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### Understanding [Embedded - DSP \(Digital Signal Processors\)](#)

[Embedded - DSP \(Digital Signal Processors\)](#) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

### Applications of [Embedded - DSP \(Digital Signal Processors\)](#)

#### Details

Product Status	Obsolete
Type	Fixed Point
Interface	Host Interface, I <sup>2</sup> C, UART
Clock Rate	300MHz
Non-Volatile Memory	ROM (8kB)
On-Chip RAM	464kB
Voltage - I/O	3.30V
Voltage - Core	1.20V
Operating Temperature	-40°C ~ 105°C (TJ)
Mounting Type	Surface Mount
Package / Case	400-LFBGA
Supplier Device Package	400-LFBGA (17x17)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/nxp-semiconductors/kmsc7118vm1200">https://www.e-xfl.com/product-detail/nxp-semiconductors/kmsc7118vm1200</a>

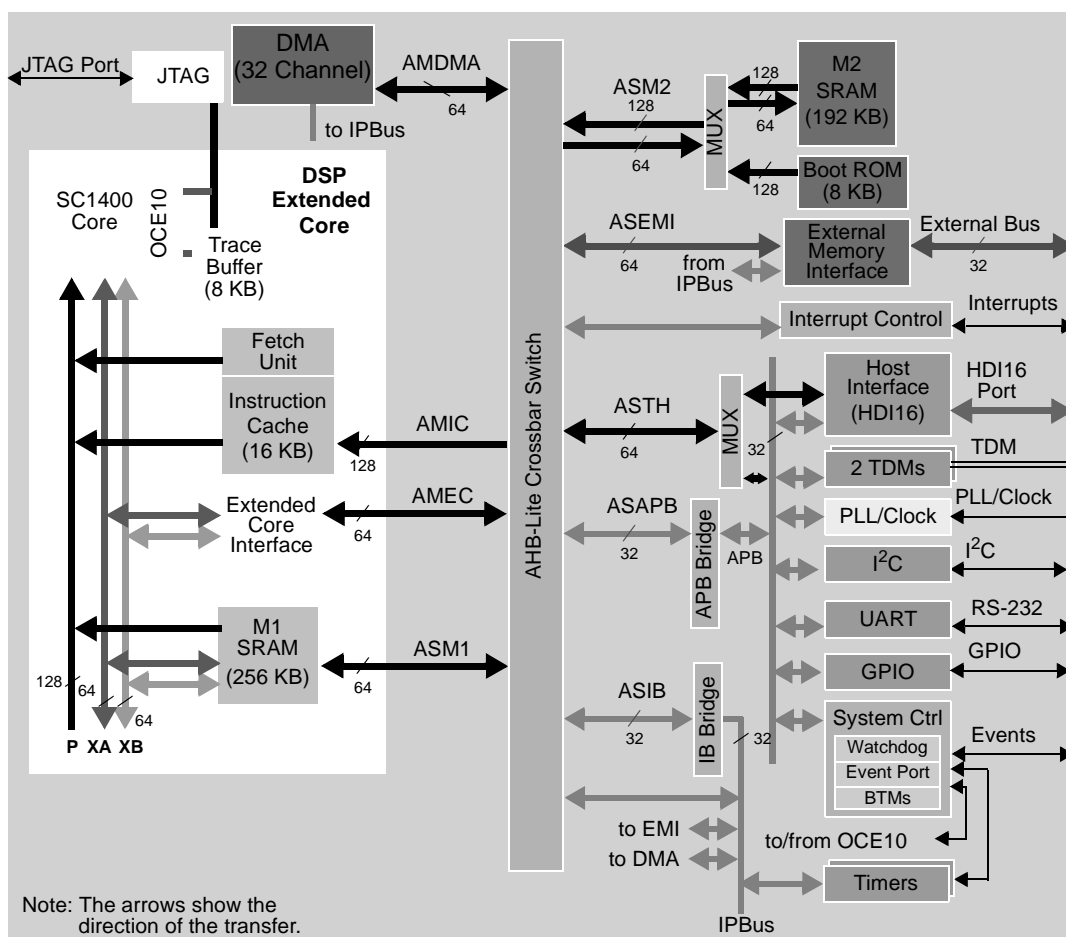


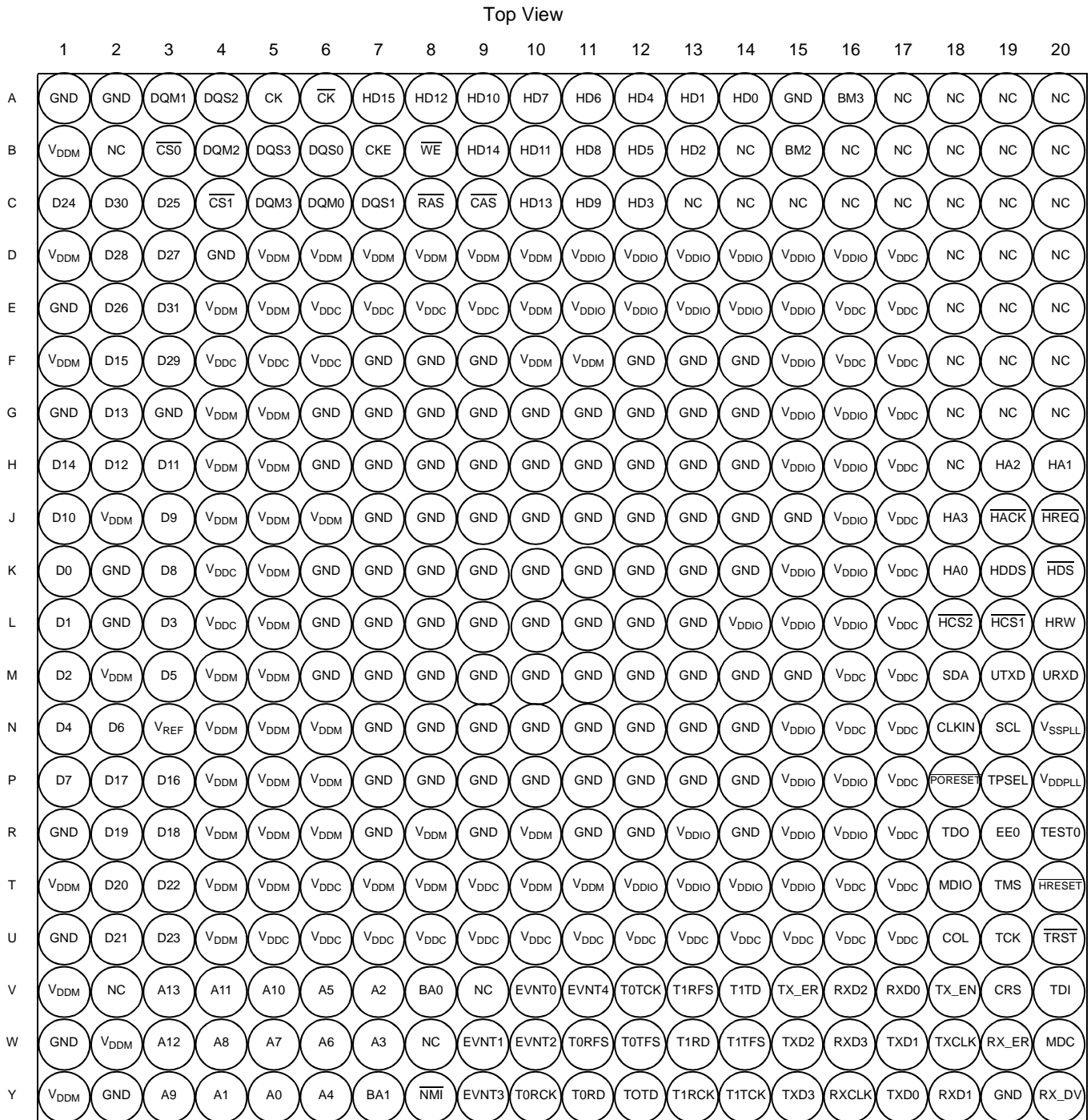
Figure 1. MSC7118 Block Diagram

# 1 Pin Assignments

This section includes diagrams of the MSC7118 package ball grid array layouts and pinout allocation tables.

## 1.1 MAP-BGA Ball Layout Diagrams

Top and bottom views of the MAP-BGA package are shown in **Figure 2** and **Figure 3** with their ball location index numbers.



**Figure 2. MSC7118 Molded Array Process-Ball Grid Array (MAP-BGA), Top View**

Bottom View

	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
A	NC	NC	NC	NC	BM3	GND	HD0	HD1	HD4	HD6	HD7	HD10	HD12	HD15	$\overline{CK}$	CK	DQS2	DQM1	GND	GND
B	NC	NC	NC	NC	NC	BM2	NC	HD2	HD5	HD8	HD11	HD14	$\overline{WE}$	CKE	DQS0	DQS3	DQM2	$\overline{CS0}$	NC	V <sub>DDM</sub>
C	NC	NC	NC	NC	NC	NC	NC	HD3	HD9	HD13	$\overline{CAS}$	$\overline{RAS}$	DQS1	DQM0	DQM3	$\overline{CS1}$	D25	D30	D24	
D	NC	NC	NC	V <sub>DD</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	GND	D27	D28	V <sub>DDM</sub>
E	NC	NC	NC	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDM</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	D31	D26	GND
F	NC	NC	NC	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DDIO</sub>	GND	GND	GND	V <sub>DDM</sub>	V <sub>DDM</sub>	GND	GND	GND	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	D29	D15	V <sub>DDM</sub>
G	NC	NC	NC	V <sub>DD</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	GND	GND	GND	GND	GND	GND	GND	GND	GND	V <sub>DDM</sub>	V <sub>DDM</sub>	GND	D13	GND
H	HA1	HA2	NC	V <sub>DD</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	GND	GND	GND	GND	GND	GND	GND	GND	GND	V <sub>DDM</sub>	V <sub>DDM</sub>	D11	D12	D14
J	$\overline{HREQ}$	$\overline{HACK}$	HA3	V <sub>DD</sub>	V <sub>DDIO</sub>	GND	GND	GND	GND	GND	GND	GND	GND	GND	V <sub>DDM</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	D9	V <sub>DDM</sub>	D10
