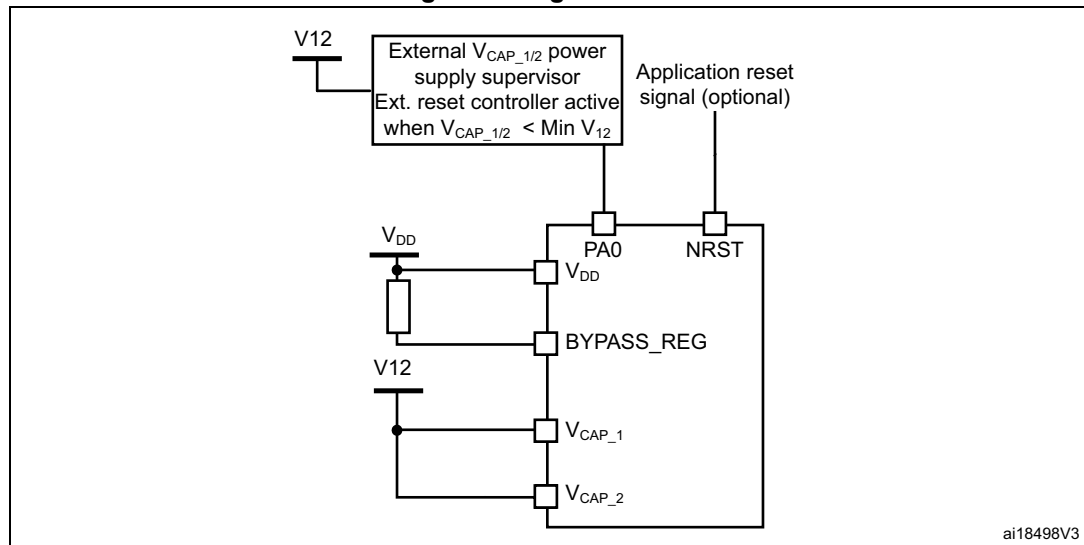
Several prescalers allow the configuration of the two AHB buses, the high-speed APB (APB2) and the low-speed APB (APB1) domains. The maximum frequency of the two AHB buses is 216 MHz while the maximum frequency of the high-speed APB domains is 108 MHz. The maximum allowed frequency of the low-speed APB domain is 54 MHz.

The devices embed two dedicated PLL (PLLI2S and PLLSAI) which allow to achieve audio class performance. In this case, the I²S and SAI master clock can generate all standard sampling frequencies from 8 kHz to 192 kHz.

In regulator OFF mode, the following features are no more supported:

- PA0 cannot be used as a GPIO pin since it allows to reset a part of the V_{12} logic power domain which is not reset by the NRST pin.
- As long as PA0 is kept low, the debug mode cannot be used under power-on reset. As a consequence, PA0 and NRST pins must be managed separately if the debug connection under reset or pre-reset is required.
- The over-drive and under-drive modes are not available.
- The Standby mode is not available.

Figure 8. Regulator OFF



The following conditions must be respected:

- V_{DD} should always be higher than V_{CAP_1} and V_{CAP_2} to avoid current injection between power domains.
- If the time for V_{CAP_1} and V_{CAP_2} to reach V_{12} minimum value is faster than the time for V_{DD} to reach 1.7 V, then PA0 should be kept low to cover both conditions: until V_{CAP_1} and V_{CAP_2} reach V_{12} minimum value and until V_{DD} reaches 1.7 V (see [Figure 9](#)).
- Otherwise, if the time for V_{CAP_1} and V_{CAP_2} to reach V_{12} minimum value is slower than the time for V_{DD} to reach 1.7 V, then PA0 could be asserted low externally (see [Figure 10](#)).
- If V_{CAP_1} and V_{CAP_2} go below V_{12} minimum value and V_{DD} is higher than 1.7 V, then a reset must be asserted on PA0 pin.

Note: The minimum value of V_{12} depends on the maximum frequency targeted in the application.

2.22.1 Advanced-control timers (TIM1, TIM8)

The advanced-control timers (TIM1, TIM8) can be seen as three-phase PWM generators multiplexed on 6 channels. They have complementary PWM outputs with programmable inserted dead times. They can also be considered as complete general-purpose timers. Their 4 independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge- or center-aligned modes)
- One-pulse mode output

If configured as standard 16-bit timers, they have the same features as the general-purpose TIMx timers. If configured as 16-bit PWM generators, they have full modulation capability (0-100%).

The advanced-control timer can work together with the TIMx timers via the Timer Link feature for synchronization or event chaining.

TIM1 and TIM8 support independent DMA request generation.

2.22.2 General-purpose timers (TIMx)

There are ten synchronizable general-purpose timers embedded in the STM32F74xxx devices (see [Table 6](#) for differences).

- **TIM2, TIM3, TIM4, TIM5**

The STM32F74xxx include 4 full-featured general-purpose timers: TIM2, TIM5, TIM3, and TIM4. The TIM2 and TIM5 timers are based on a 32-bit auto-reload up/downcounter and a 16-bit prescaler. The TIM3 and TIM4 timers are based on a 16-bit auto-reload up/downcounter and a 16-bit prescaler. They all feature 4 independent channels for input capture/output compare, PWM or one-pulse mode output. This gives up to 16 input capture/output compare/PWMs on the largest packages.

The TIM2, TIM3, TIM4, TIM5 general-purpose timers can work together, or with the other general-purpose timers and the advanced-control timers TIM1 and TIM8 via the Timer Link feature for synchronization or event chaining.

Any of these general-purpose timers can be used to generate PWM outputs.

TIM2, TIM3, TIM4, TIM5 all have independent DMA request generation. They are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 4 hall-effect sensors.

