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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Active
Core Processor	ARM® Cortex®-M3
Core Size	32-Bit Single-Core
Speed	48MHz
Connectivity	I²C, IrDA, SmartCard, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, DMA, POR, PWM, WDT
Number of I/O	56
Program Memory Size	1MB (1M x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	128K x 8
Voltage - Supply (Vcc/Vdd)	1.98V ~ 3.8V
Data Converters	A/D 8x12b; D/A 2x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	64-VFQFN Exposed Pad
Supplier Device Package	64-QFN (9x9)
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/efm32gg230f1024g-e-qfn64r

process and close to automatic transfers. Automatic recognition of slave addresses is provided in all energy modes.

2.1.11 Universal Synchronous/Asynchronous Receiver/Transmitter (USART)

The Universal Synchronous Asynchronous serial Receiver and Transmitter (USART) is a very flexible serial I/O module. It supports full duplex asynchronous UART communication as well as RS-485, SPI, MicroWire and 3-wire. It can also interface with ISO7816 SmartCards, IrDA and I2S devices.

2.1.12 Pre-Programmed UART Bootloader

The bootloader presented in application note AN0003 is pre-programmed in the device at factory. Auto-baud and destructive write are supported. The autobaud feature, interface and commands are described further in the application note.

2.1.13 Low Energy Universal Asynchronous Receiver/Transmitter (LEUART)

The unique LEUARTTM, the Low Energy UART, is a UART that allows two-way UART communication on a strict power budget. Only a 32.768 kHz clock is needed to allow UART communication up to 9600 baud/s. The LEUART includes all necessary hardware support to make asynchronous serial communication possible with minimum of software intervention and energy consumption.

2.1.14 Timer/Counter (TIMER)

The 16-bit general purpose Timer has 3 compare/capture channels for input capture and compare/Pulse-Width Modulation (PWM) output. TIMERO also includes a Dead-Time Insertion module suitable for motor control applications.

2.1.15 Real Time Counter (RTC)

The Real Time Counter (RTC) contains a 24-bit counter and is clocked either by a 32.768 kHz crystal oscillator, or a 32.768 kHz RC oscillator. In addition to energy modes EM0 and EM1, the RTC is also available in EM2. This makes it ideal for keeping track of time since the RTC is enabled in EM2 where most of the device is powered down.

2.1.16 Backup Real Time Counter (BURTC)

The Backup Real Time Counter (BURTC) contains a 32-bit counter and is clocked either by a 32.768 kHz crystal oscillator, a 32.768 kHz RC oscillator or a 1 kHz ULFRCO. The BURTC is available in all Energy Modes and it can also run in backup mode, making it operational even if the main power should drain out.

2.1.17 Low Energy Timer (LETIMER)

The unique LETIMERTM, the Low Energy Timer, is a 16-bit timer that is available in energy mode EM2 in addition to EM1 and EM0. Because of this, it can be used for timing and output generation when most of the device is powered down, allowing simple tasks to be performed while the power consumption of the system is kept at an absolute minimum. The LETIMER can be used to output a variety of waveforms with minimal software intervention. It is also connected to the Real Time Counter (RTC), and can be configured to start counting on compare matches from the RTC.

2.1.18 Pulse Counter (PCNT)

The Pulse Counter (PCNT) can be used for counting pulses on a single input or to decode quadrature encoded inputs. It runs off either the internal LFACLK or the PCNTn_S0IN pin as external clock source. The module may operate in energy mode EM0 - EM3.

and key registers. All write accesses to the AES module must be 32-bit operations, i.e. 8- or 16-bit operations are not supported.

2.1.27 General Purpose Input/Output (GPIO)

In the EFM32GG230, there are 56 General Purpose Input/Output (GPIO) pins, which are divided into ports with up to 16 pins each. These pins can individually be configured as either an output or input. More advanced configurations like open-drain, filtering and drive strength can also be configured individually for the pins. The GPIO pins can also be overridden by peripheral pin connections, like Timer PWM outputs or USART communication, which can be routed to several locations on the device. The GPIO supports up to 16 asynchronous external pin interrupts, which enables interrupts from any pin on the device. Also, the input value of a pin can be routed through the Peripheral Reflex System to other peripherals.

2.2 Configuration Summary

The features of the EFM32GG230 is a subset of the feature set described in the EFM32GG Reference Manual. Table 2.1 (p. 7) describes device specific implementation of the features.

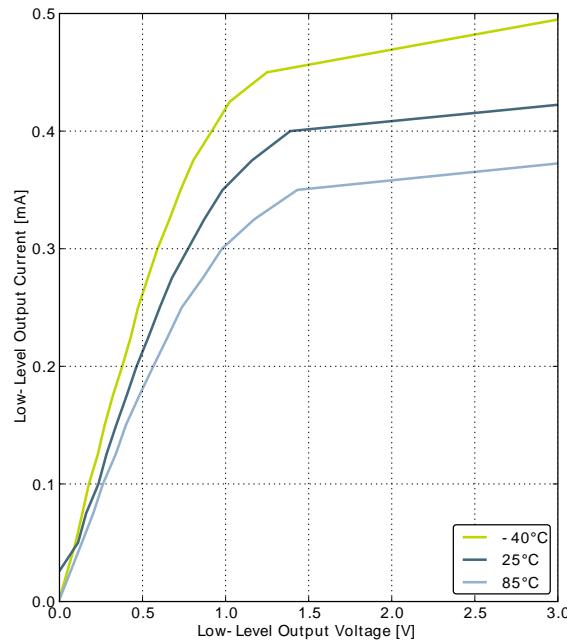
Table 2.1. Configuration Summary

Module	Configuration	Pin Connections
Cortex-M3	Full configuration	NA
DBG	Full configuration	DBG_SWCLK, DBG_SWDIO, DBG_SWO
MSC	Full configuration	NA
DMA	Full configuration	NA
RMU	Full configuration	NA
EMU	Full configuration	NA
CMU	Full configuration	CMU_OUT0, CMU_OUT1
WDOG	Full configuration	NA
PRS	Full configuration	NA
I2C0	Full configuration	I2C0_SDA, I2C0_SCL
I2C1	Full configuration	I2C1_SDA, I2C1_SCL
USART0	Full configuration with IrDA	US0_TX, US0_RX, US0_CLK, US0_CS
USART1	Full configuration with I2S	US1_TX, US1_RX, US1_CLK, US1_CS
USART2	Full configuration with I2S	US2_TX, US2_RX, US2_CLK, US2_CS
LEUART0	Full configuration	LEU0_TX, LEU0_RX
LEUART1	Full configuration	LEU1_TX, LEU1_RX
TIMER0	Full configuration with DTI	TIM0_CC[2:0], TIM0_CDTI[2:0]
TIMER1	Full configuration	TIM1_CC[2:0]
TIMER2	Full configuration	TIM2_CC[2:0]
TIMER3	Full configuration	TIM3_CC[2:0]
RTC	Full configuration	NA
BURTC	Full configuration	NA
LETIMER0	Full configuration	LET0_O[1:0]

