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Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	600
Number of Logic Elements/Cells	2700
Total RAM Bits	40960
Number of I/O	94
Number of Gates	108904
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	144-TFBGA, CSPBGA
Supplier Device Package	144-LCSBGA (12x12)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcv100-6cs144c

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong



Virtex Architecture

Virtex devices feature a flexible, regular architecture that comprises an array of configurable logic blocks (CLBs) surrounded by programmable input/output blocks (IOBs), all interconnected by a rich hierarchy of fast, versatile routing resources. The abundance of routing resources permits the Virtex family to accommodate even the largest and most complex designs.

Virtex FPGAs are SRAM-based, and are customized by loading configuration data into internal memory cells. In some modes, the FPGA reads its own configuration data from an external PROM (master serial mode). Otherwise, the configuration data is written into the FPGA (Select-MAPTM, slave serial, and JTAG modes).

The standard Xilinx Foundation™ and Alliance Series™ Development systems deliver complete design support for Virtex, covering every aspect from behavioral and schematic entry, through simulation, automatic design translation and implementation, to the creation, downloading, and readback of a configuration bit stream.

Higher Performance

Virtex devices provide better performance than previous generations of FPGA. Designs can achieve synchronous system clock rates up to 200 MHz including I/O. Virtex inputs and outputs comply fully with PCI specifications, and interfaces can be implemented that operate at 33 MHz or 66 MHz. Additionally, Virtex supports the hot-swapping requirements of Compact PCI.

Xilinx thoroughly benchmarked the Virtex family. While performance is design-dependent, many designs operated internally at speeds in excess of 100 MHz and can achieve 200 MHz. Table 2 shows performance data for representative circuits, using worst-case timing parameters.

Table 2: Performance for Common Circuit Functions

Function	Bits	Virtex -6
Register-to-Register		
Adder	16	5.0 ns
Audei	64	7.2 ns
Pipelined Multiplier	8 x 8	5.1 ns
	16 x 16	6.0 ns
Address Decoder	16	4.4 ns
	64	6.4 ns
16:1 Multiplexer		5.4 ns
Parity Tree	9	4.1 ns
	18	5.0 ns
	36	6.9 ns
Chip-to-Chip		
HSTL Class IV		200 MHz
LVTTL,16mA, fast slew		180 MHz



Virtex Device/Package Combinations and Maximum I/O

Table 3: Virtex Family Maximum User I/O by Device/Package (Excluding Dedicated Clock Pins)

Package	XCV50	XCV100	XCV150	XCV200	XCV300	XCV400	XCV600	XCV800	XCV1000
CS144	94	94							
TQ144	98	98							
PQ240	166	166	166	166	166				
HQ240						166	166	166	
BG256	180	180	180	180					
BG352			260	260	260				
BG432					316	316	316	316	
BG560						404	404	404	404
FG256	176	176	176	176					
FG456			260	284	312				
FG676						404	444	444	
FG680							512	512	512

Virtex Ordering Information

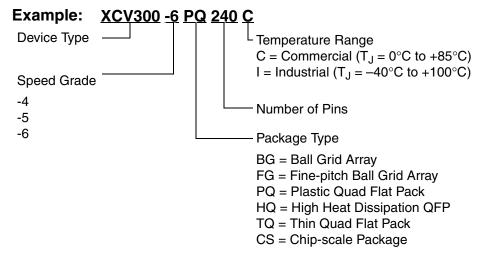


Figure 1: Virtex Ordering Information

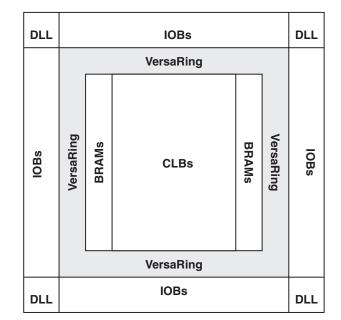


DS003-2 (v4.0) March 1, 2013

Virtex[™] 2.5 V Field Programmable Gate Arrays

Product Specification

The output buffer and all of the IOB control signals have independent polarity controls.



vao_b.eps

Figure 1: Virtex Architecture Overview

All pads are protected against damage from electrostatic discharge (ESD) and from over-voltage transients. Two forms of over-voltage protection are provided, one that permits 5 V compliance, and one that does not. For 5 V compliance, a Zener-like structure connected to ground turns on when the output rises to approximately 6.5 V. When PCI 3.3 V compliance is required, a conventional clamp diode is connected to the output supply voltage, $V_{\rm CCO}$.

Optional pull-up and pull-down resistors and an optional weak-keeper circuit are attached to each pad. Prior to configuration, all pins not involved in configuration are forced into their high-impedance state. The pull-down resistors and the weak-keeper circuits are inactive, but inputs can optionally be pulled up.

The activation of pull-up resistors prior to configuration is controlled on a global basis by the configuration mode pins. If the pull-up resistors are not activated, all the pins will float. Consequently, external pull-up or pull-down resistors must be provided on pins required to be at a well-defined logic level prior to configuration.

All Virtex IOBs support IEEE 1149.1-compatible boundary scan testing.

Architectural Description

Virtex Array

The Virtex user-programmable gate array, shown in Figure 1, comprises two major configurable elements: configurable logic blocks (CLBs) and input/output blocks (IOBs).

- CLBs provide the functional elements for constructing logic
- IOBs provide the interface between the package pins and the CLBs

CLBs interconnect through a general routing matrix (GRM). The GRM comprises an array of routing switches located at the intersections of horizontal and vertical routing channels. Each CLB nests into a VersaBlock™ that also provides local routing resources to connect the CLB to the GRM.

The VersaRing[™] I/O interface provides additional routing resources around the periphery of the device. This routing improves I/O routability and facilitates pin locking.

The Virtex architecture also includes the following circuits that connect to the GRM.

- Dedicated block memories of 4096 bits each
- Clock DLLs for clock-distribution delay compensation and clock domain control
- 3-State buffers (BUFTs) associated with each CLB that drive dedicated segmentable horizontal routing resources

Values stored in static memory cells control the configurable logic elements and interconnect resources. These values load into the memory cells on power-up, and can reload if necessary to change the function of the device.

Input/Output Block

The Virtex IOB, Figure 2, features SelectIO™ inputs and outputs that support a wide variety of I/O signalling standards, see Table 1.

The three IOB storage elements function either as edge-triggered D-type flip-flops or as level sensitive latches. Each IOB has a clock signal (CLK) shared by the three flip-flops and independent clock enable signals for each flip-flop.

In addition to the CLK and CE control signals, the three flip-flops share a Set/Reset (SR). For each flip-flop, this signal can be independently configured as a synchronous Set, a synchronous Reset, an asynchronous Preset, or an asynchronous Clear.

