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Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	600
Number of Logic Elements/Cells	2700
Total RAM Bits	40960
Number of I/O	180
Number of Gates	108904
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	256-BBGA
Supplier Device Package	256-PBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcv100-5bg256i

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

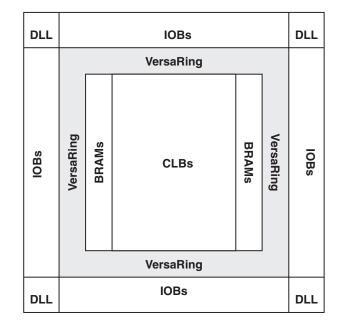


DS003-2 (v4.0) March 1, 2013

Virtex[™] 2.5 V Field Programmable Gate Arrays

Product Specification

The output buffer and all of the IOB control signals have independent polarity controls.



vao_b.eps

Figure 1: Virtex Architecture Overview

All pads are protected against damage from electrostatic discharge (ESD) and from over-voltage transients. Two forms of over-voltage protection are provided, one that permits 5 V compliance, and one that does not. For 5 V compliance, a Zener-like structure connected to ground turns on when the output rises to approximately 6.5 V. When PCI 3.3 V compliance is required, a conventional clamp diode is connected to the output supply voltage, $V_{\rm CCO}$.

Optional pull-up and pull-down resistors and an optional weak-keeper circuit are attached to each pad. Prior to configuration, all pins not involved in configuration are forced into their high-impedance state. The pull-down resistors and the weak-keeper circuits are inactive, but inputs can optionally be pulled up.

The activation of pull-up resistors prior to configuration is controlled on a global basis by the configuration mode pins. If the pull-up resistors are not activated, all the pins will float. Consequently, external pull-up or pull-down resistors must be provided on pins required to be at a well-defined logic level prior to configuration.

All Virtex IOBs support IEEE 1149.1-compatible boundary scan testing.

Architectural Description

Virtex Array

The Virtex user-programmable gate array, shown in Figure 1, comprises two major configurable elements: configurable logic blocks (CLBs) and input/output blocks (IOBs).

- CLBs provide the functional elements for constructing logic
- IOBs provide the interface between the package pins and the CLBs

CLBs interconnect through a general routing matrix (GRM). The GRM comprises an array of routing switches located at the intersections of horizontal and vertical routing channels. Each CLB nests into a VersaBlock™ that also provides local routing resources to connect the CLB to the GRM.

The VersaRing[™] I/O interface provides additional routing resources around the periphery of the device. This routing improves I/O routability and facilitates pin locking.

The Virtex architecture also includes the following circuits that connect to the GRM.

- Dedicated block memories of 4096 bits each
- Clock DLLs for clock-distribution delay compensation and clock domain control
- 3-State buffers (BUFTs) associated with each CLB that drive dedicated segmentable horizontal routing resources

Values stored in static memory cells control the configurable logic elements and interconnect resources. These values load into the memory cells on power-up, and can reload if necessary to change the function of the device.

Input/Output Block

The Virtex IOB, Figure 2, features SelectIO™ inputs and outputs that support a wide variety of I/O signalling standards, see Table 1.

The three IOB storage elements function either as edge-triggered D-type flip-flops or as level sensitive latches. Each IOB has a clock signal (CLK) shared by the three flip-flops and independent clock enable signals for each flip-flop.

In addition to the CLK and CE control signals, the three flip-flops share a Set/Reset (SR). For each flip-flop, this signal can be independently configured as a synchronous Set, a synchronous Reset, an asynchronous Preset, or an asynchronous Clear.

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Input Path

A buffer In the Virtex IOB input path routes the input signal either directly to internal logic or through an optional input flip-flop.

An optional delay element at the D-input of this flip-flop eliminates pad-to-pad hold time. The delay is matched to the internal clock-distribution delay of the FPGA, and when used, assures that the pad-to-pad hold time is zero.

Each input buffer can be configured to conform to any of the low-voltage signalling standards supported. In some of these standards the input buffer utilizes a user-supplied threshold voltage, V_{REF}. The need to supply V_{REF} imposes constraints on which standards can used in close proximity to each other. See I/O Banking, page 3.

There are optional pull-up and pull-down resistors at each user I/O input for use after configuration. Their value is in the range 50 k Ω – 100 k Ω .

Output Path

The output path includes a 3-state output buffer that drives the output signal onto the pad. The output signal can be routed to the buffer directly from the internal logic or through an optional IOB output flip-flop.

The 3-state control of the output can also be routed directly from the internal logic or through a flip-flip that provides synchronous enable and disable.

Each output driver can be individually programmed for a wide range of low-voltage signalling standards. Each output buffer can source up to 24 mA and sink up to 48mA. Drive strength and slew rate controls minimize bus transients.

In most signalling standards, the output High voltage depends on an externally supplied V_{CCO} voltage. The need to supply V_{CCO} imposes constraints on which standards can be used in close proximity to each other. See **I/O Banking**, page 3.

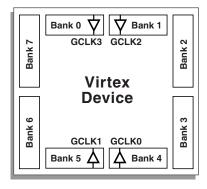
An optional weak-keeper circuit is connected to each output. When selected, the circuit monitors the voltage on the pad and weakly drives the pin High or Low to match the input signal. If the pin is connected to a multiple-source signal, the weak keeper holds the signal in its last state if all drivers are disabled. Maintaining a valid logic level in this way eliminates bus chatter.

Because the weak-keeper circuit uses the IOB input buffer to monitor the input level, an appropriate V_{REF} voltage must be provided if the signalling standard requires one. The provision of this voltage must comply with the I/O banking rules.

I/O Banking

Some of the I/O standards described above require V_{CCO} and/or V_{REF} voltages. These voltages externally and connected to device pins that serve groups of IOBs, called banks. Consequently, restrictions exist about which I/O standards can be combined within a given bank.

Eight I/O banks result from separating each edge of the FPGA into two banks, as shown in Figure 3. Each bank has multiple $V_{\rm CCO}$ pins, all of which must be connected to the same voltage. This voltage is determined by the output standards in use.



X8778_b

Figure 3: Virtex I/O Banks

Within a bank, output standards can be mixed only if they use the same V_{CCO} . Compatible standards are shown in Table 2. GTL and GTL+ appear under all voltages because their open-drain outputs do not depend on V_{CCO} .

