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Understanding <u>Embedded - FPGAs (Field</u> <u>Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	1164
Number of Logic Elements/Cells	10476
Total RAM Bits	368640
Number of I/O	158
Number of Gates	500000
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xa3s500e-4pqg208i

Email: info@E-XFL.COM

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Key Feature Differences from Commercial XC Devices

- AEC-Q100 device qualification and full production part approval process (PPAP) documentation support available in both extended temperature I- and Q-Grades
- Guaranteed to meet full electrical specification over the $T_J = -40^{\circ}$ C to +125°C temperature range (Q-Grade)
- XA Spartan-3E devices are available in the -4 speed grade only.
- PCI-66 is not supported in the XA Spartan-3E FPGA product line.
- The readback feature is not supported in the XA

Table 1: Summary of XA Spartan-3E FPGA Attributes

Spartan-3E FPGA product line.

- XA Spartan-3E devices are available in Step 1 only.
- JTAG configuration frequency reduced from 30 MHz to 25 MHz.
- Platform Flash is not supported within the XA family.
- XA Spartan-3E devices are available in Pb-free packaging only.
- MultiBoot is not supported in XA versions of this product.
- The XA Spartan-3E device must be power cycled prior to reconfiguration.

		Equivalent	(CLB Array (One CLB = Four Slices)			Block				Maximum	
Device	System Gates	Logic Cells	Rows	Columns	Total CLBs	Total Slices	Distributed RAM bits ⁽¹⁾	RAM bits ⁽¹⁾	Dedicated Multipliers	DCMs	Maximum User I/O	Differential I/O Pairs
XA3S100E	100K	2,160	22	16	240	960	15K	72K	4	2	108	40
XA3S250E	250K	5,508	34	26	612	2,448	38K	216K	12	4	172	68
XA3S500E	500K	10,476	46	34	1,164	4,656	73K	360K	20	4	190	77
XA3S1200E	1200K	19,512	60	46	2,168	8,672	136K	504K	28	8	304	124
XA3S1600E	1600K	33,192	76	58	3,688	14,752	231K	648K	36	8	376	156

Notes:

1. By convention, one Kb is equivalent to 1,024 bits.

Architectural Overview

The XA Spartan-3E family architecture consists of five fundamental programmable functional elements:

- Configurable Logic Blocks (CLBs) contain flexible Look-Up Tables (LUTs) that implement logic plus storage elements used as flip-flops or latches. CLBs perform a wide variety of logical functions as well as store data.
- Input/Output Blocks (IOBs) control the flow of data between the I/O pins and the internal logic of the device. Each IOB supports bidirectional data flow plus 3-state operation. Supports a variety of signal standards, including four high-performance differential standards. Double Data-Rate (DDR) registers are included.
- **Block RAM** provides data storage in the form of 18-Kbit dual-port blocks.
- **Multiplier Blocks** accept two 18-bit binary numbers as inputs and calculate the product.

 Digital Clock Manager (DCM) Blocks provide self-calibrating, fully digital solutions for distributing, delaying, multiplying, dividing, and phase-shifting clock signals.

These elements are organized as shown in Figure 1. A ring of IOBs surrounds a regular array of CLBs. Each device has two columns of block RAM except for the XA3S100E, which has one column. Each RAM column consists of several 18-Kbit RAM blocks. Each block RAM is associated with a dedicated multiplier. The DCMs are positioned in the center with two at the top and two at the bottom of the device. The XA3S100E has only one DCM at the top and bottom, while the XA3S1200E and XA3S1600E add two DCMs in the middle of the left and right sides.

The XA Spartan-3E family features a rich network of traces that interconnect all five functional elements, transmitting signals among them. Each functional element has an associated switch matrix that permits multiple connections to the routing.



Notes:

1. The XA3S1200E and XA3S1600E have two additional DCMs on both the left and right sides as indicated by the dashed lines. The XA3S100E has only one DCM at the top and one at the bottom.

Figure 1: XA Spartan-3E Family Architecture

Configuration

XA Spartan-3E FPGAs are programmed by loading configuration data into robust, reprogrammable, static CMOS configuration latches (CCLs) that collectively control all functional elements and routing resources. The FPGA's configuration data is stored externally in a PROM or some other non-volatile medium, either on or off the board. After applying power, the configuration data is written to the FPGA using any of five different modes:

- Serial Peripheral Interface (SPI) from an industry-standard SPI serial Flash
- Byte Peripheral Interface (BPI) Up or Down from an industry-standard x8 or x8/x16 parallel NOR Flash
- Slave Serial, typically downloaded from a processor
- Slave Parallel, typically downloaded from a processor
- Boundary Scan (JTAG), typically downloaded from a processor or system tester.

I/O Capabilities

The XA Spartan-3E FPGA SelectIO interface supports many popular single-ended and differential standards. Table 2 shows the number of user I/Os as well as the number of differential I/O pairs available for each device/package combination.

XA Spartan-3E FPGAs support the following single-ended standards:

- 3.3V low-voltage TTL (LVTTL)
- Low-voltage CMOS (LVCMOS) at 3.3V, 2.5V, 1.8V, 1.5V, or 1.2V
- 3V PCI at 33 MHz
- HSTL I and III at 1.8V, commonly used in memory applications
- SSTL I at 1.8V and 2.5V, commonly used for memory applications

XA Spartan-3E FPGAs support the following differential standards:

- LVDS
- Bus LVDS
- mini-LVDS
- RSDS

Table 2: Available User I/Os and Differential (Diff) I/O Pairs

Package	VQC	G100	CPG	a132	TQC	à144	PQG	208	FTG	i256	FGG	3400	FGG	i484
Size (mm)	16 :	c 16	8	x 8	22 >	c 22	28 3	c 28	17 >	c 17	21 :	x 21	23 x	23
Device	User	Diff	User	Diff	User	Diff	User	Diff	User	Diff	User	Diff	User	Diff
XA3S100E	66 (7)	30 (2)	83 (11)	35 (2)	108 (28)	40 (4)	-	-	-	-	-	-	-	-
XA3S250E	66 (7)	30 (2)	92 (7)	41 <i>(2)</i>	108 (28)	40 (4)	158 (32)	65 (5)	172 (40)	68 (8)	-	-	-	-
XA3S500E	-	-	92 (7)	41 <i>(2)</i>	-	-	158 (32)	65 (5)	190 (41)	77 (8)	-	-	-	-
XA3S1200E	-	-	-	-	-	-	-	-	190 (40)	77 (8)	304 (72)	124 (20)	-	-
XA3S1600E	-	-	-	-	-	-	-	-	-	-	304 (72)	124 (20)	376 (82)	156 (21)

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Differential HSTL (1.8V, Types I and III)

2.5V LVPECL inputs

Differential SSTL (2.5V and 1.8V, Type I)

Notes:

1. All XA Spartan-3E devices provided in the same package are pin-compatible as further described in Module 4: Pinout Descriptions of DS312.

2. The number shown in **bold** indicates the maximum number of I/O and input-only pins. The number shown in (*italics*) indicates the number of input-only pins.



Package Marking

Figure 2 provides a top marking example for XA Spartan-3E FPGAs in the quad-flat packages. Figure 3 shows the top marking for XA Spartan-3E FPGAs in BGA packages except the 132-ball chip-scale package (CPG132). The markings for the BGA packages are nearly identical to those

for the quad-flat packages, except that the marking is rotated with respect to the ball A1 indicator. Figure 4 shows the top marking for XA Spartan-3E FPGAs in the CPG132 package.





Figure 2: XA Spartan-3E FPGA QFP Package Marking Example



Figure 3: XA Spartan-3E FPGA BGA Package Marking Example



Figure 4: XA Spartan-3E FPGA CPG132 Package Marking Example

DC Specifications

Table 6: General Recommended Operating Conditions

Symbol	Descriptio	n	Min	Nominal	Мах	Units
TJ	Junction temperature	I-Grade	-40	25	100	°C
		Q-Grade		25	125	°C
V _{CCINT}	Internal supply voltage		1.140	1.200	1.260	V
V _{CCO} ⁽¹⁾	Output driver supply voltage	1.100	-	3.465	V	
V _{CCAUX}	Auxiliary supply voltage	2.375	2.500	2.625	V	
$\Delta V_{CCAUX}^{(2)}$	Voltage variance on V_{CCAUX} whe	en using a DCM	-	-	10	mV/ms
V _{IN} ^(3,4,5,6)	Input voltage extremes to avoid turning on I/O protection diodes	I/O, Input-only, and Dual-Purpose pins ⁽³⁾	-0.5	-	V _{CCO} + 0.5	V
		Dedicated pins ⁽⁴⁾	-0.5	_	V _{CCAUX} + 0.5	V
T _{IN}	Input signal transition time ⁽⁷⁾	·	-	-	500	ns

Notes:

- 1. This V_{CCO} range spans the lowest and highest operating voltages for all supported I/O standards. Table 9 lists the recommended V_{CCO} range specific to each of the single-ended I/O standards, and Table 11 lists that specific to the differential standards.
- 2. Only during DCM operation is it recommended that the rate of change of V_{CCAUX} not exceed 10 mV/ms.
- Each of the User I/O and Dual-Purpose pins is associated with one of the four banks' V_{CCO} rails. Meeting the V_{IN} limit ensures that the internal diode junctions that exist between these pins and their associated V_{CCO} and GND rails do not turn on. See Absolute Maximum Ratings in <u>DS312</u>).
- 4. All Dedicated pins (PROG_B, DONE, TCK, TDI, TDO, and TMS) draw power from the V_{CCAUX} rail (2.5V). Meeting the V_{IN} max limit ensures that the internal diode junctions that exist between each of these pins and the V_{CCAUX} and GND rails do not turn on.
- 5. Input voltages outside the recommended range is permissible provided that the I_{IK} input clamp diode rating is met and no more than 100 pins exceed the range simultaneously. See Absolute Maximum Ratings in <u>DS312</u>).
- 6. See XAPP459, "Eliminating I/O Coupling Effects when Interfacing Large-Swing Single-Ended Signals to User I/O Pins."
- 7. Measured between 10% and 90% V_{CCO} . Follow Signal Integrity recommendations.

General DC Characteristics for I/O Pins

Table 7: General DC Characteristics of User I/O, Dual-Purpose, and Dedicated Pins

Symbol	Description	Test Conditions	Min	Тур	Мах	Units
ι _L	Leakage current at User I/O, Input-only, Dual-Purpose, and Dedicated pins	Driver is in a high-impedance state, $V_{IN} = 0V$ or V_{CCO} max, sample-tested	-10	_	+10	μA
I _{RPU} ⁽²⁾	Current through pull-up resistor at	$V_{IN} = 0V, V_{CCO} = 3.3V$	-0.36	-	-1.24	mA
	User I/O, Dual-Purpose, Input-only, and Dedicated pins	$V_{IN} = 0V, V_{CCO} = 2.5V$	-0.22	-	-0.80	mA
		$V_{IN} = 0V, V_{CCO} = 1.8V$	-0.10	-	-0.42	mA
		$V_{IN} = 0V, V_{CCO} = 1.5V$	-0.06	-	-0.27	mA
		$V_{IN} = 0V, V_{CCO} = 1.2V$	-0.04	-	-0.22	mA
R _{PU} ⁽²⁾	Equivalent pull-up resistor value at	$V_{IN} = 0V, V_{CCO} = 3.0V \text{ to } 3.465V$	2.4	-	10.8	kΩ
	and Dedicated pins (based on I _{BPU}	$V_{IN} = 0V, V_{CCO} = 2.3V \text{ to } 2.7V$	2.7	_	11.8	kΩ
	per Note 2)	$V_{IN} = 0V, V_{CCO} = 1.7V \text{ to } 1.9V$	4.3	_	20.2	kΩ
		V _{IN} = 0V, V _{CCO} =1.4V to 1.6V	5.0	_	25.9	kΩ
		$V_{\rm IN} = 0$ V, $V_{\rm CCO} = 1.14$ V to 1.26V	5.5	_	32.0	kΩ