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General Purpose Routing

Most Virtex signals are routed on the general purpose routing, and consequently, the majority of interconnect resources are associated with this level of the routing hierarchy. The general routing resources are located in horizontal and vertical routing channels associated with the rows and columns CLBs. The general-purpose routing resources are listed below.

- Adjacent to each CLB is a General Routing Matrix (GRM). The GRM is the switch matrix through which horizontal and vertical routing resources connect, and is also the means by which the CLB gains access to the general purpose routing.
- 24 single-length lines route GRM signals to adjacent GRMs in each of the four directions.
- 12 buffered Hex lines route GRM signals to another GRMs six-blocks away in each one of the four directions. Organized in a staggered pattern, Hex lines can be driven only at their endpoints. Hex-line signals can be accessed either at the endpoints or at the midpoint (three blocks from the source). One third of the Hex lines are bidirectional, while the remaining ones are uni-directional.

 12 Longlines are buffered, bidirectional wires that distribute signals across the device quickly and efficiently. Vertical Longlines span the full height of the device, and horizontal ones span the full width of the device.

I/O Routing

Virtex devices have additional routing resources around their periphery that form an interface between the CLB array and the IOBs. This additional routing, called the VersaRing, facilitates pin-swapping and pin-locking, such that logic redesigns can adapt to existing PCB layouts. Time-to-market is reduced, since PCBs and other system components can be manufactured while the logic design is still in progress.

Dedicated Routing

Some classes of signal require dedicated routing resources to maximize performance. In the Virtex architecture, dedicated routing resources are provided for two classes of signal.

- Horizontal routing resources are provided for on-chip 3-state busses. Four partitionable bus lines are provided per CLB row, permitting multiple busses within a row, as shown in Figure 8.
- Two dedicated nets per CLB propagate carry signals vertically to the adjacent CLB.

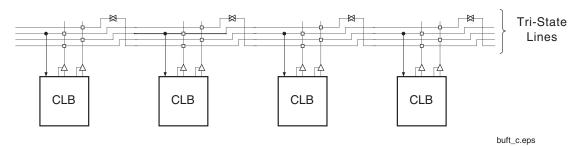


Figure 8: BUFT Connections to Dedicated Horizontal Bus Lines

Global Routing

Global Routing resources distribute clocks and other signals with very high fanout throughout the device. Virtex devices include two tiers of global routing resources referred to as primary global and secondary local clock routing resources.

• The primary global routing resources are four dedicated global nets with dedicated input pins that are designed to distribute high-fanout clock signals with minimal skew. Each global clock net can drive all CLB, IOB, and block RAM clock pins. The primary global nets can only be driven by global buffers. There are four global buffers, one for each global net. The secondary local clock routing resources consist of 24 backbone lines, 12 across the top of the chip and 12 across bottom. From these lines, up to 12 unique signals per column can be distributed via the 12 longlines in the column. These secondary resources are more flexible than the primary resources since they are not restricted to routing only to clock pins.

Clock Distribution

Virtex provides high-speed, low-skew clock distribution through the primary global routing resources described above. A typical clock distribution net is shown in Figure 9.

Four global buffers are provided, two at the top center of the device and two at the bottom center. These drive the four primary global nets that in turn drive any clock pin.

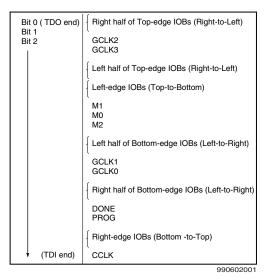


Figure 11: Boundary Scan Bit Sequence

Table 5: Boundary Scan Instructions

Boundary-Scan Command	Binary Code(4:0)	Description
EXTEST	00000	Enables boundary-scan EXTEST operation
SAMPLE/PRELOAD	00001	Enables boundary-scan SAMPLE/PRELOAD operation
USER 1	00010	Access user-defined register 1
USER 2	00011	Access user-defined register 2
CFG_OUT	00100	Access the configuration bus for read operations.
CFG_IN	00101	Access the configuration bus for write operations.
INTEST	00111	Enables boundary-scan INTEST operation
USERCODE	01000	Enables shifting out USER code
IDCODE	01001	Enables shifting out of ID Code
HIGHZ	01010	3-states output pins while enabling the Bypass Register
JSTART	01100	Clock the start-up sequence when StartupClk is TCK
BYPASS	11111	Enables BYPASS
RESERVED	All other codes	Xilinx reserved instructions

Identification Registers

The IDCODE register is supported. By using the IDCODE, the device connected to the JTAG port can be determined.

The IDCODE register has the following binary format:

vvvv:ffff:fffa:aaaa:aaaa:cccc:cccc1

where

v = the die version number

f = the family code (03h for Virtex family)

a = the number of CLB rows (ranges from 010h for XCV50 to 040h for XCV1000)

c = the company code (49h for Xilinx)

The USERCODE register is supported. By using the USER-CODE, a user-programmable identification code can be loaded and shifted out for examination. The identification code is embedded in the bitstream during bitstream generation and is valid only after configuration.

Table 6: IDCODEs Assigned to Virtex FPGAs

FPGA	IDCODE
XCV50	v0610093h
XCV100	v0614093h
XCV150	v0618093h
XCV200	v061C093h
XCV300	v0620093h
XCV400	v0628093h
XCV600	v0630093h
XCV800	v0638093h
XCV1000	v0640093h

Including Boundary Scan in a Design

Since the boundary scan pins are dedicated, no special element needs to be added to the design unless an internal data register (USER1 or USER2) is desired.

If an internal data register is used, insert the boundary scan symbol and connect the necessary pins as appropriate.

Development System

Virtex FPGAs are supported by the Xilinx Foundation and Alliance CAE tools. The basic methodology for Virtex design consists of three interrelated steps: design entry, implementation, and verification. Industry-standard tools are used for design entry and simulation (for example, Synopsys FPGA Express), while Xilinx provides proprietary architecture-specific tools for implementation.

The Xilinx development system is integrated under the Xilinx Design Manager (XDM™) software, providing design-



ers with a common user interface regardless of their choice of entry and verification tools. The XDM software simplifies the selection of implementation options with pull-down menus and on-line help.

Application programs ranging from schematic capture to Placement and Routing (PAR) can be accessed through the XDM software. The program command sequence is generated prior to execution, and stored for documentation.

Several advanced software features facilitate Virtex design. RPMs, for example, are schematic-based macros with relative location constraints to guide their placement. They help ensure optimal implementation of common functions.

For HDL design entry, the Xilinx FPGA Foundation development system provides interfaces to the following synthesis design environments.

- Synopsys (FPGA Compiler, FPGA Express)
- Exemplar (Spectrum)
- Synplicity (Synplify)

For schematic design entry, the Xilinx FPGA Foundation and alliance development system provides interfaces to the following schematic-capture design environments.