Table 2: Compatible Output Standards

V _{CCO}	Compatible Standards
3.3 V	PCI, LVTTL, SSTL3 I, SSTL3 II, CTT, AGP, GTL, GTL+
2.5 V	SSTL2 I, SSTL2 II, LVCMOS2, GTL, GTL+
1.5 V	HSTL I, HSTL III, HSTL IV, GTL, GTL+

Some input standards require a user-supplied threshold voltage, V_{REF} In this case, certain user-I/O pins are automatically configured as inputs for the V_{REF} voltage. Approximately one in six of the I/O pins in the bank assume this role

The V_{REF} pins within a bank are interconnected internally and consequently only one V_{REF} voltage can be used within each bank. All V_{REF} pins in the bank, however, must be connected to the external voltage source for correct operation.

Within a bank, inputs that require V_{REF} can be mixed with those that do not. However, only one V_{REF} voltage can be used within a bank. Input buffers that use V_{REF} are not 5 V tolerant. LVTTL, LVCMOS2, and PCI 33 MHz 5 V, are 5 V tolerant.

The V_{CCO} and V_{REF} pins for each bank appear in the device Pinout tables and diagrams. The diagrams also show the bank affiliation of each I/O.

Within a given package, the number of V_{REF} and V_{CCO} pins can vary depending on the size of device. In larger devices,



more I/O pins convert to V_{REF} pins. Since these are always a superset of the V_{REF} pins used for smaller devices, it is possible to design a PCB that permits migration to a larger device if necessary. All the V_{REF} pins for the largest device anticipated must be connected to the V_{REF} voltage, and not used for I/O.

In smaller devices, some V_{CCO} pins used in larger devices do not connect within the package. These unconnected pins can be left unconnected externally, or can be connected to the V_{CCO} voltage to permit migration to a larger device if necessary.

In TQ144 and PQ/HQ240 packages, all V_{CCO} pins are bonded together internally, and consequently the same V_{CCO} voltage must be connected to all of them. In the CS144 package, bank pairs that share a side are interconnected internally, permitting four choices for V_{CCO} . In both cases, the V_{REF} pins remain internally connected as eight banks, and can be used as described previously.

Configurable Logic Block

The basic building block of the Virtex CLB is the logic cell (LC). An LC includes a 4-input function generator, carry logic, and a storage element. The output from the function generator in each LC drives both the CLB output and the D input of the flip-flop. Each Virtex CLB contains four LCs, organized in two similar slices, as shown in Figure 4.

Figure 5 shows a more detailed view of a single slice.

In addition to the four basic LCs, the Virtex CLB contains logic that combines function generators to provide functions

of five or six inputs. Consequently, when estimating the number of system gates provided by a given device, each CLB counts as 4.5 LCs.

Look-Up Tables

Virtex function generators are implemented as 4-input look-up tables (LUTs). In addition to operating as a function generator, each LUT can provide a 16 x 1-bit synchronous RAM. Furthermore, the two LUTs within a slice can be combined to create a 16 x 2-bit or 32 x 1-bit synchronous RAM, or a 16x1-bit dual-port synchronous RAM.

The Virtex LUT can also provide a 16-bit shift register that is ideal for capturing high-speed or burst-mode data. This mode can also be used to store data in applications such as Digital Signal Processing.

Storage Elements

The storage elements in the Virtex slice can be configured either as edge-triggered D-type flip-flops or as level-sensitive latches. The D inputs can be driven either by the function generators within the slice or directly from slice inputs, bypassing the function generators.

In addition to Clock and Clock Enable signals, each Slice has synchronous set and reset signals (SR and BY). SR forces a storage element into the initialization state specified for it in the configuration. BY forces it into the opposite state. Alternatively, these signals can be configured to operate asynchronously. All of the control signals are independently invertible, and are shared by the two flip-flops within the slice.

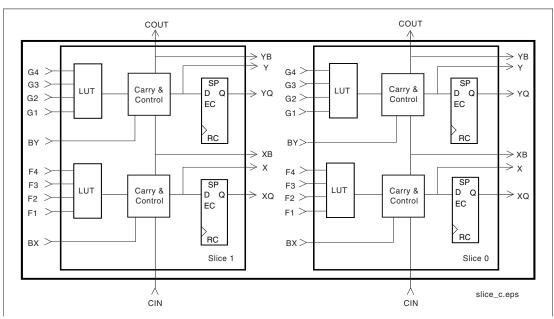


Figure 4: 2-Slice Virtex CLB

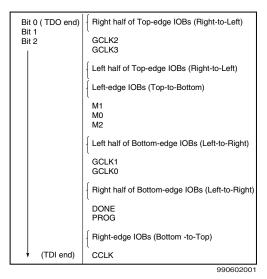


Figure 11: Boundary Scan Bit Sequence

Table 5: Boundary Scan Instructions

Boundary-Scan Command	Binary Code(4:0)	Description
EXTEST	00000	Enables boundary-scan EXTEST operation
SAMPLE/PRELOAD	00001	Enables boundary-scan SAMPLE/PRELOAD operation
USER 1	00010	Access user-defined register 1
USER 2	00011	Access user-defined register 2
CFG_OUT	00100	Access the configuration bus for read operations.
CFG_IN	00101	Access the configuration bus for write operations.
INTEST	00111	Enables boundary-scan INTEST operation
USERCODE	01000	Enables shifting out USER code
IDCODE	01001	Enables shifting out of ID Code
HIGHZ	01010	3-states output pins while enabling the Bypass Register
JSTART	01100	Clock the start-up sequence when StartupClk is TCK
BYPASS	11111	Enables BYPASS
RESERVED	All other codes	Xilinx reserved instructions

Identification Registers

The IDCODE register is supported. By using the IDCODE, the device connected to the JTAG port can be determined.

The IDCODE register has the following binary format:

vvvv:ffff:fffa:aaaa:aaaa:cccc:cccc1

where

v = the die version number

f = the family code (03h for Virtex family)

a = the number of CLB rows (ranges from 010h for XCV50 to 040h for XCV1000)

c = the company code (49h for Xilinx)

The USERCODE register is supported. By using the USER-CODE, a user-programmable identification code can be loaded and shifted out for examination. The identification code is embedded in the bitstream during bitstream generation and is valid only after configuration.