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Single-Ended I/O Standards

Table	<u>9</u> :	Recommended O	perating	Conditions for	or User I/Os	Usina Sin	ale-Ended	Standards
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IOSTANDARD	V _{CC}	_{CO} for Drive	rs ⁽²⁾		V _{REF}		V _{IL}	V _{IH}
Attribute	Min (V)	Nom (V)	Max (V)	Min (V)	Nom (V)	Max (V)	Max (V)	Min (V)
LVTTL	3.0	3.3	3.465				0.8	2.0
LVCMOS33 ⁽⁴⁾	3.0	3.3	3.465				0.8	2.0
LVCMOS25 ^(4,5)	2.3	2.5	2.7	0.7		0.7	1.7	
LVCMOS18	1.65	1.8	1.95	V _{RE} the	_{EF} is not use se I/O stand	d for ards	0.4	0.8
LVCMOS15	1.4	1.5	1.6				0.4	0.8
LVCMOS12	1.1	1.2	1.3				0.4	0.7
PCI33_3	3.0	3.3	3.465				0.3 * V _{CCO}	0.5 * V _{CCO}
HSTL_I_18	1.7	1.8	1.9	0.8	0.9	1.1	V _{REF} - 0.1	V _{REF} + 0.1
HSTL_III_18	1.7	1.8	1.9	-	1.1	-	V _{REF} - 0.1	V _{REF} + 0.1
SSTL18_I	1.7	1.8	1.9	0.833 0.900 0.969			V _{REF} - 0.125	V _{REF} + 0.125
SSTL2_I	2.3	2.5	2.7	1.15	1.25	1.35	V _{REF} - 0.125	V _{REF} + 0.125

Notes:

- 1. Descriptions of the symbols used in this table are as follows:

 - $\label{eq:V_CCO} V_{CCO} \text{the supply voltage for output drivers} \\ V_{REF} \text{the reference voltage for setting the input switching threshold} \\ V_{IL} \text{the input voltage that indicates a Low logic level} \\ V_{IH} \text{the input voltage that indicates a High logic level} \\ \end{array}$
- 2. The V_{CCO} rails supply only output drivers, not input circuits.
- For device operation, the maximum signal voltage (V_{IH} max) may be as high as V_{IN} max. See Table 72 in DS312. З.
- There is approximately 100 mV of hysteresis on inputs using LVCMOS33 and LVCMOS25 I/O standards. 4.
- All Dedicated pins (PROG_B, DONE, TCK, TDI, TDO, and TMS) use the LVCMOS25 standard and draw power from the V_{CCAUX} rail (2.5V). 5. The Dual-Purpose configuration pins use the LVCMOS standard before the User mode. When using these pins as part of a standard 2.5V configuration interface, apply 2.5V to the V_{CCO} lines of Banks 0, 1, and 2 at power-on as well as throughout configuration.
- For information on PCI IP solutions, see www.xilinx.com/pci. 6.



Table 10: DC Characteristics of User I/Os UsingSingle-Ended Standards

		Te Cond	est itions	Logic Charac	: Level teristics
IOSTANDAR Attribute	D	I _{OL} (mA)	I _{OH} (mA)	V _{OL} Max (V)	V _{OH} Min (V)
LVTTL ⁽³⁾	2	2	-2	0.4	2.4
	4	4	-4		
	6	6	-6		
	8	8	-8		
	12	12	-12		
	16	16	-16		
LVCMOS33 ⁽³⁾	2	2	-2	0.4	V _{CCO} – 0.4
	4	4	-4		
	6	6	-6		
	8	8	-8		
	12	12	-12		
	16	16	-16		
LVCMOS25 ⁽³⁾	2	2	-2	0.4	V _{CCO} – 0.4
	4	4	-4		
	6	6	-6		
	8	8	-8		
	12	12	-12		
LVCMOS18 ⁽³⁾	2	2	-2	0.4	$V_{CCO} - 0.4$
	4	4	-4		
	6	6	-6		
	8	8	-8		
LVCMOS15 ⁽³⁾	2	2	-2	0.4	V _{CCO} – 0.4
	4	4	-4		
	6	6	-6		

Table 10: DC Characteristics of User I/Os Using Single-Ended Standards (Continued)

		Te Cond	est itions	Logic Level Characteristics			
IOSTANDAR Attribute	D	I _{OL} (mA)	I _{OH} (mA)	V _{OL} Max (V)	V _{OH} Min (V)		
LVCMOS12 ⁽³⁾	2	2	-2	0.4	V _{CCO} - 0.4		
PCI33_3 ⁽⁴⁾		1.5	-0.5	10% V _{CCO}	90% V _{CCO}		
HSTL_I_18		8	-8	0.4	V _{CCO} - 0.4		
HSTL_III_18		24	-8	0.4	V _{CCO} - 0.4		
SSTL18_I		6.7	-6.7	V _{TT} – 0.475	V _{TT} + 0.475		
SSTL2_I		8.1	-8.1	V _{TT} – 0.61	V _{TT} + 0.61		

Notes:

1. The numbers in this table are based on the conditions set forth in Table 6 and Table 9.

2. Descriptions of the symbols used in this table are as follows: I_{OL} — the output current condition under which V_{OL} is tested I_{OH} — the output current condition under which V_{OH} is tested V_{OL} — the output voltage that indicates a Low logic level V_{OH} — the output voltage that indicates a High logic level V_{CCO} — the supply voltage for output drivers V_{TT} — the voltage applied to a resistor termination