- **TIM9, TIM10, TIM11, TIM12, TIM13, and TIM14**

These timers are based on a 16-bit auto-reload upcounter and a 16-bit prescaler. TIM10, TIM11, TIM13, and TIM14 feature one independent channel, whereas TIM9 and TIM12 have two independent channels for input capture/output compare, PWM or one-pulse mode output. They can be synchronized with the TIM2, TIM3, TIM4, TIM5 full-featured general-purpose timers. They can also be used as simple time bases.

2.38 General-purpose input/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain, with or without pull-up or pull-down), as input (floating, with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. All GPIOs are high-current-capable and have speed selection to better manage internal noise, power consumption and electromagnetic emission.

The I/O configuration can be locked if needed by following a specific sequence in order to avoid spurious writing to the I/Os registers.

Fast I/O handling allowing maximum I/O toggling up to 108 MHz.

2.39 Analog-to-digital converters (ADCs)

Three 12-bit analog-to-digital converters are embedded and each ADC shares up to 16 external channels, performing conversions in the single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

Additional logic functions embedded in the ADC interface allow:

- Simultaneous sample and hold
- Interleaved sample and hold

The ADC can be served by the DMA controller. An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

To synchronize A/D conversion and timers, the ADCs could be triggered by any of TIM1, TIM2, TIM3, TIM4, TIM5, or TIM8 timer.

2.40 Temperature sensor

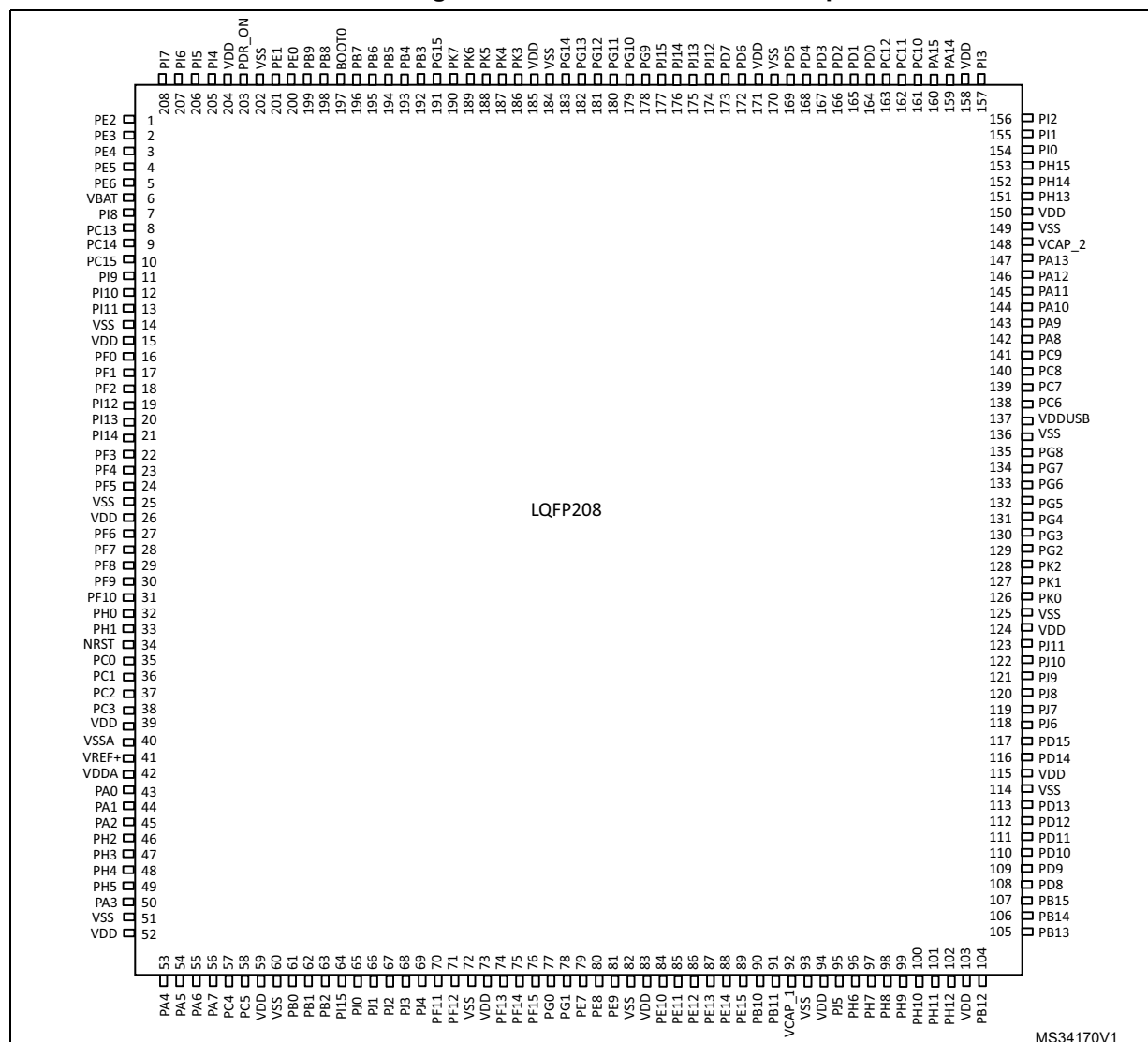
The temperature sensor has to generate a voltage that varies linearly with temperature. The conversion range is between 1.7 V and 3.6 V. The temperature sensor is internally connected to the same input channel as V_{BAT} , ADC1_IN18, which is used to convert the sensor output voltage into a digital value. When the temperature sensor and V_{BAT} conversion are enabled at the same time, only V_{BAT} conversion is performed.