Table 6: IDCODEs Assigned to Virtex FPGAs

FPGA	IDCODE
XCV50	v0610093h
XCV100	v0614093h
XCV150	v0618093h
XCV200	v061C093h
XCV300	v0620093h
XCV400	v0628093h
XCV600	v0630093h
XCV800	v0638093h
XCV1000	v0640093h

Including Boundary Scan in a Design

Since the boundary scan pins are dedicated, no special element needs to be added to the design unless an internal data register (USER1 or USER2) is desired.

If an internal data register is used, insert the boundary scan symbol and connect the necessary pins as appropriate.

Development System

Virtex FPGAs are supported by the Xilinx Foundation and Alliance CAE tools. The basic methodology for Virtex design consists of three interrelated steps: design entry, implementation, and verification. Industry-standard tools are used for design entry and simulation (for example, Synopsys FPGA Express), while Xilinx provides proprietary architecture-specific tools for implementation.

The Xilinx development system is integrated under the Xilinx Design Manager (XDM™) software, providing design-



ers with a common user interface regardless of their choice of entry and verification tools. The XDM software simplifies the selection of implementation options with pull-down menus and on-line help.

Application programs ranging from schematic capture to Placement and Routing (PAR) can be accessed through the XDM software. The program command sequence is generated prior to execution, and stored for documentation.

Several advanced software features facilitate Virtex design. RPMs, for example, are schematic-based macros with relative location constraints to guide their placement. They help ensure optimal implementation of common functions.

For HDL design entry, the Xilinx FPGA Foundation development system provides interfaces to the following synthesis design environments.

- Synopsys (FPGA Compiler, FPGA Express)
- Exemplar (Spectrum)
- Synplicity (Synplify)

For schematic design entry, the Xilinx FPGA Foundation and alliance development system provides interfaces to the following schematic-capture design environments.

- Mentor Graphics V8 (Design Architect, QuickSim II)
- Viewlogic Systems (Viewdraw)

Third-party vendors support many other environments.

A standard interface-file specification, Electronic Design Interchange Format (EDIF), simplifies file transfers into and out of the development system.

Virtex FPGAs supported by a unified library of standard functions. This library contains over 400 primitives and macros, ranging from 2-input AND gates to 16-bit accumulators, and includes arithmetic functions, comparators, counters, data registers, decoders, encoders, I/O functions, latches, Boolean functions, multiplexers, shift registers, and barrel shifters.

The "soft macro" portion of the library contains detailed descriptions of common logic functions, but does not contain any partitioning or placement information. The performance of these macros depends, therefore, on the partitioning and placement obtained during implementation.

RPMs, on the other hand, do contain predetermined partitioning and placement information that permits optimal implementation of these functions. Users can create their own library of soft macros or RPMs based on the macros and primitives in the standard library.

The design environment supports hierarchical design entry, with high-level schematics that comprise major functional blocks, while lower-level schematics define the logic in these blocks. These hierarchical design elements are automatically combined by the implementation tools. Different design entry tools can be combined within a hierarchical

design, thus allowing the most convenient entry method to be used for each portion of the design.

Design Implementation

The place-and-route tools (PAR) automatically provide the implementation flow described in this section. The partitioner takes the EDIF net list for the design and maps the logic into the architectural resources of the FPGA (CLBs and IOBs, for example). The placer then determines the best locations for these blocks based on their interconnections and the desired performance. Finally, the router interconnects the blocks.

The PAR algorithms support fully automatic implementation of most designs. For demanding applications, however, the user can exercise various degrees of control over the process. User partitioning, placement, and routing information is optionally specified during the design-entry process. The implementation of highly structured designs can benefit greatly from basic floor planning.

The implementation software incorporates Timing Wizard® timing-driven placement and routing. Designers specify timing requirements along entire paths during design entry. The timing path analysis routines in PAR then recognize these user-specified requirements and accommodate them.

Timing requirements are entered on a schematic in a form directly relating to the system requirements, such as the targeted clock frequency, or the maximum allowable delay between two registers. In this way, the overall performance of the system along entire signal paths is automatically tailored to user-generated specifications. Specific timing information for individual nets is unnecessary.

Design Verification

In addition to conventional software simulation, FPGA users can use in-circuit debugging techniques. Because Xilinx devices are infinitely reprogrammable, designs can be verified in real time without the need for extensive sets of software simulation vectors.

The development system supports both software simulation and in-circuit debugging techniques. For simulation, the system extracts the post-layout timing information from the design database, and back-annotates this information into the net list for use by the simulator. Alternatively, the user can verify timing-critical portions of the design using the TRACE® static timing analyzer.

For in-circuit debugging, the development system includes a download and readback cable. This cable connects the FPGA in the target system to a PC or workstation. After downloading the design into the FPGA, the designer can single-step the logic, readback the contents of the flip-flops, and so observe the internal logic state. Simple modifications can be downloaded into the system in a matter of minutes.



Virtex[™] 2.5 V Field Programmable Gate Arrays

DS003-3 (v4.0) March 1, 2013

Production Product Specification

Virtex Electrical Characteristics Definition of Terms

Electrical and switching characteristics are specified on a per-speed-grade basis and can be designated as Advance, Preliminary, or Production. Each designation is defined as follows:

Advance: These speed files are based on simulations only and are typically available soon after device design specifications are frozen. Although speed grades with this designation are considered relatively stable and conservative, some under-reporting might still occur.

Preliminary: These speed files are based on complete ES (engineering sample) silicon characterization. Devices and speed grades with this designation are intended to give a better indication of the expected performance of production silicon. The probability of under-reporting delays is greatly reduced as compared to Advance data.

Production: These speed files are released once enough production silicon of a particular device family member has been characterized to provide full correlation between speed files and devices over numerous production lots. There is no under-reporting of delays, and customers receive formal notification of any subsequent changes. Typically, the slowest speed grades transition to Production before faster speed grades.

All specifications are representative of worst-case supply voltage and junction temperature conditions. The parameters included are common to popular designs and typical applications. Contact the factory for design considerations requiring more detailed information.

Table 1 correlates the current status of each Virtex device with a corresponding speed file designation.