K	$\overline{HDS}$	HDDS	HA0	V <sub>DD</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	GND	GND	GND	GND	GND	GND	GND	GND	GND	V <sub>DDM</sub>	V <sub>DD</sub>	D8	GND	D0
L	HRW	$\overline{HCS1}$	$\overline{HCS2}$	V <sub>DD</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	GND	GND	GND	GND	GND	GND	GND	GND	V <sub>DDM</sub>	V <sub>DD</sub>	D3	GND	D1
M	URXD	UTXD	SDA	V <sub>DD</sub>	V <sub>DD</sub>	GND	GND	GND	GND	GND	GND	GND	GND	GND	GND	V <sub>DDM</sub>	V <sub>DDM</sub>	D5	V <sub>DDM</sub>	D2
N	V <sub>SSPLL</sub>	SCL	CLKIN	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DDIO</sub>	GND	GND	GND	GND	GND	GND	GND	GND	V <sub>DDM</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	V <sub>REF</sub>	D6	D4
P	V <sub>DDPLL</sub>	TPSEL	$\overline{PORESET}$	V <sub>DD</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	GND	GND	GND	GND	GND	GND	GND	GND	V <sub>DDM</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	D16	D17	D7
R	TEST0	EE0	TDO	V <sub>DD</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	GND	V <sub>DDIO</sub>	GND	GND	V <sub>DDM</sub>	GND	V <sub>DDM</sub>	GND	V <sub>DDM</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	D18	D19	GND
T	$\overline{HRESET}$	TMS	MDIO	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	V <sub>DD</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	V <sub>DD</sub>	V <sub>DDM</sub>	V <sub>DDM</sub>	D22	D20	V <sub>DDM</sub>
U	$\overline{TRST}$	TCK	COL	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DDM</sub>	D23	D21	GND
V	TDI	CRS	TX_EN	RXD0	RXD2	TX_ER	T1TD	T1RFS	T0TCK	EVNT4	EVNT0	NC	BA0	A2	A5	A10	A11	A13	NC	V <sub>DDM</sub>
W	MDC	RX_ER	TXCLK	TXD1	RXD3	TXD2	T1TFS	T1RD	T0TFS	T0RFS	EVNT2	EVNT1	NC	A3	A6	A7	A8	A12	V <sub>DDM</sub>	GND
Y	RX_DV	GND	RXD1	TXD0	RXCLK	TXD3	T1TCK	T1RCK	T0TD	T0RD	T0RCK	EVNT3	$\overline{NMI}$	BA1	A4	A0	A1	A9	GND	V <sub>DDM</sub>

**Figure 3. MSC7118 Molded Array Process-Ball Grid Array (MAP-BGA), Bottom View**

Table 1. MSC7118 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
B14	NC					
B15	BM2	GPID7		GPOD7	reserved	
B16	NC					
B17	NC					
B18	NC					
B19	NC					
B20	NC					
C1	D24					
C2	D30					
C3	D25					
C4	$\overline{\text{CS1}}$					
C5	DQM3					
C6	DQM0					
C7	DQS1					
C8	$\overline{\text{RAS}}$					
C9	$\overline{\text{CAS}}$					
C10	GPIC5			GPOC5	HD13	
C11	GPIC1			GPOC1	HD9	
C12	reserved				HD3	
C13	NC					
C14	NC					
C15	NC					
C16	NC					
C17	NC					
C18	NC					
C19	NC					
C20	NC					
D1	$V_{\text{DDM}}$					
D2	D28					
D3	D27					
D4	GND					
D5	$V_{\text{DDM}}$					
D6	$V_{\text{DDM}}$					
D7	$V_{\text{DDM}}$					
D8	$V_{\text{DDM}}$					
D9	$V_{\text{DDM}}$					