- Mentor Graphics V8 (Design Architect, QuickSim II)
- Viewlogic Systems (Viewdraw)

Third-party vendors support many other environments.

A standard interface-file specification, Electronic Design Interchange Format (EDIF), simplifies file transfers into and out of the development system.

Virtex FPGAs supported by a unified library of standard functions. This library contains over 400 primitives and macros, ranging from 2-input AND gates to 16-bit accumulators, and includes arithmetic functions, comparators, counters, data registers, decoders, encoders, I/O functions, latches, Boolean functions, multiplexers, shift registers, and barrel shifters.

The "soft macro" portion of the library contains detailed descriptions of common logic functions, but does not contain any partitioning or placement information. The performance of these macros depends, therefore, on the partitioning and placement obtained during implementation.

RPMs, on the other hand, do contain predetermined partitioning and placement information that permits optimal implementation of these functions. Users can create their own library of soft macros or RPMs based on the macros and primitives in the standard library.

The design environment supports hierarchical design entry, with high-level schematics that comprise major functional blocks, while lower-level schematics define the logic in these blocks. These hierarchical design elements are automatically combined by the implementation tools. Different design entry tools can be combined within a hierarchical

design, thus allowing the most convenient entry method to be used for each portion of the design.

Design Implementation

The place-and-route tools (PAR) automatically provide the implementation flow described in this section. The partitioner takes the EDIF net list for the design and maps the logic into the architectural resources of the FPGA (CLBs and IOBs, for example). The placer then determines the best locations for these blocks based on their interconnections and the desired performance. Finally, the router interconnects the blocks.

The PAR algorithms support fully automatic implementation of most designs. For demanding applications, however, the user can exercise various degrees of control over the process. User partitioning, placement, and routing information is optionally specified during the design-entry process. The implementation of highly structured designs can benefit greatly from basic floor planning.

The implementation software incorporates Timing Wizard® timing-driven placement and routing. Designers specify timing requirements along entire paths during design entry. The timing path analysis routines in PAR then recognize these user-specified requirements and accommodate them.

Timing requirements are entered on a schematic in a form directly relating to the system requirements, such as the targeted clock frequency, or the maximum allowable delay between two registers. In this way, the overall performance of the system along entire signal paths is automatically tailored to user-generated specifications. Specific timing information for individual nets is unnecessary.

Design Verification

In addition to conventional software simulation, FPGA users can use in-circuit debugging techniques. Because Xilinx devices are infinitely reprogrammable, designs can be verified in real time without the need for extensive sets of software simulation vectors.

The development system supports both software simulation and in-circuit debugging techniques. For simulation, the system extracts the post-layout timing information from the design database, and back-annotates this information into the net list for use by the simulator. Alternatively, the user can verify timing-critical portions of the design using the TRACE® static timing analyzer.

For in-circuit debugging, the development system includes a download and readback cable. This cable connects the FPGA in the target system to a PC or workstation. After downloading the design into the FPGA, the designer can single-step the logic, readback the contents of the flip-flops, and so observe the internal logic state. Simple modifications can be downloaded into the system in a matter of minutes.

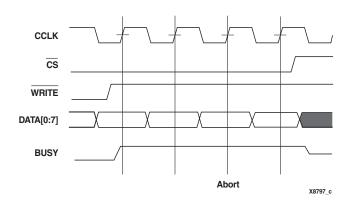


Figure 18: SelectMAP Write Abort Waveforms

Boundary-Scan Mode

In the boundary-scan mode, configuration is done through the IEEE 1149.1 Test Access Port. Note that the PROGRAM pin must be pulled High prior to reconfiguration. A Low on the PROGRAM pin resets the TAP controller and no JTAG operations can be performed.

Configuration through the TAP uses the CFG_IN instruction. This instruction allows data input on TDI to be converted into data packets for the internal configuration bus.

The following steps are required to configure the FPGA through the boundary-scan port (when using TCK as a start-up clock).

- Load the CFG_IN instruction into the boundary-scan instruction register (IR)
- 2. Enter the Shift-DR (SDR) state
- 3. Shift a configuration bitstream into TDI
- 4. Return to Run-Test-Idle (RTI)
- 5. Load the JSTART instruction into IR
- 6. Enter the SDR state
- 7. Clock TCK through the startup sequence
- 8. Return to RTI

Configuration and readback via the TAP is always available. The boundary-scan mode is selected by a <101> or 001> on the mode pins (M2, M1, M0). For details on TAP characteristics, refer to XAPP139.

Configuration Sequence

The configuration of Virtex devices is a three-phase process. First, the configuration memory is cleared. Next, configuration data is loaded into the memory, and finally, the logic is activated by a start-up process.

Configuration is automatically initiated on power-up unless it is delayed by the user, as described below. The configuration process can also be initiated by asserting $\overline{\mathsf{PROGRAM}}$.

The end of the memory-clearing phase is signalled by INIT going High, and the completion of the entire process is signalled by DONE going High.

The power-up timing of configuration signals is shown in Figure 19. The corresponding timing characteristics are listed in Table 10.

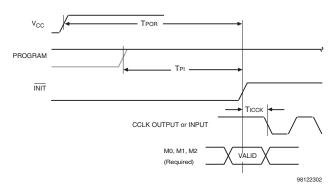


Figure 19: Power-Up Timing Configuration Signals

Table 10: Power-up Timing Characteristics

Description	Symbol	Value	Units
Power-on Reset	T _{POR}	2.0	ms, max
Program Latency	T _{PL}	100.0	μs, max
CCLK (output) Delay	T _{ICCK}	0.5	μs, min
		4.0	μs, max
Program Pulse Width	T _{PROGRAM}	300	ns, min

Delaying Configuration

INIT can be held Low using an open-drain driver. An open-drain is required since INIT is a bidirectional open-drain pin that is held Low by the FPGA while the configuration memory is being cleared. Extending the time that the pin is Low causes the configuration sequencer to wait. Thus, configuration is delayed by preventing entry into the phase where data is loaded.

Start-Up Sequence

The default Start-up sequence is that one CCLK cycle after DONE goes High, the global 3-state signal (GTS) is released. This permits device outputs to turn on as necessary.

One CCLK cycle later, the Global Set/Reset (GSR) and Global Write Enable (GWE) signals are released. This permits the internal storage elements to begin changing state in response to the logic and the user clock.

The relative timing of these events can be changed. In addition, the GTS, GSR, and GWE events can be made dependent on the DONE pins of multiple devices all going High, forcing the devices to start in synchronism. The sequence can also be paused at any stage until lock has been achieved on any or all DLLs.