Table 1: Virtex Device Speed Grade Designations

	Speed	d Grade Design	ations
Device	Advance	Preliminary	Production
XCV50			-6, -5, -4
XCV100			-6, -5, -4
XCV150			-6, -5, -4
XCV200			-6, -5, -4
XCV300			-6, -5, -4
XCV400			-6, -5, -4
XCV600			-6, -5, -4
XCV800			-6, -5, -4
XCV1000			-6, -5, -4

All specifications are subject to change without notice.



Virtex Switching Characteristics

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE

in the Xilinx Development System) and back-annotated to the simulation net list. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature). Values apply to all Virtex devices unless otherwise noted.

IOB Input Switching Characteristics

Input delays associated with the pad are specified for LVTTL levels. For other standards, adjust the delays with the values shown in , page 6.

				Speed	Grade		
Description	Device	Symbol	Min	-6	-5	-4	Units
Propagation Delays							
Pad to I output, no delay	All	T _{IOPI}	0.39	0.8	0.9	1.0	ns, max
Pad to I output, with delay	XCV50	T _{IOPID}	0.8	1.5	1.7	1.9	ns, max
	XCV100		0.8	1.5	1.7	1.9	ns, max
	XCV150		0.8	1.5	1.7	1.9	ns, max
	XCV200		0.8	1.5	1.7	1.9	ns, max
	XCV300		0.8	1.5	1.7	1.9	ns, max
	XCV400		0.9	1.8	2.0	2.3	ns, max
	XCV600		0.9	1.8	2.0	2.3	ns, max
	XCV800		1.1	2.1	2.4	2.7	ns, max
	XCV1000		1.1	2.1	2.4	2.7	ns, max
Pad to output IQ via transparent latch, no delay	All	T _{IOPLI}	0.8	1.6	1.8	2.0	ns, max
Pad to output IQ via transparent	XCV50	T _{IOPLID}	1.9	3.7	4.2	4.8	ns, max
latch, with delay	XCV100		1.9	3.7	4.2	4.8	ns, max
	XCV150		2.0	3.9	4.3	4.9	ns, max
	XCV200		2.0	4.0	4.4	5.1	ns, max
	XCV300		2.0	4.0	4.4	5.1	ns, max
	XCV400		2.1	4.1	4.6	5.3	ns, max
	XCV600		2.1	4.2	4.7	5.4	ns, max
	XCV800		2.2	4.4	4.9	5.6	ns, max
	XCV1000		2.3	4.5	5.1	5.8	ns, max
Sequential Delays			·				
Clock CLK	All						
Minimum Pulse Width, High		T _{CH}	0.8	1.5	1.7	2.0	ns, min
Minimum Pulse Width, Low		T _{CL}	0.8	1.5	1.7	2.0	ns, min
Clock CLK to output IQ		T _{IOCKIQ}	0.2	0.7	0.7	8.0	ns, max



Description	Symbol	Min	-6	-5	-4	Units
Clock CLK to Pad delay with OBUFT enabled (non-3-state)	T _{IOCKP}	1.0	2.9	3.2	3.5	ns, max
Clock CLK to Pad high-impedance (synchronous) ⁽¹⁾	T _{IOCKHZ}	1.1	2.3	2.5	2.9	ns, max
Clock CLK to valid data on Pad delay, plus enable delay for OBUFT	T _{IOCKON}	1.5	3.4	3.7	4.1	ns, max
Setup and Hold Times before/after Clock	CLK ⁽²⁾		Setup	Time / Hold	Time	1
O input	T _{IOOCK} /T _{IOCKO}	0.51 / 0	1.1 / 0	1.2 / 0	1.3 / 0	ns, min
OCE input	T _{IOOCECK} /T _{IOCKOCE}	0.37 / 0	0.8 / 0	0.9 / 0	1.0 / 0	ns, min
SR input (OFF)	T _{IOSRCKO} /T _{IOCKOSR}	0.52 / 0	1.1 / 0	1.2 / 0	1.4 / 0	ns, min
3-State Setup Times, T input	T _{IOTCK} /T _{IOCKT}	0.34 / 0	0.7 / 0	0.8 / 0	0.9 / 0	ns, min
3-State Setup Times, TCE input	T _{IOTCECK} /T _{IOCKTCE}	0.41 / 0	0.9 / 0	0.9 / 0	1.1 / 0	ns, min
3-State Setup Times, SR input (TFF)	T _{IOSRCKT} /T _{IOCKTSR}	0.49 / 0	1.0 / 0	1.1 / 0	1.3 / 0	ns, min
Set/Reset Delays						
SR input to Pad (asynchronous)	T _{IOSRP}	1.6	3.8	4.1	4.6	ns, max
SR input to Pad high-impedance (asynchronous) ⁽¹⁾	T _{IOSRHZ}	1.6	3.1	3.4	3.9	ns, max
SR input to valid data on Pad (asynchronous)	T _{IOSRON}	2.0	4.2	4.6	5.1	ns, max
GSR to Pad	T _{IOGSRQ}	4.9	9.7	10.9	12.5	ns, max

- 1. 3-state turn-off delays should not be adjusted.
- 2. A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.



Clock Distribution Guidelines

			Speed Grade			
Description	Device	Symbol	-6	-5	-4	Units
Global Clock Skew ⁽¹⁾						
Global Clock Skew between IOB Flip-flops	XCV50	T _{GSKEWIOB}	0.10	0.12	0.14	ns, max
	XCV100		0.12	0.13	0.15	ns, max
	XCV150		0.12	0.13	0.15	ns, max
	XCV200		0.13	0.14	0.16	ns, max
	XCV300		0.14	0.16	0.18	ns, max
	XCV400		0.13	0.13	0.14	ns, max
	XCV600		0.14	0.15	0.17	ns, max
	XCV800		0.16	0.17	0.20	ns, max
	XCV1000		0.20	0.23	0.25	ns, max

Notes:

Clock Distribution Switching Characteristics

			Speed Grade			
Description	Symbol	Min	-6	- 5	-4	Units
GCLK IOB and Buffer						
Global Clock PAD to output.	T _{GPIO}	0.33	0.7	0.8	0.9	ns, max
Global Clock Buffer I input to O output	T _{GIO}	0.34	0.7	0.8	0.9	ns, max

^{1.} These clock-skew delays are provided for guidance only. They reflect the delays encountered in a typical design under worst-case conditions. Precise values for a particular design are provided by the timing analyzer.