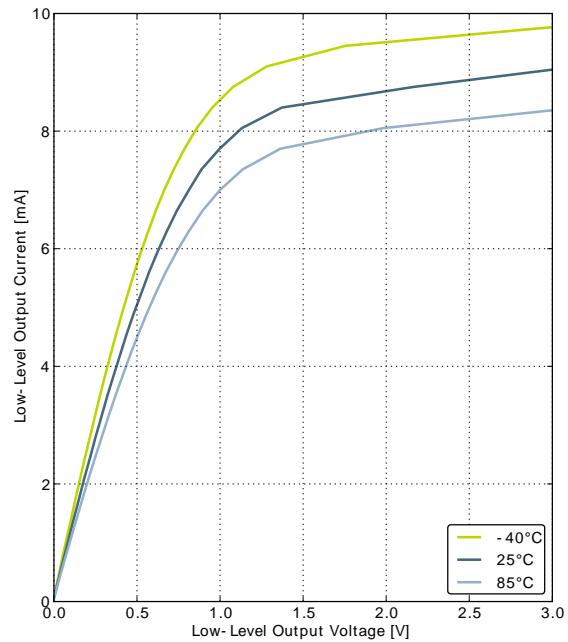
3.8 General Purpose Input Output

Table 3.7. GPIO

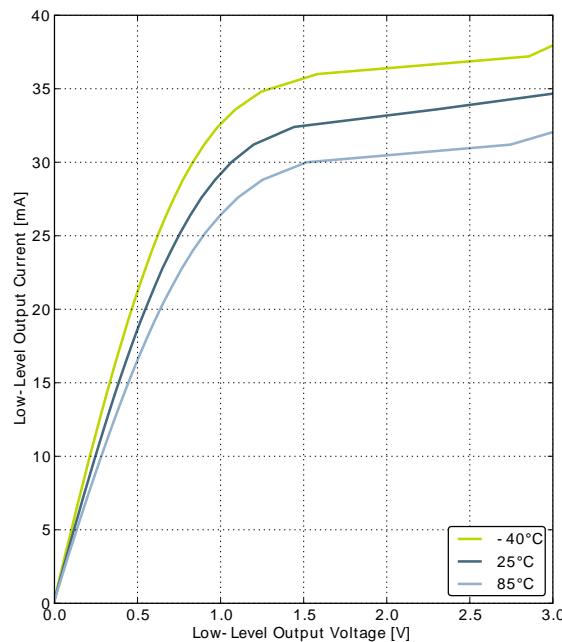
Symbol	Parameter	Condition	Min	Typ	Max	Unit
V_{IOIL}	Input low voltage				$0.30V_{DD}$	V
V_{IOIH}	Input high voltage		$0.70V_{DD}$			V
V_{IOOH}	Output high voltage (Production test condition = 3.0V, DRIVEMODE = STANDARD)	Sourcing 0.1 mA, $V_{DD}=1.98$ V, GPIO_Px_CTRL DRIVEMODE = LOWEST		$0.80V_{DD}$		V
		Sourcing 0.1 mA, $V_{DD}=3.0$ V, GPIO_Px_CTRL DRIVEMODE = LOWEST		$0.90V_{DD}$		V
		Sourcing 1 mA, $V_{DD}=1.98$ V, GPIO_Px_CTRL DRIVEMODE = LOW		$0.85V_{DD}$		V
		Sourcing 1 mA, $V_{DD}=3.0$ V, GPIO_Px_CTRL DRIVEMODE = LOW		$0.90V_{DD}$		V
		Sourcing 6 mA, $V_{DD}=1.98$ V, GPIO_Px_CTRL DRIVEMODE = STANDARD	$0.75V_{DD}$			V
		Sourcing 6 mA, $V_{DD}=3.0$ V, GPIO_Px_CTRL DRIVEMODE = STANDARD	$0.85V_{DD}$			V
		Sourcing 20 mA, $V_{DD}=1.98$ V, GPIO_Px_CTRL DRIVEMODE = HIGH	$0.60V_{DD}$			V
		Sourcing 20 mA, $V_{DD}=3.0$ V, GPIO_Px_CTRL DRIVEMODE = HIGH	$0.80V_{DD}$			V
V_{IOOL}	Output low voltage (Production test condition = 3.0V, DRIVEMODE = STANDARD)	Sinking 0.1 mA, $V_{DD}=1.98$ V, GPIO_Px_CTRL DRIVEMODE = LOWEST		$0.20V_{DD}$		V
		Sinking 0.1 mA, $V_{DD}=3.0$ V, GPIO_Px_CTRL DRIVEMODE = LOWEST		$0.10V_{DD}$		V
		Sinking 1 mA, $V_{DD}=1.98$ V, GPIO_Px_CTRL DRIVEMODE = LOW		$0.10V_{DD}$		V
		Sinking 1 mA, $V_{DD}=3.0$ V, GPIO_Px_CTRL DRIVEMODE = LOW		$0.05V_{DD}$		V
		Sinking 6 mA, $V_{DD}=1.98$ V, GPIO_Px_CTRL DRIVEMODE = STANDARD			$0.30V_{DD}$	V
		Sinking 6 mA, $V_{DD}=3.0$ V, GPIO_Px_CTRL DRIVEMODE = STANDARD			$0.20V_{DD}$	V
		Sinking 20 mA, $V_{DD}=1.98$ V, GPIO_Px_CTRL DRIVEMODE = HIGH			$0.35V_{DD}$	V