As the offset of the temperature sensor varies from chip to chip due to process variation, the internal temperature sensor is mainly suitable for applications that detect temperature changes instead of absolute temperatures. If an accurate temperature reading is needed, then an external temperature sensor part should be used.

2.41 Digital-to-analog converter (DAC)

The two 12-bit buffered DAC channels can be used to convert two digital signals into two analog voltage signal outputs.

Figure 16. STM32F74xBx LQFP208 pinout



1. The above figure shows the package top view.

Table 9. Legend/abbreviations used in the pinout table

Name	Abbreviation	Definition
Pin name	Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name	
Pin type	S	Supply pin
	I	Input only pin
	I/O	Input / output pin
I/O structure	FT	5 V tolerant I/O
	TTa	3.3 V tolerant I/O directly connected to ADC
	B	Dedicated BOOT pin
	RST	Bidirectional reset pin with weak pull-up resistor
Notes	Unless otherwise specified by a note, all I/Os are set as floating inputs during and after reset	
Alternate functions	Functions selected through GPIOx_AFR registers	
Additional functions	Functions directly selected/enabled through peripheral registers	

Table 10. STM32F745xx and STM32F746xx pin and ball definition

Pin Number								Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
LQFP100	TFBGA100	WLCSP143	LQFP144	UFBGA176	LQFP176	LQFP208	TFBGA216						
1	A3	D8	1	A2	1	1	A3	PE2	I/O	FT	-	TRACECLK, SPI4_SCK, SAI1_MCLK_A, QUADSPI_BK1_IO2, ETH_MII_TXD3, FMC_A23, EVENTOUT	-
2	B3	C10	2	A1	2	2	A2	PE3	I/O	FT	-	TRACED0, SAI1_SD_B, FMC_A19, EVENTOUT	-
3	C3	B11	3	B1	3	3	A1	PE4	I/O	FT	-	TRACED1, SPI4_NSS, SAI1_FS_A, FMC_A20, DCMI_D4, LCD_B0, EVENTOUT	-



Table 12. STM32F745xx and STM32F746xx alternate function mapping (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	TIM1/2	TIM3/4/5	TIM8/9/10/ 11/LPTIM 1/CEC	I2C1/2/3/ 4/CEC	SPI1/2/3/ 4/5/6	SPI3/ SAI1	SPI2/3/U SART1/2/ 3/UART5/ SPDIFRX	SAI2/US ART6/UA RT4/5/7/8 /SPDIFR X	CAN1/2/T IM12/13/ 14/QUAD SPI/LCD	SAI2/QU ADSPI/O TG2_HS/ OTG1_FS	ETH/ OTG1_FS	FMC/SD MMC1/O TG2_FS	DCMI	LCD	SYS
Port G	PG11	-	-	-	-	-	-	-	SPDIFRX _IN0	-	-	-	ETH_MII_ TX_EN/E TH_RMII_ TX_EN	-	DCMI_D 3	LCD_B3	EVEN TOUT
	PG12	-	-	-	LPTIM1_I N1	-	SPI6_MI SO	-	SPDIFRX _IN1	USART6 _RTS	LCD_B4	-	-	FMC_NE 4	-	LCD_B1	EVEN TOUT
	PG13	TRACE D0	-	-	LPTIM1_ OUT	-	SPI6_SC K	-	-	USART6 _CTS	-	-	ETH_MII_ TXD0/ET H_RMII_T XD0	FMC_A2 4	-	LCD_R0	EVEN TOUT
	PG14	TRACE D1	-	-	LPTIM1_E TR	-	SPI6_M OSI	-	-	USART6 _TX	QUADSP I_BK2_IO 3	-	ETH_MII_ TXD1/ET H_RMII_T XD1	FMC_A2 5	-	LCD_B0	EVEN TOUT
	PG15	-	-	-	-	-	-	-	-	USART6 _CTS	-	-	-	FMC_SD NCAS	DCMI_D 13	-	EVEN TOUT
Port H	PH0	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVEN TOUT
	PH1	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVEN TOUT
	PH2	-	-	-	LPTIM1_I N2	-	-	-	-	-	QUADSP I_BK2_IO 0	SAI2_SC K_B	ETH_MII_ CRS	FMC_SD CKE0	-	LCD_R0	EVEN TOUT
	PH3	-	-	-	-	-	-	-	-	-	QUADSP I_BK2_IO 1	SAI2_MC K_B	ETH_MII_ COL	FMC_SD NE0	-	LCD_R1	EVEN TOUT
	PH4	-	-	-	-	I2C2_SC L	-	-	-	-	-	OTG_HS_ ULPI_NX T	-	-	-	-	EVEN TOUT
	PH5	-	-	-	-	I2C2_SD A	SPI5_NS S	-	-	-	-	-	-	FMC_SD NWE	-	-	EVEN TOUT
	PH6	-	-	-	-	I2C2_SM BA	SPI5_SC K	-	-	-	TIM12_C H1	-	ETH_MII_ RXD2	FMC_SD NE1	DCMI_D 8	-	EVEN TOUT
	PH7	-	-	-	-	I2C3_SC L	SPI5_MI SO	-	-	-	-	-	ETH_MII_ RXD3	FMC_SD CKE1	DCMI_D 9	-	EVEN TOUT

Table 12. STM32F745xx and STM32F746xx alternate function mapping (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	TIM1/2	TIM3/4/5	TIM8/9/10/ 11/LPTIM 1/CEC	I2C1/2/3/ 4/CEC	SPI1/2/3/ 4/5/6	SPI3/ SAI1	SPI2/3/U SART1/2/ 3/UART5/ SPDIFRX	SAI2/US ART6/UA RT4/5/7/8 /SPDIFR X	CAN1/2/T IM12/13/ 14/QUAD SPI/LCD	SAI2/QU ADSPI/O TG2_HS/ OTG1_FS	ETH/ OTG1_FS	FMC/SD MMC1/O TG2_FS	DCMI	LCD	SYS
Port J	PJ7	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G0	EVEN TOUT
	PJ8	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G1	EVEN TOUT
	PJ9	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G2	EVEN TOUT
	PJ10	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G3	EVEN TOUT
	PJ11	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G4	EVEN TOUT
	PJ12	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B0	EVEN TOUT
	PJ13	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B1	EVEN TOUT
	PJ14	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B2	EVEN TOUT
	PJ15	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B3	EVEN TOUT

4 Memory mapping

The memory map is shown in [Figure 19](#).