I/O Standard Global Clock Input Adjustments

Description	Symbol	Symbol Standard ⁽¹⁾		-6	-5	-4	Units
Data Input Delay Adjustments							
Standard-specific global clock input delay adjustments	T _{GPLVTTL}	LVTTL	0	0	0	0	ns, max
	T _{GPLVCMOS}	LVCMOS2	-0.02	-0.04	-0.04	-0.05	ns, max
	T _{GPPCl33_3}	PCI, 33 MHz, 3.3 V	-0.05	-0.11	-0.12	-0.14	ns, max
	T _{GPPCl33_5}	PCI, 33 MHz, 5.0 V	0.13	0.25	0.28	0.33	ns, max
	T _{GPPCl66_3}	PCI, 66 MHz, 3.3 V	-0.05	-0.11	-0.12	-0.14	ns, max
	T _{GPGTL}	GTL	0.7	0.8	0.9	0.9	ns, max
	T _{GPGTLP}	GTL+	0.7	0.8	0.8	0.8	ns, max
	T _{GPHSTL}	HSTL	0.7	0.7	0.7	0.7	ns, max
	T _{GPSSTL2}	SSTL2	0.6	0.52	0.51	0.50	ns, max
	T _{GPSSTL3}	SSTL3	0.6	0.6	0.55	0.54	ns, max
	T _{GPCTT}	СТТ	0.7	0.7	0.7	0.7	ns, max
	T _{GPAGP}	AGP	0.6	0.54	0.53	0.52	ns, max

^{1.} Input timing for GPLVTTL is measured at 1.4 V. For other I/O standards, see Table 3.



Minimum Clock-to-Out for Virtex Devices

	With DLL		Without DLL								
I/O Standard	All Devices	V50	V100	V150	V200	V300	V400	V600	V800	V1000	Units
*LVTTL_S2	5.2	6.0	6.0	6.0	6.0	6.1	6.1	6.1	6.1	6.1	ns
*LVTTL_S4	3.5	4.3	4.3	4.3	4.3	4.4	4.4	4.4	4.4	4.4	ns
*LVTTL_S6	2.8	3.6	3.6	3.6	3.6	3.7	3.7	3.7	3.7	3.7	ns
*LVTTL_S8	2.2	3.1	3.1	3.1	3.1	3.1	3.1	3.2	3.2	3.2	ns
*LVTTL_S12	2.0	2.9	2.9	2.9	2.9	2.9	2.9	3.0	3.0	3.0	ns
*LVTTL_S16	1.9	2.8	2.8	2.8	2.8	2.8	2.8	2.9	2.9	2.9	ns
*LVTTL_S24	1.8	2.6	2.6	2.7	2.7	2.7	2.7	2.7	2.7	2.8	ns
*LVTTL_F2	2.9	3.8	3.8	3.8	3.8	3.8	3.8	3.9	3.9	3.9	ns
*LVTTL_F4	1.7	2.6	2.6	2.6	2.6	2.6	2.6	2.7	2.7	2.7	ns
*LVTTL_F6	1.2	2.0	2.0	2.0	2.1	2.1	2.1	2.1	2.1	2.2	ns
*LVTTL_F8	1.1	1.9	1.9	1.9	1.9	2.0	2.0	2.0	2.0	2.0	ns
*LVTTL_F12	1.0	1.8	1.8	1.8	1.8	1.9	1.9	1.9	1.9	1.9	ns
*LVTTL_F16	0.9	1.7	1.8	1.8	1.8	1.8	1.8	1.8	1.9	1.9	ns
*LVTTL_F24	0.9	1.7	1.7	1.7	1.8	1.8	1.8	1.8	1.8	1.9	ns
LVCMOS2	1.1	1.9	1.9	1.9	2.0	2.0	2.0	2.0	2.0	2.1	ns
PCI33_3	1.5	2.4	2.4	2.4	2.4	2.4	2.4	2.5	2.5	2.5	ns
PCI33_5	1.4	2.2	2.2	2.3	2.3	2.3	2.3	2.3	2.3	2.4	ns
PCI66_3	1.1	1.9	1.9	2.0	2.0	2.0	2.0	2.0	2.1	2.1	ns
GTL	1.6	2.5	2.5	2.5	2.5	2.5	2.5	2.6	2.6	2.6	ns
GTL+	1.7	2.5	2.5	2.6	2.6	2.6	2.6	2.6	2.6	2.7	ns
HSTL I	1.1	1.9	1.9	1.9	1.9	2.0	2.0	2.0	2.0	2.0	ns
HSTL III	0.9	1.7	1.7	1.8	1.8	1.8	1.8	1.8	1.8	1.9	ns
HSTL IV	0.8	1.6	1.6	1.6	1.7	1.7	1.7	1.7	1.7	1.8	ns
SSTL2 I	0.9	1.7	1.7	1.7	1.7	1.8	1.8	1.8	1.8	1.8	ns
SSTL2 II	0.8	1.6	1.6	1.6	1.6	1.7	1.7	1.7	1.7	1.7	ns
SSTL3 I	0.8	1.6	1.7	1.7	1.7	1.7	1.7	1.7	1.8	1.8	ns
SSTL3 II	0.7	1.5	1.5	1.6	1.6	1.6	1.6	1.6	1.6	1.7	ns
CTT	1.0	1.8	1.8	1.8	1.9	1.9	1.9	1.9	1.9	2.0	ns
AGP	1.0	1.8	1.8	1.9	1.9	1.9	1.9	1.9	1.9	2.0	ns

^{*}S = Slow Slew Rate, F = Fast Slew Rate

^{1.} Listed above are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.

^{2.} Input and output timing is measured at 1.4 V for LVTTL. For other I/O standards, see Table 3. In all cases, an 8 pF external capacitive load is used.