Table 1. MSC7118 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
F6						$V_{DDC}$
F7						GND
F8						GND
F9						GND
F10						$V_{DDM}$
F11						$V_{DDM}$
F12						GND
F13						GND
F14						GND
F15						$V_{DDIO}$
F16						$V_{DDC}$
F17						$V_{DDC}$
F18						NC
F19						NC
F20						NC
G1						GND
G2						D13
G3						GND
G4						$V_{DDM}$
G5						$V_{DDM}$
G6						GND
G7						GND
G8						GND
G9						GND
G10						GND
G11						GND
G12						GND
G13						GND
G14						GND
G15						$V_{DDIO}$
G16						$V_{DDIO}$
G17						$V_{DDC}$
G18						NC
G19						NC
G20						NC
H1						D14

Table 1. MSC7118 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
N10						GND
N11						GND
N12						GND
N13						GND
N14						GND
N15						V <sub>DDIO</sub>
N16						V <sub>DDC</sub>
N17						V <sub>DDC</sub>
N18						CLKIN
N19		GPIA15	$\overline{\text{IRQ14}}$	GPOA15		SCL
N20						V <sub>SSPLL</sub>
P1						D7
P2						D17
P3						D16
P4						V <sub>DDM</sub>
P5						V <sub>DDM</sub>
P6						V <sub>DDM</sub>
P7						GND
P8						GND
P9						GND
P10						GND
P11						GND
P12						GND
P13						GND
P14						GND
P15						V <sub>DDIO</sub>
P16						V <sub>DDIO</sub>
P17						V <sub>DDC</sub>
P18						$\overline{\text{PORESET}}$
P19						TPSEL
P20						V <sub>DDPLL</sub>
R1						GND
R2						D19
R3						D18
R4						V <sub>DDM</sub>
R5						V <sub>DDM</sub>

Table 1. MSC7118 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
V18	GPIA24		$\overline{\text{IRQ}}_{24}$	GPOA24	TX_EN	
V19	reserved				CRS	
V20	TDI					
W1	GND					
W2	$V_{\text{DDM}}$					
W3	A12					
W4	A8					
W5	A7					
W6	A6					
W7	A3					
W8	NC					
W9	GPIA17		$\overline{\text{IRQ}}_{13}$	GPOA17	EVNT1	CLKO
W10	BM0	GPIC14		GPOC14	EVNT2	
W11	GPIA10		$\overline{\text{IRQ}}_5$	GPOA10	T0RFS	
W12	GPIA7		$\overline{\text{IRQ}}_7$	GPOA7	T0TFS	
W13	GPIA3		$\overline{\text{IRQ}}_8$	GPOA3	T1RD	
W14	GPIA1		$\overline{\text{IRQ}}_{10}$	GPOA1	T1TFS	
W15	GPID4			GPOD4	TXD2	reserved
W16	GPIA27		$\overline{\text{IRQ}}_{18}$	GPOA27	RXD3	reserved
W17	GPIA19		$\overline{\text{IRQ}}_{19}$	GPOA19	TXD1	
W18	GPIA23		$\overline{\text{IRQ}}_{23}$	GPOA23	TXCLK or REFCLK	
W19	GPIA26		$\overline{\text{IRQ}}_{26}$	GPOA26	RX_ER	
W20	H8BIT	reserved			MDC	
Y1	$V_{\text{DDM}}$					
Y2	GND					
Y3	A9					
Y4	A1					
Y5	A0					
Y6	A4					
Y7	BA1					
Y8	reserved		$\overline{\text{NMI}}$	reserved		
Y9	BM1	GPIC15		GPOC15	EVNT3	
Y10	GPIA11		$\overline{\text{IRQ}}_4$	GPOA11	T0RCK	
Y11	GPIA9			GPOA9	T0RD	
Y12	GPIA6			GPOA6	T0TD	
Y13	GPIA5		$\overline{\text{IRQ}}_0$	GPOA5	T1RCK	

Table 2 describes the maximum electrical ratings for the MSC7118.

**Table 2. Absolute Maximum Ratings**

Rating	Symbol	Value	Unit
Core supply voltage	$V_{DDC}$	1.5	V
Memory supply voltage	$V_{DDM}$	4.0	V
PLL supply voltage	$V_{DDPLL}$	1.5	V
I/O supply voltage	$V_{DDIO}$	−0.2 to 4.0	V
Input voltage	$V_{IN}$	(GND − 0.2) to 4.0	V
Reference voltage	$V_{REF}$	4.0	V
Maximum operating temperature	$T_J$	105	°C
Minimum operating temperature	$T_A$	−40	°C
Storage temperature range	$T_{STG}$	−55 to +150	°C
<b>Notes:</b> <ol style="list-style-type: none"> <li>1. Functional operating conditions are given in <b>Table 3</b>.</li> <li>2. Absolute maximum ratings are stress ratings only, and functional operation at the maximum is not guaranteed. Stress beyond the listed limits may affect device reliability or cause permanent damage.</li> <li>3. <b>Section 3.1, Thermal Design Considerations</b> includes a formula for computing the chip junction temperature (<math>T_J</math>).</li> </ol>			

## 2.2 Recommended Operating Conditions

Table 3 lists recommended operating conditions. Proper device operation outside of these conditions is not guaranteed.

**Table 3. Recommended Operating Conditions**

Rating	Symbol	Value	Unit
Core supply voltage	$V_{DDC}$	1.14 to 1.26	V
Memory supply voltage	$V_{DDM}$	2.38 to 2.63	V
PLL supply voltage	$V_{DDPLL}$	1.14 to 1.26	V
I/O supply voltage	$V_{DDIO}$	3.14 to 3.47	V
Reference voltage	$V_{REF}$	1.19 to 1.31	V
Operating temperature range	$T_J$	maximum: 105	°C
	$T_A$	minimum: −40	°C

## 2.5 AC Timings

This section presents timing diagrams and specifications for individual signals and parallel I/O outputs and inputs. All AC timings are based on a 30 pF load, except where noted otherwise, and a 50  $\Omega$  transmission line. For any additional pF, use the following equations to compute the delay:

- Standard interface:  $2.45 + (0.054 \times C_{load})$  ns
- DDR interface:  $1.6 + (0.002 \times C_{load})$  ns

### 2.5.1 Clock and Timing Signals

The following tables describe clock signal characteristics. **Table 6** shows the maximum frequency values for internal (core, reference, and peripherals) and external (CLKO) clocks. You must ensure that maximum frequency values are not exceeded (see **Section 2.5.2** for the allowable ranges when using the PLL).