Virtex[™] 2.5 V Field Programmable Gate Arrays

DS003-3 (v4.0) March 1, 2013

Production Product Specification

Virtex Electrical Characteristics Definition of Terms

Electrical and switching characteristics are specified on a per-speed-grade basis and can be designated as Advance, Preliminary, or Production. Each designation is defined as follows:

Advance: These speed files are based on simulations only and are typically available soon after device design specifications are frozen. Although speed grades with this designation are considered relatively stable and conservative, some under-reporting might still occur.

Preliminary: These speed files are based on complete ES (engineering sample) silicon characterization. Devices and speed grades with this designation are intended to give a better indication of the expected performance of production silicon. The probability of under-reporting delays is greatly reduced as compared to Advance data.

Production: These speed files are released once enough production silicon of a particular device family member has been characterized to provide full correlation between speed files and devices over numerous production lots. There is no under-reporting of delays, and customers receive formal notification of any subsequent changes. Typically, the slowest speed grades transition to Production before faster speed grades.

All specifications are representative of worst-case supply voltage and junction temperature conditions. The parameters included are common to popular designs and typical applications. Contact the factory for design considerations requiring more detailed information.

Table 1 correlates the current status of each Virtex device with a corresponding speed file designation.

Table 1: Virtex Device Speed Grade Designations

	Speed	Speed Grade Designations						
Device	Advance	Preliminary	Production					
XCV50			-6, -5, -4					
XCV100			-6, -5, -4					
XCV150			-6, -5, -4					
XCV200			-6, -5, -4					
XCV300			-6, -5, -4					
XCV400			-6, -5, -4					
XCV600			-6, -5, -4					
XCV800			-6, -5, -4					
XCV1000			-6, -5, -4					

All specifications are subject to change without notice.



Virtex DC Characteristics

Absolute Maximum Ratings

Symbol	Description ⁽¹⁾		Units	
V _{CCINT}	Supply voltage relative to GND ⁽²⁾	Supply voltage relative to GND ⁽²⁾		
V _{CCO}	Supply voltage relative to GND ⁽²⁾		-0.5 to 4.0	V
V _{REF}	Input Reference Voltage	-0.5 to 3.6	V	
V	Input voltage relative to GND ⁽³⁾	Using V _{REF}	-0.5 to 3.6	V
V _{IN}		Internal threshold	-0.5 to 5.5	V
V _{TS}	Voltage applied to 3-state output		-0.5 to 5.5	V
V _{CC}	Longest Supply Voltage Rise Time from 1V-2.375V	50	ms	
T _{STG}	Storage temperature (ambient)	-65 to +150	°C	
TJ	Junction temperature ⁽⁴⁾	Plastic Packages	+125	°C

Notes:

- Stresses beyond those listed under Absolute Maximum Ratings can cause permanent damage to the device. These are stress
 ratings only, and functional operation of the device at these or any other conditions beyond those listed under Operating Conditions
 is not implied. Exposure to Absolute Maximum Ratings conditions for extended periods of time can affect device reliability.
- 2. Power supplies can turn on in any order.
- 3. For protracted periods (e.g., longer than a day), V_{IN} should not exceed V_{CCO} by more than 3.6 V.
- 4. For soldering guidelines and thermal considerations, see the "Device Packaging" information on www.xilinx.com.

Recommended Operating Conditions

Symbol	Description	Min	Max	Units	
V _{CCINT} ⁽¹⁾	Input Supply voltage relative to GND, $T_J = 0$ °C to +85°C	2.5 – 5%	2.5 + 5%	V	
VCCINT'''	Input Supply voltage relative to GND, $T_J = -40^{\circ}\text{C}$ to $+100^{\circ}\text{C}$	2.5 – 5%	2.5 + 5%	V	
V _{CCO} ⁽⁴⁾	Supply voltage relative to GND, T _J = 0 °C to +85°C	Commercial	1.4	3.6	V
	Supply voltage relative to GND, $T_J = -40^{\circ}C$ to $+100^{\circ}C$	Industrial	1.4	3.6	V
T _{IN}	Input signal transition time			250	ns

- Correct operation is guaranteed with a minimum V_{CCINT} of 2.375 V (Nominal V_{CCINT} -5%). Below the minimum value, all delay parameters increase by 3% for each 50-mV reduction in V_{CCINT} below the specified range.
- 2. At junction temperatures above those listed as Operating Conditions, delay parameters do increase. Please refer to the TRCE report.
- 3. Input and output measurement threshold is \sim 50% of V_{CC} .
- Min and Max values for V_{CCO} are I/O Standard dependant.



DC Characteristics Over Recommended Operating Conditions

Symbol	Description	1	Device	Min	Max	Units
V _{DRINT}	Data Retention V _{CCINT} Voltage		All	2.0		V
21	(below which configuration data can be	e lost)				
V_{DRIO}	Data Retention V _{CCO} Voltage (below which configuration data can be	e lost)	All	1.2		V
I _{CCINTQ}	Quiescent V _{CCINT} supply current ^(1,3)		XCV50		50	mA
			XCV100		50	mA
			XCV150		50	mA
			XCV200		75	mA
			XCV300		75	mA
			XCV400		75	mA
			XCV600		100	mA
			XCV800		100	mA
			XCV1000		100	mA
Iccoq	Quiescent V _{CCO} supply current ⁽¹⁾		XCV50		2	mA
			XCV100		2	mA
			XCV150		2	mA
			XCV200		2	mA
			XCV300		2	mA
			XCV400		2	mA
			XCV600		2	mA
			XCV800		2	mA
			XCV1000		2	mA
I _{REF}	V _{REF} current per V _{REF} pin		All		20	μΑ
ΙL	Input or output leakage current		All	-10	+10	μΑ
C _{IN}	Input capacitance (sample tested)	BGA, PQ, HQ, packages	All		8	pF
I _{RPU}	Pad pull-up (when selected) @ V _{in} = 0 tested)	V, V _{CCO} = 3.3 V (sample	All	Note (2)	0.25	mA
I _{RPD}	Pad pull-down (when selected) @ V _{in} =	= 3.6 V (sample tested)		Note (2)	0.15	mA

- 1. With no output current loads, no active input pull-up resistors, all I/O pins 3-stated and floating.
- 2. Internal pull-up and pull-down resistors guarantee valid logic levels at unconnected input pins. These pull-up and pull-down resistors do not guarantee valid logic levels when input pins are connected to other circuits.
- 3. Multiply I_{CCINTQ} limit by two for industrial grade.



Power-On Power Supply Requirements

Xilinx FPGAs require a certain amount of supply current during power-on to insure proper device operation. The actual current consumed depends on the power-on ramp rate of the power supply. This is the time required to reach the nominal power supply voltage of the device⁽¹⁾ from 0 V. The current is highest at the fastest suggested ramp rate (0 V to nominal voltage in 2 ms) and is lowest at the slowest allowed ramp rate (0 V to nominal voltage in 50 ms). For more details on power supply requirements, see Application Note XAPP158 on www.xilinx.com.

