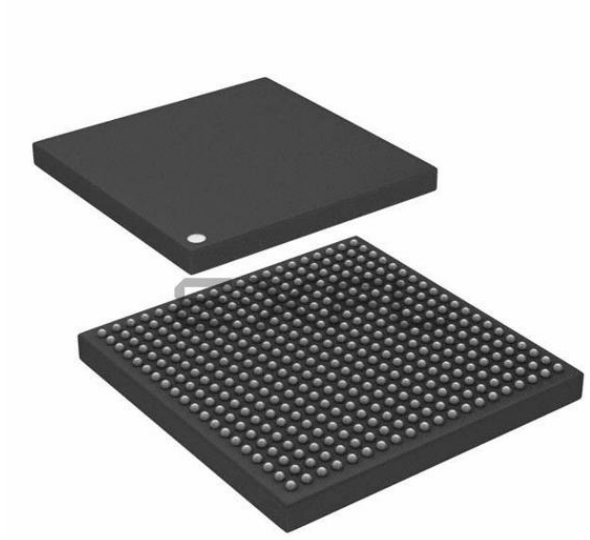

Welcome to E-XFL.COM

Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Obsolete
Type	Fixed Point
Interface	Host Interface, I²C, UART
Clock Rate	300MHz
Non-Volatile Memory	ROM (8kB)
On-Chip RAM	464kB
Voltage - I/O	3.30V
Voltage - Core	1.20V
Operating Temperature	-40°C ~ 105°C (TJ)
Mounting Type	Surface Mount
Package / Case	400-LFBGA
Supplier Device Package	400-LFBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmsc7118vf1200

Table of Contents

List of Figures

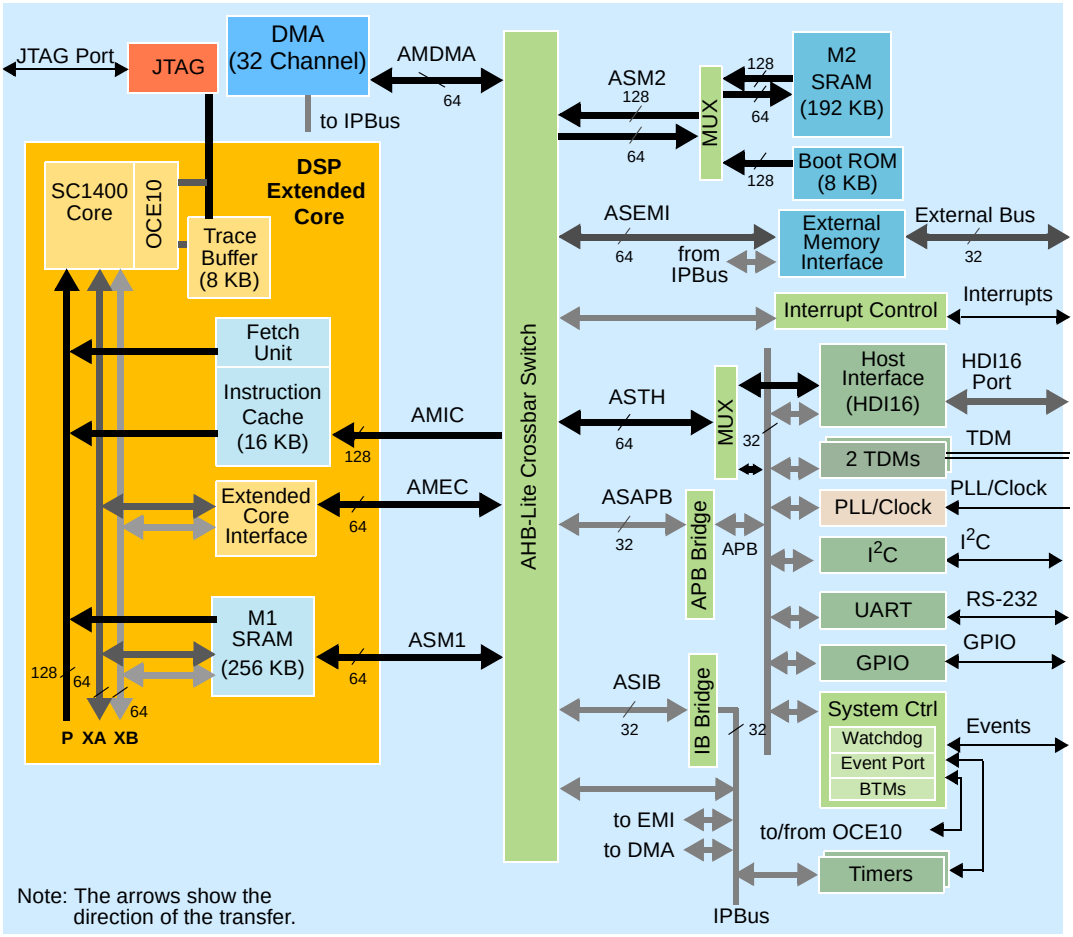

Figure 1. MSC7118 Block Diagram

Table 1. MSC7118 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
D10	V_{DDM}					
D11	V_{DDIO}					
D12	V_{DDIO}					
D13	V_{DDIO}					
D14	V_{DDIO}					
D15	V_{DDIO}					
D16	V_{DDIO}					
D17	V_{DDC}					
D18	NC					
D19	NC					
D20	NC					
E1	GND					
E2	D26					
E3	D31					
E4	V_{DDM}					
E5	V_{DDM}					
E6	V_{DDC}					
E7	V_{DDC}					
E8	V_{DDC}					
E9	V_{DDC}					
E10	V_{DDM}					
E11	V_{DDIO}					
E12	V_{DDIO}					
E13	V_{DDIO}					
E14	V_{DDIO}					
E15	V_{DDIO}					
E16	V_{DDC}					
E17	V_{DDC}					
E18	NC					
E19	NC					
E20	NC					
F1	V_{DDM}					
F2	D15					
F3	D29					
F4	V_{DDC}					
F5	V_{DDC}					