Figure 3.6. Typical Low-Level Output Current, 3V Supply Voltage

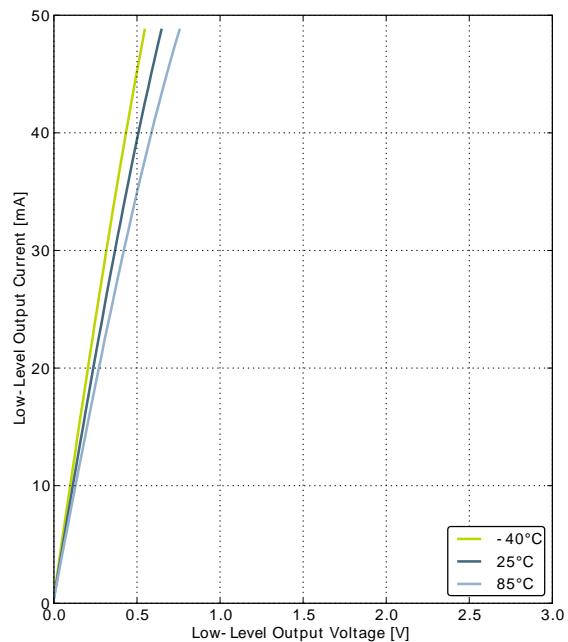
GPIO_Px_CTRL DRIVEMODE = LOWEST



GPIO_Px_CTRL DRIVEMODE = LOW



GPIO_Px_CTRL DRIVEMODE = STANDARD



GPIO_Px_CTRL DRIVEMODE = HIGH

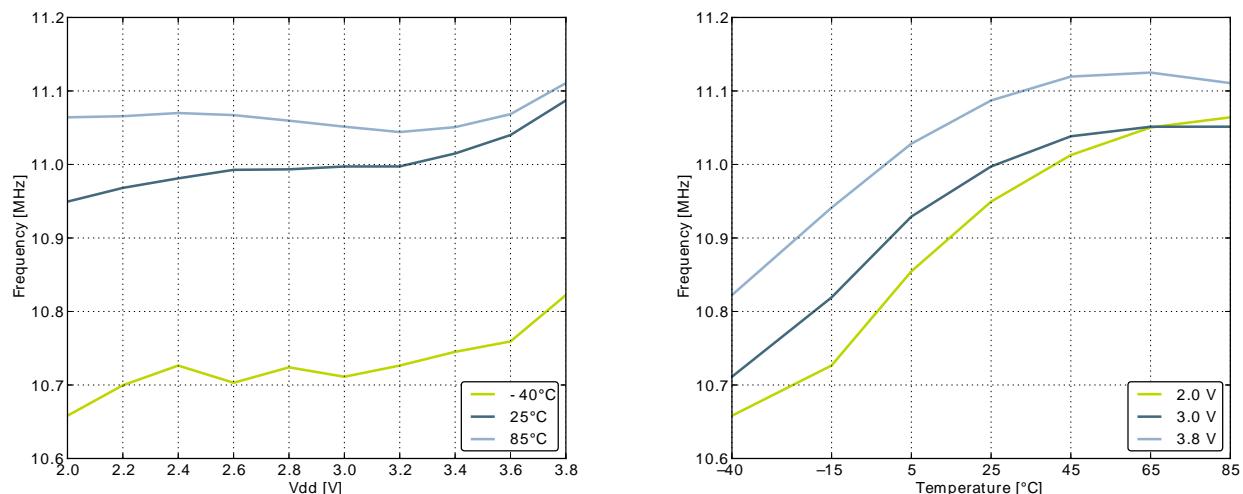
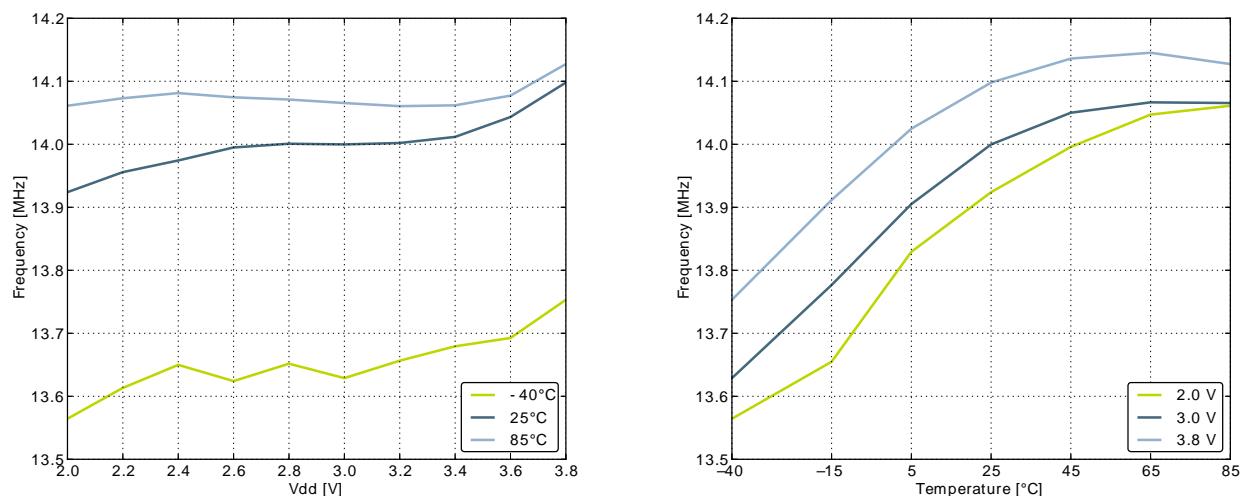
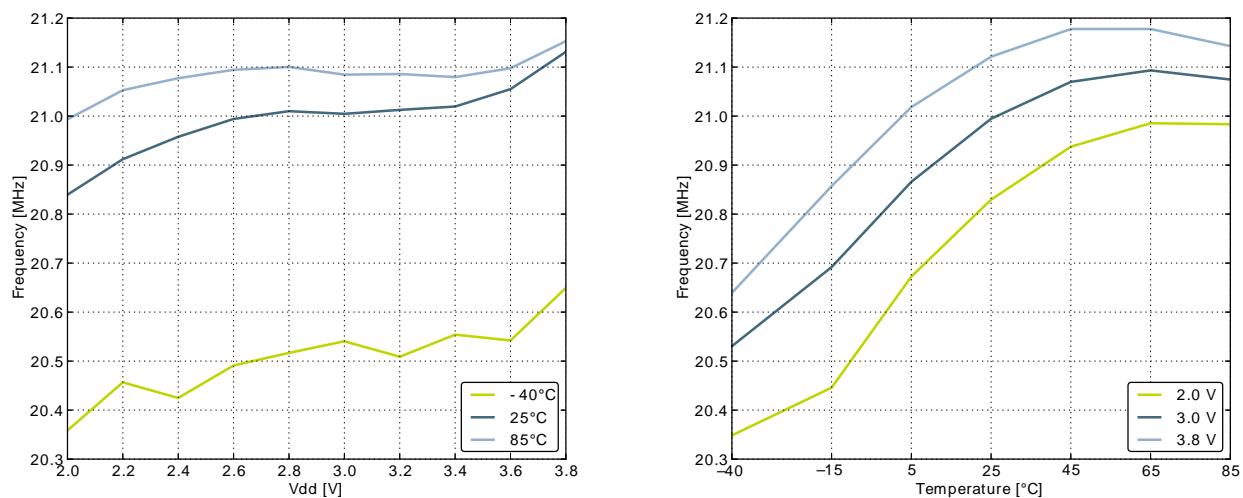
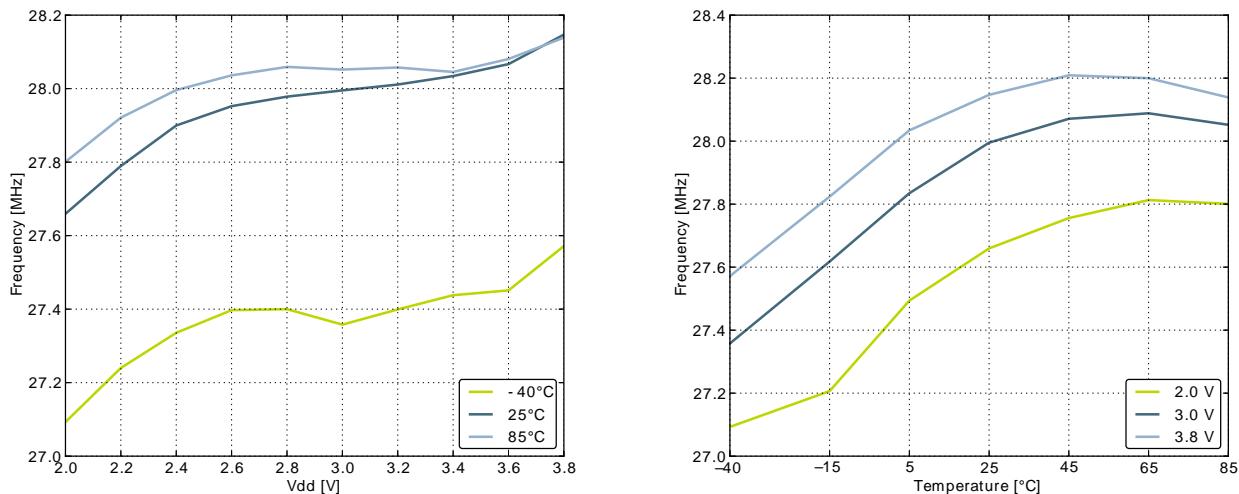
Figure 3.13. Calibrated HFRCO 11 MHz Band Frequency vs Supply Voltage and Temperature**Figure 3.14. Calibrated HFRCO 14 MHz Band Frequency vs Supply Voltage and Temperature****Figure 3.15. Calibrated HFRCO 21 MHz Band Frequency vs Supply Voltage and Temperature**