Product	Description ⁽²⁾	Current Requirement ^(1,3)		
Virtex Family, Commercial Grade	Minimum required current supply	500 mA		
Virtex Family, Industrial Grade	Minimum required current supply	2 A		

Notes:

- Ramp rate used for this specification is from 0 2.7 VDC. Peak current occurs on or near the internal power-on reset threshold of 1.0V and lasts for less than 3 ms.
- Devices are guaranteed to initialize properly with the minimum current available from the power supply as noted above.
- 3. Larger currents can result if ramp rates are forced to be faster.

DC Input and Output Levels

Values for V_{IL} and V_{IH} are recommended input voltages. Values for I_{OL} and I_{OH} are guaranteed output currents over the recommended operating conditions at the V_{OL} and V_{OH} test points. Only selected standards are tested. These are chosen to ensure that all standards meet their specifications. The selected standards are tested at minimum V_{CCO} for each standard with the respective V_{OL} and V_{OH} voltage levels shown. Other standards are sample tested.

Input/Output		V _{IL}	VI	Н	V _{OL}	V _{OH}	I _{OL}	I _{OH}
Standard	V, min	V, max	V, min	V, max	V, Max	V, Min	mA	mA
LVTTL ⁽¹⁾	- 0.5	0.8	2.0	5.5	0.4	2.4	24	-24
LVCMOS2	- 0.5	.7	1.7	5.5	0.4	1.9	12	-12
PCI, 3.3 V	- 0.5	44% V _{CCINT}	60% V _{CCINT}	V _{CCO} + 0.5	10% V _{CCO}	90% V _{CCO}	Note 2	Note 2
PCI, 5.0 V	- 0.5	0.8	2.0	5.5	0.55	2.4	Note 2	Note 2
GTL	- 0.5	V _{REF} - 0.05	V _{REF} + 0.05	3.6	0.4	n/a	40	n/a
GTL+	- 0.5	V _{REF} – 0.1	V _{REF} + 0.1	3.6	0.6	n/a	36	n/a
HSTL I ⁽³⁾	- 0.5	V _{REF} – 0.1	V _{REF} + 0.1	3.6	0.4	V _{CCO} - 0.4	8	-8
HSTL III	- 0.5	V _{REF} – 0.1	V _{REF} + 0.1	3.6	0.4	V _{CCO} - 0.4	24	-8
HSTL IV	- 0.5	V _{REF} – 0.1	V _{REF} + 0.1	3.6	0.4	V _{CCO} - 0.4	48	-8
SSTL3 I	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	V _{REF} - 0.6	V _{REF} + 0.6	8	-8
SSTL3 II	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	V _{REF} - 0.8	V _{REF} + 0.8	16	-16
SSTL2 I	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	V _{REF} - 0.61	V _{REF} + 0.61	7.6	-7.6
SSTL2 II	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	V _{REF} - 0.80	V _{REF} + 0.80	15.2	-15.2
CTT	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	V _{REF} - 0.4	V _{REF} + 0.4	8	-8
AGP	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	10% V _{CCO}	90% V _{CCO}	Note 2	Note 2

- V_{OL} and V_{OH} for lower drive currents are sample tested.
- 2. Tested according to the relevant specifications.
- DC input and output levels for HSTL18 (HSTL I/O standard with V_{CCO} of 1.8 V) are provided in an HSTL white paper on www.xilinx.com.



Virtex Switching Characteristics

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE

in the Xilinx Development System) and back-annotated to the simulation net list. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature). Values apply to all Virtex devices unless otherwise noted.

IOB Input Switching Characteristics

Input delays associated with the pad are specified for LVTTL levels. For other standards, adjust the delays with the values shown in , page 6.

				Speed	Grade		
Description	Device	Symbol	Min	-6	-5	-4	Units
Propagation Delays							
Pad to I output, no delay	All	T _{IOPI}	0.39	0.8	0.9	1.0	ns, max
Pad to I output, with delay	XCV50	T _{IOPID}	0.8	1.5	1.7	1.9	ns, max
	XCV100		0.8	1.5	1.7	1.9	ns, max
	XCV150		0.8	1.5	1.7	1.9	ns, max
	XCV200		0.8	1.5	1.7	1.9	ns, max
	XCV300		0.8	1.5	1.7	1.9	ns, max
	XCV400		0.9	1.8	2.0	2.3	ns, max
	XCV600		0.9	1.8	2.0	2.3	ns, max
	XCV800		1.1	2.1	2.4	2.7	ns, max
	XCV1000		1.1	2.1	2.4	2.7	ns, max
Pad to output IQ via transparent latch, no delay	All	T _{IOPLI}	0.8	1.6	1.8	2.0	ns, max
Pad to output IQ via transparent	XCV50	T _{IOPLID}	1.9	3.7	4.2	4.8	ns, max
latch, with delay	XCV100		1.9	3.7	4.2	4.8	ns, max
	XCV150		2.0	3.9	4.3	4.9	ns, max
	XCV200		2.0	4.0	4.4	5.1	ns, max
	XCV300		2.0	4.0	4.4	5.1	ns, max
	XCV400		2.1	4.1	4.6	5.3	ns, max
	XCV600		2.1	4.2	4.7	5.4	ns, max
	XCV800		2.2	4.4	4.9	5.6	ns, max
	XCV1000		2.3	4.5	5.1	5.8	ns, max
Sequential Delays			·				
Clock CLK	All						
Minimum Pulse Width, High		T _{CH}	0.8	1.5	1.7	2.0	ns, min
Minimum Pulse Width, Low		T _{CL}	0.8	1.5	1.7	2.0	ns, min
Clock CLK to output IQ		T _{IOCKIQ}	0.2	0.7	0.7	8.0	ns, max



			Speed Grade				
Description	Device	Symbol	Min	-6	-5	-4	Units
Setup and Hold Times with respect to Clock CLK at IOB input register ⁽¹⁾		CLK at IOB input		Setup	Time / Hol	d Time	
Pad, no delay	All	T _{IOPICK} /T _{IOICKP}	0.8 / 0	1.6 / 0	1.8 / 0	2.0 / 0	ns, min
Pad, with delay	XCV50	T _{IOPICKD} /T _{IOICKPD}	1.9 / 0	3.7 / 0	4.1 / 0	4.7 / 0	ns, min
	XCV100		1.9 / 0	3.7 / 0	4.1 / 0	4.7 / 0	ns, min
	XCV150		1.9 / 0	3.8 / 0	4.3 / 0	4.9 / 0	ns, min
	XCV200		2.0 / 0	3.9 / 0	4.4 / 0	5.0 / 0	ns, min
	XCV300		2.0 / 0	3.9 / 0	4.4 / 0	5.0 / 0	ns, min
	XCV400		2.1 / 0	4.1 / 0	4.6 / 0	5.3 / 0	ns, min
	XCV600		2.1 / 0	4.2 / 0	4.7 / 0	5.4 / 0	ns, min
	XCV800		2.2 / 0	4.4 / 0	4.9 / 0	5.6 / 0	ns, min
	XCV1000		2.3 / 0	4.5 / 0	5.0 / 0	5.8 / 0	ns, min
ICE input	All	T _{IOICECK} /T _{IOCKICE}	0.37/ 0	0.8 / 0	0.9 / 0	1.0 / 0	ns, max
Set/Reset Delays							
SR input (IFF, synchronous)	All	T _{IOSRCKI}	0.49	1.0	1.1	1.3	ns, max
SR input to IQ (asynchronous)	All	T _{IOSRIQ}	0.70	1.4	1.6	1.8	ns, max
GSR to output IQ	All	T _{GSRQ}	4.9	9.7	10.9	12.5	ns, max

^{1.} A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values cannot be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.