Table 1. MSC7118 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
F6	V_{DDC}					
F7	GND					
F8	GND					
F9	GND					
F10	V_{DDM}					
F11	V_{DDM}					
F12	GND					
F13	GND					
F14	GND					
F15	V_{DDIO}					
F16	V_{DDC}					
F17	V_{DDC}					
F18	NC					
F19	NC					
F20	NC					
G1	GND					
G2	D13					
G3	GND					
G4	V_{DDM}					
G5	V_{DDM}					
G6	GND					
G7	GND					
G8	GND					
G9	GND					
G10	GND					
G11	GND					
G12	GND					
G13	GND					
G14	GND					
G15	V_{DDIO}					
G16	V_{DDIO}					
G17	V_{DDC}					
G18	NC					
G19	NC					
G20	NC					
H1	D14					

Table 1. MSC7118 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
V18	GPIA24		$\overline{\text{IRQ24}}$	GPOA24	TX_EN	
V19	reserved				CRS	
V20	TDI					
W1	GND					
W2	V_{DDM}					
W3	A12					
W4	A8					
W5	A7					
W6	A6					
W7	A3					
W8	NC					
W9	GPIA17		$\overline{\text{IRQ13}}$	GPOA17	EVNT1	CLKO
W10	BM0	GPIC14		GPOC14	EVNT2	
W11	GPIA10		$\overline{\text{IRQ5}}$	GPOA10	T0RFS	
W12	GPIA7		$\overline{\text{IRQ7}}$	GPOA7	T0TFS	
W13	GPIA3		$\overline{\text{IRQ8}}$	GPOA3	T1RD	
W14	GPIA1		$\overline{\text{IRQ10}}$	GPOA1	T1TFS	
W15	GPID4			GPOD4	TXD2	reserved
W16	GPIA27		$\overline{\text{IRQ18}}$	GPOA27	RXD3	reserved
W17	GPIA19		$\overline{\text{IRQ19}}$	GPOA19	TXD1	
W18	GPIA23		$\overline{\text{IRQ23}}$	GPOA23	TXCLK or REFCLK	
W19	GPIA26		$\overline{\text{IRQ26}}$	GPOA26	RX_ER	
W20	H8BIT	reserved			MDC	
Y1	V_{DDM}					
Y2	GND					
Y3	A9					
Y4	A1					
Y5	A0					
Y6	A4					
Y7	BA1					
Y8	reserved		$\overline{\text{NMI}}$	reserved		
Y9	BM1	GPIC15		GPOC15	EVNT3	
Y10	GPIA11		$\overline{\text{IRQ4}}$	GPOA11	T0RCK	
Y11	GPIA9			GPOA9	T0RD	
Y12	GPIA6			GPOA6	T0TD	
Y13	GPIA5		$\overline{\text{IRQ0}}$	GPOA5	T1RCK	

Table 2 describes the maximum electrical ratings for the MSC7118.

Table 2. Absolute Maximum Ratings

Rating	Symbol	Value	Unit
Core supply voltage	V_{DDC}	1.5	V
Memory supply voltage	V_{DDM}	4.0	V
PLL supply voltage	V_{DDPLL}	1.5	V
I/O supply voltage	V_{DDIO}	–0.2 to 4.0	V
Input voltage	V_{IN}	(GND – 0.2) to 4.0	V
Reference voltage	V_{REF}	4.0	V
Maximum operating temperature	T_J	105	°C
Minimum operating temperature	T_A	–40	°C
Storage temperature range	T_{STG}	–55 to +150	°C

Notes:
1. Functional operating conditions are given in **Table 3.**
2. Absolute maximum ratings are stress ratings only, and functional operation at the maximum is not guaranteed. Stress beyond the listed limits may affect device reliability or cause permanent damage.
3. **Section 3.1**, *Thermal Design Considerations* includes a formula for computing the chip junction temperature (T_J).

2.2 Recommended Operating Conditions

Table 3 lists recommended operating conditions. Proper device operation outside of these conditions is not guaranteed.

Table 3. Recommended Operating Conditions

Rating	Symbol	Value	Unit
Core supply voltage	V_{DDC}	1.14 to 1.26	V
Memory supply voltage	V_{DDM}	2.38 to 2.63	V
PLL supply voltage	V_{DDPLL}	1.14 to 1.26	V
I/O supply voltage	V_{DDIO}	3.14 to 3.47	V
Reference voltage	V_{REF}	1.19 to 1.31	V
Operating temperature range	T_J T_A	maximum: 105 minimum: –40	°C °C

2.5 AC Timings

This section presents timing diagrams and specifications for individual signals and parallel I/O outputs and inputs. All AC timings are based on a 30 pF load, except where noted otherwise, and a 50 Ω transmission line. For any additional pF, use the following equations to compute the delay:

— Standard interface: 2.45 + (0.054 × C_{load}) ns
— DDR interface: 1.6 + (0.002 × C_{load}) ns

2.5.1 Clock and Timing Signals

The following tables describe clock signal characteristics. **Table 6** shows the maximum frequency values for internal (core, reference, and peripherals) and external (CLKO) clocks. You must ensure that maximum frequency values are not exceeded (see **Section 2.5.2** for the allowable ranges when using the PLL).