**Table 6. Maximum Frequencies**

Characteristic	Maximum in MHz
Core clock frequency (CLOCK)	300
External output clock frequency (CLKO)	75
Memory clock frequency (CK, $\overline{CK}$ )	150
TDM clock frequency (TxRCK, TxTCK)	50

**Table 7. Clock Frequencies in MHz**

Characteristic	Symbol	Min	Max
CLKIN frequency	$F_{CLKIN}$	10	100
CLOCK frequency	$F_{CORE}$	—	300
CK, $\overline{CK}$ frequency	$F_{CK}$	—	150
TDMxRCK, TDMxTCK frequency	$F_{TDMCK}$	—	50
CLKO frequency	$F_{CKO}$	—	75
AHB/IPBus/APB clock frequency	$F_{BCK}$	—	150
<b>Note:</b> The rise and fall time of external clocks should be 5 ns maximum			

**Table 8. System Clock Parameters**

Characteristic	Min	Max	Unit
CLKIN frequency	10	100	MHz
CLKIN slope	—	5	ns
CLKIN frequency jitter (peak-to-peak)	—	1000	ps
CLKO frequency jitter (peak-to-peak)	—	150	ps

## 2.5.2 Configuring Clock Frequencies

This section describes important requirements for configuring clock frequencies in the MSC7118 device when using the PLL block. To configure the device clocking, you must program four fields in the Clock Control Register (CLKCTL):

- *PLLDVF field*. Specifies the PLL division factor ( $PLLDVF + 1$ ) to divide the input clock frequency  $F_{CLKIN}$ . The output of the divider block is the input to the multiplier block.
- *PLLMLTF field*. Specifies the PLL multiplication factor ( $PLLMLTF + 1$ ). The output from the multiplier block is the loop frequency  $F_{LOOP}$ .
- *RNG field*. Selects the available PLL frequency range for  $F_{VCO}$ , either  $F_{LOOP}$  when the RNG bit is set (1) or  $F_{LOOP}/2$  when the RNG bit is cleared (0).
- *CKSEL field*. Selects  $F_{CLKIN}$ ,  $F_{VCO}$ , or  $F_{VCO}/2$  as the source for the core clock.

There are restrictions on the frequency range permitted at the beginning of the multiplication portion of the PLL that affect the allowable values for the PLLDVF and PLLMLTF fields. The following sections define these restrictions and provide guidelines to configure the device clocking when using the PLL. Refer to the Clock and Power Management chapter in the *MSC711x Reference Manual* for details on the clock programming model.

### 2.5.2.1 PLL Multiplier Restrictions

There are two restrictions for correct usage of the PLL block:

- The input frequency to the PLL multiplier block (that is, the output of the divider) must be in the range 10–25 MHz.
- The output frequency of the PLL multiplier must be in the range 266–532 MHz.

When programming the PLL for a desired output frequency using the PLLDVF, PLLMLTF, and RNG fields, you must meet these constraints.

### 2.5.2.2 Input Division Factors and Corresponding CLKIN Frequency Range

The value of the PLLDVF field determines the allowable CLKIN frequency range, as shown in **Table 9**.

**Table 9. CLKIN Frequency Ranges by Divide Factor Value**

PLLDVF Field Value	Input Divide Factor	CLKIN Frequency Range	Comments
0x00	1	10 to 25 MHz	Input Division by 1
0x01	2	20 to 50 MHz	Input Division by 2
0x02	3	30 to 75 MHz	Input Division by 3
0x03	4	40 to 100 MHz	Input Division by 4
0x04	5	50 to 100 MHz	Input Division by 5
0x05	6	60 to 100 MHz	Input Division by 6
0x06	7	70 to 100 MHz	Input Division by 7
0x07	8	80 to 100 MHz	Input Division by 8
0x08	9	90 to 100 MHz	Input Division by 9
0x09	10	100 MHz	Input Division by 10
<b>Note:</b> The maximum CLKIN frequency is 100 MHz. Therefore, the PLLDVF value must be in the range from 1–10.			

### 2.5.2.3 Multiplication Factor Range

The multiplier block output frequency ranges depend on the divided input clock frequency as shown in **Table 10**.

**Table 10. PLLMLTF Ranges**

Multiplier Block (Loop) Output Range	Minimum PLLMLTF Value	Maximum PLLMLTF Value
$266 \leq [\text{Divided Input Clock} \times (\text{PLLMLTF} + 1)] \leq 532 \text{ MHz}$	266/Divided Input Clock	532/Divided Input Clock
<b>Note:</b> This table results from the allowed range for $F_{\text{Loop}}$ . The minimum and maximum multiplication factors are dependent on the frequency of the Divided Input Clock.		

### 2.5.2.4 Allowed Core Clock Frequency Range

The frequency delivered to the core, extended core, and peripherals depends on the value of the CLKCTRL[RNG] bit as shown in **Table 11**.

**Table 11.  $F_{\text{vco}}$  Frequency Ranges**

CLKCTRL[RNG] Value	Allowed Range of $F_{\text{vco}}$
1	$266 \leq F_{\text{vco}} \leq 532 \text{ MHz}$
0	$133 \leq F_{\text{vco}} \leq 266 \text{ MHz}$
<b>Note:</b> This table results from the allowed range for $F_{\text{vco}}$ , which is $F_{\text{Loop}}$ modified by CLKCTRL[RNG].	

This bit along with the CKSEL determines the frequency range of the core clock.

**Table 12. Resulting Ranges Permitted for the Core Clock**

CLKCTRL[CKSEL]	CLKCTRL[RNG]	Resulting Division Factor	Allowed Range of Core Clock	Comments
11	1	1	$266 \leq \text{core clock} \leq 300 \text{ MHz}$	Limited by maximum core frequency
11	0	2	$133 \leq \text{core clock} \leq 266 \text{ MHz}$	Limited by range of PLL
01	1	2	$133 \leq \text{core clock} \leq 266 \text{ MHz}$	Limited by range of PLL
01	0	4	$66.5 \leq \text{core clock} \leq 133 \text{ MHz}$	Limited by range of PLL
<b>Note:</b> This table results from the allowed range for $F_{\text{OUT}}$ , which depends on clock selected via CLKCTRL[CKSEL].				

### 2.5.2.5 Core Clock Frequency Range When Using DDR Memory

The core clock can also be limited by the frequency range of the DDR devices in the system. **Table 13** summarizes this restriction.

**Table 13. Core Clock Ranges When Using DDR**

DDR Type	Allowed Frequency Range for DDR CK	Corresponding Range for the Core Clock	Comments
DDR 200 (PC-1600)	83–100 MHz	$166 \leq \text{core clock} \leq 200 \text{ MHz}$	Core limited to $2 \times$ maximum DDR frequency
DDR 266 (PC-2100)	83–133 MHz	$166 \leq \text{core clock} \leq 266 \text{ MHz}$	Core limited to $2 \times$ maximum DDR frequency
DDR 333 (PC-2600)	83–150 MHz	$166 \leq \text{core clock} \leq 300 \text{ MHz}$	Core limited to $2 \times$ maximum DDR frequency

## 2.5.3 Reset Timing

The MSC7118 device has several inputs to the reset logic. All MSC7118 reset sources are fed into the reset controller, which takes different actions depending on the source of the reset. The reset status register indicates the most recent sources to cause a reset. **Table 14** describes the reset sources.