Figure 19. Memory map

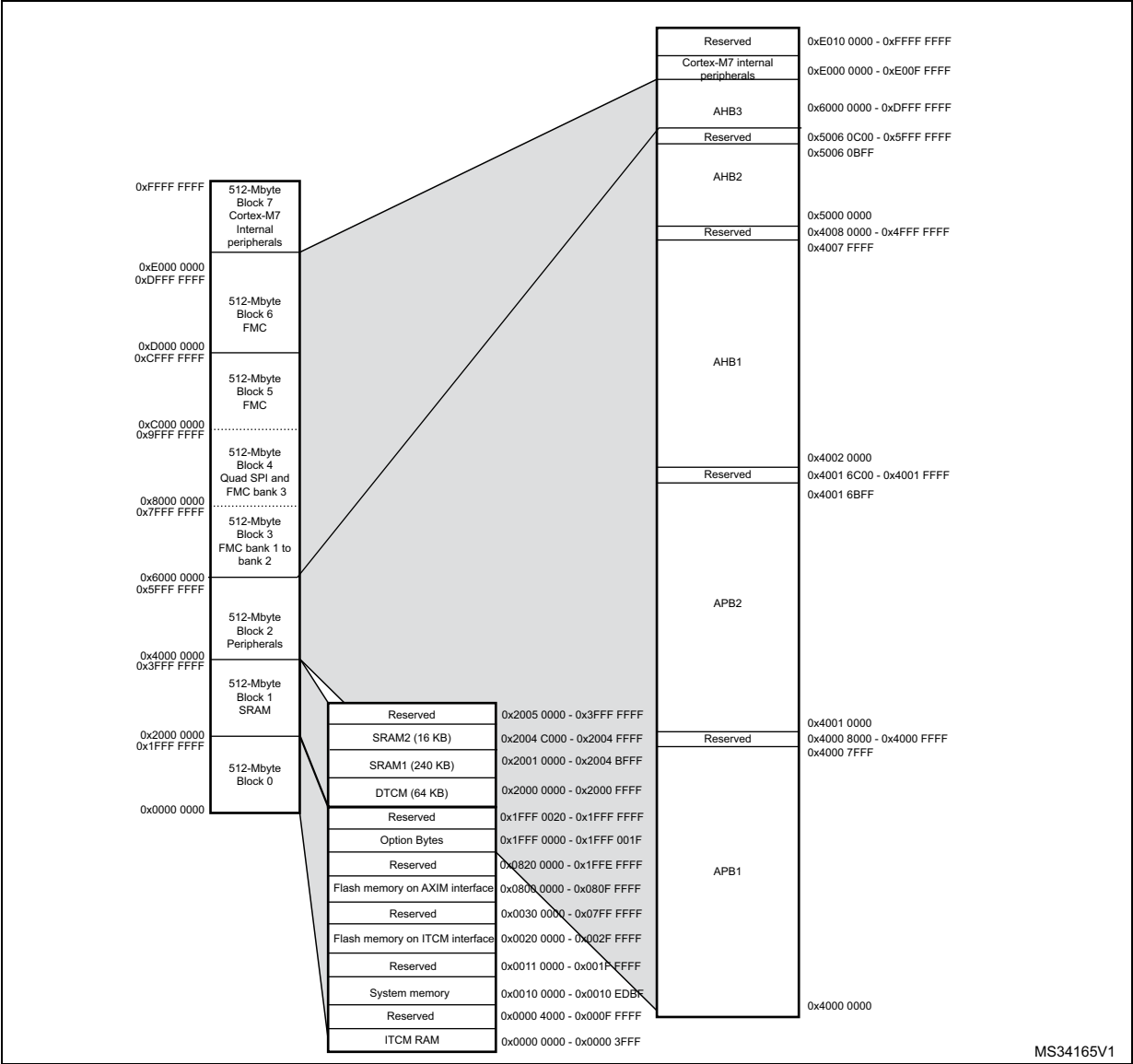


Table 13. STM32F745xx and STM32F746xx register boundary addresses (continued)