Table 6. Maximum Frequencies

Characteristic	Maximum in MHz
Core clock frequency (CLOCK)	300
External output clock frequency (CLKO)	75
Memory clock frequency (CK, $\overline{CK}$)	150
TDM clock frequency (TxRCK, TxTCK)	50

Table 7. Clock Frequencies in MHz

Characteristic	Symbol	Min	Max
CLKIN frequency	F_{CLKIN}	10	100
CLOCK frequency	F_{CORE}	—	300
CK, $\overline{CK}$ frequency	F_{CK}	—	150
TDMxRCK, TDMxTCK frequency	F_{TDMCK}	—	50
CLKO frequency	F_{CKO}	—	75
AHB/IPBus/APB clock frequency	F_{BCK}	—	150
Note: The rise and fall time of external clocks should be 5 ns maximum			

Table 8. System Clock Parameters

Characteristic	Min	Max	Unit
CLKIN frequency	10	100	MHz
CLKIN slope	—	5	ns
CLKIN frequency jitter (peak-to-peak)	—	1000	ps
CLKO frequency jitter (peak-to-peak)	—	150	ps

2.5.3 Reset Timing

The MSC7118 device has several inputs to the reset logic. All MSC7118 reset sources are fed into the reset controller, which takes different actions depending on the source of the reset. The reset status register indicates the most recent sources to cause a reset. **Table 14** describes the reset sources.

Table 14. Reset Sources

Name	Direction	Description
Power-on reset (PORESET)	Input	Initiates the power-on reset flow that resets the MSC7118 and configures various attributes of the MSC7118. On PORESET, the entire MSC7118 device is reset. SPLL and DLL states are reset, HRESET is driven, the SC1400 extended core is reset, and system configuration is sampled. The system is configured only when PORESET is asserted.
External Hard reset (HRESET)	Input/ Output	Initiates the hard reset flow that configures various attributes of the MSC7118. While HRESET is asserted, HRESET is an open-drain output. Upon hard reset, HRESET is driven and the SC1400 extended core is reset.
Software watchdog reset	Internal	When the MSC7118 watchdog count reaches zero, a software watchdog reset is signalled. The enabled software watchdog event then generates an internal hard reset sequence.
Bus monitor reset	Internal	When the MSC7118 bus monitor count reaches zero, a bus monitor hard reset is asserted. The enabled bus monitor event then generates an internal hard reset sequence.
JTAG EXTEST, CLAMP, or HIGHZ command	Internal	When a Test Access Port (TAP) executes an EXTEST, CLAMP, or HIGHZ command, the TAP logic asserts an internal reset signal that generates an internal soft reset sequence.

Table 15 summarizes the reset actions that occur as a result of the different reset sources.

Table 15. Reset Actions for Each Reset Source

Reset Action/Reset Source	Power-On Reset (PORESET)	Hard Reset (HRESET)	Soft Reset (SRESET)
	External only	External or Internal (Software Watchdog or Bus Monitor)	JTAG Command: EXTEST, CLAMP, or HIGHZ
Configuration pins sampled (refer to **Section 2.5.3.1** for details).	Yes	No	No
PLL and clock synthesis states Reset	Yes	No	No
HRESET Driven	Yes	Yes	No
Software watchdog and bus time-out monitor registers	Yes	Yes	Yes
Clock synthesis modules (STOPCTRL, HLTREQ, and HLTACK) reset	Yes	Yes	Yes
Extended core reset	Yes	Yes	Yes
Peripheral modules reset	Yes	Yes	Yes

2.5.3.1 Power-On Reset (PORESET) Pin

Asserting PORESET initiates the power-on reset flow. PORESET must be asserted externally for at least 16 CLKIN cycles after external power to the MSC7118 reaches at least 2/3 V_{DD}.

2.5.3.2 Reset Configuration

The MSC7118 has two mechanisms for writing the reset configuration:

- From a host through the host interface (HDI16)
- From memory through the I^2C interface

Five signal levels (see **Chapter 1** for signal description details) are sampled on $\overline{\text{PORESET}}$ deassertion to define the boot and operating conditions:

- BM[0–1]
- SWTE
- H8BIT
- HDSP

2.5.3.3 Reset Timing Tables

Table 16 and **Figure 4** describe the reset timing for a reset configuration write.

Table 16. Timing for a Reset Configuration Write

No.	Characteristics	Expression	Unit
1	Required external $\overline{\text{PORESET}}$ duration minimum	$16/F_{CLKIN}$	clocks
2	Delay from $\overline{\text{PORESET}}$ deassertion to $\overline{\text{HRESET}}$ deassertion	$521/F_{CLKIN}$	clocks
Note:	Timings are not tested, but are guaranteed by design.		