^{2.} Input timing for LVTTL is measured at 1.4 V. For other I/O standards, see Table 3.



IOB Output Switching Characteristics Standard Adjustments

Output delays terminating at a pad are specified for LVTTL with 12 mA drive and fast slew rate. For other standards, adjust the delays by the values shown.

				Speed	Grade		Unit
Description	Symbol	Standard ⁽¹⁾	Min	-6	-5	-4	s
Output Delay Adjustments							
Standard-specific adjustments for	T _{OLVTTL_S2}	LVTTL, Slow, 2 mA	4.2	14.7	15.8	17.0	ns
output delays terminating at pads (based on standard capacitive load,	T _{OLVTTL_S4}	4 mA	2.5	7.5	8.0	8.6	ns
Csl)	T _{OLVTTL_S6}	6 mA	1.8	4.8	5.1	5.6	ns
	T _{OLVTTL_S8}	8 mA	1.2	3.0	3.3	3.5	ns
	T _{OLVTTL_S12}	12 mA	1.0	1.9	2.1	2.2	ns
	T _{OLVTTL_S16}	16 mA	0.9	1.7	1.9	2.0	ns
	T _{OLVTTL_S24}	24 mA	0.8	1.3	1.4	1.6	ns
	T _{OLVTTL_F2}	LVTTL, Fast, 2mA	1.9	13.1	14.0	15.1	ns
	T _{OLVTTL_F4}	4 mA	0.7	5.3	5.7	6.1	ns
	T _{OLVTTL_F6}	6 mA	0.2	3.1	3.3	3.6	ns
	T _{OLVTTL_F8}	8 mA	0.1	1.0	1.1	1.2	ns
	T _{OLVTTL_F12}	12 mA	0	0	0	0	ns
	T _{OLVTTL_F16}	16 mA	-0.10	-0.05	-0.05	-0.05	ns
	T _{OLVTTL_F24}	24 mA	-0.10	-0.20	-0.21	-0.23	ns
	T _{OLVCMOS2}	LVCMOS2	0.10	0.10	0.11	0.12	ns
	T _{OPCl33_3}	PCI, 33 MHz, 3.3 V	0.50	2.3	2.5	2.7	ns
	T _{OPCl33_5}	PCI, 33 MHz, 5.0 V	0.40	2.8	3.0	3.3	ns
	T _{OPCI66_3}	PCI, 66 MHz, 3.3 V	0.10	-0.40	-0.42	-0.46	ns
	T _{OGTL}	GTL	0.6	0.50	0.54	0.6	ns
	T _{OGTLP}	GTL+	0.7	0.8	0.9	1.0	ns
	T _{OHSTL_I}	HSTL I	0.10	-0.50	-0.53	-0.5	ns
	T _{OHSTL_III}	HSTL III	-0.10	-0.9	-0.9	-1.0	ns
	T _{OHSTL_IV}	HSTL IV	-0.20	-1.0	-1.0	-1.1	ns
	T _{OSSTL2_I}	SSTL2 I	-0.10	-0.50	-0.53	-0.5	ns
	T _{OSSLT2_II}	SSTL2 II	-0.20	-0.9	-0.9	-1.0	ns
	T _{OSSTL3_I}	SSTL3 I	-0.20	-0.50	-0.53	-0.5	ns
	T _{OSSTL3_II}	SSTL3 II	-0.30	-1.0	-1.0	-1.1	ns
	T _{OCTT}	СТТ	0	-0.6	-0.6	-0.6	ns
	T _{OAGP}	AGP	0	-0.9	-0.9	-1.0	ns

^{1.} Output timing is measured at 1.4 V with 35 pF external capacitive load for LVTTL. For other I/O standards and different loads, see Table 2 and Table 3.



Calculation of T_{ioop} as a Function of Capacitance

 T_{ioop} is the propagation delay from the O Input of the IOB to the pad. The values for T_{ioop} were based on the standard capacitive load (CsI) for each I/O standard as listed in Table 2.

Table 2: Constants for Calculating T_{ioop}

Standard	Csl (pF)	fl (ns/pF)
LVTTL Fast Slew Rate, 2mA drive	35	0.41
LVTTL Fast Slew Rate, 4mA drive	35	0.20
LVTTL Fast Slew Rate, 6mA drive	35	0.13
LVTTL Fast Slew Rate, 8mA drive	35	0.079
LVTTL Fast Slew Rate, 12mA drive	35	0.044
LVTTL Fast Slew Rate, 16mA drive	35	0.043
LVTTL Fast Slew Rate, 24mA drive	35	0.033
LVTTL Slow Slew Rate, 2mA drive	35	0.41
LVTTL Slow Slew Rate, 4mA drive	35	0.20
LVTTL Slow Slew Rate, 6mA drive	35	0.100
LVTTL Slow Slew Rate, 8mA drive	35	0.086
LVTTL Slow Slew Rate, 12mA drive	35	0.058
LVTTL Slow Slew Rate, 16mA drive	35	0.050
LVTTL Slow Slew Rate, 24mA drive	35	0.048
LVCMOS2	35	0.041
PCI 33MHz 5V	50	0.050
PCI 33MHZ 3.3 V	10	0.050
PCI 66 MHz 3.3 V	10	0.033
GTL	0	0.014
GTL+	0	0.017
HSTL Class I	20	0.022
HSTL Class III	20	0.016
HSTL Class IV	20	0.014
SSTL2 Class I	30	0.028
SSTL2 Class II	30	0.016
SSTL3 Class I	30	0.029
SSTL3 Class II	30	0.016
СТТ	20	0.035
AGP	10	0.037

Notes:

- I/O parameter measurements are made with the capacitance values shown above. See Application Note XAPP133 on <u>www.xilinx.com</u> for appropriate terminations.
- I/O standard measurements are reflected in the IBIS model information except where the IBIS format precludes it.