- 3. For the LVCMOS and LVTTL standards: the same $\rm V_{OL}$ and $\rm V_{OH}$ limits apply for both the Fast and Slow slew attributes.
- Tested according to the relevant PCI specifications. For information on PCI IP solutions, see <u>www.xilinx.com/pci.</u>

Table	15:	Setup and H	Hold Times	for the	IOB Input	Path
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			IFD_ DELAY		-4 Speed Grade	
Symbol	Description	Conditions	VALUE	Device	Min	Units
Setup Tim	es					
T _{IOPICK}	Time from the setup of data at the Input pin to the active transition at the ICLK input of the Input Flip-Flop (IFF). No Input Delay is programmed.	LVCMOS25 ⁽²⁾ , IFD_DELAY_VALUE = 0	0	All	2.12	ns
T _{IOPICKD}	Time from the setup of data at the Input	LVCMOS25 ⁽²⁾ ,	2	XA3S100E	6.49	ns
	pin to the active transition at the IFF's ICLK	IFD_DELAY_VALUE = default software setting	3	XA3S250E	6.85	ns
		de la dificienta le certing	2	XA3S500E	7.01	ns
			5	XA3S1200E	8.67	ns
			4	XA3S1600E	7.69	ns
Hold Time	S			-		
T _{IOICKP}	Time from the active transition at the IFF's ICLK input to the point where data must be held at the Input pin. No Input Delay is programmed.	LVCMOS25 ⁽²⁾ , IFD_DELAY_VALUE = 0	0	All	-0.76	ns
T _{IOICKPD}	Time from the active transition at the IFF's	LVCMOS25 ⁽²⁾ ,	2	XA3S100E	-3.93	ns
	ICLK input to the point where data must be held at the Input pin. The Input Delay is	IFD_DELAY_VALUE =	3	XA3S250E	-3.51	ns
	programmed.	doladit ootware ootting	2	XA3S500E	-3.74	ns
			5	XA3S1200E	-4.30	ns
			4	XA3S1600E	-4.14	ns
Set/Reset	Pulse Width				•	•
T _{RPW_IOB}	Minimum pulse width to SR control input on IOB			All	1.80	ns

Notes:

1. The numbers in this table are tested using the methodology presented in Table 19 and are based on the operating conditions set forth in Table 6 and Table 9.

2. This setup time requires adjustment whenever a signal standard other than LVCMOS25 is assigned to the data Input. If this is true, add the appropriate Input adjustment from Table 17.

3. These hold times require adjustment whenever a signal standard other than LVCMOS25 is assigned to the data Input. If this is true, subtract the appropriate Input adjustment from Table 17. When the hold time is negative, it is possible to change the data before the clock's active edge.



Table 16: Propagation Times for the IOB Input Path

			IFD_ DELAY		-4 Speed Grade	
Symbol	Description	Conditions	VALUE	Device	Max	Units
Propagatio	on Times					
T _{IOPLI}	The time it takes for data to travel from the Input pin through the IFF latch to the I output with no input delay programmed	LVCMOS25 ⁽²⁾ , IFD_DELAY_VALUE = 0	0	All	2.25	ns
T _{IOPLID}	The time it takes for data to	LVCMOS25 ⁽²⁾ ,	2	XA3S100E	5.97	ns
	travel from the Input pin through the IFF latch to the I output with	IFD_DELAY_VALUE = default software setting	3	XA3S250E	6.33	ns
	the input delay programmed	g	2	XA3S500E	6.49	ns
			5	XA3S1200E	8.15	ns
			4	XA3S1600E	7.16	ns

Notes:

1. The numbers in this table are tested using the methodology presented in Table 19 and are based on the operating conditions set forth in Table 6 and Table 9.

2. This propagation time requires adjustment whenever a signal standard other than LVCMOS25 is assigned to the data Input. When this is true, *add* the appropriate Input adjustment from Table 17.

Table 17: Input Timing Adjustments by IOSTANDARD

Convert Input Time from LVCMOS25 to the Following	Add the Adjustment Below	
(IOSTANDARD)	-4 Speed Grade	Units
Single-Ended Standards		
LVTTL	0.43	ns
LVCMOS33	0.43	ns
LVCMOS25	0	ns
LVCMOS18	0.98	ns
LVCMOS15	0.63	ns
LVCMOS12	0.27	ns
PCI33_3	0.42	ns
HSTL_I_18	0.12	ns
HSTL_III_18	0.17	ns
SSTL18_I	0.30	ns
SSTL2_I	0.15	ns

Table 17: Input Timing Adjustments by IOSTANDARD

Convert Input Time from LVCMOS25 to the Following Signal Standard	Add the Adjustment Below	
(IOSTANDARD)	-4 Speed Grade	Units
Differential Standards		
LVDS_25	0.49	ns
BLVDS_25	0.39	ns
MINI_LVDS_25	0.49	ns
LVPECL_25	0.27	ns
RSDS_25	0.49	ns
DIFF_HSTL_I_18	0.49	ns
DIFF_HSTL_III_18	0.49	ns
DIFF_SSTL18_I	0.30	ns
DIFF_SSTL2_I	0.32	ns

Notes:

- 1. The numbers in this table are tested using the methodology presented in Table 19 and are based on the operating conditions set forth in Table 6, Table 9, and Table 11.
- 2. These adjustments are used to convert input path times originally specified for the LVCMOS25 standard to times that correspond to other signal standards.