**Table 14. Reset Sources**

Name	Direction	Description
Power-on reset (PORESET)	Input	Initiates the power-on reset flow that resets the MSC7118 and configures various attributes of the MSC7118. On PORESET, the entire MSC7118 device is reset. SPL and DLL states are reset, HRESET is driven, the SC1400 extended core is reset, and system configuration is sampled. The system is configured only when PORESET is asserted.
External Hard reset (HRESET)	Input/ Output	Initiates the hard reset flow that configures various attributes of the MSC7118. While HRESET is asserted, HRESET is an open-drain output. Upon hard reset, HRESET is driven and the SC1400 extended core is reset.
Software watchdog reset	Internal	When the MSC7118 watchdog count reaches zero, a software watchdog reset is signalled. The enabled software watchdog event then generates an internal hard reset sequence.
Bus monitor reset	Internal	When the MSC7118 bus monitor count reaches zero, a bus monitor hard reset is asserted. The enabled bus monitor event then generates an internal hard reset sequence.
JTAG EXTEST, CLAMP, or HIGHZ command	Internal	When a Test Access Port (TAP) executes an EXTEST, CLAMP, or HIGHZ command, the TAP logic asserts an internal reset signal that generates an internal soft reset sequence.

**Table 15** summarizes the reset actions that occur as a result of the different reset sources.

**Table 15. Reset Actions for Each Reset Source**

Reset Action/Reset Source	Power-On Reset (PORESET)	Hard Reset (HRESET)	Soft Reset (SRESET)
	External only	External or Internal (Software Watchdog or Bus Monitor)	JTAG Command: EXTEST, CLAMP, or HIGHZ
Configuration pins sampled (refer to <b>Section 2.5.3.1</b> for details).	Yes	No	No
PLL and clock synthesis states Reset	Yes	No	No
HRESET Driven	Yes	Yes	No
Software watchdog and bus time-out monitor registers	Yes	Yes	Yes
Clock synthesis modules (STOPCTRL, HLTREQ, and HLTACK) reset	Yes	Yes	Yes
Extended core reset	Yes	Yes	Yes
Peripheral modules reset	Yes	Yes	Yes

### 2.5.3.1 Power-On Reset (PORESET) Pin

Asserting PORESET initiates the power-on reset flow. PORESET must be asserted externally for at least 16 CLKIN cycles after external power to the MSC7118 reaches at least  $\frac{2}{3} V_{DD}$ .

## 2.5.4 DDR DRAM Controller Timing

This section provides the AC electrical characteristics for the DDR DRAM interface.

### 2.5.4.1 DDR DRAM Input AC Timing Specifications

Table 17 provides the input AC timing specifications for the DDR DRAM interface.

Table 17. DDR DRAM Input AC Timing

No.	Parameter	Symbol	Min	Max	Unit
—	AC input low voltage	$V_{IL}$	—	$V_{REF} - 0.31$	V
—	AC input high voltage	$V_{IH}$	$V_{REF} + 0.31$	$V_{DDM} + 0.3$	V
201	Maximum Dn input setup skew relative to DQSn input	—	—	900	ps
202	Maximum Dn input hold skew relative to DQSn input	—	—	900	ps
<b>Notes:</b> <ol style="list-style-type: none"> <li>Maximum possible skew between a data strobe (DQSn) and any corresponding bit of data (<math>D[8n + \{0...7\}]</math> if <math>0 \leq n \leq 7</math>).</li> <li>See Table 18 for <math>t_{CK}</math> value.</li> <li>Dn should be driven at the same time as DQSn. This is necessary because the DQSn centering on the DQn data tenure is done internally.</li> </ol>					

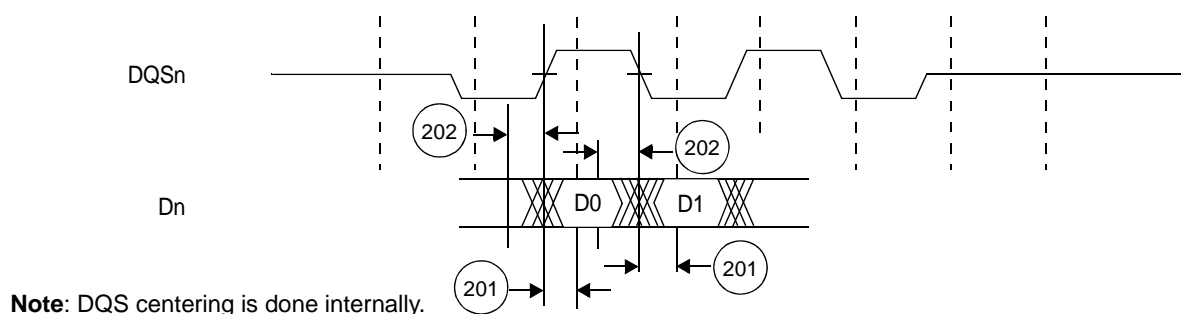


Figure 5. DDR DRAM Input Timing Diagram

### 2.5.4.2 DDR DRAM Output AC Timing Specifications

Table 18 and Table 19 list the output AC timing specifications and measurement conditions for the DDR DRAM interface.

Table 18. DDR DRAM Output AC Timing

No.	Parameter	Symbol	Min	Max	Unit
200	CK cycle time, (CK/CK crossing) <sup>1</sup> <ul style="list-style-type: none"> <li>100 MHz (DDR200)</li> <li>150 MHz (DDR300)</li> </ul>	$t_{CK}$	10 6.67	— —	ns ns
204	$\overline{An}/\overline{RAS}/\overline{CAS}/\overline{WE}/\overline{CKE}$ output setup with respect to CK	$t_{DDKHAS}$	$0.5 \times t_{CK} - 1000$	—	ps
205	$\overline{An}/\overline{RAS}/\overline{CAS}/\overline{WE}/\overline{CKE}$ output hold with respect to CK	$t_{DDKHAX}$	$0.5 \times t_{CK} - 1000$	—	ps
206	$\overline{CSn}$ output setup with respect to CK	$t_{DDKHCS}$	$0.5 \times t_{CK} - 1000$	—	ps
207	$\overline{CSn}$ output hold with respect to CK	$t_{DDKHCSX}$	$0.5 \times t_{CK} - 1000$	—	ps
208	CK to DQSn <sup>2</sup>	$t_{DDKMHM}$	-600	600	ps

Figure 10 and Figure 11 show HDI16 read signal timing. Figure 12 and Figure 13 show HDI16 write signal timing.