For other capacitive loads, use the formulas below to calculate the corresponding T_{ioop} .

$$T_{ioop} = T_{ioop} + T_{opadjust} + (C_{load} - C_{sl}) * fl$$

Where:

 $T_{opadjust}$ is reported above in the Output Delay Adjustment section.

C_{load} is the capacitive load for the design.

Table 3: Delay Measurement Methodology

Standard	ν _L ⁽¹⁾	V _H ⁽¹⁾	Meas. Point	V _{REF} Typ ⁽²⁾
LVTTL	0	3	1.4	-
LVCMOS2	0	2.5	1.125	-
PCI33_5	Pe	er PCI Spec		-
PCl33_3	Pe	er PCI Spec		-
PCI66_3	Pe	er PCI Spec		-
GTL	V _{REF} -0.2	V _{REF} +0.2	V _{REF}	0.80
GTL+	V _{REF} -0.2	V _{REF} +0.2	V_{REF}	1.0
HSTL Class I	V _{REF} -0.5	V _{REF} +0.5	V _{REF}	0.75
HSTL Class III	V _{REF} -0.5	V _{REF} +0.5	V _{REF}	0.90
HSTL Class IV	V _{REF} -0.5	V _{REF} +0.5	V _{REF}	0.90
SSTL3 I & II	V _{REF} -1.0	V _{REF} +1.0	V _{REF}	1.5
SSTL2 I & II	V _{REF} -0.75	V _{REF} +0.75	V _{REF}	1.25
CTT	V _{REF} -0.2	V _{REF} +0.2	V_{REF}	1.5
AGP	V _{REF} – (0.2xV _{CCO})	V _{REF} + (0.2xV _{CCO})	V _{REF}	Per AGP Spec

- Input waveform switches between V_Land V_H.
- 2. Measurements are made at VREF (Typ), Maximum, and Minimum. Worst-case values are reported.
- I/O parameter measurements are made with the capacitance values shown in Table 2. See Application Note XAPP133 on www.xilinx.com for appropriate terminations.
- 4. I/O standard measurements are reflected in the IBIS model information except where the IBIS format precludes it.



Clock Distribution Guidelines

			Speed Grade			
Description	Device	Symbol	-6	-5	-4	Units
Global Clock Skew ⁽¹⁾						
Global Clock Skew between IOB Flip-flops	XCV50	T _{GSKEWIOB}	0.10	0.12	0.14	ns, max
	XCV100		0.12	0.13	0.15	ns, max
	XCV150		0.12	0.13	0.15	ns, max
	XCV200		0.13	0.14	0.16	ns, max
	XCV300		0.14	0.16	0.18	ns, max
	XCV400		0.13	0.13	0.14	ns, max
	XCV600		0.14	0.15	0.17	ns, max
	XCV800		0.16	0.17	0.20	ns, max
	XCV1000		0.20	0.23	0.25	ns, max

Notes:

Clock Distribution Switching Characteristics

		Speed Grade				
Description	Symbol	Min	-6	- 5	-4	Units
GCLK IOB and Buffer						
Global Clock PAD to output.	T _{GPIO}	0.33	0.7	0.8	0.9	ns, max
Global Clock Buffer I input to O output	T _{GIO}	0.34	0.7	0.8	0.9	ns, max

^{1.} These clock-skew delays are provided for guidance only. They reflect the delays encountered in a typical design under worst-case conditions. Precise values for a particular design are provided by the timing analyzer.



CLB Switching Characteristics

Delays originating at F/G inputs vary slightly according to the input used. The values listed below are worst-case. Precise values are provided by the timing analyzer.

		Speed Grade				
Description	Symbol	Min	-6	-5	-4	Units
Combinatorial Delays		•				
4-input function: F/G inputs to X/Y outputs	T _{ILO}	0.29	0.6	0.7	0.8	ns, max
5-input function: F/G inputs to F5 output	T _{IF5}	0.32	0.7	0.8	0.9	ns, max
5-input function: F/G inputs to X output	T _{IF5X}	0.36	0.8	0.8	1.0	ns, max
6-input function: F/G inputs to Y output via F6 MUX	T _{IF6Y}	0.44	0.9	1.0	1.2	ns, max
6-input function: F5IN input to Y output	T _{F5INY}	0.17	0.32	0.36	0.42	ns, max
Incremental delay routing through transparent latch to XQ/YQ outputs	T _{IFNCTL}	0.31	0.7	0.7	0.8	ns, max
BY input to YB output	T _{BYYB}	0.27	0.53	0.6	0.7	ns, max
Sequential Delays						1
FF Clock CLK to XQ/YQ outputs	T _{CKO}	0.54	1.1	1.2	1.4	ns, max
Latch Clock CLK to XQ/YQ outputs	T _{CKLO}	0.6	1.2	1.4	1.6	ns, max
Setup and Hold Times before/after Clock CLK ⁽¹⁾		Setup T	ime / Hol	d Time		
4-input function: F/G Inputs	T _{ICK} /T _{CKI}	0.6 / 0	1.2 / 0	1.4 / 0	1.5 / 0	ns, min
5-input function: F/G inputs	T _{IF5CK} /T _{CKIF5}	0.7 / 0	1.3 / 0	1.5 / 0	1.7 / 0	ns, min
6-input function: F5IN input	T _{F5INCK} /T _{CKF5IN}	0.46 / 0	1.0 / 0	1.1 / 0	1.2 / 0	ns, min
6-input function: F/G inputs via F6 MUX	T _{IF6CK} /T _{CKIF6}	0.8 / 0	1.5 / 0	1.7 / 0	1.9 / 0	ns, min
BX/BY inputs	T_{DICK}/T_{CKDI}	0.30 / 0	0.6 / 0	0.7 / 0	0.8 / 0	ns, min
CE input	T_{CECK}/T_{CKCE}	0.37 / 0	0.8 / 0	0.9 / 0	1.0 / 0	ns, min
SR/BY inputs (synchronous)	$T_{RCK}T_{CKR}$	0.33 / 0	0.7 / 0	0.8 / 0	0.9 / 0	ns, min
Clock CLK						
Minimum Pulse Width, High	T _{CH}	0.8	1.5	1.7	2.0	ns, min
Minimum Pulse Width, Low	T_CL	0.8	1.5	1.7	2.0	ns, min
Set/Reset						
Minimum Pulse Width, SR/BY inputs	T _{RPW}	1.3	2.5	2.8	3.3	ns, min
Delay from SR/BY inputs to XQ/YQ outputs (asynchronous)	T _{RQ}	0.54	1.1	1.3	1.4	ns, max
Delay from GSR to XQ/YQ outputs	T _{IOGSRQ}	4.9	9.7	10.9	12.5	ns, max
Toggle Frequency (MHz) (for export control)	F _{TOG} (MHz)	625	333	294	250	MHz

^{1.} A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values cannot be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.