Convert C LVCMOS25 w	Putput Time vith 12mA [e from Drive and	Add the Adjustment Below	-
Signal Stand	ard (IOSTA	NDARD)	-4 Speed Grade	Units
Single-Ended	Standards		I	
LVTTL	Slow	2 mA	5.41	ns
		4 mA	2.41	ns
		6 mA	1.90	ns
		8 mA	0.67	ns
		12 mA	0.70	ns
		16 mA	0.43	ns
	Fast	2 mA	5.00	ns
		4 mA	1.96	ns
		6 mA	1.45	ns
		8 mA	0.34	ns
		12 mA	0.30	ns
		16 mA	0.30	ns
LVCMOS33	Slow	2 mA	5.29	ns
		4 mA	1.89	ns
		6 mA	1.04	ns
		8 mA	0.69	ns
		12 mA	0.42	ns
		16 mA	0.43	ns
	Fast	2 mA	4.87	ns
		4 mA	1.52	ns
		6 mA	0.39	ns
		8 mA	0.34	ns
		12 mA	0.30	ns
		16 mA	0.30	ns
LVCMOS25	Slow	2 mA	4.21	ns
		4 mA	2.26	ns
		6 mA	1.52	ns
		8 mA	1.08	ns
		12 mA	0.68	ns
	Fast	2 mA	3.67	ns
		4 mA	1.72	ns
		6 mA	0.46	ns
		8 mA	0.21	ns
		12 mA	0	ns

Table 18: Output Timing Adjustments for IOB

Table 18: Output Timing Adjustments for IOB (Continued)

Convert Output Time from LVCMOS25 with 12mA Drive and Fast Slew Rate to the Following		Add the Adjustment Below		
Signal Standa	rd (IOSTA	NDARD)	Grade	Units
LVCMOS18	Slow	2 mA	5.24	ns
		4 mA	3.21	ns
		6 mA	2.49	ns
		8 mA	1.90	ns
	Fast	2 mA	4.15	ns
		4 mA	2.13	ns
		6 mA	1.14	ns
		8 mA	0.75	ns
LVCMOS15	Slow	2 mA	4.68	ns
		4 mA	3.97	ns
		6 mA	3.11	ns
	Fast	2 mA	3.38	ns
		4 mA	2.70	ns
		6 mA	1.53	ns
LVCMOS12	Slow	2 mA	6.63	ns
	Fast	2 mA	4.44	ns
HSTL_I_18			0.34	ns
HSTL_III_18			0.55	ns
PCI33_3			0.46	ns
SSTL18_I			0.25	ns
SSTL2_I			-0.20	ns
Differential Star	ndards			
LVDS_25			-0.55	ns
BLVDS_25			0.04	ns
MINI_LVDS_25		-0.56	ns	
LVPECL_25		Input Only	ns	
RSDS_25		-0.48	ns	
DIFF_HSTL_I_18		0.42	ns	
DIFF_HSTL_III_	18		0.55	ns
DIFF_SSTL18_I			0.40	ns
DIFF_SSTL2_I			0.44	ns

Notes:

1. The numbers in this table are tested using the methodology presented in Table 19 and are based on the operating conditions set forth in Table 6, Table 9, and Table 11.

 These adjustments are used to convert output- and three-state-path times originally specified for the LVCMOS25 standard with 12 mA drive and Fast slew rate to times that correspond to other signal standards. Do not adjust times that measure when outputs go into a high-impedance state.

Table 24: 18 x 18 Embedded Multiplier Timing (Continued)

		-4 Speed Grade			
Symbol	Description	Min	Мах	Units	
Clock Frequency					
F _{MULT}	Internal operating frequency for a two-stage 18x18 multiplier using the AREG and BREG input registers and the PREG output register ⁽¹⁾	0	240	MHz	

Notes:

1. Combinatorial delay is less and pipelined performance is higher when multiplying input data with less than 18 bits.

2. The PREG register is typically used in both single-stage and two-stage pipelined multiplier implementations.

3. Input registers AREG or BREG are typically used when inferring a two-stage multiplier.

Block RAM Timing

Table 25: Block RAM Timing

	-4 Speed Grade		d Grade	
Symbol	Description	Min	Max	Units
Clock-to-Out	put Times			
Т _{ВСКО}	When reading from block RAM, the delay from the active transition at the CLK input to data appearing at the DOUT output	-	2.82	ns
Setup Times			I	I
T _{BACK}	Setup time for the ADDR inputs before the active transition at the CLK input of the block RAM	0.38	-	ns
T _{BDCK}	Setup time for data at the DIN inputs before the active transition at the CLK input of the block RAM	0.23	-	ns
T _{BECK}	Setup time for the EN input before the active transition at the CLK input of the block RAM	0.77	-	ns
T _{BWCK}	Setup time for the WE input before the active transition at the CLK input of the block RAM	1.26	-	ns
Hold Times				1
Т _{ВСКА}	Hold time on the ADDR inputs after the active transition at the CLK input	0.14	-	ns
T _{BCKD}	Hold time on the DIN inputs after the active transition at the CLK input	0.13	-	ns
T _{BCKE}	Hold time on the EN input after the active transition at the CLK input	0	-	ns
T _{BCKW}	Hold time on the WE input after the active transition at the CLK input	0	-	ns

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Table 25: Block RAM Timing (Continued)

		-4 Speed Grade		
Symbol	Description	Min	Max	Units
Clock Timing		,	,	,
T _{BPWH}	High pulse width of the CLK signal	1.59	-	ns
T _{BPWL}	Low pulse width of the CLK signal		-	ns
Clock Frequency				
F _{BRAM}	Block RAM clock frequency. RAM read output value written back into RAM, for shift registers and circular buffers. Write-only or read-only performance is faster.	0	230	MHz

Notes:

1. The numbers in this table are based on the operating conditions set forth in Table 6.

Digital Clock Manager Timing

For specification purposes, the DCM consists of three key components: the Delay-Locked Loop (DLL), the Digital Frequency Synthesizer (DFS), and the Phase Shifter (PS).

Aspects of DLL operation play a role in all DCM applications. All such applications inevitably use the CLKIN and the CLKFB inputs connected to either the CLK0 or the CLK2X feedback, respectively. Thus, specifications in the DLL tables (Table 26 and Table 27) apply to any application that only employs the DLL component. When the DFS and/or the PS components are used together with the DLL, then the specifications listed in the DFS and PS tables (Table 28 through Table 31) supersede any corresponding ones in the DLL tables. DLL specifications that do not change with the addition of DFS or PS functions are presented in Table 26 and Table 27.