Figure 3.16. Calibrated HFRCO 28 MHz Band Frequency vs Supply Voltage and Temperature

3.9.5 AUXHFRCO

Table 3.12. AUXHFRCO

Symbol	Parameter	Condition	Min	Typ	Max	Unit
f_{AUXHFRCO}	Oscillation frequency, $V_{\text{DD}} = 3.0 \text{ V}$, $T_{\text{AMB}} = 25^\circ\text{C}$	28 MHz frequency band	27.5	28.0	28.5	MHz
		21 MHz frequency band	20.6	21.0	21.4	MHz
		14 MHz frequency band	13.7	14.0	14.3	MHz
		11 MHz frequency band	10.8	11.0	11.2	MHz
		7 MHz frequency band	6.48 ¹	6.60 ¹	6.72 ¹	MHz
		1 MHz frequency band	1.15 ²	1.20 ²	1.25 ²	MHz
$t_{\text{AUXHFRCO_settling}}$	Settling time after start-up	$f_{\text{AUXHFRCO}} = 14 \text{ MHz}$		0.6		Cycles
$\text{DC}_{\text{AUXHFRCO}}$	Duty cycle	$f_{\text{AUXHFRCO}} = 14 \text{ MHz}$	48.5	50	51	%
$\text{TUNESTEP}_{\text{AUXHFRCO}}$	Frequency step for LSB change in TUNING value			0.3 ³		%

¹For devices with prod. rev. < 19, Typ = 7MHz and Min/Max values not applicable.

²For devices with prod. rev. < 19, Typ = 1MHz and Min/Max values not applicable.

³The TUNING field in the CMU_AUXHFRCOCTRL register may be used to adjust the AUXHFRCO frequency. There is enough adjustment range to ensure that the frequency bands above 7 MHz will always have some overlap across supply voltage and temperature. By using a stable frequency reference such as the LFXO or HFXO, a firmware calibration routine can vary the TUNING bits and the frequency band to maintain the AUXHFRCO frequency at any arbitrary value between 7 MHz and 28 MHz across operating conditions.