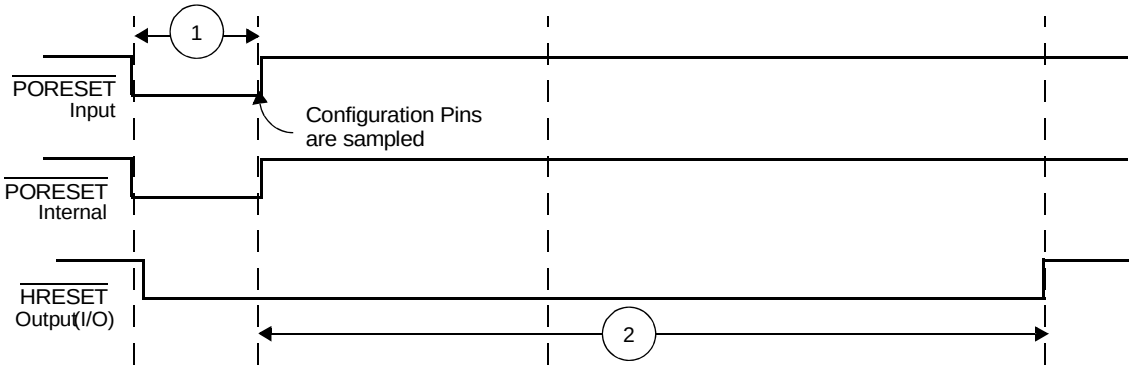

Figure 4. Timing Diagram for a Reset Configuration Write

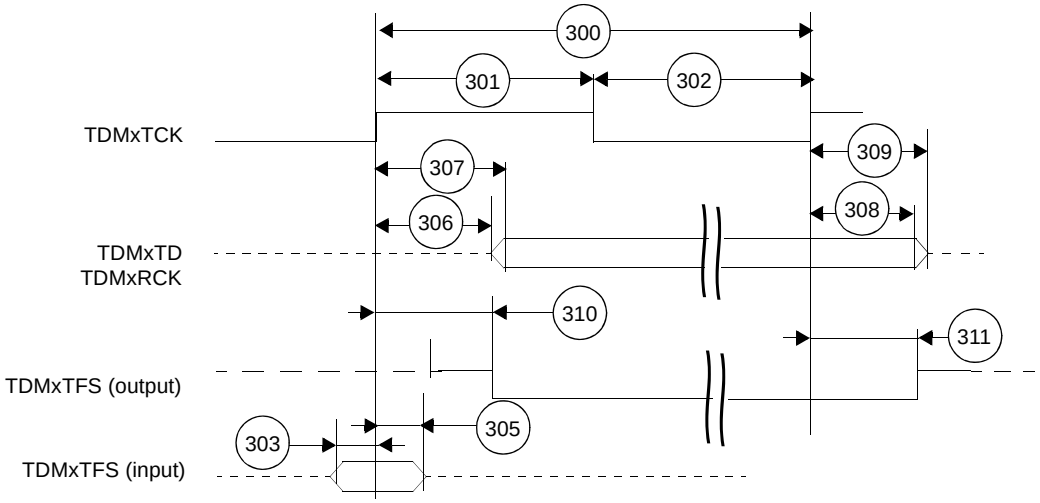

Figure 9. TDM Transmit Signals

Figure 10 and **Figure 11** show HDI16 read signal timing. **Figure 12** and **Figure 13** show HDI16 write signal timing.