Bus	Boundary address	Peripheral
	0x4000 8000 - 0x4000 FFFF	Reserved
APB1	0x4000 7C00 - 0x4000 7FFF	UART8
	0x4000 7800 - 0x4000 7BFF	UART7
	0x4000 7400 - 0x4000 77FF	DAC
	0x4000 7000 - 0x4000 73FF	PWR
	0x4000 6C00 - 0x4000 6FFF	HDMI-CEC
	0x4000 6800 - 0x4000 6BFF	CAN2
	0x4000 6400 - 0x4000 67FF	CAN1
	0x4000 6000 - 0x4000 63FF	I2C4
	0x4000 5C00 - 0x4000 5FFF	I2C3
	0x4000 5800 - 0x4000 5BFF	I2C2
	0x4000 5400 - 0x4000 57FF	I2C1
	0x4000 5000 - 0x4000 53FF	UART5
	0x4000 4C00 - 0x4000 4FFF	UART4
	0x4000 4800 - 0x4000 4BFF	USART3
	0x4000 4400 - 0x4000 47FF	USART2
	0x4000 4000 - 0x4000 43FF	SPDIFRX
	0x4000 3C00 - 0x4000 3FFF	SPI3 / I2S3
	0x4000 3800 - 0x4000 3BFF	SPI2 / I2S2
	0x4000 3400 - 0x4000 37FF	Reserved
	0x4000 3000 - 0x4000 33FF	IWDG
	0x4000 2C00 - 0x4000 2FFF	WWDG
	0x4000 2800 - 0x4000 2BFF	RTC & BKP Registers
	0x4000 2400 - 0x4000 27FF	LPTIM1
	0x4000 2000 - 0x4000 23FF	TIM14
	0x4000 1C00 - 0x4000 1FFF	TIM13
	0x4000 1800 - 0x4000 1BFF	TIM12
	0x4000 1400 - 0x4000 17FF	TIM7
	0x4000 1000 - 0x4000 13FF	TIM6
	0x4000 0C00 - 0x4000 0FFF	TIM5
	0x4000 0800 - 0x4000 0BFF	TIM4
	0x4000 0400 - 0x4000 07FF	TIM3
	0x4000 0000 - 0x4000 03FF	TIM2

Equation 2

Equation 2 allows to calculate the increment step (INCSTEP):

$$\text{INCSTEP} = \text{round}[(2^{15} - 1) \times \text{md} \times \text{PLLN}] / (100 \times 5 \times \text{MODEPER})$$

$f_{\text{VCO_OUT}}$ must be expressed in MHz.

With a modulation depth (md) = $\pm 2\%$ (4 % peak to peak), and PLLN = 240 (in MHz):

$$\text{INCSTEP} = \text{round}[(2^{15} - 1) \times 2 \times 240] / (100 \times 5 \times 250) = 126 \text{md}(\text{quantitized})\%$$

An amplitude quantization error may be generated because the linear modulation profile is obtained by taking the quantized values (rounded to the nearest integer) of MODPER and INCSTEP. As a result, the achieved modulation depth is quantized. The percentage quantized modulation depth is given by the following formula:

$$\text{md}_{\text{quantized}}\% = (\text{MODEPER} \times \text{INCSTEP} \times 100 \times 5) / ((2^{15} - 1) \times \text{PLLN})$$

As a result:

$$\text{md}_{\text{quantized}}\% = (250 \times 126 \times 100 \times 5) / ((2^{15} - 1) \times 240) = 2.002\%(\text{peak})$$

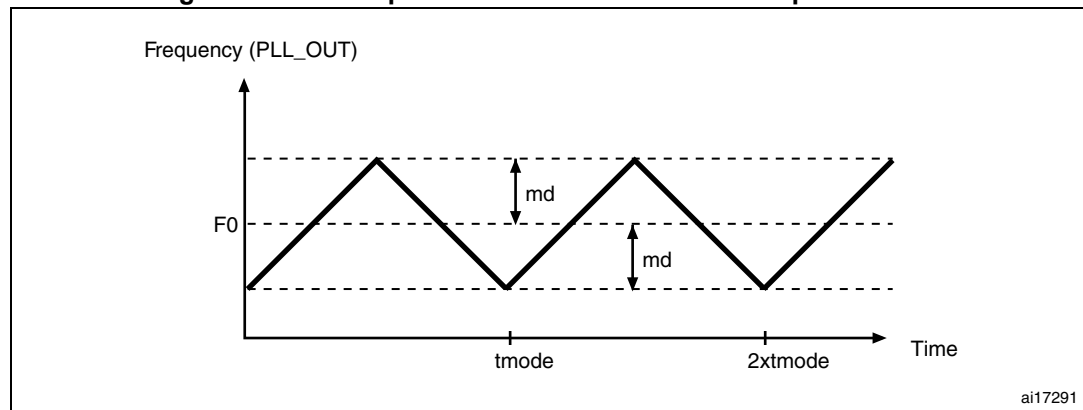
Figure 36 and Figure 37 show the main PLL output clock waveforms in center spread and down spread modes, where:

F0 is $f_{\text{PLL_OUT}}$ nominal.

T_{mode} is the modulation period.

md is the modulation depth.

Figure 36. PLL output clock waveforms in center spread mode



The 20mA output drive requirement in Fast-mode Plus is not supported. This limits the maximum load C_{load} supported in Fm+, which is given by these formulas:

- $T_r(SDA/SCL) = 0.8473 \times R_p \times C_{load}$
- $R_p(\min) = (V_{DD} - V_{OL}(\max)) / I_{OL}(\max)$

Where R_p is the I2C lines pull-up. Refer to [Section 5.3.17: I/O port characteristics](#) for the I2C I/Os characteristics.