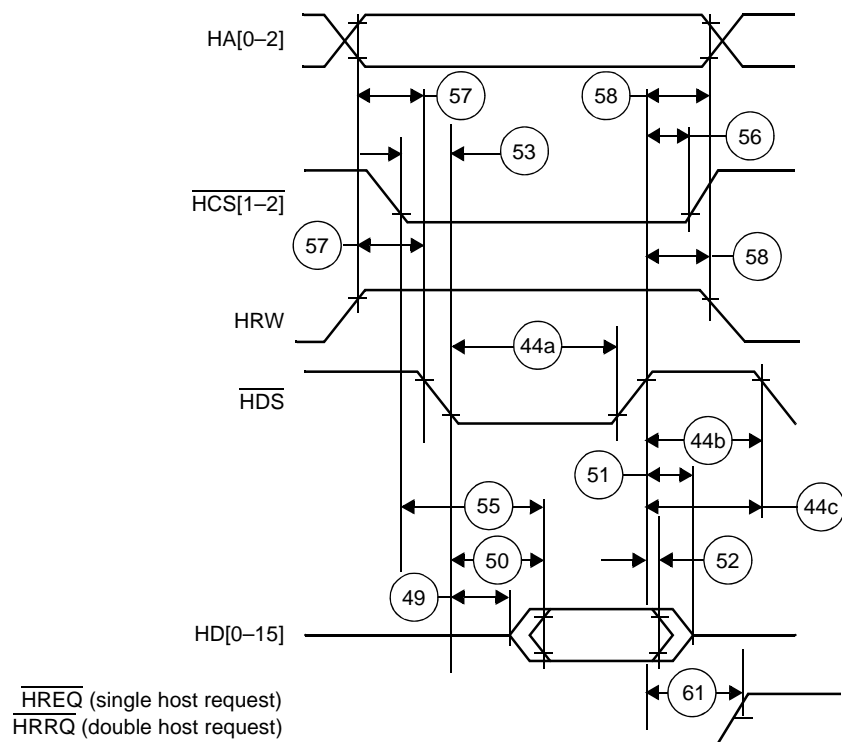


Figure 10. Read Timing Diagram, Single Data Strobe

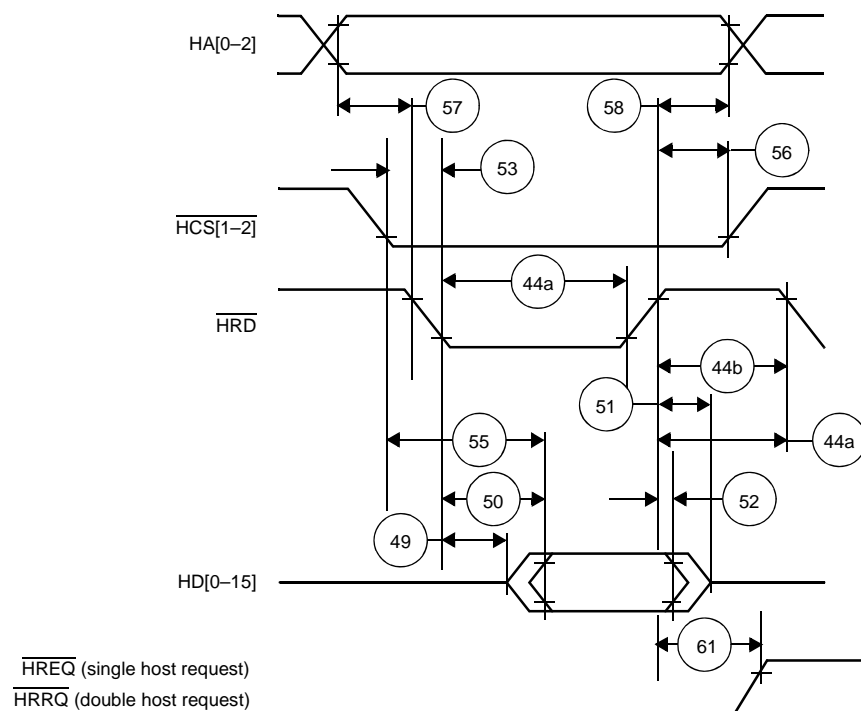
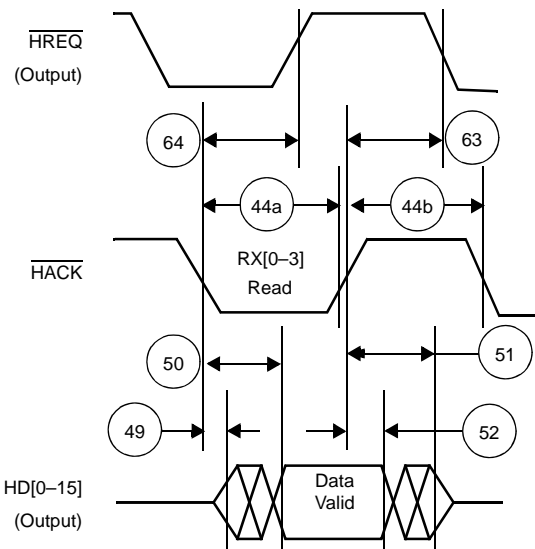
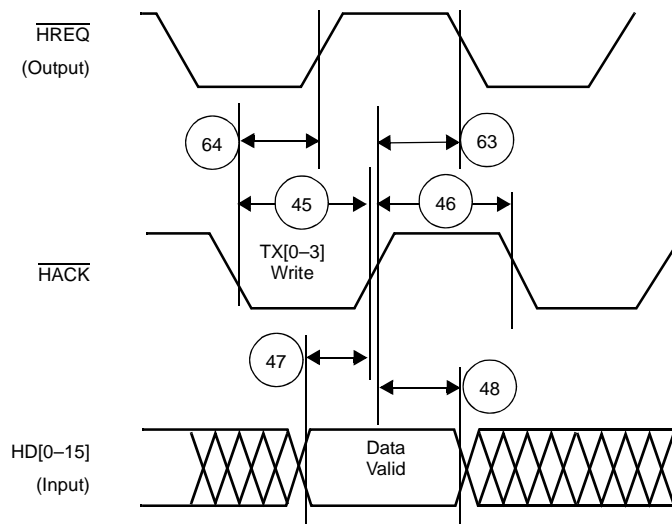


Figure 11. Read Timing Diagram, Double Data Strobe



**Figure 14. Host DMA Read Timing Diagram, HPCR[OAD] = 0**



**Figure 15. Host DMA Write Timing Diagram, HPCR[OAD] = 0**

## 2.5.8 UART Timing

Table 23. UART Timing

No.	Characteristics	Expression	Min	Max	Unit
—	Internal bus clock (APBCLK)	$F_{\text{CORE}}/2$	—	150	MHz
—	Internal bus clock period (1/APBCLK)	$T_{\text{APBCLK}}$	6.67	—	ns
400	URXD and UTXD inputs high/low duration	$16 \times T_{\text{APBCLK}}$	106.67	—	ns
401	URXD and UTXD inputs rise/fall time	—	—	5	ns
402	UTXD output rise/fall time	—	—	5	ns

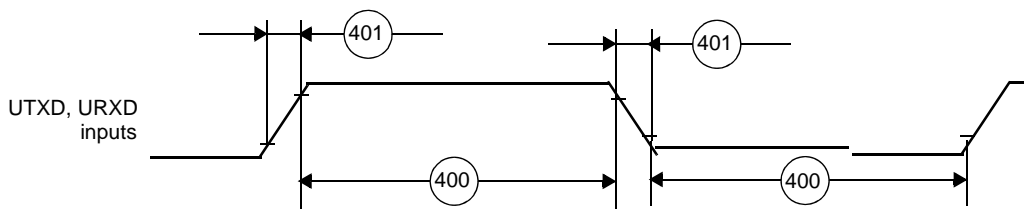


Figure 17. UART Input Timing

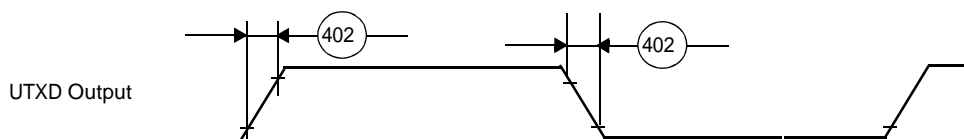


Figure 18. UART Output Timing

## 2.5.9 EE Timing

Table 24. EE0 Timing

Number	Characteristics	Type	Min
65	EE0 input to the core	Asynchronous	4 core clock periods
66	EE0 output from the core	Synchronous to core clock	1 core clock period

**Notes:**

1. The core clock is the SC1400 core clock. The ratio between the core clock and CLKOUT is configured during power-on-reset.
2. Configure the direction of the EE pin in the EE\_CTRL register (see the *SC140/SC1400 Core Reference Manual* for details).
3. Refer to **Table 1-11** on page 1-16 for details on EE pin functionality.