Virtex Pin-to-Pin Output Parameter Guidelines

All devices are 100% functionally tested. Listed below are representative values for typical pin locations and normal clock loading. Values are expressed in nanoseconds unless otherwise noted.

Global Clock Input to Output Delay for LVTTL, 12 mA, Fast Slew Rate, with DLL

				Speed	Grade		
Description	Symbol	Device	Min	-6	-5	-4	Units
LVTTL Global Clock Input to Output Delay using	T _{ICKOFDLL}	XCV50	1.0	3.1	3.3	3.6	ns, max
Output Flip-flop, 12 mA, Fast Slew Rate, with DLL. For data output with different standards, adjust delays with the values shown in Output Delay		XCV100	1.0	3.1	3.3	3.6	ns, max
		XCV150	1.0	3.1	3.3	3.6	ns, max
Adjustments.		XCV200	1.0	3.1	3.3	3.6	ns, max
		XCV300	1.0	3.1	3.3	3.6	ns, max
		XCV400	1.0	3.1	3.3	3.6	ns, max
		XCV600	1.0	3.1	3.3	3.6	ns, max
		XCV800	1.0	3.1	3.3	3.6	ns, max
		XCV1000	1.0	3.1	3.3	3.6	ns, max

Notes:

- 1. Listed above are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.
- Output timing is measured at 1.4 V with 35 pF external capacitive load for LVTTL. The 35 pF load does not apply to the Min values. For other I/O standards and different loads, see Table 2 and Table 3.
- 3. DLL output jitter is already included in the timing calculation.

Global Clock Input-to-Output Delay for LVTTL, 12 mA, Fast Slew Rate, without DLL

			Speed Grade				
Description	Symbol	Device	Min	-6	-5	-4	Units
LVTTL Global Clock Input to Output Delay using Output Flip-flop, 12 mA, Fast Slew Rate, without DLL. For data output with different standards, adjust delays with the values shown in Input and Output	T _{ICKOF}	XCV50	1.5	4.6	5.1	5.7	ns, max
		XCV100	1.5	4.6	5.1	5.7	ns, max
		XCV150	1.5	4.7	5.2	5.8	ns, max
Delay Adjustments. For I/O standards requiring V _{RFF} , such as GTL,		XCV200	1.5	4.7	5.2	5.8	ns, max
GTL+, SSTL, HSTL, CTT, and AGO, an additional		XCV300	1.5	4.7	5.2	5.9	ns, max
600 ps must be added.		XCV400	1.5	4.8	5.3	6.0	ns, max
		XCV600	1.6	4.9	5.4	6.0	ns, max
		XCV800	1.6	4.9	5.5	6.2	ns, max
		XCV1000	1.7	5.0	5.6	6.3	ns, max

- Listed above are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.
- 2. Output timing is measured at 1.4 V with 35 pF external capacitive load for LVTTL. The 35 pF load does not apply to the Min values. For other I/O standards and different loads, see Table 2 and Table 3.



Global Clock Set-Up and Hold for LVTTL Standard, without DLL

				Speed	Grade		
Description	Symbol	Device	Min	-6	-5	-4	Units
Input Setup and Hold Time Relat standards, adjust the setup time of					For data inp	ut with diffe	rent
Full Delay Global Clock and IFF, without	T _{PSFD} /T _{PHFD}	XCV50	0.6 / 0	2.3 / 0	2.6 / 0	2.9 / 0	ns, min
DLL		XCV100	0.6 / 0	2.3 / 0	2.6 / 0	3.0 / 0	ns, min
		XCV150	0.6 / 0	2.4 / 0	2.7 / 0	3.1 / 0	ns, min
		XCV200	0.7 / 0	2.5 / 0	2.8 / 0	3.2 / 0	ns, min
		XCV300	0.7 / 0	2.5 / 0	2.8 / 0	3.2 / 0	ns, min
		XCV400	0.7 / 0	2.6 / 0	2.9 / 0	3.3 / 0	ns, min
		XCV600	0.7 / 0	2.6 / 0	2.9 / 0	3.3 / 0	ns, min
		XCV800	0.7 / 0	2.7 / 0	3.1 / 0	3.5 / 0	ns, min
		XCV1000	0.7 / 0	2.8 / 0	3.1 / 0	3.6 / 0	ns, min

IFF = Input Flip-Flop or Latch

Notes: Notes:

- 1. Set-up time is measured relative to the Global Clock input signal with the fastest route and the lightest load. Hold time is measured relative to the Global Clock input signal with the slowest route and heaviest load.
- 2. A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.



Table 2: Virtex Pinout Tables (Chip-Scale and QFP Packages) (Continued)

Pin Name	Device	CS144	TQ144	PQ/HQ240
V _{cco}	All	Banks 0 and 1: A2, A13, D7 Banks 2 and 3: B12, G11, M13 Banks 4 and 5: N1, N7, N13 Banks 6 and 7: B2, G2, M2	No I/O Banks in this package: 1, 17, 37, 55, 73, 92, 109, 128	No I/O Banks in this package: 15, 30, 44, 61, 76, 90, 105, 121, 136, 150, 165, 180, 197, 212, 226, 240
V _{RFF} Bank 0	XCV50	C4, D6	5, 13	218, 232
(V _{REF} pins are listed	XCV100/150	+ B4	+ 7	+ 229
incrementally. Connect	XCV200/300	N/A	N/A	+ 236
all pins listed for both the required device	XCV400	N/A	N/A	+ 215
and all smaller devices	XCV600	N/A	N/A	+ 230
listed in the same package.)	XCV800	N/A	N/A	+ 222
Within each bank, if input reference voltage is not required, all V _{REF} pins are general I/O.				
V _{REF} , Bank 1	XCV50	A10, B8	22, 30	191, 205
(V _{REF} pins are listed	XCV100/150	+ D9	+ 28	+ 194
incrementally. Connect all pins listed for both	XCV200/300	N/A	N/A	+ 187
the required device	XCV400	N/A	N/A	+ 208
and all smaller devices listed in the same	XCV600	N/A	N/A	+ 193
package.) Within each bank, if input reference voltage is not required, all V _{REF} pins are general I/O.	XCV800	N/A	N/A	+ 201
V _{REF} , Bank 2	XCV50	D11, F10	42, 50	157, 171
(V _{REF} pins are listed	XCV100/150	+ D13	+ 44	+ 168
incrementally. Connect all pins listed for both	XCV200/300	N/A	N/A	+ 175
the required device	XCV400	N/A	N/A	+ 154
and all smaller devices listed in the same	XCV600	N/A	N/A	+ 169
package.) Within each bank, if input reference voltage is not required, all V _{REF} pins are general I/O.	XCV800	N/A	N/A	+ 161