Virtex Pin-to-Pin Input Parameter Guidelines

All devices are 100% functionally tested. Listed below are representative values for typical pin locations and normal clock loading. Values are expressed in nanoseconds unless otherwise noted

Global Clock Set-Up and Hold for LVTTL Standard, with DLL

Description	Symbol	Device	Min	-6	-5	-4	Units		
Input Setup and Hold Time Relative to Global Clock Input Signal for LVTTL Standard. For data input with different standards, adjust the setup time delay by the values shown in Input Delay Adjustments.									
No Delay Global Clock and IFF, with DLL	T _{PSDLL} /T _{PHDLL}	XCV50	0.40 / -0.4	1.7 /-0.4	1.8 /0.4	2.1 /-0.4	ns, min		
		XCV100	0.40 /0.4	1.7 /-0.4	1.9 /0.4	2.1 /-0.4	ns, min		
		XCV150	0.40 /0.4	1.7 /-0.4	1.9 /0.4	2.1 /-0.4	ns, min		
		XCV200	0.40 /0.4	1.7 /-0.4	1.9 /0.4	2.1 /-0.4	ns, min		
		XCV300	0.40 /0.4	1.7 /-0.4	1.9 /0.4	2.1 /-0.4	ns, min		
		XCV400	0.40 /0.4	1.7 /-0.4	1.9 /0.4	2.1 /-0.4	ns, min		
		XCV600	0.40 /0.4	1.7 /-0.4	1.9 /0.4	2.1 /-0.4	ns, min		
		XCV800	0.40 /-0.4	1.7 /-0.4	1.9 /-0.4	2.1 /-0.4	ns, min		
		XCV1000	0.40 /-0.4	1.7 /-0.4	1.9 /0.4	2.1 /-0.4	ns, min		

IFF = Input Flip-Flop or Latch

- 2. DLL output jitter is already included in the timing calculation.
- 3. A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.

^{1.} Set-up time is measured relative to the Global Clock input signal with the fastest route and the lightest load. Hold time is measured relative to the Global Clock input signal with the slowest route and heaviest load.



Period Tolerance: the allowed input clock period change in nanoseconds.

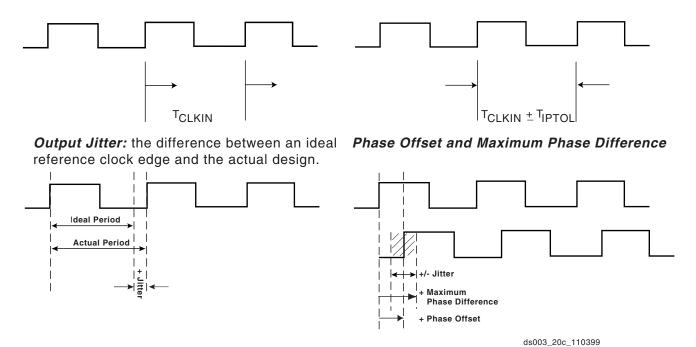


Figure 1: Frequency Tolerance and Clock Jitter

Revision History

Date	Version	Revision
11/98	1.0	Initial Xilinx release.
01/99	1.2	Updated package drawings and specs.
02/99	1.3	Update of package drawings, updated specifications.
05/99	1.4	Addition of package drawings and specifications.
05/99	1.5	Replaced FG 676 & FG680 package drawings.
07/99	1.6	Changed Boundary Scan Information and changed Figure 11, Boundary Scan Bit Sequence. Updated IOB Input & Output delays. Added Capacitance info for different I/O Standards. Added 5 V tolerant information. Added DLL Parameters and waveforms and new Pin-to-pin Input and Output Parameter tables for Global Clock Input to Output and Setup and Hold. Changed Configuration Information including Figures 12, 14, 17 & 19. Added device-dependent listings for quiescent currents ICCINTQ and ICCOQ. Updated IOB Input and Output Delays based on default standard of LVTTL, 12 mA, Fast Slew Rate. Added IOB Input Switching Characteristics Standard Adjustments.
09/99	1.7	Speed grade update to preliminary status, Power-on specification and Clock-to-Out Minimums additions, "0" hold time listing explanation, quiescent current listing update, and Figure 6 ADDRA input label correction. Added T _{IJITCC} parameter, changed T _{OJIT} to T _{OPHASE} .
01/00	1.8	Update to speed.txt file 1.96. Corrections for CRs 111036,111137, 112697, 115479, 117153, 117154, and 117612. Modified notes for Recommended Operating Conditions (voltage and temperature). Changed Bank information for V _{CCO} in CS144 package on p.43.



Virtex[™] 2.5 V Field Programmable Gate Arrays

DS003-4 (v4.0) March 1, 2013

Production Product Specification

Virtex Pin Definitions

Table 1: Special Purpose Pins

Pin Name	Dedicated Pin	Direction	Description		
GCK0, GCK1, GCK2, GCK3	Yes	Input	Clock input pins that connect to Global Clock Buffers. These pins become user inputs when not needed for clocks.		
M0, M1, M2	Yes	Input	Mode pins are used to specify the configuration mode.		
CCLK	Yes	Input or Output	The configuration Clock I/O pin: it is an input for SelectMAP and slave-serial modes, and output in master-serial mode. After configuration, it is input only, logic level = Don't Care.		
PROGRAM	Yes	Input	Initiates a configuration sequence when asserted Low.		
DONE	Yes	Bidirectional	Indicates that configuration loading is complete, and that the start-up sequence is in progress. The output can be open drain.		
INIT	No	Bidirectional (Open-drain)	When Low, indicates that the configuration memory is being cleared. The pin becomes a user I/O after configuration.		
BUSY/ DOUT	No	Output	In SelectMAP mode, BUSY controls the rate at which configuration data is loaded. The pin becomes a user I/O after configuration unless the SelectMAP port is retained.		
			In bit-serial modes, DOUT provides header information to downstream devices in a daisy-chain. The pin becomes a user I/O after configuration.		
D0/DIN, D1, D2, D3, D4, D5, D6,	No	Input or Output	In SelectMAP mode, D0 - D7 are configuration data pins. These pins become user I/Os after configuration unless the SelectMAP port is retained. In bit-serial modes, DIN is the single data input. This pin becomes a user		
D7			I/O after configuration.		
WRITE	No	Input	In SelectMAP mode, the active-low Write Enable signal. The pin becomes a user I/O after configuration unless the SelectMAP port is retained.		
CS	No	Input	In SelectMAP mode, the active-low Chip Select signal. The pin becomes a user I/O after configuration unless the SelectMAP port is retained.		
TDI, TDO, TMS, TCK	Yes	Mixed	Boundary-scan Test-Access-Port pins, as defined in IEEE 1149.1.		
DXN, DXP	Yes	N/A	Temperature-sensing diode pins. (Anode: DXP, cathode: DXN)		
V _{CCINT}	Yes	Input	Power-supply pins for the internal core logic.		
V _{cco}	Yes	Input	Power-supply pins for the output drivers (subject to banking rules)		
V _{REF}	No	Input	Input threshold voltage pins. Become user I/Os when an external threshold voltage is not needed (subject to banking rules).		
GND	Yes	Input	Ground		