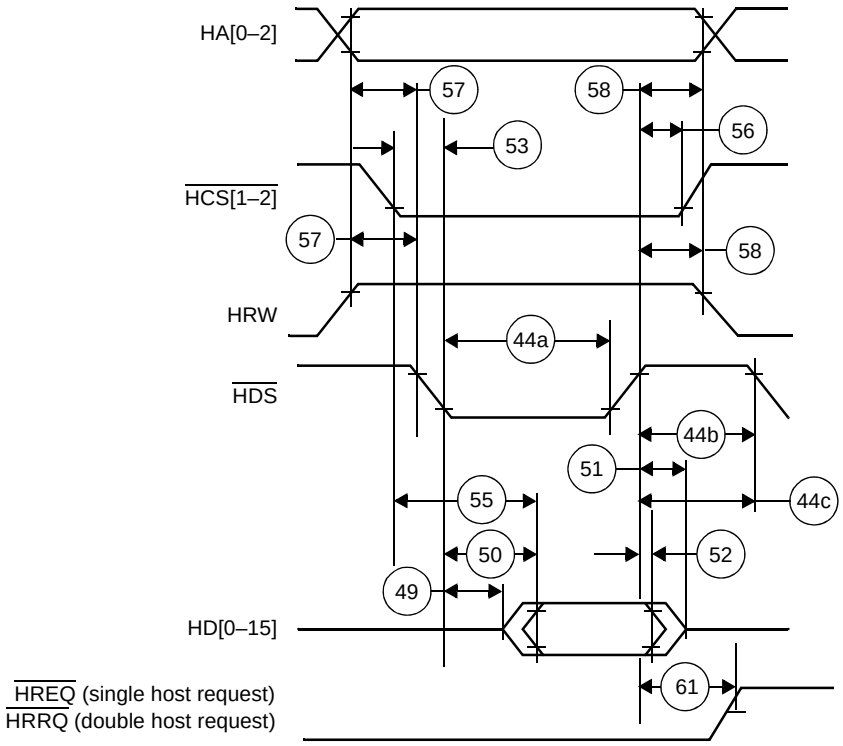

Figure 10. Read Timing Diagram, Single Data Strobe

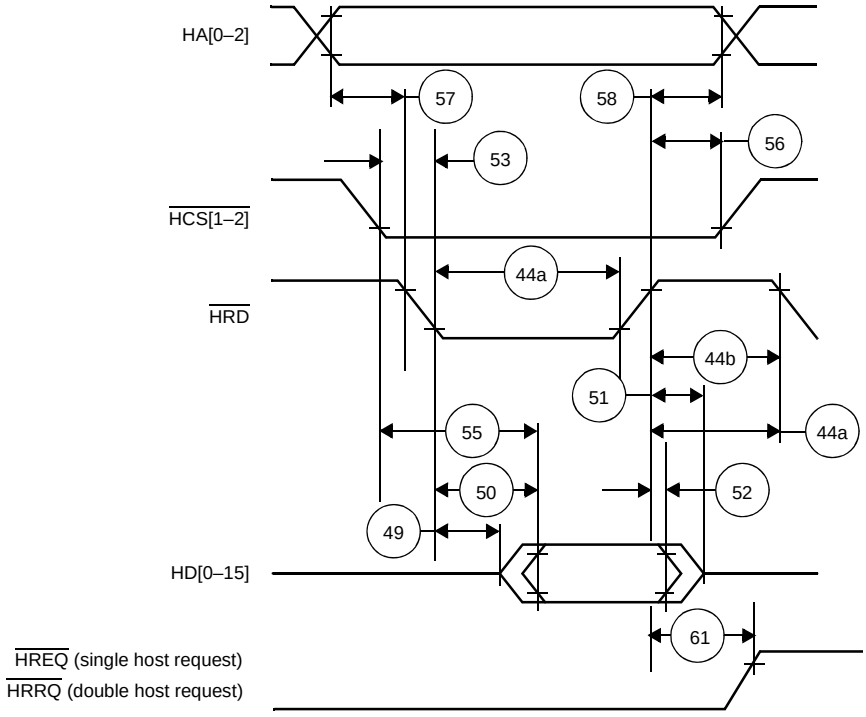

Figure 11. Read Timing Diagram, Double Data Strobe

2.5.12 JTAG Signals

Table 27. JTAG Timing

No.	Characteristics	All frequencies		Unit
		Min	Max	
700	TCK frequency of operation (1/($T_C \times 3$) **Note:** T_C = 1/CLOCK which is the period of the core clock. The TCK frequency must less than 1/3 of the core frequency with an absolute maximum limit of 40 MHz.	0.0	40.0	MHz
701	TCK cycle time	25.0	—	ns
702	TCK clock pulse width measured at V_M = 1.6 V	11.0	—	ns
703	TCK rise and fall times	0.0	3.0	ns
704	Boundary scan input data set-up time	5.0	—	ns
705	Boundary scan input data hold time	14.0	—	ns
706	TCK low to output data valid	0.0	20.0	ns
707	TCK low to output high impedance	0.0	20.0	ns
708	TMS, TDI data set-up time	5.0	—	ns
709	TMS, TDI data hold time	14.0	—	ns
710	TCK low to TDO data valid	0.0	24.0	ns
711	TCK low to TDO high impedance	0.0	10.0	ns
712	$\overline{\text{TRST}}$ assert time	100.0	—	ns

Note: All timings apply to OCE module data transfers as the OCE module uses the JTAG port as an interface.