Figure 20 shows the signal behavior of the EE pin.

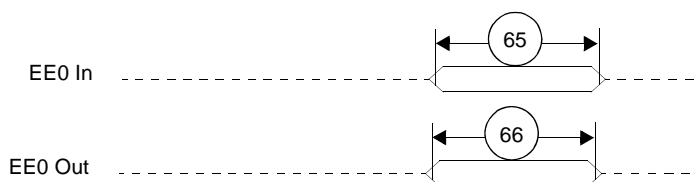


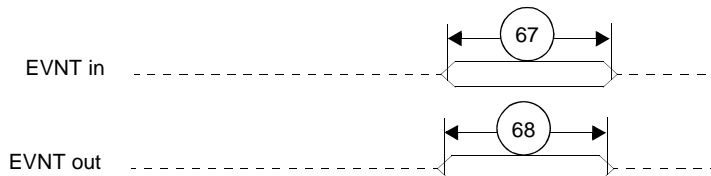
Figure 19. EE Pin Timing

## 2.5.10 Event Timing

**Table 25. EVNT Signal Timing**

Number	Characteristics	Type	Min
67	EVNT as input	Asynchronous	$1.5 \times \text{APBCLK periods}$
68	EVNT as output	Synchronous to core clock	1 APBCLK period
<b>Notes:</b> <ol style="list-style-type: none"> <li>1. Refer to <b>Table 23</b> for a definition of the APBCLK period.</li> <li>2. Direction of the EVNT signal is configured through the GPIO and Event port registers.</li> <li>3. Refer to the signal chapter in the <i>MSC711x Reference Manual</i> for details on EVNT pin functionality.</li> </ol>			

**Figure 20** shows the signal behavior of the EVNT pins.



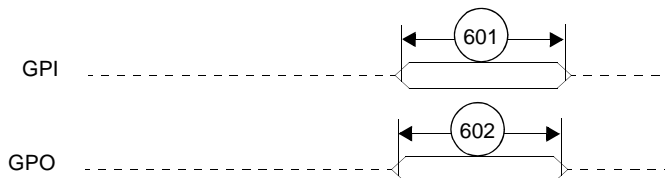
**Figure 20. EVNT Pin Timing**

## 2.5.11 GPIO Timing

**Table 26. GPIO Signal Timing<sup>1,2,3</sup>**

Number	Characteristics	Type	Min
601	GPI <sup>4,5</sup>	Asynchronous	$1.5 \times \text{APBCLK periods}$
602	GPO <sup>5</sup>	Synchronous to core clock	1 APBCLK period
603	Port A edge-sensitive interrupt	Asynchronous	$1.5 \times \text{APBCLK periods}$
604	Port A level-sensitive interrupt	Asynchronous	$3 \times \text{APBCLK periods}^6$
<b>Notes:</b> <ol style="list-style-type: none"> <li>1. Refer to <b>Table 23</b> for a definition of the APBCLK period.</li> <li>2. Direction of the GPIO signal is configured through the GPIO port registers.</li> <li>3. Refer to <b>Section 1.5</b> for details on GPIO pin functionality.</li> <li>4. GPI data is synchronized to the APBCLK internally and the minimum listed is the capability of the hardware to capture data into a register when the GPADR is read. The specification is not tested due to the asynchronous nature of the input and dependence on the state of the DSP core. It is guaranteed by design.</li> <li>5. The output signals cannot toggle faster than 75 MHz.</li> <li>6. Level-sensitive interrupts should be held low until the system determines (via the service routine) that the interrupt is acknowledged.</li> </ol>			

**Figure 21** shows the signal behavior of the GPI/GPO pins.



**Figure 21. GPI/GPO Pin Timing**

## 3.2 Power Supply Design Considerations

This section outlines the MSC7118 power considerations: power supply, power sequencing, power planes, decoupling, power supply filtering, and power consumption. It also presents a recommended power supply design and options for low-power consumption. For information on AC/DC electrical specifications and thermal characteristics, refer to **Section 2**.

### 3.2.1 Power Supply

The MSC7118 requires four input voltages, as shown in **Table 28**.

**Table 28. MSC7118 Voltages**

Voltage	Symbol	Value
Core	$V_{DDC}$	1.2 V
Memory	$V_{DDM}$	2.5 V
Reference	$V_{REF}$	1.25 V
I/O	$V_{DDIO}$	3.3 V

You should supply the MSC7118 core voltage via a variable switching supply or regulator to allow for compatibility with possible core voltage changes on future silicon revisions. The core voltage is supplied with 1.2 V (+5% and –10%) across  $V_{DDC}$  and GND and the I/O section is supplied with 3.3 V ( $\pm 10\%$ ) across  $V_{DDIO}$  and GND. The memory and reference voltages supply the DDR memory controller block. The memory voltage is supplied with 2.5 V across  $V_{DDM}$  and GND. The reference voltage is supplied across  $V_{REF}$  and GND and must be between  $0.49 \times V_{DDM}$  and  $0.51 \times V_{DDM}$ . Refer to the JEDEC standard JESD8 (*Stub Series Terminated Logic for 2.5 Volts (STTL\_2)*) for memory voltage supply requirements.

### 3.2.2 Power Sequencing

One consequence of multiple power supplies is that the voltage rails ramp up at different rates when power is initially applied. The rates depend on the power supply, the type of load on each power supply, and the way different voltages are derived. It is extremely important to observe the power up and power down sequences at the board level to avoid latch-up, forward biasing of ESD devices, and excessive currents, which all lead to severe device damage.

**Note:** There are five possible power-up/power-down sequence cases. The first four cases listed in the following sections are recommended for new designs. The fifth case is not recommended for new designs and must be carefully evaluated for current spike risks based on actual information for the specific application.

### 3.2.2.1 Case 1

The power-up sequence is as follows:

1. Turn on the  $V_{DDIO}$  (3.3 V) supply first.
2. Turn on the  $V_{DDC}$  (1.2 V) supply second.
3. Turn on the  $V_{DDM}$  (2.5 V) supply third.
4. Turn on the  $V_{REF}$  (1.25 V) supply fourth (last).