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Table 2: Virtex Pinout Tables (Chip-Scale and QFP Packages) (Continued)

Pin Name	Device	CS144	TQ144	PQ/HQ240
V _{REF} , Bank 3	XCV50	H11, K12	60, 68	130, 144
(V _{REF} pins are listed incrementally. Connect all pins listed for both	XCV100/150	+ J10	+ 66	+ 133
	XCV200/300	N/A	N/A	+ 126
the required device	XCV400	N/A	N/A	+ 147
and all smaller devices listed in the same	XCV600	N/A	N/A	+ 132
package.)	XCV800	N/A	N/A	+ 140
Within each bank, if input reference voltage is not required, all V _{REF} pins are general I/O.				
V _{REF} , Bank 4	XCV50	L8, L10	79, 87	97, 111
(V _{REF} pins are listed	XCV100/150	+ N10	+ 81	+ 108
incrementally. Connect all pins listed for both	XCV200/300	N/A	N/A	+ 115
the required device and all smaller devices	XCV400	N/A	N/A	+ 94
listed in the same	XCV600	N/A	N/A	+ 109
package.)	XCV800	N/A	N/A	+ 101
Within each bank, if input reference voltage is not required, all V _{REF} pins are general I/O.				
V _{REF} , Bank 5	XCV50	L4, L6	96, 104	70, 84
(V _{REF} pins are listed	XCV100/150	+ N4	+ 102	+ 73
incrementally. Connect all pins listed for both	XCV200/300	N/A	N/A	+ 66
the required device	XCV400	N/A	N/A	+ 87
and all smaller devices listed in the same package.)	XCV600	N/A	N/A	+ 72
	XCV800	N/A	N/A	+ 80
Within each bank, if input reference voltage is not required, all V _{REF} pins are general I/O.				



Table 2: Virtex Pinout Tables (Chip-Scale and QFP Packages) (Continued)

Pin Name	Device	CS144	TQ144	PQ/HQ240
V _{REF} , Bank 6	XCV50	H2, K1	116, 123	36, 50
(V _{REF} pins are listed	XCV100/150	+ J3	+ 118	+ 47
incrementally. Connect all pins listed for both	XCV200/300	N/A	N/A	+ 54
the required device	XCV400	N/A	N/A	+ 33
and all smaller devices listed in the same	XCV600	N/A	N/A	+ 48
package.)	XCV800	N/A	N/A	+ 40
Within each bank, if input reference voltage is not required, all V _{REF} pins are general I/O.				
V _{REF} , Bank 7	XCV50	D4, E1	133, 140	9, 23
(V _{REF} pins are listed	XCV100/150	+ D2	+ 138	+ 12
incrementally. Connect all pins listed for both	XCV200/300	N/A	N/A	+ 5
the required device	XCV400	N/A	N/A	+ 26
and all smaller devices listed in the same	XCV600	N/A	N/A	+ 11
package.)	XCV800	N/A	N/A	+ 19
Within each bank, if input reference voltage is not required, all V _{REF} pins are general I/O.				
GND	All	A1, B9, B11, C7, D5, E4, E11, F1, G10, J1, J12, L3, L5, L7, L9, N12	9, 18, 26, 35, 46, 54, 64, 75, 83, 91, 100, 111, 120, 129, 136, 144,	1, 8, 14, 22, 29, 37, 45, 51, 59, 69, 75, 83, 91, 98, 106, 112, 119, 129, 135, 143, 151, 158, 166, 172, 182, 190, 196, 204, 211, 219, 227, 233



Table 3: Virtex Pinout Tables (BGA) (Continued)

Pin Name	Device	BG256	BG352	BG432	BG560
VCCINT Notes: Superset includes all pins, including the ones in bold type. Subset excludes pins in bold type. In BG352, for XCV300 all the V _{CCINT} pins in the superset must be connected. For XCV150/200, V _{CCINT} pins in the subset must be connected, and pins in bold type can be left unconnected (these unconnected pins cannot be used as user I/O.) In BG432, for XCV400/600/800 all V _{CCINT} pins in the superset must be connected. For XCV300, V _{CCINT} pins in the superset must be connected. For XCV300, V _{CCINT} pins in the subset must be connected, and pins in bold type can be left unconnected (these unconnected pins cannot be used as user I/O.) In BG560, for XCV800/1000 all V _{CCINT} pins in the superset must be connected. For XCV400/600, V _{CCINT} pins in the superset must be connected. For XCV400/600, V _{CCINT} pins in the subset must be connected, and pins in bold type can be left unconnected (these unconnected pins cannot be used as user I/O.)	XCV50/100	C10, D6, D15, F4, F17, L3, L18, R4, R17, U6, U15, V10	N/A	N/A	N/A
	XCV150/200/300	Same as above	A20, C14, D10, J24, K4, P2, P25, V24, W2, AC10, AE14, AE19, B16, D12, L1, L25, R23, T1, AF11, AF16	A10, A17, B23, C14, C19, K3, K29, N2, N29, T1, T29, W2, W31, AB2, AB30, AJ10, AJ16, AK13, AK19, AK22, B26, C7, F1, F30, AE29, AF1, AH8, AH24	N/A
	XCV400/600/800/1000	N/A	N/A	Same as above	A21, B14, B18, B28, C24, E9, E12, F2, H30, J1, K32, N1, N33, U5, U30, Y2, Y31, AD2, AD32, AG3, AG31, AK8, AK11, AK17, AK20, AL14, AL27, AN25, B12, C22, M3, N29, AB2, AB32, AJ13, AL22
V _{CCO} , Bank 0	All	D7, D8	A17, B25, D19	A21, C29, D21	A22, A26, A30, B19, B32
V _{CCO} , Bank 1	All	D13, D14	A10, D7, D13	A1, A11, D11	A10, A16, B13, C3, E5
V _{CCO} , Bank 2	All	G17, H17	B2, H4, K1	C3, L1, L4	B2, D1, H1, M1, R2
V _{CCO} , Bank 3	All	N17, P17	P4, U1, Y4	AA1, AA4, AJ3	V1, AA2, AD1, AK1, AL2
V _{CCO} , Bank 4	All	U13, U14	AC8, AE2, AF10	AH11, AL1, AL11	AM2, AM15, AN4, AN8, AN12
V _{CCO} , Bank 5	All	U7, U8	AC14, AC20, AF17	AH21, AJ29, AL21	AL31, AM21, AN18, AN24, AN30
V _{CCO} , Bank 6	All	N4, P4	U26, W23, AE25	AA28, AA31, AL31	W32, AB33, AF33, AK33, AM32