Period jitter and cycle-cycle jitter are two of many different ways of specifying clock jitter. Both specifications describe statistical variation from a mean value.

Period jitter is the worst-case deviation from the ideal clock period over a collection of millions of samples. In a histogram of period jitter, the mean value is the clock period.

Cycle-cycle jitter is the worst-case difference in clock period between adjacent clock cycles in the collection of clock periods sampled. In a histogram of cycle-cycle jitter, the mean value is zero.

Spread Spectrum

DCMs accept typical spread spectrum clocks as long as they meet the input requirements. The DLL will track the frequency changes created by the spread spectrum clock to drive the global clocks to the FPGA logic. See <u>XAPP469</u>, *Spread-Spectrum Clocking Reception for Displays* for details.

Delay-Locked Loop

Table 26: Recommended Operating Conditions for the DLL

				-4 Speed Grade		
	Symbol	Des	scription	Min	Max	Units
Input Frequency Ranges						
F _{CLKIN}	CLKIN_FREQ_DLL	Frequency of the CLKIN clock in	nput	5 ⁽²⁾	240 ⁽³⁾	MHz
Input Pulse Requirements					•	
CLKIN_PULSE CLKIN percent period		CLKIN pulse width as a percentage of the CLKIN period	F _{CLKIN} ≤ 150 MHz	40%	60%	-
			F _{CLKIN} > 150 MHz	45%	55%	-
Input Cl	ock Jitter Tolerance and	I Delay Path Variation ⁽⁴⁾				
CLKIN_C	CYC_JITT_DLL_LF	Cycle-to-cycle jitter at the	F _{CLKIN} ≤ 150 MHz	-	±300	ps
CLKIN_C	CYC_JITT_DLL_HF	CLKIN input	F _{CLKIN} > 150 MHz	-	±150	ps
CLKIN_F	PER_JITT_DLL	Period jitter at the CLKIN input		-	±1	ns
CLKFB_	DELAY_VAR_EXT	Allowable variation of off-chip fee the CLKFB input	Allowable variation of off-chip feedback delay from the DCM output to the CLKFB input		±1	ns

Notes:

1. DLL specifications apply when any of the DLL outputs (CLK0, CLK90, CLK180, CLK270, CLK2X, CLK2X180, or CLKDV) are in use.

2. The DFS, when operating independently of the DLL, supports lower FCLKIN frequencies. See Table 28.

3. To support double the maximum effective FCLKIN limit, set the CLKIN_DIVIDE_BY_2 attribute to TRUE. This attribute divides the incoming clock frequency by two as it enters the DCM. The CLK2X output reproduces the clock frequency provided on the CLKIN input.

4. CLKIN input jitter beyond these limits might cause the DCM to lose lock.

Table 27: Switching Characteristics for the DLL

		-4 Spe		
Symbol	Description	Min	Max	Units
Output Frequency Ranges	•			
CLKOUT_FREQ_CLK0	Frequency for the CLK0 and CLK180 outputs	5	240	MHz
CLKOUT_FREQ_CLK90	Frequency for the CLK90 and CLK270 outputs	5	200	MHz
CLKOUT_FREQ_2X	Frequency for the CLK2X and CLK2X180 outputs	10	311	MHz
CLKOUT_FREQ_DV	Frequency for the CLKDV output	0.3125	160	MHz
Output Clock Jitter ^(2,3,4)	1	L	I	
CLKOUT_PER_JITT_0	Period jitter at the CLK0 output	-	±100	ps
CLKOUT_PER_JITT_90	Period jitter at the CLK90 output	-	±150	ps
CLKOUT_PER_JITT_180	Period jitter at the CLK180 output	-	±150	ps
CLKOUT_PER_JITT_270	Period jitter at the CLK270 output	-	±150	ps
CLKOUT_PER_JITT_2X	Period jitter at the CLK2X and CLK2X180 outputs	-	±[1% of CLKIN period + 150]	ps
CLKOUT_PER_JITT_DV1	Period jitter at the CLKDV output when performing integer division	-	±150	ps
CLKOUT_PER_JITT_DV2	Period jitter at the CLKDV output when performing non-integer division	-	±[1% of CLKIN period + 200]	ps

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Table 27: Switching Characteristics for the DLL (Continued)

			-4 Spe	ed Grade	
Symbol	Description	-	Min	Max	Units
Duty Cycle ⁽⁴⁾	·				
CLKOUT_DUTY_CYCLE_DLL	Duty cycle variation for the CLK0, CLK90, CLK180, CLK270, CLK2X, CLK2X180, and CLKDV outputs, including the BUFGMUX and clock tree duty-cycle distortion		-	±[1% of CLKIN period + 400]	ps
Phase Alignment ⁽⁴⁾					
CLKIN_CLKFB_PHASE	Phase offset between the CLKIN and	I CLKFB inputs	-	±200	ps
CLKOUT_PHASE_DLL	Phase offset between DLL outputs	CLK0 to CLK2X (not CLK2X180)	-	±[1% of CLKIN period + 100]	ps
		All others	-	±[1% of CLKIN period + 200]	ps
Lock Time		· · · · ·			
LOCK_DLL ⁽³⁾	When using the DLL alone: The time from deassertion at the DCM's Reset input to the rising transition at its LOCKED output. When the DCM is locked, the CLKIN and CLKFB signals are in phase	$\begin{array}{c} 5 \text{ MHz} \leq \text{F}_{\text{CLKIN}} \leq \\ 15 \text{ MHz} \end{array}$	-	5	ms
		F _{CLKIN} > 15 MHz	-	600	μs
Delay Lines					
DCM_DELAY_STEP	Finest delay resolution		20	40	ps

Notes:

1. The numbers in this table are based on the operating conditions set forth in Table 6 and Table 26.

- 2. Indicates the maximum amount of output jitter that the DCM adds to the jitter on the CLKIN input.
- 3. For optimal jitter tolerance and faster lock time, use the CLKIN_PERIOD attribute.
- Some jitter and duty-cycle specifications include 1% of input clock period or 0.01 UI. *Example:* The data sheet specifies a maximum jitter of "±[1% of CLKIN period + 150]". Assume the CLKIN frequency is 100 MHz. The equivalent CLKIN period is 10 ns and 1% of 10 ns is 0.1 ns or 100 ps. According to the data sheet, the maximum jitter is ±[100 ps + 150 ps] = ±250ps.