Symbol	Parameter	Condition	Min	Typ	Max	Unit
C_{ADCIN}	Input capacitance			2		pF
R_{ADCIN}	Input ON resistance		1			MΩ
$R_{ADCFILT}$	Input RC filter resistance			10		kΩ
$C_{ADCFILT}$	Input RC filter/de-coupling capacitance			250		fF
f_{ADCCLK}	ADC Clock Frequency				13	MHz
$t_{ADCCONV}$	Conversion time	6 bit	7			ADC-CLK Cycles
		8 bit	11			ADC-CLK Cycles
		12 bit	13			ADC-CLK Cycles
t_{ADCACQ}	Acquisition time	Programmable	1		256	ADC-CLK Cycles
$t_{ADCACQVDD3}$	Required acquisition time for VDD/3 reference		2			μs
$t_{ADCSTART}$	Startup time of reference generator and ADC core in NORMAL mode			5		μs
	Startup time of reference generator and ADC core in KEEPADCWARM mode			1		μs
SNR_{ADC}	Signal to Noise Ratio (SNR)	1 MSamples/s, 12 bit, single ended, internal 1.25V reference		59		dB
		1 MSamples/s, 12 bit, single ended, internal 2.5V reference		63		dB
		1 MSamples/s, 12 bit, single ended, V_{DD} reference		65		dB
		1 MSamples/s, 12 bit, differential, internal 1.25V reference		60		dB
		1 MSamples/s, 12 bit, differential, internal 2.5V reference		65		dB
		1 MSamples/s, 12 bit, differential, 5V reference		54		dB
		1 MSamples/s, 12 bit, differential, V_{DD} reference		67		dB
		1 MSamples/s, 12 bit, differential, $2 \times V_{DD}$ reference		69		dB

Symbol	Parameter	Condition	Min	Typ	Max	Unit
		200 kSamples/s, 12 bit, single ended, internal 1.25V reference		62		dB
		200 kSamples/s, 12 bit, single ended, internal 2.5V reference		63		dB
		200 kSamples/s, 12 bit, single ended, V _{DD} reference		67		dB
		200 kSamples/s, 12 bit, differential, internal 1.25V reference		63		dB
		200 kSamples/s, 12 bit, differential, internal 2.5V reference		66		dB
		200 kSamples/s, 12 bit, differential, 5V reference		66		dB
		200 kSamples/s, 12 bit, differential, V _{DD} reference	63	66		dB
		200 kSamples/s, 12 bit, differential, 2xV _{DD} reference		70		dB
SINAD _{ADC}	Signal-to-Noise And Distortion-ratio (SINAD)	1 MSamples/s, 12 bit, single ended, internal 1.25V reference		58		dB
		1 MSamples/s, 12 bit, single ended, internal 2.5V reference		62		dB
		1 MSamples/s, 12 bit, single ended, V _{DD} reference		64		dB
		1 MSamples/s, 12 bit, differential, internal 1.25V reference		60		dB
		1 MSamples/s, 12 bit, differential, internal 2.5V reference		64		dB
		1 MSamples/s, 12 bit, differential, 5V reference		54		dB
		1 MSamples/s, 12 bit, differential, V _{DD} reference		66		dB
		1 MSamples/s, 12 bit, differential, 2xV _{DD} reference		68		dB
		200 kSamples/s, 12 bit, single ended, internal 1.25V reference		61		dB
		200 kSamples/s, 12 bit, single ended, internal 2.5V reference		65		dB
		200 kSamples/s, 12 bit, single ended, V _{DD} reference		66		dB
		200 kSamples/s, 12 bit, differential, internal 1.25V reference		63		dB
		200 kSamples/s, 12 bit, differential, internal 2.5V reference		66		dB
		200 kSamples/s, 12 bit, differential, 5V reference		66		dB
		200 kSamples/s, 12 bit, differential, V _{DD} reference	62	65		dB

Symbol	Parameter	Condition	Min	Typ	Max	Unit
GAIN _{ED}	Gain error drift	1.25V reference		0.01 ²	0.033 ³	%/°C
		2.5V reference		0.01 ²	0.03 ³	%/°C
OFFSET _{ED}	Offset error drift	1.25V reference		0.2 ²	0.7 ³	LSB/°C
		2.5V reference		0.2 ²	0.62 ³	LSB/°C

¹On the average every ADC will have one missing code, most likely to appear around $2048 +/ - n \cdot 512$ where n can be a value in the set $\{-3, -2, -1, 1, 2, 3\}$. There will be no missing code around 2048, and in spite of the missing code the ADC will be monotonic at all times so that a response to a slowly increasing input will always be a slowly increasing output. Around the one code that is missing, the neighbour codes will look wider in the DNL plot. The spectra will show spurs on the level of -78dBc for a full scale input for chips that have the missing code issue.

²Typical numbers given by $\text{abs}(\text{Mean}) / (85 - 25)$.

³Max number given by $(\text{abs}(\text{Mean}) + 3 \times \text{stddev}) / (85 - 25)$.

The integral non-linearity (INL) and differential non-linearity parameters are explained in Figure 3.17 (p. 31) and Figure 3.18 (p. 32), respectively.

Figure 3.17. Integral Non-Linearity (INL)