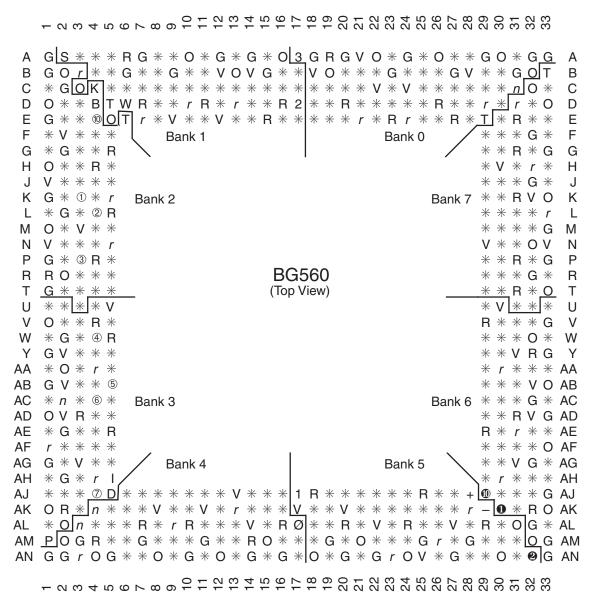


Table 3: Virtex Pinout Tables (BGA) (Continued)

Pin Name	Device	BG256	BG352	BG432	BG560
V _{CCO} , Bank 7	All	G4, H4	G23, K26, N23	A31, L28, L31	C32, D33, K33, N32, T33
V _{REF} , Bank 0	XCV50	A8, B4	N/A	N/A	N/A
(VREF pins are listed incrementally. Connect all	XCV100/150	+ A4	A16,C19, C21	N/A	N/A
pins listed for both the required device and all smaller devices listed in the	XCV200/300	+ A2	+ D21	B19, D22, D24, D26	N/A
same package.)	XCV400	N/A	N/A	+ C18	A19, D20,
Within each bank, if input					D26, E23, E27
reference voltage is not required, all V _{REF} pins are	XCV600	N/A	N/A	+ C24	+ E24
general I/O.	XCV800	N/A	N/A	+ B21	+ E21
	XCV1000	N/A	N/A	N/A	+ D29
V _{REF} , Bank 1	XCV50	A17, B12	N/A	N/A	N/A
(VREF pins are listed incrementally. Connect all	XCV100/150	+ B15	B6, C9, C12	N/A	N/A
pins listed for both the required device and all smaller devices listed in the	XCV200/300	+ B17	+ D6	A13, B7, C6, C10	N/A
same package.) Within each bank, if input reference voltage is not	XCV400	N/A	N/A	+ B15	A6, D7, D11, D16, E15
required, all V _{REF} pins are	XCV600	N/A	N/A	+ D10	+ D10
general I/O.	XCV800	N/A	N/A	+ B12	+ D13
	XCV1000	N/A	N/A	N/A	+ E7
V _{REF} , Bank 2	XCV50	C20, J18	N/A	N/A	N/A
(V _{REF} pins are listed incrementally. Connect all pins listed for both the	XCV100/150	+ F19	E2, H2, M4	N/A	N/A
required device and all smaller devices listed in the	XCV200/300	+ G18	+ D2	E2, G3, J2, N1	N/A
same package.)	XCV400	N/A	N/A	+ R3	G5, H4,
Within each bank, if input reference voltage is not					L5, P4, R1
required, all V _{REF} pins are	XCV600	N/A	N/A	+ H1	+ K5
general I/O.	XCV800	N/A	N/A	+ M3	+ N5
	XCV1000	N/A	N/A	N/A	+ B3



BG560 Pin Function Diagram



DS003_22_100300

Figure 7: BG560 Pin Function Diagram



FG680 Pin Function Diagram

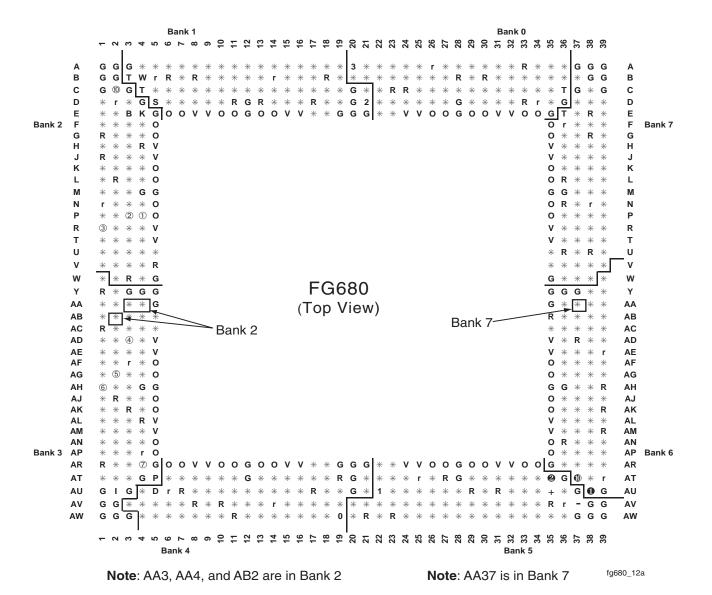


Figure 11: FG680 Pin Function Diagram