Digital Frequency Synthesizer

Table 28: Recommended Operating Conditions for the DFS

				-4 Spee	d Grade	
	Symbol Description		Min	Max	Units	
Input Frequency Ranges ⁽²⁾						
F _{CLKIN}	CLKIN_FREQ_FX	Frequency for the CLKIN input		0.200	333 ⁽⁴⁾	MHz
Input Clock	Input Clock Jitter Tolerance ⁽³⁾					
CLKIN_CYC	_JITT_FX_LF	Cycle-to-cycle jitter at the CLKIN	F _{CLKFX} ≤ 150 MHz	-	±300	ps
CLKIN_CYC	_JITT_FX_HF	input, based on CLKFX output frequency	F _{CLKFX} > 150 MHz	-	±150	ps
CLKIN_PER	_JITT_FX	Period jitter at the CLKIN input		-	±1	ns

Notes:

1. DFS specifications apply when either of the DFS outputs (CLKFX or CLKFX180) are used.

2. If both DFS and DLL outputs are used on the same DCM, follow the more restrictive CLKIN_FREQ_DLL specifications in Table 26.

3. CLKIN input jitter beyond these limits may cause the DCM to lose lock.

 To support double the maximum effective FCLKIN limit, set the CLKIN_DIVIDE_BY_2 attribute to TRUE. This attribute divides the incoming clock frequency by two as it enters the DCM.

Table 31: Switching Characteristics for the PS in Variable Phase Mode

Symbol	Description			Units
Phase Shifting Range				
MAX_STEPS ⁽²⁾	Maximum allowed number of DCM_DELAY_STEP steps for a given CLKIN clock period, where T = CLKIN clock period in ns. If using CLKIN_DIVIDE_BY_2 = TRUE, double the clock effective clock period.	CLKIN < 60 MHz	±[INTEGER(10 ● (T _{CLKIN} – 3 ns))]	steps
		CLKIN <u>></u> 60 MHz	±[INTEGER(15 ● (T _{CLKIN} – 3 ns))]	steps
FINE_SHIFT_RANGE_MIN	Minimum guaranteed delay for variable phase shifting	±[MAX_STEPS ● DCM_DELAY_STEP_MIN]		ns
FINE_SHIFT_RANGE_MAX	Maximum guaranteed delay for variable phase shifting	±[MAX_STEPS ● DCM_DELAY_STEP_MAX]		ns

Notes:

- 1. The numbers in this table are based on the operating conditions set forth in Table 6 and Table 30.
- 2. The maximum variable phase shift range, MAX_STEPS, is only valid when the DCM is has no initial fixed phase shifting, i.e., the PHASE_SHIFT attribute is set to 0.
- 3. The DCM_DELAY_STEP values are provided at the bottom of Table 27.

Miscellaneous DCM Timing

Table 32: Miscellaneous DCM Timing

Symbol	Description	Min	Max	Units
DCM_RST_PW_MIN ⁽¹⁾	Minimum duration of a RST pulse width	3	-	CLKIN cycles
DCM_RST_PW_MAX ⁽²⁾	Maximum duration of a RST pulse width	N/A	N/A	seconds
		N/A	N/A	seconds
DCM_CONFIG_LAG_TIME ⁽³⁾ Maximum duration from V _{CCINT} applied to FPGA		N/A	N/A	minutes
	configuration successfully completed (DONE pin goes High) and clocks applied to DCM DLL		N/A	minutes

Notes:

1. This limit only applies to applications that use the DCM DLL outputs (CLK0, CLK90, CLK180, CLK270, CLK2X, CLK2X180, and CLKDV). The DCM DFS outputs (CLKFX, CLKFX180) are unaffected.

- 2. This specification is equivalent to the Virtex-4 DCM_RESET specification. This specification does not apply for Spartan-3E FPGAs.
- 3. This specification is equivalent to the Virtex-4 TCONFIG specification. This specification does not apply for Spartan-3E FPGAs.



Configuration and JTAG Timing

Table 33: Power-On Timing and the Beginning of Configuration

			-4 Spee	d Grade	
Symbol	Description	Device	Min	Max	Units
T _{POR} ⁽²⁾	The time from the application of $V_{CCINT}\!, V_{CCAUX}\!,$ and V_{CCO}	XA3S100E	-	5	ms
	Bank 2 supply voltage ramps (whichever occurs last) to the	XA3S250E	-	5	ms
		XA3S500E	-	5	ms
		XA3S1200E	-	5	ms
		XA3S1600E	-	7	ms
T _{PROG}	The width of the low-going pulse on the PROG_B pin	All	0.5	-	μs
T _{PL} ⁽²⁾	The time from the rising edge of the PROG_B pin to the rising transition on the INIT_B pin	XA3S100E	-	0.5	ms
		XA3S250E	-	0.5	ms
		XA3S500E	-	1	ms
		XA3S1200E	-	2	ms
		XA3S1600E	-	2	ms
T _{INIT}	Minimum Low pulse width on INIT_B output	All	250	-	ns
T _{ICCK} ⁽³⁾	The time from the rising edge of the INIT_B pin to the generation of the configuration clock signal at the CCLK output pin	All	0.5	4.0	μs

Notes:

1. The numbers in this table are based on the operating conditions set forth in Table 6. This means power must be applied to all V_{CCINT} , V_{CCO} , and V_{CCAUX} lines.