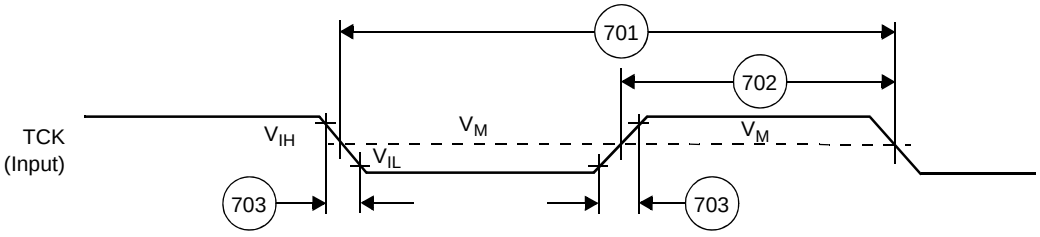

Figure 22. Test Clock Input Timing Diagram

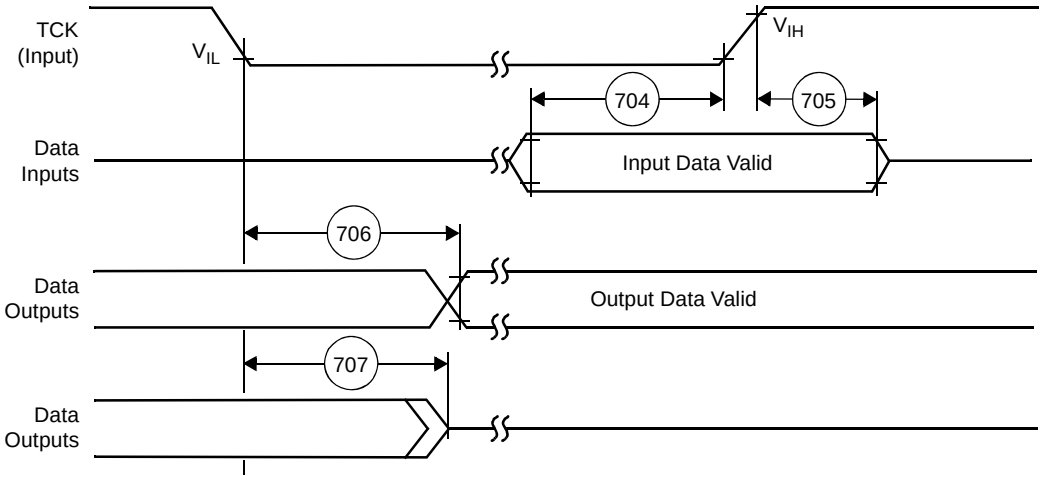

Figure 23. Boundary Scan (JTAG) Timing Diagram

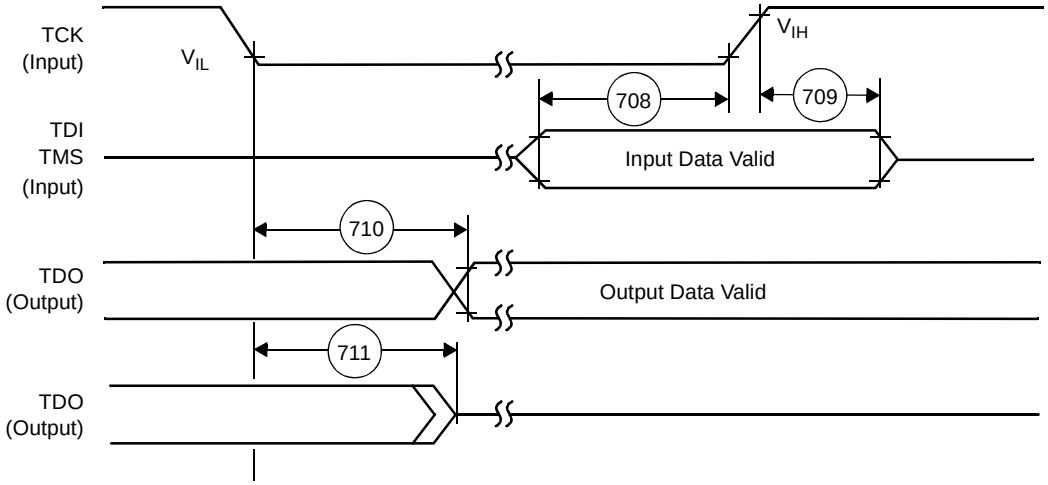

Figure 24. Test Access Port Timing Diagram

Figure 25. $\overline{\text{TRST}}$ Timing Diagram

3.3.2 Peripheral Power

Peripherals include the DDR memory controller, DMA controller, HDI16, TDM, UART, timers, GPIOs, and the I^2C module. Basic power consumption by each module is assumed to be the same and is computed by using the following equation which assumes an effective load of 20 pF, core voltage swing of 1.2 V, and a switching frequency of 100 MHz. This yields:

$$P_{PERIPHERAL} = 20\ pF \times (1.2\ V)^2 \times 150\ MHz \times 10^{-3} = 4.32\ mW\ per\ peripheral \quad \text{Eqn. 6}$$

Multiply this value by the number of peripherals used in the application to compute the total peripheral power consumption.

3.3.3 External Memory Power

Estimation of power consumption by the DDR memory system is complex. It varies based on overall system signal line usage, termination and load levels, and switching rates. Because the DDR memory includes terminations external to the MSC7118 device, the 2.5 V power source provides the power for the termination, which is a static value of 16 mA per signal driven high. The dynamic power is computed, however, using a differential voltage swing of ±0.200 V, yielding a peak-to-peak swing of 0.4 V. The equations for computing the DDR power are:

$$P_{DDRIO} = P_{STATIC} + P_{DYNAMIC} \quad \text{Eqn. 7}$$

$$P_{STATIC} = (unused\ pins \times \%\ driven\ high) \times 16\ mA \times 2.5\ V \quad \text{Eqn. 8}$$

$$P_{DYNAMIC} = (pin\ activity\ value) \times 20\ pF \times (0.4\ V)^2 \times 300\ MHz \times 10^{-3}\ mW \quad \text{Eqn. 9}$$

$$pin\ activity\ value = (active\ data\ lines \times \%\ activity \times \%\ data\ switching) + (active\ address\ lines \times \%\ activity) \quad \text{Eqn. 10}$$

As an example, assume the following:

unused pins = 16 (DDR uses 16-pin mode)
% driven high = 50%
active data lines = 16
% activity = 60%
% data switching = 50%
active address lines = 3

In this example, the DDR memory power consumption is:

$$P_{DDRIO} = ((16 \times 0.5) \times 16 \times 2.5) + (((16 \times 0.6 \times 0.5) + (3 \times 0.6)) \times 20 \times (0.4)^2 \times 300 \times 10^{-3}) = 326.3\ mW \quad \text{Eqn. 11}$$

3.3.4 External I/O Power

The estimation of the I/O power is similar to the computation of the peripheral power estimates. The power consumption per signal line is computed assuming a maximum load of 20 pF, a voltage swing of 3.3 V, and a switching frequency of 25 MHz, which yields:

$$P_{IO} = 20\ pF \times (3.3\ V)^2 \times 25\ MHz \times 10^{-3} = 5.44\ mW\ per\ I/O\ line \quad \text{Eqn. 12}$$

Multiply this number by the number of I/O signal lines used in the application design to compute the total I/O power.

Note: The signal loading depends on the board routing. For systems using a single DDR device, the load could be as low as 7 pF.

3.3.5 Leakage Power

The leakage power is for all power supplies combined at a specific temperature. The value is temperature dependent. The observed leakage value at room temperature is 64 mW.

3.3.6 Example Total Power Consumption

Using the examples in this section and assuming four peripherals and 10 I/O lines active, a total power consumption value is estimated as the following:

$$P_{TOTAL} = 324.0 + (4 \times 4.32) + 326.3 + (10 \times 5.44) + 64 = 784.98\ mW \qquad \text{Eqn. 13}$$

3.4 Reset and Boot

This section describes the recommendations for configuring the MSC7118 at reset and boot.