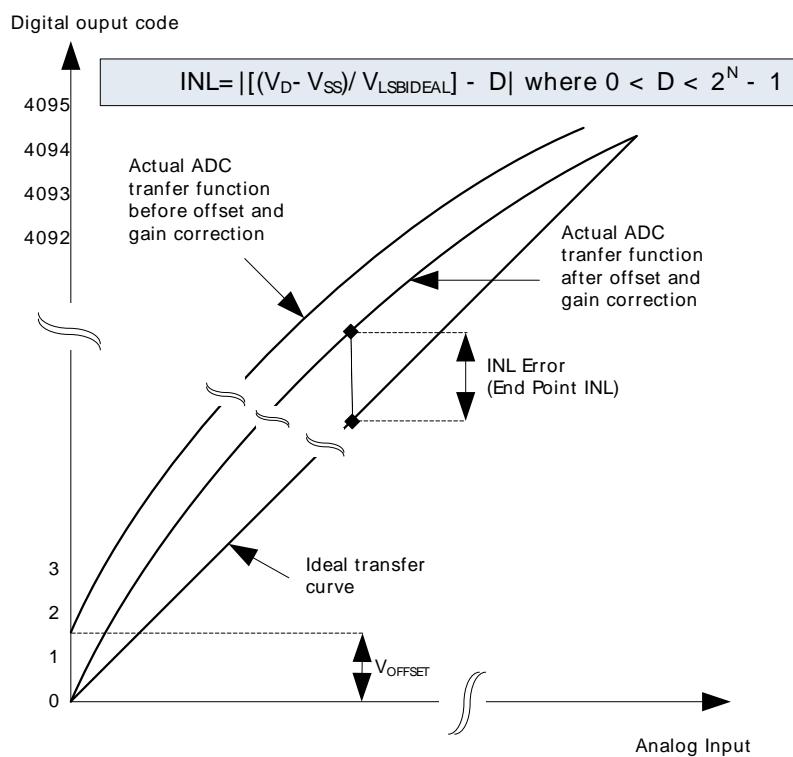
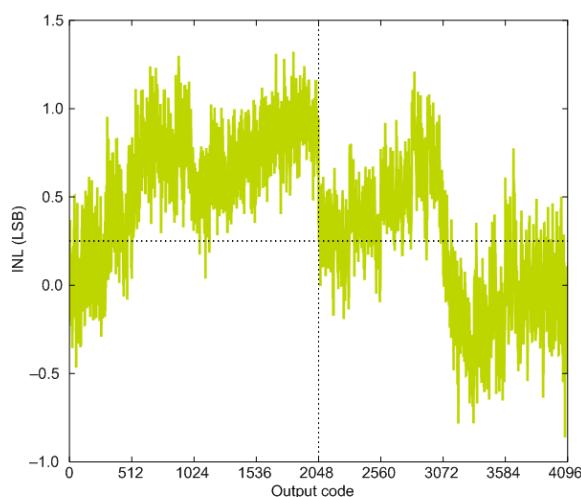
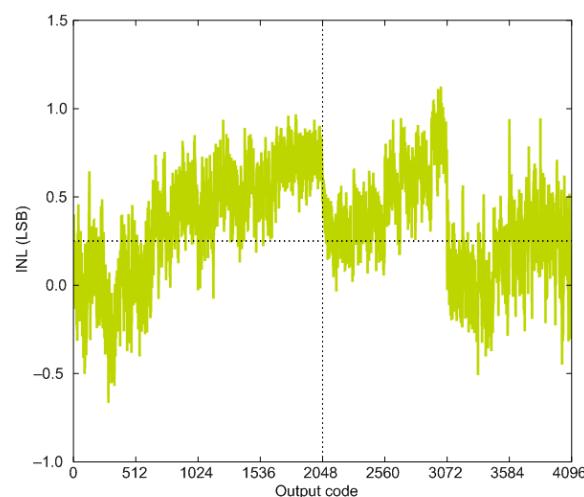
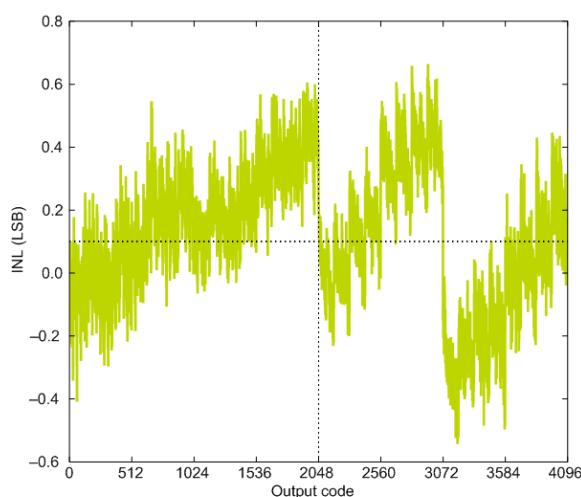