2. Power-on reset and the clearing of configuration memory occurs during this period.

3. This specification applies only to the Master Serial, SPI, BPI-Up, and BPI-Down modes.



Master Serial and Slave Serial Mode Timing

Table	38:	Timing for	r the Master	Serial and	Slave Serial	Configuration	Modes
-------	-----	------------	--------------	------------	---------------------	---------------	-------

			Slave/	-4 Speed Grade		
Symbol	Descri	ption	Master	Min	Max	Units
Clock-to-Output Times						
T _{CCO}	The time from the falling transition on the CCLK pin to data appearing at the DOUT pin		Both	1.5	10.0	ns
Setup Tim	es					
T _{DCC}	The time from the setup of data at the DIN pin to the active edge of the CCLK pin			11.0	-	ns
Hold Time	S					
T _{CCD}	The time from the active edge of the CCLK pin to the point when data is last held at the DIN pin			0	-	ns
Clock Tim	ing					
т _{ссн}	High pulse width at the CCLK input pin		Master	See Table 36		
				See Table 37		
T _{CCL}	Low pulse width at the CCLK input pin		Master	See Table 36		
				Se	e Table 37	
F _{CCSER} Frequency of the clock signal at the CCLK input pin		No bitstream compression	Slave	0	66 ⁽²⁾	MHz
		With bitstream compression		0	20	MHz

Notes:

1. The numbers in this table are based on the operating conditions set forth in Table 6.

2. For serial configuration with a daisy-chain of multiple FPGAs, the maximum limit is 25 MHz.



Slave Parallel Mode Timing

Table 39: Timing for the Slave Parallel Configuration Mode

			-4 Speed Grad			
Symbol	Description			Min	Max	Units
Clock-to-Out	out Times					
T _{SMCKBY}	The time from the rising tran BUSY pin	nsition on the CCL	K pin to a signal transition at the	-	12.0	ns
Setup Times	•				•	+
T _{SMDCC}	The time from the setup of o pin	data at the D0-D7 p	ins to the active edge the CCLK	11.0	-	ns
T _{SMCSCC}	Setup time on the CSI_B p	in before the active	edge of the CCLK pin	10.0	-	ns
T _{SMCCW} ⁽²⁾	Setup time on the RDWR_	B pin before active	edge of the CCLK pin	23.0	-	ns
Hold Times						
T _{SMCCD}	The time from the active edge of the CCLK pin to the point when data is last held at the D0-D7 pins			1.0	-	ns
T _{SMCCCS}	The time from the active edge of the CCLK pin to the point when a logic level is last held at the CSO_B pin				-	ns
T _{SMWCC}	The time from the active edge of the CCLK pin to the point when a logic level is last held at the RDWR_B pin			0	-	ns
Clock Timing						
Т _{ССН}	The High pulse width at the	e CCLK input pin		5	-	ns
T _{CCL}	The Low pulse width at the CCLK input pin			5	-	ns
F _{CCPAR}	Frequency of the clock signal at the CCLK input pin	No bitstream compression	Not using the BUSY pin ⁽²⁾	0	50	MHz
			Using the BUSY pin	0	66	MHz
		With bitstream compression		0	20	MHz

Notes:

1. The numbers in this table are based on the operating conditions set forth in Table 6.

2. In the Slave Parallel mode, it is necessary to use the BUSY pin when the CCLK frequency exceeds this maximum specification.

3. Some Xilinx documents refer to Parallel modes as "SelectMAP" modes.

Byte Peripheral Interface Configuration Timing

Table 42: Timing for BPI Configuration Mode

Symbol	Description			Maximum	Units
T _{CCLK1}	Initial CCLK clock period			ee Table 34)	
T _{CCLKn}	CCLK clock period after FPGA loads ConfigRate setting			ee Table 34)	
T _{MINIT}	Setup time on CSI_B, RDWR_B, and M[2:0] mode pins before the rising edge of INIT_B			-	ns
T _{INITM}	Hold time on CSI_B, RDWR_B, and M[2:0] mode pins after the rising edge of INIT_B			-	ns
T _{INITADDR}	Minimum period of initial A[23:0] address cycle;BPI-UP:LDC[2:0] and HDC are asserted and valid(M[2:0]=<0:1:0>)		5	5	T _{CCLK1} cycles
		BPI-DN: (M[2:0]=<0:1:1>)	2	2	
T _{CCO}	Address A[23:0] outputs valid after CCLK falling edge			ee Table 38	
T _{DCC}	Setup time on D[7:0] data inputs before CCLK rising edge			ee Table 38	
T _{CCD}	Hold time on D[7:0] data inputs after CCLK rising edg	ge	See Table 38		

Table 43: Configuration Timing Requirements for Attached Parallel NOR Flash

Symbol	Description	Requirement	Units
T _{CE} (t _{ELQV})	Parallel NOR Flash PROM chip-select time	T _{CE} ≤ T _{INITADDR}	ns
T _{OE} (t _{GLQV})	Parallel NOR Flash PROM output-enable time	T _{OE} ≤ T _{INITADDR}	ns
T _{ACC} (t _{AVQV})	Parallel NOR Flash PROM read access time	$T_{ACC} \leq 0.5T_{CCLKn(min)} - T_{CCO} - T_{DCC} - PCB$	ns
T _{BYTE} (t _{FLQV,} t _{FHQV})	For x8/x16 PROMs only: BYTE# to output valid time ⁽³⁾	T _{byte} ≤T _{initaddr}	ns

Notes:

1. These requirements are for successful FPGA configuration in BPI mode, where the FPGA provides the CCLK frequency. The post configuration timing can be different to support the specific needs of the application loaded into the FPGA and the resulting clock source.

2. Subtract additional printed circuit board routing delay as required by the application.

3. The initial BYTE# timing can be extended using an external, appropriately sized pull-down resistor on the FPGA's LDC2 pin. The resistor value also depends on whether the FPGA's HSWAP pin is High or Low.



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