3.4.1 Reset Circuit

$\overline{\text{HRESET}}$ is a bidirectional signal and, if driven as an input, should be driven with an open collector or open-drain device. For an open-drain output such as $\overline{\text{HRESET}}$, take care when driving many buffers that implement input bus-hold circuitry. The bus-hold currents can cause enough voltage drop across the pull-up resistor to change the logic level to low. Either a smaller value of pull-up or less current loading from the bus-hold drivers overcomes this issue. To avoid exceeding the MSC7118 output current, the pull-up value should not be too small (a 1 KΩ pull-up resistor is used in the MSC711xADS reference design).

3.4.2 Reset Configuration Pins

Table 30 shows the MSC7118 reset configuration signals. These signals are sampled at the deassertion (rising edge) of $\overline{\text{PORESET}}$. For details, refer to the Reset chapter of the *MSC711x Reference Manual*.

Table 30. Reset Configuration Signals

Signal	Description	Settings
BM[3–0]	Determines boot mode.	See **Table 31** for details.
SWTE	Determines watchdog functionality.	0 Watchdog timer disabled. 1 Watchdog timer enabled.
HDSP	Configures HDI16 strobe polarity.	0 Host Data strobes active low. 1 Host Data strobes active high.
H8BIT	Configures HDI16 operation mode.	0 HDI16 port configured for 16-bit operation. 1 HDI16 port configured for 8-bit operation.

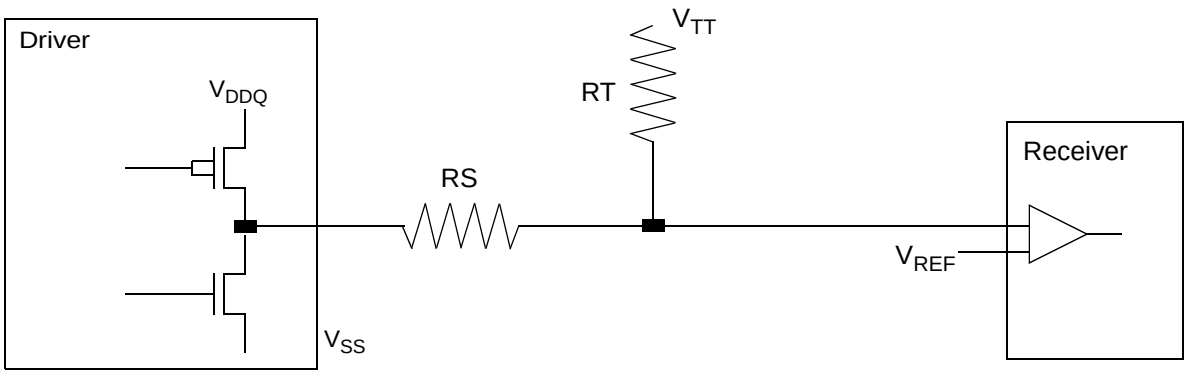

Figure 33. SSTL Power Value

3.5.1 V_{REF} and V_{TT} Design Constraints

V_{TT} and V_{REF} are isolated power supplies at the same voltage, with V_{TT} as a high current power source. This section outlines the voltage supply design needs and goals:

- Minimize the noise on both rails.
- V_{TT} must track variation in the V_{REF} DC offsets. Although they are isolated supplies, one possible solution is to use a single IC to generate both signals.
- Both references should have minimal drift over temperature and source supply.
- It is important to minimize the noise from coupling onto V_{REF} as follows:
  - Isolate V_{REF} and shield it with a ground trace.
  - Use 15–20 mm track.
  - Use 20–30 mm clearance between other traces for isolating.
  - Use the outer layer route when possible.
  - Use distributed decoupling to localize transient currents and return path and decouple with an inductance less than 3 nH.
- Max source/sink transient currents of up to 1.8 A for a 32-bit data bus.
- Use a wide island trace on the outer layer:
  - Place the island at the end of the bus.
  - Decouple both ends of the bus.
  - Use distributed decoupling across the island.
  - Place SSTL termination resistors inside the V_{TT} island and ensure a good, solid connection.
- Place the V_{TT} regulator as closely as possible to the termination island.
  - Reduce inductance and return path.
  - Tie current sense pin at the midpoint of the island.

3.5.2 Decoupling

The DDR decoupling considerations are as follows:

- DDR memory requires significantly more burst current than previous SDRAMs.
- In the worst case, up to 64 drivers may be switching states.
- Pay special attention and decouple discrete ICs per manufacturer guidelines.
- Leverage V_{TT} island topology to minimize the number of capacitors required to supply the burst current needs of the termination rail.
- See the Micron DesignLine publication entitled *Decoupling Capacitor Calculation for a DDR Memory Channel* (`http://download.micron.com/pdf/pubs/designline/3Q00dl1-4.pdf`).

3.5.3 General Routing

The general routing considerations for the DDR are as follows:

- All DDR signals must be routed next to a solid reference:
  - For data, next to solid ground planes.
  - For address/command, power planes if necessary.
- All DDR signals must be impedance controlled. This is system dependent, but typical values are 50–60 ohm.
- Minimize other cross-talk opportunities. As possible, maintain at least a four times the trace width spacing between all DDR signals to non-DDR signals.
- Keep the number of vias to a minimum to eliminate additional stubs and capacitance.
- Signal group routing priorities are as follows:
  - DDR clocks.
  - Route MVTT/MVREF.
  - Data group.
  - Command/address.
- Minimize data bit jitter by trace matching.