All I²C SDA and SCL I/Os embed an analog filter. Refer to the table below for the analog filter characteristics:

Table 75. I2C analog filter characteristics⁽¹⁾

Symbol	Parameter	Min	Max	Unit
t_{AF}	Maximum pulse width of spikes that are suppressed by the analog filter	50 ⁽²⁾	150 ⁽³⁾	ns

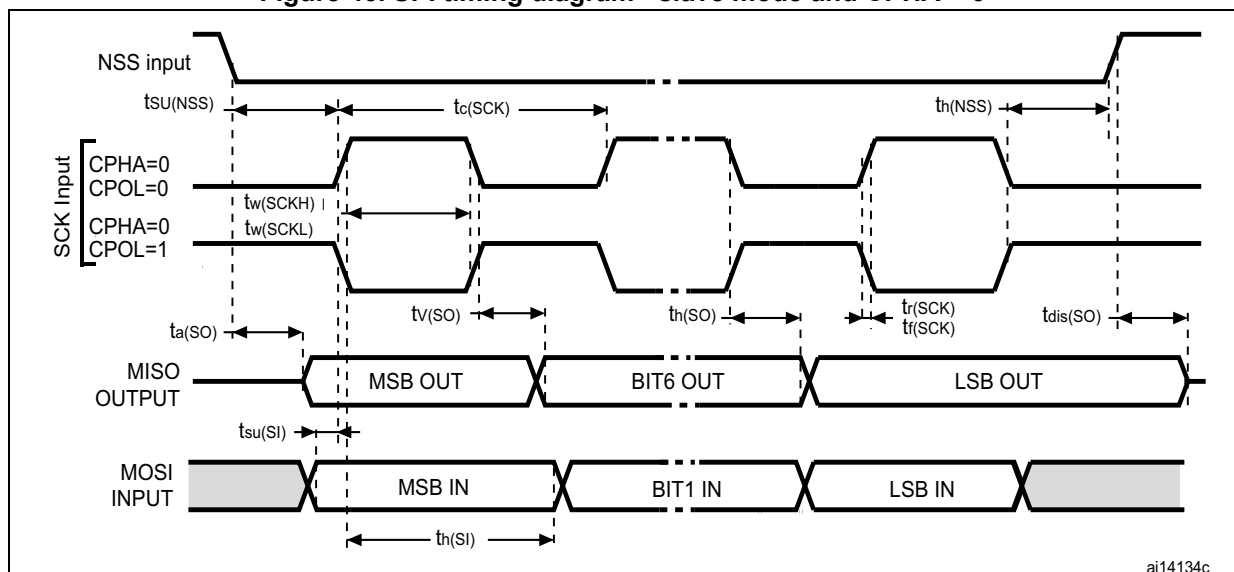
1. Guaranteed by characterization results.
2. Spikes with widths below $t_{AF(\min)}$ are filtered.
3. Spikes with widths above $t_{AF(\max)}$ are not filtered

Table 76. SPI dynamic characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
tsu(MI)	Data input setup time	Master mode	5.5	-	-	ns
tsu(SI)		Slave mode	4	-	-	
th(MI)	Data input hold time	Master mode	4	-	-	
th(SI)		Slave mode	2	-	-	
ta(SO)	Data output access time	Slave mode	7	-	21	
tdis(SO)	Data output disable time	Slave mode	5	-	12	
tv(SO)	Data output valid time	Slave mode $2.7 \leq VDD \leq 3.6V$	-	6.5	10	
		Slave mode $1.71 \leq VDD \leq 3.6V$	-	6.5	13	
tv(MO)		Master mode	-	2	4	
th(SO)	Data output hold time	Slave mode $1.71 \leq VDD \leq 3.6V$	5.5	-	-	
th(MO)		Master mode	0	-	-	

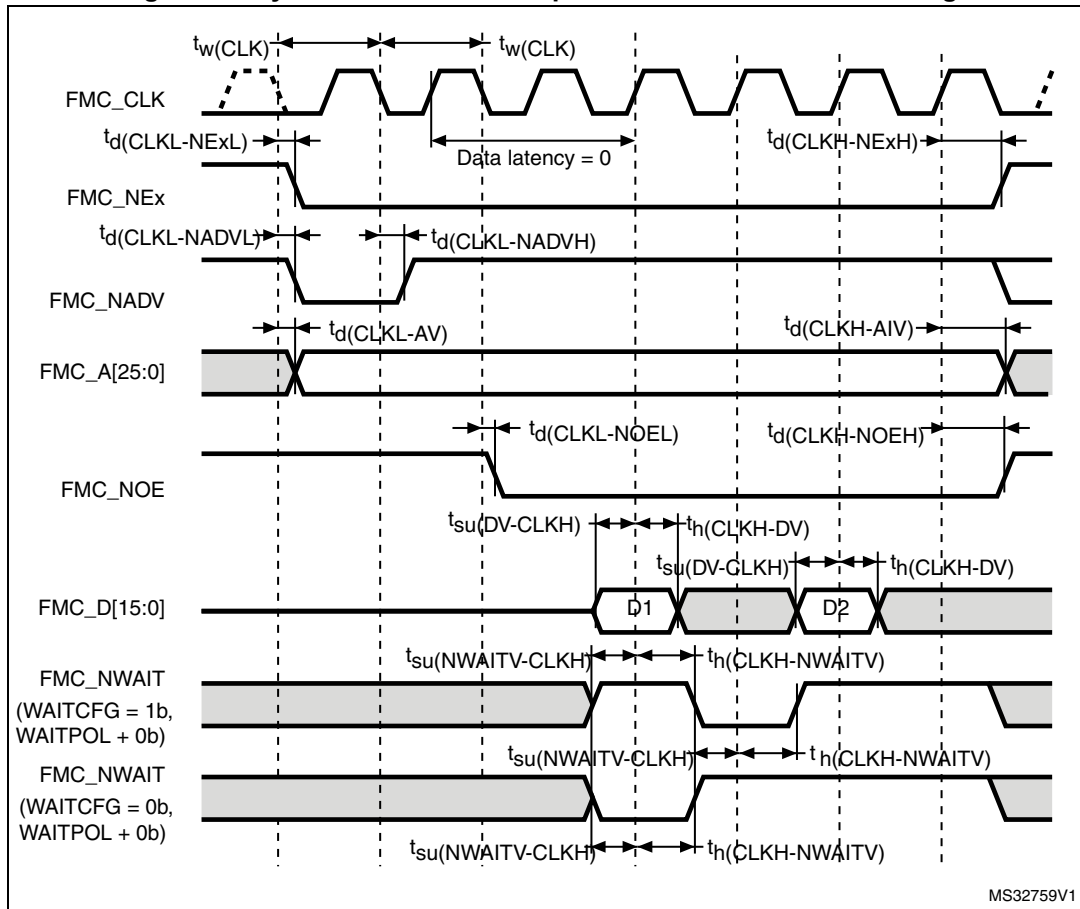
1. Guaranteed by characterization results.
2. Excepting SPI1 with SCK IO pin mapped on PA5. In this configuration, Maximum achievable frequency is 40MHz.
3. Maximum Frequency of Slave Transmitter is determined by sum of Tv(SO) and Tsu(MI) intervals which has to fit into SCK level phase preceding the SCK sampling edge. This value can be achieved when it communicates with a Master having Tsu(MI)=0 while signal Duty(SCK)=50%.