The power-down sequence is as follows:

1. Turn off the  $V_{REF}$  (1.25 V) supply first.
2. Turn off the  $V_{DDM}$  (2.5 V) supply second.
3. Turn off the  $V_{DDC}$  (1.2 V) supply third.
4. Turn of the  $V_{DDIO}$  (3.3 V) supply fourth (last).

Use the following guidelines:

- Make sure that the time interval between the ramp-down of  $V_{DDIO}$  and  $V_{DDC}$  is less than 10 ms.
- Make sure that the time interval between the ramp-up or ramp-down for  $V_{DDC}$  and  $V_{DDM}$  is less than 10 ms for power-up and power-down.
- Refer to **Figure 26** for relative timing for power sequencing case 1.

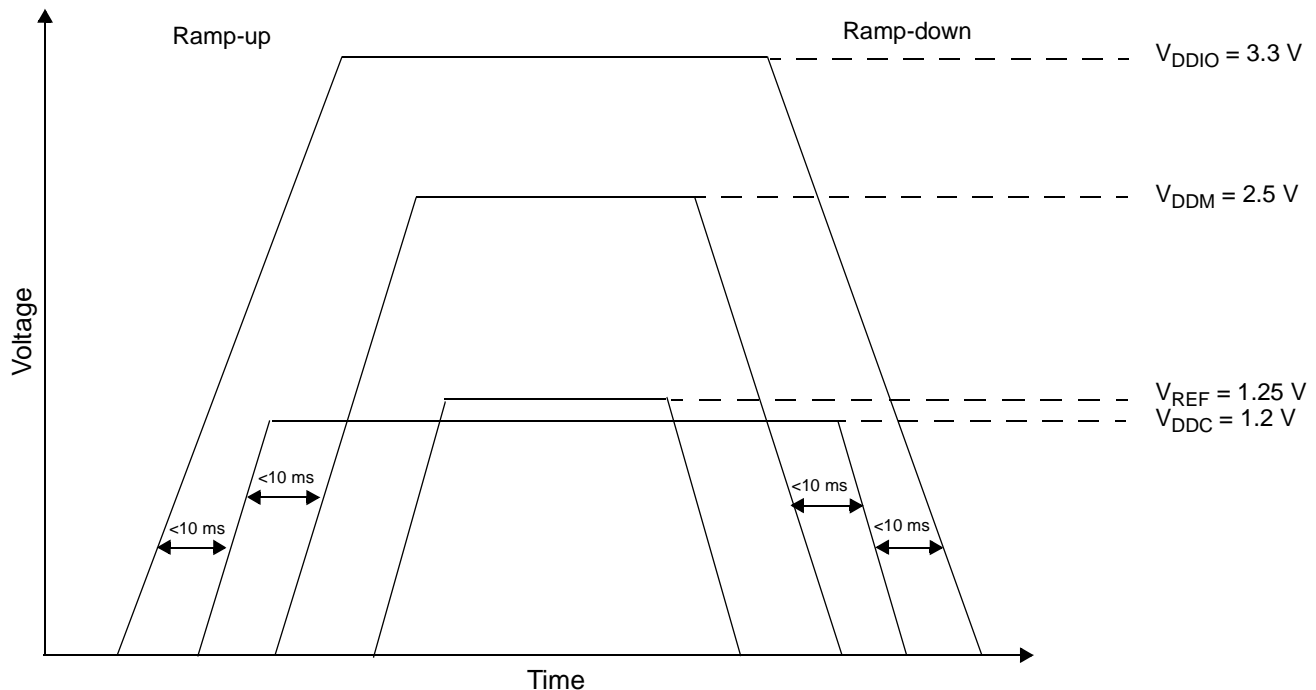


Figure 26. Voltage Sequencing Case 1

### 3.3.2 Peripheral Power

Peripherals include the DDR memory controller, DMA controller, HDI16, TDM, UART, timers, GPIOs, and the I<sup>2</sup>C module. Basic power consumption by each module is assumed to be the same and is computed by using the following equation which assumes an effective load of 20 pF, core voltage swing of 1.2 V, and a switching frequency of 100 MHz. This yields:

$$P_{PERIPHERAL} = 20 \text{ pF} \times (1.2 \text{ V})^2 \times 100 \text{ MHz} \times 10^{-3} = 4.32 \text{ mW per peripheral} \quad \text{Eqn. 6}$$

Multiply this value by the number of peripherals used in the application to compute the total peripheral power consumption.

### 3.3.3 External Memory Power

Estimation of power consumption by the DDR memory system is complex. It varies based on overall system signal line usage, termination and load levels, and switching rates. Because the DDR memory includes terminations external to the MSC7118 device, the 2.5 V power source provides the power for the termination, which is a static value of 16 mA per signal driven high. The dynamic power is computed, however, using a differential voltage swing of  $\pm 0.200$  V, yielding a peak-to-peak swing of 0.4 V. The equations for computing the DDR power are:

$$P_{DDRIO} = P_{STATIC} + P_{DYNAMIC} \quad \text{Eqn. 7}$$

$$P_{STATIC} = (\text{unused pins} \times \% \text{ driven high}) \times 16 \text{ mA} \times 2.5 \text{ V} \quad \text{Eqn. 8}$$

$$P_{DYNAMIC} = (\text{pin activity value}) \times 20 \text{ pF} \times (0.4 \text{ V})^2 \times 300 \text{ MHz} \times 10^{-3} \text{ mW} \quad \text{Eqn. 9}$$

$$\text{pin activity value} = (\text{active data lines} \times \% \text{ activity} \times \% \text{ data switching}) + (\text{active address lines} \times \% \text{ activity}) \quad \text{Eqn. 10}$$

As an example, assume the following:

unused pins = 16 (DDR uses 16-pin mode)  
 % driven high = 50%  
 active data lines = 16  
 % activity = 60%  
 % data switching = 50%  
 active address lines = 3

In this example, the DDR memory power consumption is:

$$P_{DDRIO} = ((16 \times 0.5) \times 16 \times 2.5) + (((16 \times 0.6 \times 0.5) + (3 \times 0.6)) \times 20 \times (0.4)^2 \times 300 \times 10^{-3}) = 326.3 \text{ mW} \quad \text{Eqn. 11}$$

### 3.3.4 External I/O Power

The estimation of the I/O power is similar to the computation of the peripheral power estimates. The power consumption per signal line is computed assuming a maximum load of 20 pF, a voltage swing of 3.3 V, and a switching frequency of 25 MHz, which yields:

$$P_{IO} = 20 \text{ pF} \times (3.3 \text{ V})^2 \times 25 \text{ MHz} \times 10^{-3} = 5.44 \text{ mW per I/O line} \quad \text{Eqn. 12}$$

Multiply this number by the number of I/O signal lines used in the application design to compute the total I/O power.

**Note:** The signal loading depends on the board routing. For systems using a single DDR device, the load could be as low as 7 pF.

### 3.3.5 Leakage Power

The leakage power is for all power supplies combined at a specific temperature. The value is temperature dependent. The observed leakage value at room temperature is 64 mW.