Figure 3.20. ADC Integral Linearity Error vs Code, Vdd = 3V, Temp = 25°C

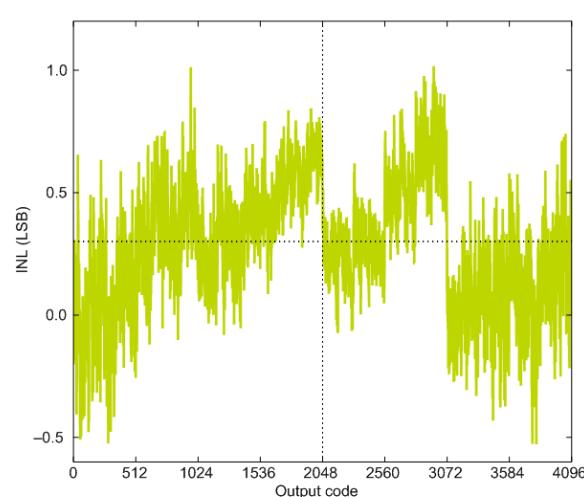
1.25V Reference



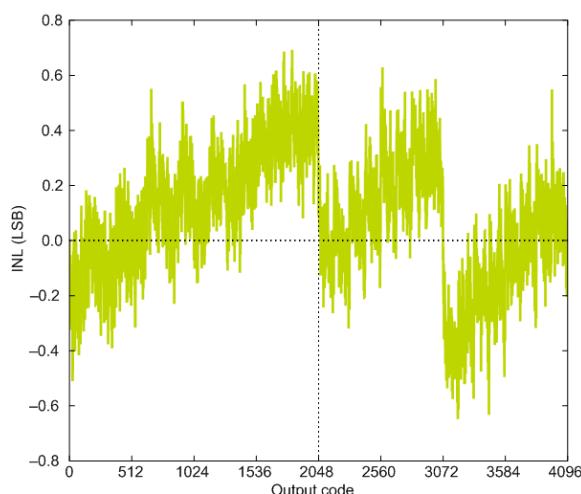
2.5V Reference



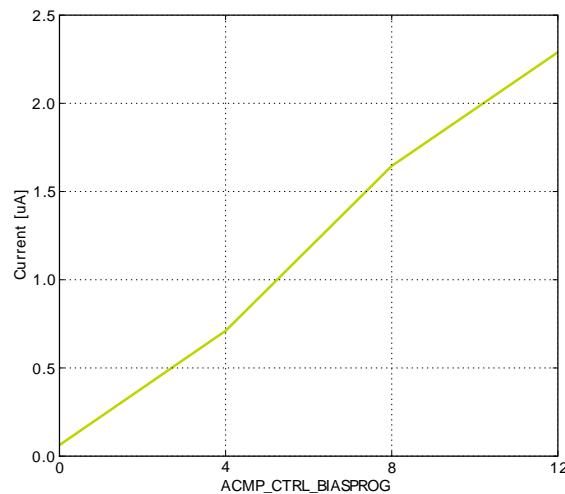
2XVDDVSS Reference



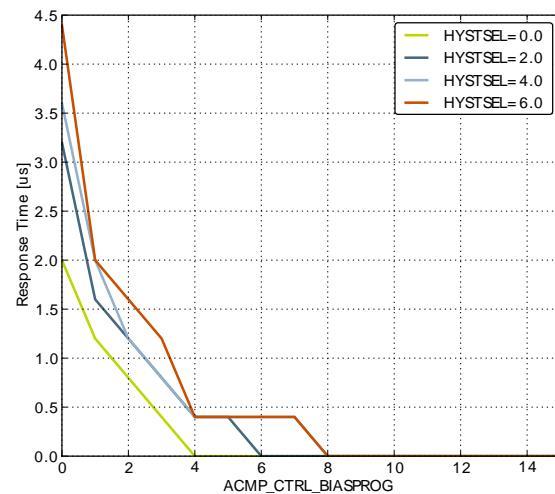
5VDIFF Reference



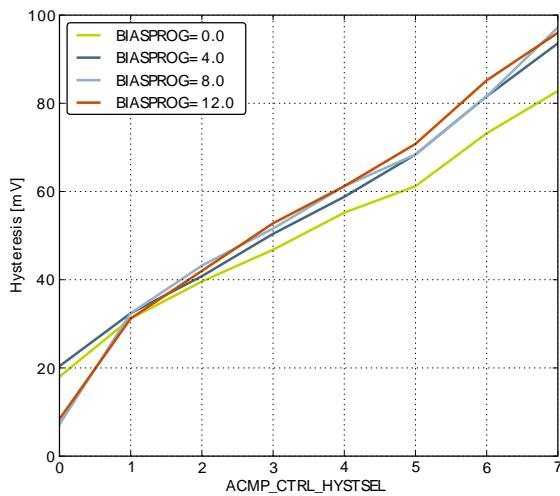
VDD Reference

Figure 3.30. ACMP Characteristics, Vdd = 3V, Temp = 25°C, FULLBIAS = 0, HALFBIAS = 1

Current consumption, HYSTSEL = 4



Response time



Hysteresis

3.14 Voltage Comparator (VCMP)

Table 3.18. VCMP

Symbol	Parameter	Condition	Min	Typ	Max	Unit
$V_{VCMPPIN}$	Input voltage range			V_{DD}		V
V_{VCMPCM}	VCMP Common Mode voltage range			V_{DD}		V
I_{VCMP}	Active current	BIASPROG=0b0000 and HALFBIAS=1 in VCMPn_CTRL register		0.3	0.6	μA
		BIASPROG=0b1111 and HALFBIAS=0 in VCMPn_CTRL register. LPREF=0.		22	30	μA
$t_{VCMPREF}$	Startup time reference generator	NORMAL		10		μs
$V_{VCMPOFFSET}$	Offset voltage	Single ended	-230	-40	190	mV
		Differential		10		mV
$V_{VCMPHYST}$	VCMP hysteresis			40		mV
$t_{VCMPSTART}$	Startup time				10	μs

The V_{DD} trigger level can be configured by setting the TRIGLEVEL field of the VCMP_CTRL register in accordance with the following equation:

VCMP Trigger Level as a Function of Level Setting

$$V_{DD \text{ Trigger Level}} = 1.667V + 0.034 \times \text{TRIGLEVEL} \quad (3.2)$$

3.15 I2C

Table 3.19. I2C Standard-mode (Sm)

Symbol	Parameter	Min	Typ	Max	Unit
f_{SCL}	SCL clock frequency	0		100^1	kHz
t_{LOW}	SCL clock low time	4.7			μs
t_{HIGH}	SCL clock high time	4.0			μs
$t_{SU,DAT}$	SDA set-up time	250			ns
$t_{HD,DAT}$	SDA hold time	8		$3450^{2,3}$	ns
$t_{SU,STA}$	Repeated START condition set-up time	4.7			μs
$t_{HD,STA}$	(Repeated) START condition hold time	4.0			μs
$t_{SU,STO}$	STOP condition set-up time	4.0			μs
t_{BUF}	Bus free time between a STOP and START condition	4.7			μs

¹For the minimum HFPERCLK frequency required in Standard-mode, see the I2C chapter in the EFM32GG Reference Manual.

²The maximum SDA hold time ($t_{HD,DAT}$) needs to be met only when the device does not stretch the low time of SCL (t_{LOW}).

³When transmitting data, this number is guaranteed only when $I2Cn_CLKDIV < ((3450 * 10^{-9}) [s] * f_{HFPERCLK} [\text{Hz}]) - 4$.

Alternate	LOCATION							
Functionality	0	1	2	3	4	5	6	Description
ACMP1_CH3	PC11							Analog comparator ACMP1, channel 3.
ACMP1_CH4	PC12							Analog comparator ACMP1, channel 4.
ACMP1_CH5	PC13							Analog comparator ACMP1, channel 5.
ACMP1_CH6	PC14							Analog comparator ACMP1, channel 6.
ACMP1_CH7	PC15							Analog comparator ACMP1, channel 7.
ACMP1_O	PF2		PD7					Analog comparator ACMP1, digital output.
ADC0_CH0	PD0							Analog to digital converter ADC0, input channel number 0.
ADC0_CH1	PD1							Analog to digital converter ADC0, input channel number 1.
ADC0_CH2	PD2							Analog to digital converter ADC0, input channel number 2.
ADC0_CH3	PD3							Analog to digital converter ADC0, input channel number 3.
ADC0_CH4	PD4							Analog to digital converter ADC0, input channel number 4.
ADC0_CH5	PD5							Analog to digital converter ADC0, input channel number 5.
ADC0_CH6	PD6							Analog to digital converter ADC0, input channel number 6.
ADC0_CH7	PD7							Analog to digital converter ADC0, input channel number 7.
BOOT_RX	PE11							Bootloader RX.
BOOT_TX	PE10							Bootloader TX.
BU_VIN	PD8							Battery input for Backup Power Domain
CMU_CLK0	PA2	PC12	PD7					Clock Management Unit, clock output number 0.
CMU_CLK1	PA1	PD8	PE12					Clock Management Unit, clock output number 1.
OPAMP_N0	PC5							Operational Amplifier 0 external negative input.
OPAMP_N1	PD7							Operational Amplifier 1 external negative input.
OPAMP_N2	PD3							Operational Amplifier 2 external negative input.
DAC0_OUT0 / OPAMP_OUT0	PB11							Digital to Analog Converter DAC0_OUT0 / OPAMP output channel number 0.
DAC0_OUT0ALT / OPAMP_OUT0ALT	PC0	PC1	PC2	PC3	PD0			Digital to Analog Converter DAC0_OUT0ALT / OPAMP alternative output for channel 0.
DAC0_OUT1 / OPAMP_OUT1	PB12							Digital to Analog Converter DAC0_OUT1 / OPAMP output channel number 1.
DAC0_OUT1ALT / OPAMP_OUT1ALT	PC12	PC13	PC14	PC15	PD1			Digital to Analog Converter DAC0_OUT1ALT / OPAMP alternative output for channel 1.
OPAMP_OUT2	PD5	PD0						Operational Amplifier 2 output.
OPAMP_P0	PC4							Operational Amplifier 0 external positive input.
OPAMP_P1	PD6							Operational Amplifier 1 external positive input.
OPAMP_P2	PD4							Operational Amplifier 2 external positive input.
DBG_SWCLK	PF0	PF0	PF0	PF0				Debug-interface Serial Wire clock input. Note that this function is enabled to pin out of reset, and has a built-in pull down.
DBG_SWDIO	PF1	PF1	PF1	PF1				Debug-interface Serial Wire data input / output. Note that this function is enabled to pin out of reset, and has a built-in pull up.
DBG_SWO	PF2	PC15	PD1	PD2				Debug-interface Serial Wire viewer Output. Note that this function is not enabled after reset, and must be enabled by software to be used.
ETM_TCLK	PD7		PC6	PA6				Embedded Trace Module ETM clock .
ETM_TD0	PD6		PC7	PA2				Embedded Trace Module ETM data 0.