Figure 46. SPI timing diagram - slave mode and CPHA = 0



ai14134c

Figure 64. Synchronous non-multiplexed NOR/PSRAM read timings

Table 98. Synchronous non-multiplexed NOR/PSRAM read timings⁽¹⁾

Symbol	Parameter	Min	Max	Unit
$t_w(\text{CLK})$	FMC_CLK period	$2T_{\text{HCLK}}-1$	-	ns
$t_{\text{d}}(\text{CLKL-NExL})$	FMC_CLK low to FMC_NEx low ($x=0..2$)	-	2.5	
$t_{\text{d}}(\text{CLKH-NExH})$	FMC_CLK high to FMC_NEx high ($x=0..2$)	$T_{\text{HCLK}}+0.5$	-	
$t_{\text{d}}(\text{CLKL-NADV})$	FMC_CLK low to FMC_NADV low	-	0	
$t_{\text{d}}(\text{CLKL-NADVH})$	FMC_CLK low to FMC_NADV high	0	-	
$t_{\text{d}}(\text{CLKL-AV})$	FMC_CLK low to FMC_Ax valid ($x=16..25$)	-	2.5	
$t_{\text{d}}(\text{CLKH-AIV})$	FMC_CLK high to FMC_Ax invalid ($x=16..25$)	T_{HCLK}	-	
$t_{\text{d}}(\text{CLKL-NOEL})$	FMC_CLK low to FMC_NOE low	-	2	
$t_{\text{d}}(\text{CLKH-NOEH})$	FMC_CLK high to FMC_NOE high	$T_{\text{HCLK}}+0.5$	-	
$t_{\text{su}}(\text{DV-CLKH})$	FMC_D[15:0] valid data before FMC_CLK high	1.5	-	
$t_{\text{h}}(\text{CLKH-DV})$	FMC_D[15:0] valid data after FMC_CLK high	1	-	
$t_{\text{su}}(\text{NWAITV-CLKH})$	FMC_NWAIT valid before FMC_CLK high	2	-	
$t_{\text{h}}(\text{CLKH-NWAITV})$	FMC_NWAIT valid after FMC_CLK high	3.5	-	