3.5.4 Routing Clock Distribution

The DDR clock distribution considerations are as follows:

- DDR controller supports six clock pairs:
  - 2 DIMM modules.
  - Up to 36 discrete chips.
- For route traces as for any other differential signals:
  - Maintain proper difference pair spacing.
  - Match pair traces within 25 mm.
- Match all clock traces to within 100 mm.
- Keep all clocks equally loaded in the system.
- Route clocks on inner critical layers.

3.5.5 Data Routing

The DDR data routing considerations are as follows:

- Route each data group (8-bits data + DQS + DM) on the same layer. Avoid switching layers within a byte group.
- Take care to match trace lengths, which is extremely important.
- To make trace matching easier, let adjacent groups be routed on alternate critical layers.
- Pin swap bits within a byte group to facilitate routing (discrete case).
- Tight trace matching is recommended within the DDR data group. Keep each 8-bit datum and its DM signal within ± 25 mm of its respective strobe.
- Minimize lengths across the entire DDR channel:
  - Between all groups maintain a delta of no more than 500 mm.
  - Allows greater flexibility in the design for readjustments as needed.
- DDR data group separation:
  - If stack-up allows, keep DDR data groups away from the address and control nets.
  - Route address and control on separate critical layers.
  - If resistor networks (RNs) are used, attempt to keep data and command lines in separate packages.

3.6 Connectivity Guidelines

This section summarizes the connections and special conditions, such as pull-up or pull-down resistors, for the MSC7118 device. Following are guidelines for signal groups and configuration settings:

- *Clock and reset signals*.
  - SWTE is used to configure the MSC7118 device and is sampled on the deassertion of $\overline{\text{PORESET}}$, so it should be tied to V_{DDC} or GND either directly or through pull-up or pull-down resistors until $\overline{\text{PORESET}}$ is deasserted. After $\overline{\text{PORESET}}$, this signal can be left floating.
  - BM[0–1] configure the MSC7118 device and are sampled until $\overline{\text{PORESET}}$ is deasserted, so they should be tied to V_{DDIO} or GND either directly or through pull-up or pull-down resistors.
  - $\overline{\text{HRESET}}$ should be pulled up.
- *Interrupt signals*. When used, $\overline{\text{IRQ}}$ pins must be pulled up.
- *HDI16 signals*.
  - When they are configured for open-drain, the $\overline{\text{HREQ}}$/HREQ or $\overline{\text{HTRQ}}$/HTRQ signals require a pull-up resistor. However, these pins are also sampled at power-on reset to determine the HDI16 boot mode and may need to be pulled down. When these pins must be pulled down on reset and pulled up otherwise, a buffer can be used with the $\overline{\text{HRESET}}$ signal as the enable.
  - When the device boots through the HDI16, the HDDS, HDSP and H8BIT pins should be pulled up or down, depending on the required boot mode settings.
- *I^2C signals*. The SCL and SDA signals, when programmed for I^2C, requires an external pull-up resistor.
- *General-purpose I/O (GPIO) signals*. An unused GPIO pin can be disconnected. After boot, program it as an output pin.
- Other signals.
  - The $\overline{\text{TEST0}}$ pin must be connected to ground.
  - The $\overline{\text{TPSEL}}$ pin should be pulled up to enable debug access via the EOnCE port and pulled down for boundary scan.
  - Pins labelled NO CONNECT (NC) must not be connected.
  - When a 16-pin double data rate (DDR) interface is used, the 16 unused data pins should be no connects (floating) if the used lines are terminated.
  - Do not connect DBREQ to DONE (as you would for the MSC8101 device). Connect DONE to one of the EVNT pins, and DBREQ to HRRQ.

4 Ordering Information

Consult a Freescale Semiconductor sales office or authorized distributor to determine product availability and place an order.

Part	Supply Voltage	Package Type	Pin Count	Core Frequency (MHz)	Solder Spheres	Order Number
MSC7118	1.2 V core 2.5 V memory 3.3 V I/O	Molded Array Process-Ball Grid Array (MAP-BGA)	400	300	Lead-free	MSC7118VM1200
					Lead-bearing	MSC7118VF1200

How to Reach Us:

Home Page:
www.freescale.com

Web Support:
http://www.freescale.com/support

USA/Europe or Locations Not Listed:
Freescale Semiconductor, Inc.
Technical Information Center, EL516
2100 East Elliot Road
Tempe, Arizona 85284
+1-800-521-6274 or
+1-480-768-2130
www.freescale.com/support

Europe, Middle East, and Africa:
Freescale Halbleiter Deutschland GmbH
Technical Information Center
Schatzbogen 7
81829 Muenchen, Germany
+44 1296 380 456 (English)
+46 8 52200080 (English)
+49 89 92103 559 (German)
+33 1 69 35 48 48 (French)
www.freescale.com/support

Japan:
Freescale Semiconductor Japan Ltd.
Headquarters
ARCO Tower 15F
1-8-1, Shimo-Meguro, Meguro-ku
Tokyo 153-0064
Japan
0120 191014 or
+81 3 5437 9125
support.japan@freescale.com

Asia/Pacific:
Freescale Semiconductor Hong Kong Ltd.
Technical Information Center
2 Dai King Street
Tai Po Industrial Estate
Tai Po, N.T., Hong Kong
+800 2666 8080
support.asia@freescale.com

For Literature Requests Only:
Freescale Semiconductor
Literature Distribution Center
P.O. Box 5405
Denver, Colorado 80217
+1-800 441-2447 or
+1-303-675-2140
Fax: +1-303-675-2150
LDCForFreescaleSemiconductor
@hibbertgroup.com

Document Number: MSC7118
Rev. 7
4/2008

freescale™
semiconductor