Alternate	LOCATION													
Functionality	0	1	2	3	4	5	6	Description						
US1_RX	PC1	PD1	PD6					USART1 Asynchronous Receive. USART1 Synchronous mode Master Input / Slave Output (MISO).						
US1_TX	PC0	PD0	PD7					USART1 Asynchronous Transmit.Also used as receive input in half duplex communication. USART1 Synchronous mode Master Output / Slave Input (MOSI).						
US2_CLK	PC4							USART2 clock input / output.						
US2_CS	PC5							USART2 chip select input / output.						
US2_RX	PC3							USART2 Asynchronous Receive. USART2 Synchronous mode Master Input / Slave Output (MISO).						
US2_TX	PC2							USART2 Asynchronous Transmit.Also used as receive input in half duplex communication. USART2 Synchronous mode Master Output / Slave Input (MOSI).						

4.3 GPIO Pinout Overview

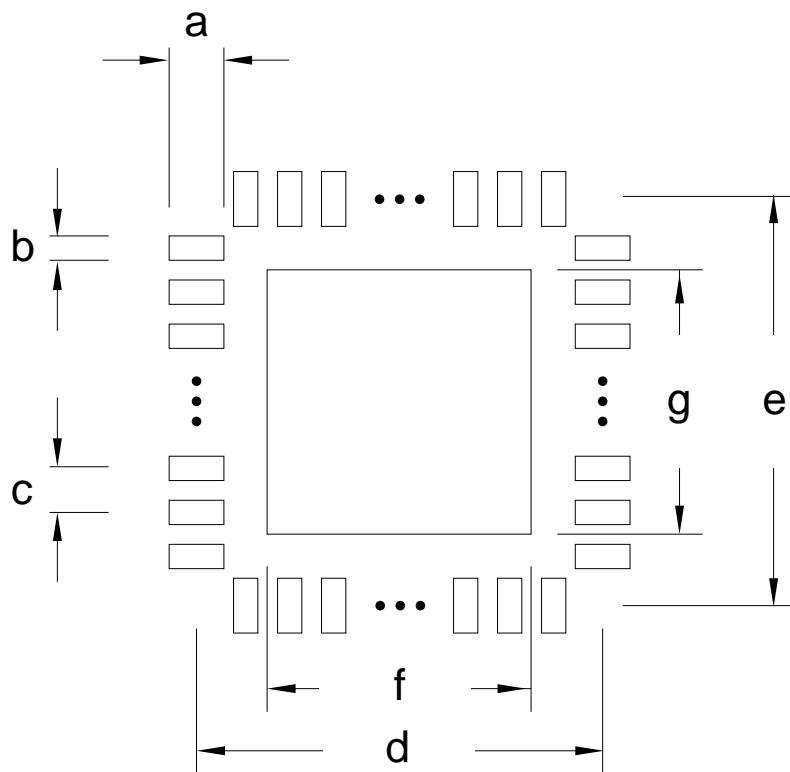
The specific GPIO pins available in *EFM32GG230* is shown in Table 4.3 (p. 55). Each GPIO port is organized as 16-bit ports indicated by letters A through F, and the individual pin on this port is indicated by a number from 15 down to 0.

Table 4.3. GPIO Pinout

Port	Pin 15	Pin 14	Pin 13	Pin 12	Pin 11	Pin 10	Pin 9	Pin 8	Pin 7	Pin 6	Pin 5	Pin 4	Pin 3	Pin 2	Pin 1	Pin 0
Port A	PA15	-	-	-	-	PA10	PA9	PA8	-	PA6	PA5	PA4	PA3	PA2	PA1	PA0
Port B	-	PB14	PB13	PB12	PB11	-	-	PB8	PB7	-	-	-	-	-	-	-
Port C	PC15	PC14	PC13	PC12	PC11	PC10	PC9	PC8	PC7	PC6	PC5	PC4	PC3	PC2	PC1	PC0
Port D	-	-	-	-	-	-	-	PD8	PD7	PD6	PD5	PD4	PD3	PD2	PD1	PD0
Port E	PE15	PE14	PE13	PE12	PE11	PE10	PE9	PE8	-	-	-	-	-	-	-	-
Port F	-	-	-	-	-	-	-	-	-	-	PF5	PF4	PF3	PF2	PF1	PF0

4.4 Opamp Pinout Overview

The specific opamp terminals available in *EFM32GG230* is shown in Figure 4.2 (p. 56) .

Figure 5.2. QFN64 PCB Solder Mask**Table 5.2. QFN64 PCB Solder Mask Dimensions (Dimensions in mm)**

Symbol	Dim. (mm)	Symbol	Dim. (mm)
a	0.97	e	8.90
b	0.42	f	7.32
c	0.50	g	7.32
d	8.90	-	-

Added new alternative locations for SWO.

Corrected slew rate data for Opamps.

7.11 Revision 0.90

February 4th, 2011

Initial preliminary release.

B Contact Information

Silicon Laboratories Inc.
400 West Cesar Chavez
Austin, TX 78701

Please visit the Silicon Labs Technical Support web page:
<http://www.silabs.com/support/pages/contacttechnicalsupport.aspx>
and register to submit a technical support request.

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