Figure 66. NAND controller waveforms for read access

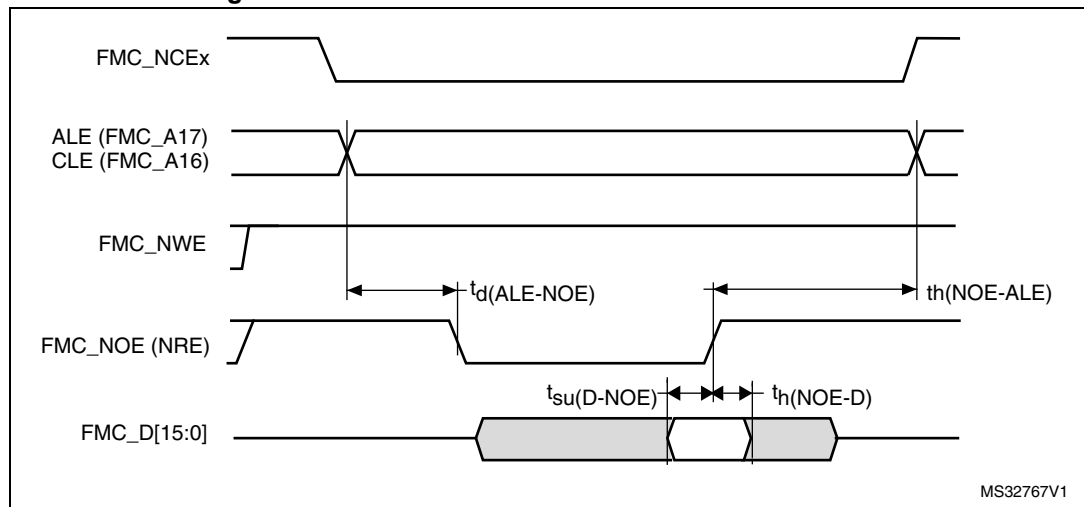


Figure 67. NAND controller waveforms for write access

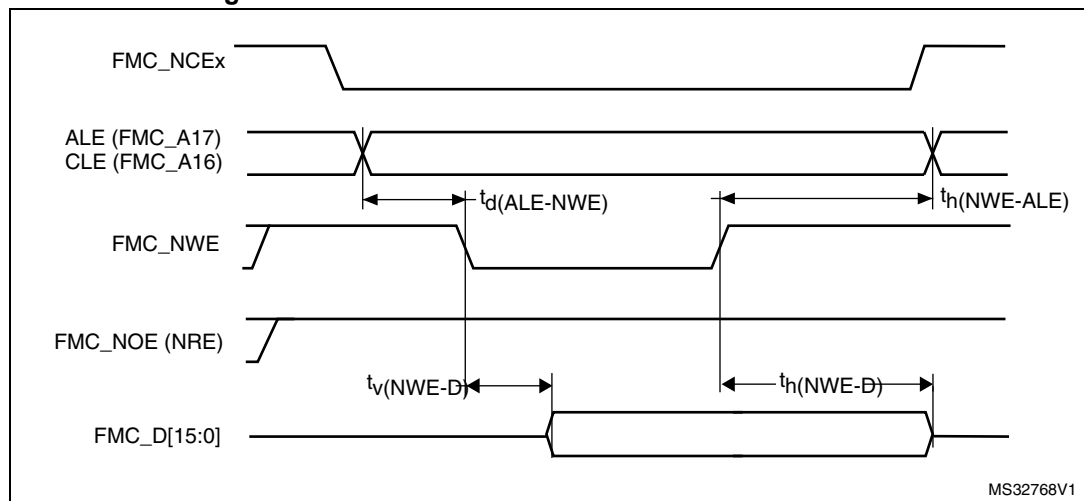
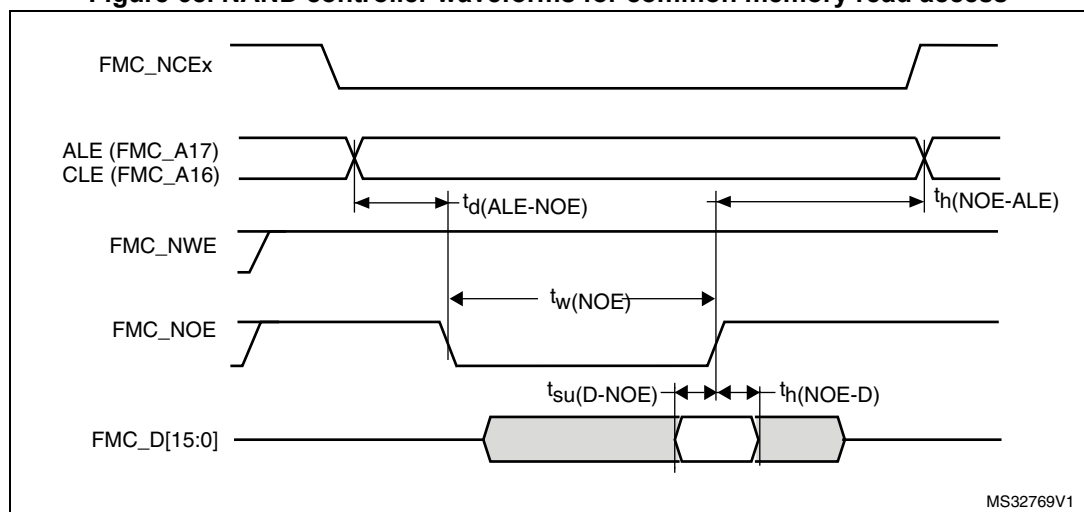


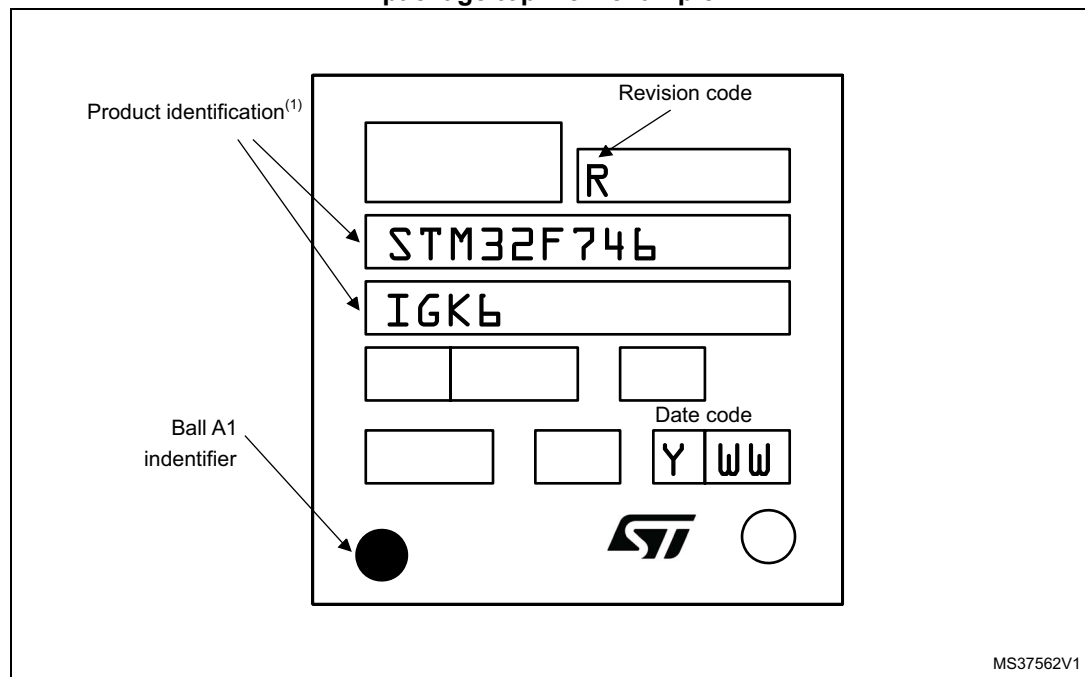
Figure 68. NAND controller waveforms for common memory read access



Marking of engineering samples

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

Figure 99. UFBGA 176+25, 10 × 10 × 0.6 mm ultra thin fine-pitch ball grid array package top view example



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